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ELECTRICAL
CHARACTERISTICS
of
TRANSMISSION CIRCUITS

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ELECTRICAL CHARACTERISTICS OF TRANSMISSION CIRCUITS

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PREFACE

THE rapid expansion in distributing and transmission systems will continue unabated until the natural power resources will have been fully developed. This expansion will necessitate a tremendous amount of arithmetical labor in connection with the proper solution and calculation of performance of projected transmission and distributing circuits. It will demand much valuable time and energy in the education of the younger engineers now going thru the technical schools and others who will follow them. It was primarily to assist these younger engineers by making their work more easy and less liable to error, and providing them with all necessary tools that the data in this book have been compiled.

Many articles each pertaining to some particular method of solution of transmission circuits have been published from time to time. This book constitutes a review of each of numerous methods perviously proposed by different authors with examples illustrating each method of solution and the accuracy which may be expected by its use. Thus by permission of various authors the reader of this book is provided with a choice of numerous methods ranging between the most simplified graphical forms of solutions and complete mathematical solutions. He is also provided with numerous and extensive tables of circuit and other constants making it unnecessary for him to lose time and risk making mistakes in calculating constants for each case in question. Much effort has been expended with a view of simplifying explanations by the aid of supplementary diagrams and tabulations. The engineer upon whose lot it only occasionally falls to determine the size of conductors and performances of circuits appreciates how easy it is to make errors in calculations which may prove very serious and should find the quick estimating tables very useful particularly for short line solutions.

For those preferring to avoid the more mathematical solutions the all graphical methods for solving long line problems including the Wilkinson & Kennelly charts for obtaining graphically the auxiliary constants should prove helpful.

When borrowed material has been used in this book full credit has been given the author at the place the material is used. It is desired, however, at this place to mention the high appreciation of assistance given by Ralph W. Atkinson, Herbert B. Dwight, Dr. A. E. Kennelly, Dr. A. S. McAllister, Ralph D. Mershon, F. W. Peak Jr., J. F. Peters, Charles R. Riker and T. A. Wilkinson.

Wm Nesbit

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ELECTRICAL CHARACTERISTICS *of* TRANSMISSION CIRCUITS

CHAPTER I RESISTANCE—SKIN EFFECT—INDUCTANCE

THE transmission of alternating-current power involves three separate circuits, one of which is composed of the wires forming the transmission line, while the others lie in the medium surrounding the wires. The constants of these circuits are interdependent; although any one may vary greatly from the others in magnitude.* There is first the electric circuit through the conductors. Then since all magnetic and dielectric lines of force are closed upon themselves forming complete circuits there is a magnetic and a dielectric circuit. The magnetic circuit consists of magnetic lines of force encircling the current carrying conductors and the dielectric circuit the dielectric lines of force terminating in the current carrying conductors. The close analogy of these is given in Table A, a careful study of which will help those not familiar with the subject to a clearer understanding of what happens in an alternating-current transmission circuit.

For a unidirectional constant current the magnetic field remains constant, and similarly for a unidirectional constant voltage the dielectric field is constant. With both the current and the voltage unidirectional and constant, the electric circuit alone enters into the calculations. A changing magnetic flux introduces a voltage into the electric circuit which modifies the initial or impressed voltage. This effect of the magnetic circuit, which is measured by the inductance L , storing the energy $0.5i^2L$, is a function of the current, and hence is of most importance in dealing with heavy current circuits. Similarly a changing electrostatic flux adds

(vectorially) a current to the main power current. This effect of the dielectric circuit, which is measured by the capacitance, storing the energy $0.5e^2C$, is a function of the voltage, and hence is of most importance in dealing with high-voltage circuits.

In an alternating-current circuit, both the voltage and the current are continually varying in magnitude, and moreover, reversing in direction for each successive half cycle. Therefore, with alternating currents, energy changes occur continuously and simultaneously in the interlinked magnetic, dielectric and electric circuits.

Figs. 1 to 5 inclusive illustrate the magnetic and dielectric field surrounding conductors carrying current. Figs. 1 and 3 represent respectively the magnetic and dielectric circuits when the conductors are far apart and Figs. 2 and 4 when they are close together. Fig. 5 represents the resultant of the superimposed magnetic and dielectric fields.

The magnetic field surrounding a conductor which is not influenced by any other field is represented by concentric circles. This field is strongest at the surface of the conductors and rapidly decreases with increasing distance from the conductor as indicated by the spacing of the lines of Figs. 1 and 2.

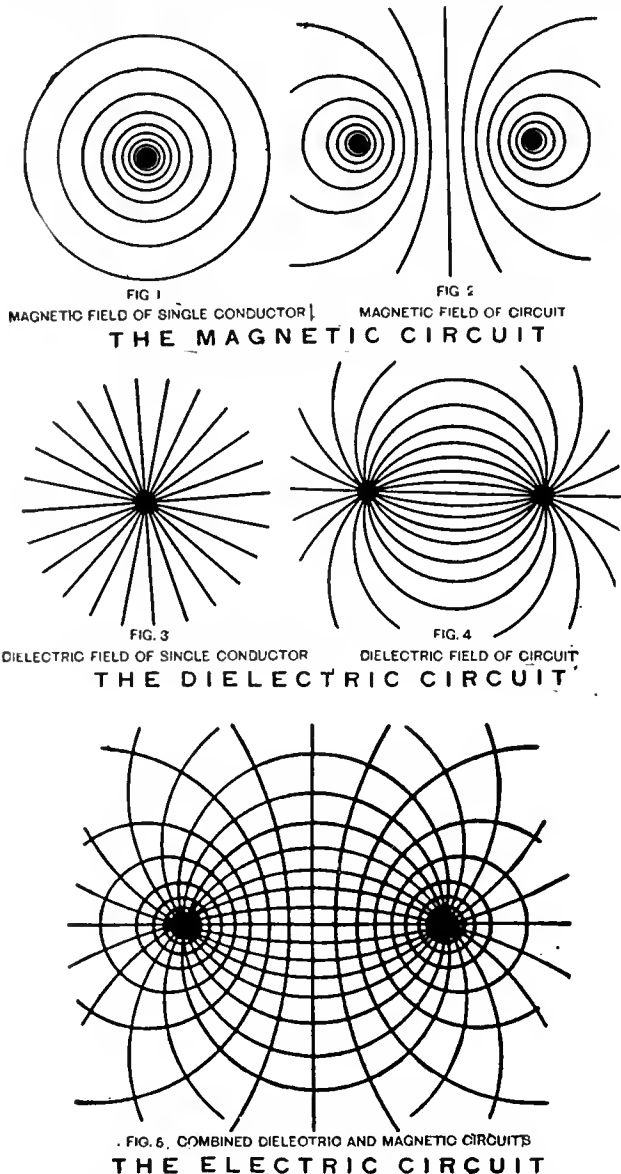
The dielectric stresses surrounding conductors are represented by lines drawn radially from the conductor. The strength of the dielectric field likewise decreases with the distance from the conductor as is indicated by the widening of the space between the lines. The magnetic and the dielectric lines of force always cross each other at right angles, as shown in Fig. 5.

*For a further description of these circuits see "Alternating Currents" by Prof. Carl E. Magnusson, from which Figs. 1 to 5 are reproduced with the permission of the author.

RESISTANCE OF COPPER CONDUCTORS

In Table I the resistance per thousand feet is listed and in Table II per mile of single conductor. Values are given for both solid and stranded copper conductors at both 100 and 97.3 percent conductivity and corresponding to various temperatures between zero and 75 degrees C. The foot notes with these tables cover all of the pertinent data upon which the values are based.

The resistance values in Table I corresponding to temperatures of 25 and 65 degrees C. were taken from



THE MAGNETIC CIRCUIT

THE DIELECTRIC CIRCUIT

THE ELECTRIC CIRCUIT

Bulletin 31 of the Bureau of Standards issued April 1st, 1912. The resistance values (taking into account the expansion of the metal with rise in temperature) for the other temperatures were calculated in accordance with the following rule from page 10 of Bulletin No. 31.

The change of resistivity of copper per degree C. is a constant, independent of the temperature of reference and of the sample of copper. This resistivity-temperature constant may be taken for general purposes as 0.0409 ohm (mil foot).

As an illustration:—A 2 000 000 circ. mil stranded copper conductor at 100 percent conductivity, has a resistance of 0.00623 ohm per 1000 feet at 65 degrees C. Required to calculate its resistance at zero degrees C.

$65 \times 0.0409 = 2.6585$ ohms (mil-foot) temperature correction or 2658.5 ohms (mil, 1000 feet).

$\frac{2658.5}{2\,000\,000} = 0.00133$ ohm change in resistance. $0.00623 - 0.00133 = 0.0049$ ohm resistance at zero degrees C.

It has been customary to publish tables of resistance values based upon a temperature of 20 degrees C. and 100 percent conductivity. The operating temperatures of conductors carrying current is usually considerably higher than 20 degrees C. and therefore calculations based upon this temperature do not often represent operating conditions. Neither does copper of 100 percent conductivity represent the usual condition for transmission circuit copper, whose average conductivity is probably nearer 97.3 percent. The values in Tables I and II furnish a comparison of resistance for annealed and hard drawn copper of stranded and solid conductors at various temperatures based upon the new "Annealed Copper Standard".

SKIN EFFECT

A solid conductor may be considered as made up of separate filaments, just as a piece of wood is made up of separate fibres. As a stranded conductor is actually made up of a number of separate wires, such a conductor will be considered in the following explanation. The inductance of the various wires of the cable will be different, due to the fact that those wires near the center of the cable will be linked by more flux lines than are the wires near the outer surface. The self-induced back e.m.f. will therefore be greater in the wires located near the center of the cable. The higher reactance of the inner wires causes the current to distribute in such a manner that the current density will be less in the interior than at the surface. This crowding of the current to the surface or "skin" of the wire is known as "skin effect".

Since the self-induced e.m.f. is proportional to the frequency as well as to the total flux linked, the skin effect becomes more pronounced at higher frequencies of the impressed e.m.f. It also becomes greater the larger the cross-section, the greater the conductivity and the greater the permeability of the conductor.

As a result the effective resistance of a conductor to alternating current is greater than to direct current. The effective resistance of nonmagnetic conductors to alternating current may be obtained by increasing their direct-current resistances by the percentages in Table B, which were derived by the formulas in Pender's Handbook. Thus the ohmic resistance of a 1 000 000 circ. mil cable is approximately 8.4 percent greater at 60 cycles than its resistance to direct current at a temperature of 25°C. If the temperature of the conductor is 65°C, its 60 cycle ohmic resistance will be approximately 6.4 percent greater than its direct-current resistance. The practical result of skin effect is to reduce the carrying capacity of large cables. As indicated by the values in Table B, skin effect may be neglected when employing non-magnetic conductors ex-

cept in the use of very large diameters. It is usual to manufacture cables of very large diameter, especially for service at high frequencies, with a non-conducting core. In case of magnetic conductors, such as steel wire or cable, as is some times used for long spans or short high voltage feeders, skin effect must be carefully considered.*

viently large, a thousandth part of it, called the millihenry, is the usual practical unit. This unit is the coefficient of self-induction and is represented by the letter *L*.

DISTRIBUTION OF FLUX

When current flows through a conductor, a magnetomotive force (m.m.f.) is established of a value proportional to the current. This m.m.f. is of zero value at the center of the conductor and increases as the square of the distance from the center until the surface is reached. (This statement as well as those following is based upon the assumption of a uniform distribution of current throughout the conductor, the conductor being of non-magnetic material and located in non-magnetic

TABLE A—COMPARISON OF THE THREE CIRCUITS

THE ELECTRIC CIRCUIT	THE MAGNETIC CIRCUIT	THE DIELECTRIC CIRCUIT
Current <i>I</i> Voltage $E=RI$ Electric Power	Magnetic Flux ϕ Magnetomotive Force $F=ni$ Magnetic Energy	Dielectric Flux ψ Electromotive Force $E=Q/C$ Dielectric Energy
Resistivity Resistance $R=W/I^2$	Reluctivity Reluctance <i>R</i> Inductance $L=\phi/i$	Elastivity $1/K$ Elastance <i>S</i> Capacitance $C=\psi/E$
Impedance $z = \sqrt{r^2 + x^2}$		
Conductivity γ Conductance $\left\{ \begin{array}{l} g=W/E^2 \\ g=r/z^2 \end{array} \right.$	Permeability $\mu = B/H$ Permeance $M = \phi/4\pi F$ Susceptance $b=x/z^2$	Permittivity <i>K</i> Permittance (Capacitance) <i>C</i>
Admittance $y=1/z=g+jb = \sqrt{g^2 + b^2}$		

INDUCTANCE

Any moving mass, for instance a flywheel in motion, will resist a change in velocity. That is, the inertia of the moving mass will tend to keep the mass moving when disconnected from the source of power. On the other hand the inertia will oppose any effort to speed up the movement of the mass.

In a similar manner, the inductance of an electric circuit resists a change in current. The cause of inductance in an electric circuit is the magnetic field which surrounds the circuit. When the current changes this magnetic field changes correspondingly, and in effect cuts the conductor, producing an e.m.f. in it. This e.m.f. of self induction has such a direction as to resist the change in current. While the current is increasing, energy is stored in the magnetic field and while the current decreases, the magnetic stored energy is returned to the electric circuit. This effect of the electric current on the surrounding space is termed magnetic induction.

Unit of Inductance—When a rate of change of current of one ampere per second produces an e.m.f. of one volt, the circuit is said to have a unit of inductance called a henry. The henry being incon-

surroundings, such as air). At the surface it becomes maximum for a given current and remains at this maximum value for all distances beyond the surface. It is customary to think of the magnetic field surrounding conductors as concentric circles of lines of force.

A physical picture of the magnetic field density surrounding a current carrying conductor A is shown by Chart I. The magnetic density due to the return circuit (conductor B) is indicated in outline by broken lines. The horizontal divisions represent the distance from the center of conductor A and the height of the

TABLE B—INCREASE OF EFFECTIVE RESISTANCE DUE TO SKIN EFFECT.

For various sizes of solid copper rods. For stranded conductors of equivalent cross sectional area the skin effect is practically the same as for the solid conductor.

Area in Circ. Mils.	Diameter in Inches of Stranded Conductor	Diameter in Inches of Solid Rod	Percent Increase of Copper Wires Above the Direct-Current Resistance Due to Alternating-Currents of Different Frequencies									
			Based Upon Direct-Current Resistance at 25 Degrees C. (77 Degrees F.)					Based Upon Direct-Current Resistance at 65 Degrees C. (149 Degrees F.)				
			15 Cycles	25 Cycles	40 Cycles	60 Cycles	133 Cycles	15 Cycles	25 Cycles	40 Cycles	60 Cycles	133 Cycles
2 000 000	1.631	1.414	2.2	6.0	14.1	28.0	78.6	1.7	4.5	10.9	22.1	67.0
1 800 000	1.548	1.342	1.8	4.9	11.7	23.7	70.4	1.3	3.7	9.0	18.5	60.0
1 600 000	1.459	1.265	1.4	3.9	9.4	19.4	61.4	1.1	3.0	7.3	15.0	51.8
1 500 000	1.412	1.225	1.3	3.4	8.4	17.4	57.3	0.9	2.6	6.4	13.5	47.4
1 200 000	1.263	1.096	0.8	2.1	5.5	11.7	42.7	0.6	1.7	4.1	9.0	34.8
1 000 000	1.152	1.000	0.6	1.5	3.8	8.4	33.8	0.4	1.1	3.0	6.4	26.2
750 000	0.998	0.866	0.3	0.9	2.2	4.9	20.6	0.2	0.7	1.7	3.7	16.4
500 000	0.815	0.707	0.1	0.4	1.0	2.2	10.1	0.1	0.3	0.7	1.7	7.7
250 000	0.575	0.500	0.0	0.1	0.3	0.6	2.7	0.0	0.1	0.2	0.4	2.0

*References:—For a bibliography on the subject of skin effect see article "Experimental Researches on Skin Effect in Conductors" by A. E. Kennelly, F. A. Laws, and P. H. Pierce in *A. I. E. E. Trans.*, Vol. 34, Part II of Sept. 1915. This article ends with a bibliography on the subject embracing a very complete list of articles.

"Calculation of Skin Effect in Strap Conductors" by H. B. Dwight in *Electrical World*, March 11, 1916.

"Skin Effect in Tubular and Flat Conductors" by H. B. Dwight in *A. I. E. E. Trans.* for 1918.

curve measured vertically the intensity of the field at the corresponding distance. The radius of the conductor has been assumed as unity, and maximum field density (always at the surface of the conductor) as 100 percent.

The intensity of the magnetic field starts at zero at the conductor center, and increases (with uniform distribution of current in the conductor) directly as the

distance from its center until its surface is reached, where it becomes maximum. For distances beyond the surface of the conductor, the field intensity varies inversely as the distance from its center.

The intensity of the magnetic field at any point is proportional to the m.m.f. acting at that point and inversely proportional to the length of its circular path (magnetic reluctance). Thus at the surface of the

TABLE I—RESISTANCE PER 1000 FEET OF COPPER CONDUCTORS AT VARIOUS TEMPERATURES STRANDED CONDUCTORS

B & S NO.	AREA CIRCULAR MILS	OHMS PER 1000 FEET OF SINGLE CONDUCTOR															
		ANNEALED COPPER 100% CONDUCTIVITY								HARD DRAWN COPPER 97.3% CONDUCTIVITY							
		0°C 32°F	15°C 59°F	20°C 68°F	25°C 77°F	35°C 95°F	50°C 122°F	65°C 149°F	75°C 167°F	0°C 32°F	15°C 59°F	20°C 68°F	25°C 77°F	35°C 95°F	50°C 122°F	65°C 149°F	75°C 167°F
		2,000,000	.00487	.00518	.00528	.00539	.00559	.00591	.00623	.00643	.00500	.00533	.00544	.00554	.00574	.00607	.00640
1,900,000	.00512	.00546	.00556	.00568	.00590	.00623	.00656	.00678	.00526	.00561	.00570	.00574	.00606	.00640	.00674	.00697	
1,800,000	.00541	.00577	.00588	.00599	.00622	.00657	.00692	.00716	.00556	.00593	.00605	.00615	.00640	.00673	.00711	.00735	
1,700,000	.00573	.00610	.00622	.00635	.00659	.00695	.00733	.00758	.00590	.00628	.00640	.00652	.00677	.00714	.00753	.00780	
1,600,000	.00607	.00647	.00660	.00674	.00700	.00740	.00779	.00805	.00626	.00665	.00678	.00693	.00720	.00760	.00800	.00827	
1,500,000	.00650	.00690	.00704	.00719	.00746	.00787	.00830	.00858	.00668	.00709	.00724	.00739	.00766	.00808	.00853	.00882	
1,400,000	.00696	.00741	.00755	.00771	.00800	.00845	.00890	.00920	.00715	.00761	.00775	.00792	.00822	.00868	.00915	.00945	
1,300,000	.00749	.00798	.00813	.00830	.00862	.00910	.00958	.00990	.00770	.00820	.00836	.00853	.00885	.00935	.00985	.0101	
1,200,000	.00812	.00864	.00880	.00899	.00933	.00985	.0104	.0107	.00835	.00888	.00905	.00924	.00958	.0101	.0107	.0110	
1,100,000	.00886	.00942	.00960	.00981	.0102	.0108	.0113	.0117	.00910	.00968	.00986	.0101	.0105	.0111	.0116	.0120	
1,000,000	.00974	.0104	.0106	.0108	.0112	.0118	.0125	.0129	.0100	.0107	.0109	.0111	.0115	.0121	.0128	.0132	
950,000	.0102	.0109	.0111	.0114	.0118	.0124	.0131	.0135	.0105	.0112	.0114	.0117	.0121	.0127	.0134	.0138	
900,000	.0108	.0115	.0117	.0120	.0124	.0131	.0138	.0142	.0111	.0118	.0120	.0123	.0127	.0134	.0142	.0146	
850,000	.0115	.0122	.0124	.0127	.0132	.0139	.0147	.0152	.0118	.0125	.0127	.0130	.0135	.0143	.0151	.0156	
800,000	.0122	.0130	.0132	.0135	.0140	.0148	.0156	.0161	.0125	.0133	.0136	.0139	.0144	.0152	.0160	.0165	
750,000	.0130	.0138	.0140	.0144	.0149	.0157	.0166	.0171	.0134	.0142	.0144	.0148	.0153	.0161	.0170	.0175	
700,000	.0139	.0148	.0151	.0154	.0160	.0169	.0178	.0184	.0143	.0152	.0155	.0158	.0164	.0173	.0183	.0189	
650,000	.0150	.0160	.0163	.0166	.0172	.0182	.0192	.0199	.0154	.0164	.0167	.0170	.0176	.0187	.0197	.0204	
600,000	.0162	.0173	.0176	.0180	.0187	.0197	.0208	.0215	.0166	.0178	.0181	.0185	.0192	.0202	.0214	.0221	
550,000	.0177	.0188	.0191	.0196	.0203	.0214	.0226	.0234	.0182	.0193	.0196	.0202	.0209	.0220	.0232	.0240	
500,000	.0195	.0207	.0211	.0216	.0224	.0236	.0249	.0258	.0200	.0213	.0217	.0222	.0230	.0242	.0256	.0265	
450,000	.0216	.0230	.0234	.0240	.0249	.0263	.0277	.0286	.0222	.0236	.0240	.0247	.0256	.0270	.0285	.0294	
400,000	.0243	.0259	.0264	.0270	.0280	.0296	.0311	.0322	.0250	.0266	.0271	.0277	.0288	.0304	.0319	.0331	
350,000	.0278	.0297	.0303	.0308	.0319	.0337	.0356	.0368	.0286	.0305	.0312	.0316	.0328	.0346	.0366	.0378	
300,000	.0324	.0344	.0353	.0360	.0373	.0394	.0415	.0428	.0333	.0356	.0363	.0370	.0383	.0405	.0427	.0440	
250,000	.0390	.0415	.0423	.0432	.0448	.0473	.0498	.0515	.0400	.0426	.0435	.0444	.0460	.0487	.0512	.0530	
200,000	.0460	.0490	.0500	.0510	.0529	.0559	.0589	.0609	.0473	.0503	.0514	.0525	.0544	.0573	.0605	.0626	
0000	167 772	0580	0618	0630	0644	0668	0706	0742	0596	0635	0648	0662	0687	0725	0762	0788	
000	133 079	0732	0778	0795	0811	0841	0888	0936	0752	0800	0815	0834	0865	0900	0962	0995	
00	105 560	0922	0982	100	102	106	112	118	0948	101	103	105	109	115	121	125	
1	83 694	116	124	126	129	134	141	149	154	119	127	129	132	138	145	153	
2	66 358	147	156	159	163	169	178	188	194	151	160	163	167	173	183	193	
3	52 624	185	197	201	205	213	225	237	245	190	202	207	211	219	231	244	
4	41 738	233	248	253	259	269	284	298	308	239	255	260	266	276	292	306	
5	33 078	294	314	320	327	339	358	376	388	302	323	328	336	348	368	386	
6	26 244	371	395	403	412	427	452	475	491	381	404	415	423	438	464	488	
7	20 822	468	497	507	519	538	569	598	618	482	512	520	533	553	585	615	
8	16 512	590	628	640	654	678	716	755	781	607	646	658	672	697	736	775	
SOLID CONDUCTORS																	
0000	211 600	0451	0480	0490	0500	0519	0548	0577	0596	0463	0493	0503	0514	0533	0563	0592	0612
000	167 772	0569	0606	0618	0630	0654	0691	0727	0752	0585	0623	0635	0647	0672	0710	0746	0772
00	133 079	0718	0769	0779	0795	0826	0871	0917	0948	0738	0785	0800	0817	0850	0895	0942	0974
0	105 560	0905	0963	0983	100	104	110	116	120	0930	0988	101	103	107	113	119	123
1	83 694	114	121	124	126	131	139	146	151	117	124	127	129	134	143	150	155
2	66 358	144	153	156	159	165	175	184	190	148	157	160	163	170	180	189	195
3	52 624	181	193	197	201	209	220	232	240	186	198	202	207	215	226	238	246
4	41 738	229	244	248	253	263	278	293	302	235	250	255	260	270	286	301	310
5	33 088	289	307	313	319	331	350	368	381	297	315	321	328	340	360	378	391
6	26 244	364	387	395	403	418	442	465	481	374	398	407	415	430	454	477	494
7	20 822	459	488	498	508	528	557	586	606	472	502	512	523	543	572	602	623
8	16 512	579	616	628	640	665	702	739	764	595	633	645	657	685	722	759	785

These resistance values do not take into account skin effect. This should be considered when the larger conductors are used, particularly at the higher frequencies. No allowance has been made for increased length due to sag when the conductors are suspended. The resistance values for the stranded conductors are two percent greater than for a solid rod of cross-section equal to the total cross-section of the wires of the cable.

The change of resistivity of copper per degree C. is a constant independent of the temperature of reference and of the sample of copper. This resistivity-temperature constant is 0.0409 ohm (mil, foot). The fundamental resistivity used in calculating this table is the annealed copper standard, viz. 0.15328 ohm (meter, gram) at 20 degrees C.

For sizes not given in the table computations may be made by the following formulas which were used in calculating the above table:—
 Ohms per 1000 feet of annealed copper at 25 degrees C = $\frac{10787}{\text{Circ. mils}}$; at 65 degrees C = $\frac{12457}{\text{Circ. mils}}$

conductor the m.m.f. reaches its maximum because all of the current of the conductor is acting to produce m.m.f. at this and all points beyond. On the other hand the circular path subject to this maximum m.m.f. is shortest at the surface, the reluctance a minimum

and consequently the field intensity is greatest. For points beyond the surface the length of the circular path through air is proportional to the distance from the center of the conductor. Thus at a distance of 2 from the center the circular path is twice as long as at

TABLE II—RESISTANCE PER MILE OF COPPER CONDUCTORS AT VARIOUS TEMPERATURES STRANDED CONDUCTORS

B & S NO.	AREA CIRCULAR MILS	OHMS PER MILE OF SINGLE CONDUCTOR															
		ANNEALED COPPER								HARD DRAWN COPPER							
		100% CONDUCTIVITY															
		97.3% CONDUCTIVITY															
		0°C	15°C	20°C	25°C	35°C	50°C	65°C	75°C	0°C	15°C	20°C	25°C	35°C	50°C	65°C	75°C
		32°F	59°F	68°F	77°F	95°F	122°F	149°F	167°F	32°F	59°F	68°F	77°F	95°F	122°F	149°F	167°F
	2 000 000	.0258	.0274	.0279	.0285	.0295	.0312	.0329	.0340	.0265	.0282	.0288	.0293	.0304	.0321	.0337	.0349
	1 900 000	.0271	.0289	.0294	.0301	.0312	.0330	.0347	.0359	.0278	.0296	.0301	.0304	.0320	.0338	.0356	.0368
	1 800 000	.0286	.0305	.0311	.0317	.0329	.0347	.0366	.0379	.0294	.0314	.0320	.0325	.0338	.0357	.0375	.0389
	1 700 000	.0303	.0323	.0329	.0336	.0348	.0368	.0388	.0400	.0312	.0331	.0339	.0344	.0358	.0377	.0398	.0412
	1 600 000	.0322	.0342	.0349	.0357	.0370	.0391	.0412	.0425	.0331	.0352	.0358	.0367	.0381	.0402	.0422	.0438
	1 500 000	.0344	.0365	.0373	.0380	.0394	.0417	.0438	.0454	.0353	.0375	.0382	.0391	.0405	.0427	.0451	.0467
	1 400 000	.0368	.0391	.0399	.0408	.0423	.0447	.0470	.0487	.0378	.0402	.0410	.0418	.0435	.0459	.0484	.0500
	1 300 000	.0396	.0422	.0430	.0439	.0456	.0482	.0507	.0523	.0407	.0433	.0442	.0451	.0468	.0495	.0521	.0539
	1 200 000	.0429	.0457	.0465	.0475	.0493	.0520	.0550	.0565	.0442	.0470	.0478	.0489	.0507	.0534	.0565	.0582
	1 100 000	.0467	.0498	.0507	.0518	.0539	.0572	.0597	.0618	.0482	.0512	.0521	.0533	.0555	.0587	.0615	.0634
	1 000 000	.0515	.0550	.0560	.0570	.0592	.0623	.0660	.0682	.0528	.0565	.0577	.0587	.0608	.0640	.0675	.0699
	950 000	.0538	.0577	.0587	.0603	.0624	.0656	.0693	.0713	.0555	.0593	.0603	.0618	.0640	.0672	.0710	.0730
	900 000	.0571	.0608	.0618	.0635	.0655	.0693	.0730	.0751	.0587	.0623	.0635	.0650	.0672	.0708	.0750	.0772
	850 000	.0608	.0645	.0655	.0672	.0698	.0735	.0778	.0803	.0623	.0660	.0672	.0688	.0713	.0755	.0795	.0825
	800 000	.0645	.0687	.0698	.0713	.0740	.0783	.0825	.0851	.0660	.0703	.0718	.0735	.0762	.0803	.0845	.0873
	750 000	.0688	.0729	.0740	.0761	.0788	.0830	.0878	.0905	.0708	.0751	.0762	.0782	.0808	.0850	.0900	.0925
	700 000	.0735	.0783	.0798	.0814	.0846	.0894	.0940	.0973	.0756	.0803	.0819	.0835	.0866	.0915	.0965	.100
	650 000	.0773	.0846	.0861	.0878	.0910	.0962	.102	.105	.0815	.0867	.0883	.0900	.0930	.0990	.104	.108
	600 000	.0857	.0915	.0930	.0952	.0988	.104	.110	.114	.0878	.0940	.0957	.0978	.102	.107	.113	.117
	550 000	.0935	.0995	.101	.104	.107	.113	.121	.124	.0963	.102	.104	.107	.111	.116	.122	.127
	500 000	.103	.110	.112	.114	.119	.125	.132	.136	.106	.113	.115	.117	.122	.128	.135	.140
	450 000	.114	.122	.124	.127	.132	.139	.146	.151	.118	.125	.127	.131	.136	.143	.150	.156
	400 000	.129	.137	.140	.143	.148	.157	.165	.170	.132	.141	.144	.147	.152	.161	.168	.175
	350 000	.147	.157	.160	.163	.169	.178	.188	.195	.151	.162	.165	.167	.174	.183	.193	.200
	300 000	.171	.183	.187	.190	.197	.208	.220	.226	.176	.188	.192	.196	.203	.214	.226	.233
	250 000	.206	.219	.224	.228	.237	.250	.263	.272	.211	.225	.230	.235	.243	.258	.270	.280
0000	211 600	.243	.259	.264	.269	.280	.296	.311	.322	.249	.266	.272	.277	.288	.303	.320	.330
000	167 772	.306	.326	.333	.341	.353	.372	.392	.405	.315	.335	.342	.350	.363	.383	.402	.416
00	133 079	.387	.412	.420	.428	.444	.470	.495	.512	.398	.423	.432	.442	.457	.476	.510	.527
0	105 560	.488	.520	.528	.540	.560	.592	.624	.645	.502	.535	.545	.555	.576	.608	.640	.661
1	83 694	.612	.655	.665	.682	.708	.745	.787	.815	.630	.672	.682	.697	.730	.766	.810	.835
2	66 358	.777	.825	.840	.862	.895	.942	.995	.1 03	.798	.845	.862	.883	.915	.968	.1 02	.1 05
3	52 624	.978	1.04	1.07	1.09	1.13	1.19	1.25	1.30	1.01	1.07	1.10	1.12	1.16	1.22	1.29	1.33
4	41 738	1.23	1.31	1.34	1.37	1.42	1.51	1.58	1.63	1.27	1.35	1.38	1.41	1.46	1.55	1.61	1.67
5	33 078	1.56	1.66	1.69	1.73	1.80	1.89	1.99	2.05	1.60	1.71	1.73	1.78	1.84	1.95	2.04	2.11
6	26 244	1.96	2.09	2.13	2.17	2.26	2.39	2.51	2.59	2.01	2.14	2.20	2.24	2.32	2.45	2.58	2.66
7	20 822	2.48	2.63	2.68	2.74	2.84	3.01	3.16	3.27	2.55	2.71	2.75	2.82	2.93	3.09	3.25	3.35
8	16 512	3.12	3.32	3.38	3.46	3.58	3.78	3.99	4.13	3.21	3.41	3.48	3.55	3.69	3.89	4.10	4.24
SOLID CONDUCTORS																	
0000	211 600	.238	.254	.259	.264	.274	.289	.305	.315	.245	.261	.266	.272	.282	.298	.312	.323
000	167 772	.301	.320	.327	.333	.346	.365	.384	.397	.309	.329	.336	.342	.355	.375	.395	.408
00	133 079	.380	.404	.412	.420	.436	.460	.485	.501	.390	.415	.423	.432	.450	.473	.497	.515
0	105 560	.478	.509	.520	.528	.550	.582	.613	.635	.492	.522	.535	.545	.565	.597	.628	.650
1	83 694	.603	.640	.655	.666	.693	.735	.772	.798	.618	.655	.672	.680	.708	.755	.793	.820
2	66 358	.760	.808	.825	.840	.872	.925	.972	1.01	.783	.830	.845	.862	.900	.950	1.00	1.03
3	52 624	.955	1.02	1.04	1.06	1.11	1.16	1.23	1.27	.983	1.05	1.07	1.10	1.14	1.19	1.26	1.30
4	41 738	1.21	1.29	1.31	1.34	1.39	1.47	1.55	1.60	1.24	1.32	1.35	1.38	1.43	1.51	1.59	1.64
5	33 078	1.53	1.62	1.66	1.69	1.75	1.85	1.95	2.02	1.57	1.67	1.70	1.73	1.80	1.90	2.00	2.07
6	26 244	1.93	2.05	2.09	2.14	2.21	2.33	2.46	2.54	1.98	2.10	2.15	2.20	2.27	2.40	2.52	2.61
7	20 822	2.43	2.58	2.63	2.69	2.79	2.94	3.10	3.20	2.49	2.65	2.71	2.77	2.87	3.02	3.18	3.29
8	16 512	3.06	3.26	3.33	3.39	3.51	3.71	3.90	4.04	3.14	3.35	3.41	3.47	3.62	3.82	4.02	4.15

These resistance values do not take into account skin effect. This should be considered when the larger conductors are used, particularly at the higher frequencies. No allowance has been made for increased length due to sag when the conductors are suspended. The resistance values for the stranded conductors are two percent greater than for a solid rod of cross-section equal to the total cross-section of the wires of the cable.

The change of resistivity of copper per degree C. is a constant independent of the temperature of reference and of the sample of copper. This resistivity-temperature constant is 0.0409 ohm (mil, foot). The fundamental resistivity used in calculating this table is the annealed copper standard, viz. 0.15328 ohm (meter, gram) at 20 degrees C.

Where L is in millihenries per 1000 feet of single conductor.

The effective flux area departs from the flux density line at E dropping down in the form of a reverse curve and terminating in zero at II . All flux to the right of II cuts the whole of both conductors producing the same amount of inductance in both of them in such a direction as to oppose or neutralize each other.

The flux cutting conductor B from q to II has its full value of effectiveness in producing inductance in conductor A . On the other hand it also produces to a less extent inductance in conductor B but in a direction to oppose that which it produces in conductor A . The difference between that produced in conductors A and B is the effective flux producing inductance in the circuit and is represented by the shaded portion through conductor B within the area $E-q-II-T-E$. To illustrate how the effective flux curved line $E-T-II$ was determined, suppose it is required to determine the effective flux at the distance IO (center of conductor B). At this point the flux density is ten percent, but since these flux density lines are actually concentric circles, having their center at the middle of conductor A this flux density curve cuts conductor B in the form of an arc (see lower right hand corner of inductance chart). The area of the shaded portion between the two arcs is a measure of the inductance in conductor B at its center. The difference between this shaded area, and the whole area of B , or the clear part to the right of the shaded portion, is a measure of the difference in inductance of the two conductors. In other words, for the spacings shown, approximately 55 percent of ten or 5.5 percent is the value of the effective flux at distance of IO from conductor A .

$$\text{If in place of } L = 0.14037 \log_{10} \frac{D-R}{R} \dots\dots (1)$$

$$\text{we take } L = 0.14037 \log_{10} \frac{D}{R} \dots\dots (2)$$

we include all of the inductance area out to the vertical line $O-IO$. This would include the area $E-O-T$ but not the area $T-IO-II$. Since these two areas are equal, the omission of one is balanced by including the other and therefore formula (2) correctly takes into account all of the effective inductance beyond the surface of conductor A .

The inductance within conductor A is determined as follows:—At a point midway between the center and its surface the flux density is 50 percent as indicated by the straight flux density line of the chart. However at this point only one-fourth of the conductor area is enclosed, so that, measured in terms of its effect if outside the conductor, its effectiveness would be only one-fourth of 50 or 12.5 percent. This is the reason that the so-called effective flux line is curved and falls to the right of the straight flux density line. The area of the triangular section $O-I-IO$ is a measure of the effective inductance within conductor A . This is a constant for all sizes of solid conductors and is represented by the

constant 0.01524 of the inductance formula based upon 1000 feet of conductor.

The fundamental formula for the total effective inductance (within and external to conductor A) of a single solid non-magnetic conductor suspended in air is therefore:

$$L = 0.01524 + 0.14037 \log_{10} \frac{D}{R} \text{ per 1000 ft.} \dots\dots (3)$$

or

$$L = 0.08017 + 0.74115 \log_{10} \frac{D}{R} \text{ per mile.} \dots\dots (4)$$

It may be interesting to note here that the above described graphical solution for inductance produces results in close agreement with those obtained by the fundamental formula for inductance. That is, lay out such a chart on cross section paper to a large scale and count the number of squares or area representing the internal and the external inductance due to current in conductor A . It will be seen that the relative values of the external and internal flux areas conform with the relative values as determined by the formula. This will also be true in the case of the conductors when so placed as to give zero separation, as illustrated by Fig. 6.

VARIATIONS FROM THE FUNDAMENTAL INDUCTANCE FORMULA

It has been proven mathematically by the Bureau of Standards and others that the fundamental formula (3) for determining inductance will give exact results for solid, round, straight, parallel conductors, provided skin and proximity effects are absent. Proximity effect is the crowding of the current to one side of a conductor, due to the proximity of another current carrying conductor. It is similar to skin effect in that it increases the resistance and decreases the inductance. Proximity effect as well as skin effect changes only the inductance due to the flux inside the conductor. Proximity effect is more pronounced for large conductors, high frequencies and close proximity.

For No. 0000 solid conductors at zero separation and 60 cycles, the error in the results (as determined by the fundamental inductance formula) due to skin effect is less than one-tenth of one percent. This error, however, increases rapidly as the size of the conductor increases. Proximity effect cannot be calculated but it is believed to be less than two percent in the above case.

Should skin and proximity effect combined, be sufficient to force all of the current out to within a very thin annulus at the surface of the conductor (a condition obviously never obtained at commercial frequencies) their combined effect would be a maximum. In such a case there would be no inductance within the conductors and the first constant 0.01524 would disappear from formula (3).

Skin and proximity effect are so small in the case of the greater spacings of conductors required for high-tension aerial transmission circuits that they may in such cases be ignored. Even in the case of the close

spacings required for three conductor cables these combined effects are usually less than four percent.

EFFECT OF STRANDING ON INDUCTANCE

The fundamental formula (3) for determining inductance is based upon a solid conductor, R being taken as the radius of the conductor. In stranded cables the effective value for R lies between the actual radius and that of a solid rod having an equivalent cross-section to that of the cable. The effective value for R varies with the stranding of the cable employed.

Formulas for use in determining the inductance of stranded cables when used for high-tension aerial transmission have been calculated by Mr. H. B. Dwight as follows:—

$$\text{For a 7-wire cable, } L = 0.741 \log_{10} \frac{2.756 D}{d} \dots\dots (5)$$

$$\text{For a 19-wire cable, } L = 0.741 \log_{10} \frac{2.640 D}{d} \dots\dots (6)$$

$$\text{For a 37-wire cable, } L = 0.741 \log_{10} \frac{2.605 D}{d} \dots\dots (7)$$

$$\text{For a 61-wire cable, } L = 0.741 \log_{10} \frac{2.590 D}{d} \dots\dots (8)$$

where L is in millihenries per mile of a single conductor, D is the spacing between centers of cables, and d is the outside diameter of the cables measured in same units as D .

SPIRALING EFFECT UPON INDUCTANCE

Spiraling of the strands of a cable and spiraling of the conductors of a three-conductor cable tend to increase the inductance. It is difficult to calculate the effect of spiraling for the various cases, but it may be considered negligible for high-tension aerial transmission circuits using non-magnetic conductors. For three-conductor cables the effect of spiraling is probably in the neighborhood of two percent.

Values for inductance per thousand feet of single conductor are given in Table III, for commercial sizes of copper and steel reinforced aluminum conductors. The formula by which the values were derived are:—

$$L = 0.01524 + 0.1403 \log_{10} \frac{D}{R} \dots\dots\dots (3)$$

where L = Millihenries per 1000 feet of single conductor of a single phase, or of a symmetrical three-phase circuit.
 D = Distance between centers of conductors.
 R = Radius (to be measured in same units as D) of solid conductor. In the case of stranded conductors, R was taken as the radius of a solid rod of equivalent cross-section to that of the stranded conductors.

Table III has been carried out to three figures only. This would seem sufficiently accurate for working values when it is considered that there are numerous sources of variation from the calculated values. In the first place formulas are based upon a uniform distribution of current throughout the cross section of the conductors, whereas the current is seldom uniform and in the larger conductors, especially at 60 cycles, may be to a large extent crowded to the outer strands as a result of skin effect. This condition is further modi-

fied when the conductors are placed close together, by the proximity effect. Stranded conductors made up of various stranding combinations result in variation of inductance of several percent. In practice the length and spacing of conductors will vary more or less from those assumed when determining the calculated values.

The values for inductance of stranded conductors in Table III, as stated above, were derived by taking R as the radius of a solid rod having an equivalent cross-section area to that of the stranded conductors. Thus for 1 000 000 circ. mil cable the outside diameter is 1.152 in. and that of an equivalent solid iron is 1.0 in. R was therefore in this case taken as 0.5 in. The effective radius is really slightly greater than that of the solid rod and less than that of the cable, varying with the stranding employed. The actual inductance of cables will therefore be slightly less (usually two or three percent) than those indicated in the table for solid rods. The table values are therefore conservative.

The steel core of steel reinforced aluminum cables carries so little current on account of its relatively greater resistance that for practical purposes it has been customary to ignore its presence and to consider such conductors as solid rods of same area as that of the aluminum strands. In the absence of accurate data this practice was followed in determining the values for inductance of such cables in Table III.

The minimum value for inductance occurs when the conductors have zero separation $\frac{D}{R} = 2$, (Fig. 6). In this case the inductance in millihenries is independent of the size of the conductor. As given by formula (3) it is $L = 0.05124 + 0.1403 \log_{10} 2 = 0.0575$ millihenries per 1000 feet of each conductor. Obviously insulation requirements will not permit of such a low value for inductance although it will be closely approached in low voltage cables.

Any given percentage difference in distance between centers of conductors represents a definite and constant value in inductance regardless of their size. These values are given in column B at the bottom of the table for various percentages increase in spacings. Thus if the distance between conductor centers is increased 50 percent the corresponding increase in inductance is 0.025 as indicated in column B , under the D/R values of 1.50. Likewise doubling the distance increases the inductance by an amount of 0.042. For instance the table value for inductance of No. 0 solid copper conductor is for one-half inch spacing 0.084, and for one inch spacing 0.126 (an increase of 0.042.) For four foot spacing the table value is 0.362, and for eight foot spacing 0.404, also a difference of 0.042.

References:—An article by Prof. Charles F. Scott, "Inductance in Transmission Circuits" in THE ELECTRIC JOURNAL for Feb. 1906 very clearly covers the field of self and mutual inductance external to the conductors.

H. B. Dwight, "Transmission Line Formulas."
 V. Karapetoff, "The Magnetic Circuit" p. 189.

CHAPTER II

REACTANCE—CAPACITANCE—CHARGING CURRENT

REACTANCE

A CONDUCTOR carrying an electric current is surrounded by a magnetic flux, whose value is proportional to the current. If the current varies, this flux also changes, thereby inducing an electromotive force in a direction which opposes the change. This counter e.m.f. is proportional to the rate of change and hence in alternating current is proportional to the frequency. It can be expressed in ohms per mile of each conductor of a single-phase or of a symmetrical three-phase circuit as follows:—

$$\text{Ohms Reactance} = 2 \pi f L \quad (9)$$

When f = Frequency in cycles per second

L = Henries per mile of single conductor.

The value for $2 \pi f$ are as follows:—

Frequency	$2 \pi f$
1	6.28
15	94.25
25	157.1
40	251.3
60	377.0
133	835.7

Tables IV and V indicate the reactance in ohms per mile, of a single conductor at 25 and 60 cycles respectively for various spacings of conductors. The foot notes to these tables cover the pertinent points relating to them. The resistance per 1000 feet, and per mile of single conductor at 25 degrees C. (77 degrees F) is given in parallel columns as a convenience for comparison of the resistance and reactance values. The resistance corresponding to other temperatures when desired may be taken from Tables I and II.

Tables VI and VII indicate the relative importance of reactance and resistance. In some cases of short lines and large single conductors, the reactance and not the resistance may determine the size and number of cables necessary. In other words, it may be necessary to keep the resistance abnormally low so that the reactance will not be so high as to result in an abnormal voltage drop in the circuit. In such cases the values in Tables VI and VII may be used for determining the permissible resistance in order not to exceed the desired reactance.

Example:—It is desired to use 1 000 000 circ. mil single conductor cables at 60 cycles, spaced two feet apart; from Table VII it is seen that the reactance drop under these conditions is 8.52 times the ohmic drop at 25 degrees C. If an ohmic drop of five percent at 25 degrees C is suggested the corresponding reactive drop would be 5×8.52 or 42.6 percent which would be excessive. If it is desired to limit the reactive drop to 10 percent in this case, the ohmic drop at 25 degrees C must be $10 \div 8.52$ or 1.18 percent.

Probably a more important use for Tables VI and VII is for determining the reactance of a conductor directly from its resistance. To do this it is only necessary to multiply its resistance (at 25 degrees C) by the

ratio value in table VI or VII corresponding to the conductor and spacing desired.

UNSYMMETRICAL SPACING

The inductance and capacitance per conductor of a three-phase circuit for symmetrical spacing of conductors is the same as the inductance and capacitance per conductor of a single-phase circuit for the same size conductor and the same spacing. For irregular spacing of conductors, the inductance and capacitance will be different. When the three conductors are placed in the same plane (flat spacing), the inductance of each of the outside conductors is greater than that of the middle conductor. By properly transposing the conductors, the inductance and capacitance may be equalized in all three conductors. However, the effect of flat spacing

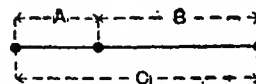


FIG. 7

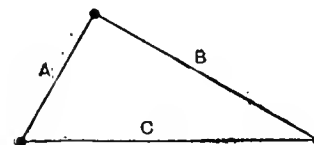


FIG. 8



FIG. 9

Conductor Spacings.

For three-phase irregular flat or triangular spacing (Figs. 7 and 8) use $D = \sqrt[3]{A B C}$.

For three-phase regular flat spacing Fig. 9 use $D = 1.26 A$.

For two-phase line the spacing is the average distance between centers of conductors of the same phase. It makes no difference whether the plane of the conductors with flat spacing is horizontal, vertical or inclined.

is equivalent to that of a symmetrical arrangement of greater spacing.

Various arrangements of conductors are indicated in Figs. 7, 8 and 9. Many three-phase high tension circuits have the three conductors regularly spaced in a common plane (regular flat spacing) Fig. 9. Beneath these figures are placed statements indicating the determination of "effective spacings" for any arrangement of conductors.

Since the so called "effective spacing" corresponding to unsymmetrical arrangements of conductors is usually a fractional number, the line constants for such effective spacing can usually not be taken directly from

the tables but may be obtained by the use of the values in columns *A* and *B* at the foot of these tables.

Example:—It is desired to determine the 60 cycle reactance per mile of a single conductor for flat spacing of 11 ft. between adjacent 0000 solid copper conductors. The effective spacing is 1.26×11 or 13.8 feet. The reactance (Table V) for this conductor at 13 feet symmetrical spacing is 0.820 ohm. The value for *A*, (bottom of Table V) = $13.8 \div 13 = 1.06$. The value of *B* corresponding to the value for *A* of 1.06 is approximately 0.006 which, added to 0.820 gives a reactance of 0.826 ohm for the effective spacing of 13.8 feet. The values of reactance for all effective spacings not included in the Table may be determined in a similar manner.

With an unsymmetrical arrangement of conductors there must be a sufficient number of transpositions of conductors to obtain balanced electrical conditions along the circuit.

CAPACITANCE

When mechanical force is exerted against a liquid or a solid mass, a displacement takes place proportional to the force exerted and inversely proportional to the resistance offered by the liquid or solid mass subjected to the force. If the mass consists of some elastic material, such as rubber, the displacement will be greater than if it consists of a more solid material, such as metal.

In a similar manner when an e.m.f. is applied to a condenser, a certain quantity of electricity will flow into it until it is charged to the same pressure as that of the applied circuit. A condenser consists of plates of conducting material separated by insulating material known as the dielectric. All electric circuits consist of conductors separated by a dielectric (usually air) and therefore act to a greater or less extent as condensers. The ability of a condenser or any electric circuit to receive the charge is a measure of its "capacity" more properly known as its "capacitance". Just as the rubber mass referred to above will, for a given force, permit of greater displacement so will circuits of greater capacitance permit more current to flow into them for a given e.m.f. impressed.

The process of charging a dielectric consists of setting up an electric strain in it similar to the mechanical strain in a liquid or mass referred to above. If an alternating voltage is impressed upon the terminals of a circuit containing capacitance, the charging current will vary directly with the impressed e.m.f. There is current to the condenser during rising and from the condenser during decreasing e.m.f. Thus the condenser is charged and then discharged in the opposite direction during the next alternation, making two complete charges and discharges for each cycle of impressed e.m.f. (Fig. 10). As long as the e.m.f. at the terminals is changing, the condenser will continue to receive or give out current. The current flowing to and from the condenser, assuming negligible resistance, leads the impressed e.m.f. by 90 electrical degrees.

DEFINITION

The capacitance of a circuit or condenser is said to be one farad when a rate of change in pressure of one volt per second at the terminals produces a current of

one ampere. Stated another way, its capacitance in farads is numerically equal to the quantity of electricity in coulombs which it will hold under a pressure of one volt. The farad being an inconveniently large unit, one millionth part of it, the microfarad, is the usual practical unit.

CAPACITANCE FORMULA

An exact formula for the capacitance between parallel conductors must take into account the nonuniformity of the distribution of charge around the conductors. Such a formula* is formed by considering the charges as concentrated at the inverse points of the conductors; thus,—

$$C = \frac{0.008467}{\cosh^{-1} \frac{D}{d}} \dots\dots\dots (10)$$

Where *C* equals the microfarads per 1000 feet of conductor between two parallel bare conductors in air, *D*, the distance between centers of the conductors and *d*, the diameter and *R* the radius of the conductors, measured in the same units as *D*.

Since $\cosh^{-1} X = \log_e (X + \sqrt{X^2 - 1}) \dots\dots\dots (11)$

$$C = \frac{0.008467}{\log_e \left(\frac{D}{d} + \sqrt{\left(\frac{D}{d}\right)^2 - 1} \right)} \dots\dots\dots (12)$$

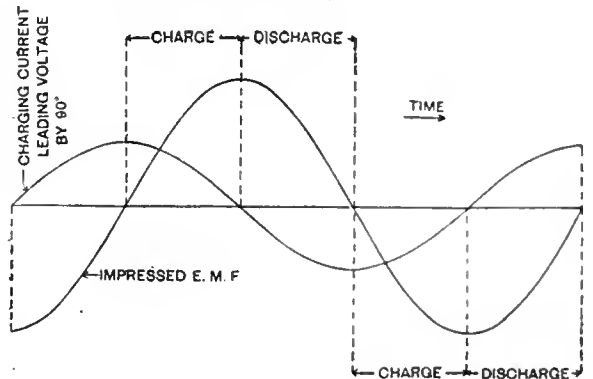


FIG. 10—CHARGING CURRENT

Reducing to common logarithms and capitivity to neutral,—

$$C = \frac{0.007354}{\log_{10} \left(\frac{D}{d} + \sqrt{\left(\frac{D}{d}\right)^2 - 1} \right)} \dots\dots\dots (13)$$

Microfarads per 1000 feet of single conductor to neutral.
or

$$C = \frac{0.038829}{\log_{10} \left(\frac{D}{d} + \sqrt{\left(\frac{D}{d}\right)^2 - 1} \right)} \dots\dots\dots (14)$$

Microfarads per mile of single conductor to neutral.

When *D* is greater than 10 *d*, which is always the case in high-tension transmission lines employing bare conductors, the following simplified formulas may be used with negligible error.—

$$C = \frac{0.007354}{\log_{10} \frac{D}{R}} \dots\dots\dots (15)$$

*See article by Pender & Osborne in *Electrical World* of Sept. 22, 1910, Vol. 56.

TABLE V—RESISTANCE AND 60 CYCLE REACTANCE OHMS PER MILE OF SINGLE CONDUCTOR

REACTANCE IN OHMS PER MILE OF EACH CONDUCTOR OF A SINGLE PHASE, OR OF A SYMMETRICAL 3 PHASE CIRCUIT. FOR OTHER ARRANGEMENTS OF CONDUCTORS SEE FOOT NOTES; X, THE TABLE VALUES WERE DERIVED FROM THE EQUATION—OHMS REACTANCE=2πFL (L BEING EXPRESSED IN HENRIES PER MILE OF SINGLE CONDUCTOR) THE REACTANCE FOR OTHER FREQUENCIES IS $\frac{F}{60}$ THE TABLE VALUES.

MATERIAL	TYPE	DIAMETER IN INCHES	B & S NO.	AREA IN CIRCULAR MILS	RESISTANCE OF A SINGLE CONDUCTOR IN OHMS AT 25° C (77° F) X X	DISTANCE (D) BETWEEN CENTERS OF CONDUCTORS X																						
						PER 1000 FEET	PER MILE	1'	2'	3'	4'	5'	6'	8'	12'	18'	2'	3'	4'	5'	6'	8'	10'	15'	17'	21'	23'	25'
								FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET	FEET
COPPER	STRAINED	1/32	1000000	1000000	0.0554	157	207	241	268	290	324	374	422	458	504	540	589	640	694	750	808	868	929	991	1053			
		1/16	1500000	1500000	0.0655	188	246	287	317	344	384	438	488	528	570	609	653	699	748	799	851	904	958	1013	1068	1124		
		1/8	2000000	2000000	0.0756	219	284	331	357	380	424	482	534	576	613	653	695	739	785	832	880	929	979	1030	1081	1133		
		3/16	2500000	2500000	0.0857	250	321	374	397	416	464	526	580	624	663	703	744	786	829	873	918	964	1010	1057	1104	1152		
		1/4	3000000	3000000	0.0958	281	357	416	436	451	503	569	624	669	708	748	789	831	874	918	963	1009	1056	1103	1150	1198		
		5/16	3500000	3500000	0.1059	312	394	458	475	488	544	614	670	716	755	794	834	874	915	956	998	1040	1082	1124	1166	1208		
		3/8	4000000	4000000	0.1160	343	430	500	515	525	585	659	716	763	801	839	877	915	953	991	1030	1069	1108	1147	1186	1225		
		7/16	4500000	4500000	0.1261	374	466	542	555	562	626	704	762	809	846	883	920	957	994	1031	1068	1105	1142	1179	1216	1253		
		1/2	5000000	5000000	0.1362	405	502	584	595	600	674	756	814	861	897	933	969	1005	1041	1076	1111	1146	1181	1216	1251	1286		
		5/8	5500000	5500000	0.1463	436	538	626	635	638	716	801	859	905	941	976	1011	1046	1081	1115	1149	1183	1217	1251	1285	1319		
	SOLID	1/32	1000000	1000000	0.0554	157	207	241	268	290	324	374	422	458	504	540	589	640	694	750	808	868	929	991	1053	1115		
		1/16	1500000	1500000	0.0655	188	246	287	317	344	384	438	488	528	570	609	653	699	748	799	851	904	958	1013	1068	1124		
		1/8	2000000	2000000	0.0756	219	284	331	357	380	424	482	534	576	613	653	695	739	785	832	880	929	979	1030	1081	1133		
		3/16	2500000	2500000	0.0857	250	321	374	397	416	464	526	580	624	663	703	744	786	829	873	918	964	1010	1057	1104	1152		
		1/4	3000000	3000000	0.0958	281	357	416	436	451	503	569	624	669	708	748	789	831	874	918	963	1009	1056	1103	1150	1198		
		5/16	3500000	3500000	0.1059	312	394	458	475	488	544	614	670	716	755	794	834	874	915	956	998	1040	1082	1124	1166	1208		
		3/8	4000000	4000000	0.1160	343	430	500	515	525	585	659	716	763	801	839	877	915	953	991	1030	1069	1108	1147	1186	1225		
		7/16	4500000	4500000	0.1261	374	466	542	555	562	626	704	762	809	846	883	920	957	994	1031	1068	1105	1142	1179	1216	1253		
		1/2	5000000	5000000	0.1362	405	502	584	595	600	674	756	814	861	897	933	969	1005	1041	1076	1111	1146	1181	1216	1251	1286		
		5/8	5500000	5500000	0.1463	436	538	626	635	638	716	801	859	905	941	976	1011	1046	1081	1115	1149	1183	1217	1251	1285	1319		
ALUMINUM	STEEL REINFORCED	1/32	1000000	1000000	0.0756	219	284	331	357	380	424	482	534	576	613	653	695	739	785	832	880	929	979	1030	1081			
		1/16	1500000	1500000	0.0857	250	321	374	397	416	464	526	580	624	663	703	744	786	829	873	918	964	1010	1057	1104			
		1/8	2000000	2000000	0.0958	281	357	416	436	451	503	569	624	669	708	748	789	831	874	918	963	1009	1056	1103	1150			
		3/16	2500000	2500000	0.1059	312	394	458	475	488	544	614	670	716	755	794	834	874	915	956	998	1040	1082	1124	1166			
		1/4	3000000	3000000	0.1160	343	430	500	515	525	585	659	716	763	801	839	877	915	953	991	1030	1069	1108	1147	1186			
	SOLID	1/32	1000000	1000000	0.0554	157	207	241	268	290	324	374	422	458	504	540	589	640	694	750	808	868	929	991	1053			
		1/16	1500000	1500000	0.0655	188	246	287	317	344	384	438	488	528	570	609	653	699	748	799	851	904	958	1013	1068			
		1/8	2000000	2000000	0.0756	219	284	331	357	380	424	482	534	576	613	653	695	739	785	832	880	929	979	1030	1081			
		3/16	2500000	2500000	0.0857	250	321	374	397	416	464	526	580	624	663	703	744	786	829	873	918	964	1010	1057	1104			
		1/4	3000000	3000000	0.0958	281	357	416	436	451	503	569	624	669	708	748	789	831	874	918	963	1009	1056	1103	1150			

The reactance for any distance D not given in the table can be found as follows: Let E = the nearest smaller distance in the table, Divido D by E and taking a value of A nearest to the quotient find the corresponding value of B, which must be added to the reactance corresponding to the size of conductor and distance E.
 For three phase regular flat spacing use D = 1.26 A. For three phase irregular flat or triangular spacing use $D = \sqrt[3]{A^2 B C}$. For a two-phase line the spacing is the average distance between centers of conductors of the same phase.
 xx At a temperature of 65° C (149° F) these resistance values would be increased by 1.5 percent. They are based upon a conductivity for copper of 97.3—for aluminum of 61 percent. They do not take into account skin effect; this should be considered when the larger conductors are used, particularly at the higher frequencies. No allowance has been made for increased length due to sag when the conductors are two percent greater than for a solid rod of cross section equal to the total cross section of the wires of the cable.
 For stranded conductors D was taken as the diameter of a solid rod of equivalent cross sectional area. Actually D for stranded conductors is slightly greater, resulting in slightly less reactance than the table values. The table values are therefore conservative. In calculating the reactance values for the steel reinforced aluminum cable the presence of the steel strands was ignored.

TABLE VI—RATIO OF 25 CYCLE REACTANCE, TO RESISTANCE AT 25° C

THE RESISTANCE VOLTS HAVING BEEN DETERMINED (AT 25° C) THE REACTANCE VOLTS MAY BE FOUND BY MULTIPLYING THE RESISTANCE VOLTS BY THE CONSTANTS GIVEN IN TABLE BELOW FOR THE SPACING AND SIZE OF CONDUCTORS CONTEMPLATED. THE RATIO FOR OTHER FREQUENCIES IS $\frac{f}{25}$ TIMES THE TABLE VALUES. FOR A TEMPERATURE OF 65° C (149° F) MULTIPLY TABLE VALUES BY .87

MATERIAL	TYPE	DIAMETER IN INCHES	B & S NO.	AREA IN CIRCULAR MILLS	RESISTANCE OF A SINGLE CONDUCTOR IN OHMS AT 25° C (77° F) X X		DISTANCE (D) BETWEEN CENTERS OF CONDUCTORS X																										
					PER 1000 FEET	PER MILE	1'	2'	3'	4'	5'	6'	8'	12'	18'	2'	3'	4'	5'	6'	7'	8'	9'	11'	13'	15'	17'	19'	21'	23'	25'		
							0.0554	0.0625	0.0700	0.0775	0.0850	0.0925	0.1000	0.1075	0.1150	0.1225	0.1300	0.1375	0.1450	0.1525	0.1600	0.1675	0.1750	0.1825	0.1900	0.1975	0.2050	0.2125	0.2200	0.2275	0.2350	0.2425	0.2500
COPPER	STRAUNDED	1/8	1000000	1000000	0.293	2.22	2.93	3.42	3.82	4.12	4.60	5.32	5.98	6.52	7.20	7.68	8.00	8.55	8.92	9.30	9.80	10.30	10.80	11.30	11.80	12.30	12.80	13.30	13.80	14.30	14.80		
		3/16	1500000	1500000	0.245	2.09	2.74	3.12	3.50	3.79	4.20	4.83	5.24	5.65	6.05	6.45	6.85	7.25	7.65	8.05	8.45	8.85	9.25	9.65	10.05	10.45	10.85	11.25	11.65	12.05	12.45		
		1/4	2000000	2000000	0.200	1.97	2.58	2.97	3.34	3.64	3.98	4.40	4.83	5.24	5.65	6.05	6.45	6.85	7.25	7.65	8.05	8.45	8.85	9.25	9.65	10.05	10.45	10.85	11.25	11.65	12.05	12.45	
		5/16	2500000	2500000	0.167	1.77	2.27	2.66	2.97	3.27	3.57	3.97	4.38	4.78	5.18	5.58	5.98	6.38	6.78	7.18	7.58	7.98	8.38	8.78	9.18	9.58	9.98	10.38	10.78	11.18	11.58	11.98	
		3/8	3000000	3000000	0.142	1.60	2.02	2.35	2.68	2.97	3.27	3.57	3.87	4.17	4.47	4.77	5.07	5.37	5.67	5.97	6.27	6.57	6.87	7.17	7.47	7.77	8.07	8.37	8.67	8.97	9.27	9.57	
		7/16	3500000	3500000	0.122	1.42	1.78	2.11	2.44	2.73	3.02	3.31	3.60	3.89	4.18	4.47	4.76	5.05	5.34	5.63	5.92	6.21	6.50	6.79	7.08	7.37	7.66	7.95	8.24	8.53	8.82	9.11	
		1/2	4000000	4000000	0.111	1.28	1.59	1.91	2.22	2.49	2.76	3.03	3.30	3.57	3.84	4.11	4.38	4.65	4.92	5.19	5.46	5.73	6.00	6.27	6.54	6.81	7.08	7.35	7.62	7.89	8.16	8.43	
		5/8	4500000	4500000	0.100	1.17	1.44	1.71	1.97	2.21	2.45	2.69	2.93	3.17	3.41	3.65	3.89	4.13	4.37	4.61	4.85	5.09	5.33	5.57	5.81	6.05	6.29	6.53	6.77	7.01	7.25	7.49	
		3/4	5000000	5000000	0.090	1.04	1.27	1.53	1.74	1.94	2.15	2.35	2.55	2.75	2.95	3.15	3.35	3.55	3.75	3.95	4.15	4.35	4.55	4.75	4.95	5.15	5.35	5.55	5.75	5.95	6.15	6.35	
		7/8	5500000	5500000	0.082	0.92	1.11	1.34	1.54	1.73	1.92	2.11	2.30	2.49	2.68	2.87	3.06	3.25	3.44	3.63	3.82	4.01	4.20	4.39	4.58	4.77	4.96	5.15	5.34	5.53	5.72	5.91	
	SOLID	1/8	1000000	1000000	0.293	2.22	2.93	3.42	3.82	4.12	4.60	5.32	5.98	6.52	7.20	7.68	8.00	8.55	8.92	9.30	9.80	10.30	10.80	11.30	11.80	12.30	12.80	13.30	13.80	14.30	14.80		
		3/16	1500000	1500000	0.245	2.09	2.74	3.12	3.50	3.79	4.20	4.83	5.24	5.65	6.05	6.45	6.85	7.25	7.65	8.05	8.45	8.85	9.25	9.65	10.05	10.45	10.85	11.25	11.65	12.05	12.45	12.85	
		1/4	2000000	2000000	0.200	1.97	2.58	2.97	3.34	3.64	3.98	4.40	4.83	5.24	5.65	6.05	6.45	6.85	7.25	7.65	8.05	8.45	8.85	9.25	9.65	10.05	10.45	10.85	11.25	11.65	12.05	12.45	
		5/16	2500000	2500000	0.167	1.77	2.27	2.66	2.97	3.27	3.57	3.97	4.38	4.78	5.18	5.58	5.98	6.38	6.78	7.18	7.58	7.98	8.38	8.78	9.18	9.58	9.98	10.38	10.78	11.18	11.58	11.98	
		3/8	3000000	3000000	0.142	1.60	2.02	2.35	2.68	2.97	3.27	3.57	3.87	4.17	4.47	4.77	5.07	5.37	5.67	5.97	6.27	6.57	6.87	7.17	7.47	7.77	8.07	8.37	8.67	8.97	9.27	9.57	
		7/16	3500000	3500000	0.122	1.42	1.78	2.11	2.44	2.73	3.02	3.31	3.60	3.89	4.18	4.47	4.76	5.05	5.34	5.63	5.92	6.21	6.50	6.79	7.08	7.37	7.66	7.95	8.24	8.53	8.82	9.11	
		1/2	4000000	4000000	0.111	1.28	1.59	1.91	2.22	2.49	2.76	3.03	3.30	3.57	3.84	4.11	4.38	4.65	4.92	5.19	5.46	5.73	6.00	6.27	6.54	6.81	7.08	7.35	7.62	7.89	8.16	8.43	
		5/8	4500000	4500000	0.100	1.17	1.44	1.71	1.97	2.21	2.45	2.69	2.93	3.17	3.41	3.65	3.89	4.13	4.37	4.61	4.85	5.09	5.33	5.57	5.81	6.05	6.29	6.53	6.77	7.01	7.25	7.49	
		3/4	5000000	5000000	0.090	1.04	1.27	1.53	1.74	1.94	2.15	2.35	2.55	2.75	2.95	3.15	3.35	3.55	3.75	3.95	4.15	4.35	4.55	4.75	4.95	5.15	5.35	5.55	5.75	5.95	6.15	6.35	
		ALUMINUM	STEEL REINFORCED	1/8	1000000	1000000	0.293	2.22	2.93	3.42	3.82	4.12	4.60	5.32	5.98	6.52	7.20	7.68	8.00	8.55	8.92	9.30	9.80	10.30	10.80	11.30	11.80	12.30	12.80	13.30	13.80	14.30	14.80
3/16	1500000			1500000	0.245	2.09	2.74	3.12	3.50	3.79	4.20	4.83	5.24	5.65	6.05	6.45	6.85	7.25	7.65	8.05	8.45	8.85	9.25	9.65	10.05	10.45	10.85	11.25	11.65	12.05	12.45	12.85	
1/4	2000000			2000000	0.200	1.97	2.58	2.97	3.34	3.64	3.98	4.40	4.83	5.24	5.65	6.05	6.45	6.85	7.25	7.65	8.05	8.45	8.85	9.25	9.65	10.05	10.45	10.85	11.25	11.65	12.05	12.45	12.85
5/16	2500000			2500000	0.167	1.77	2.27	2.66	2.97	3.27	3.57	3.97	4.38	4.78	5.18	5.58	5.98	6.38	6.78	7.18	7.58	7.98	8.38	8.78	9.18	9.58	9.98	10.38	10.78	11.18	11.58	11.98	
3/8	3000000			3000000	0.142	1.60	2.02	2.35	2.68	2.97	3.27	3.57	3.87	4.17	4.47	4.77	5.07	5.37	5.67	5.97	6.27	6.57	6.87	7.17	7.47	7.77	8.07	8.37	8.67	8.97	9.27	9.57	
7/16	3500000			3500000	0.122	1.42	1.78	2.11	2.44	2.73	3.02	3.31	3.60	3.89	4.18	4.47	4.76	5.05	5.34	5.63	5.92	6.21	6.50	6.79	7.08	7.37	7.66	7.95	8.24	8.53	8.82	9.11	
1/2	4000000			4000000	0.111	1.28	1.59	1.91	2.22	2.49	2.76	3.03	3.30	3.57	3.84	4.11	4.38	4.65	4.92	5.19	5.46	5.73	6.00	6.27	6.54	6.81	7.08	7.35	7.62	7.89	8.16	8.43	
5/8	4500000			4500000	0.100	1.17	1.44	1.71	1.97	2.21	2.45	2.69	2.93	3.17	3.41	3.65	3.89	4.13	4.37	4.61	4.85	5.09	5.33	5.57	5.81	6.05	6.29	6.53	6.77	7.01	7.25	7.49	
3/4	5000000			5000000	0.090	1.04	1.27	1.53	1.74	1.94	2.15	2.35	2.55	2.75	2.95	3.15	3.35	3.55	3.75	3.95	4.15	4.35	4.55	4.75	4.95	5.15	5.35	5.55	5.75	5.95	6.15	6.35	

For three-phase irregular flat or triangular spacing use $D = \sqrt[3]{ABC}$ For a two-phase line the spacing is the average distance between centers of conductors of the same phase.

xx. At a temperature of 65° C (149° F) these resistance values would be increased by 15 percent. They are based upon a conductivity for copper of 97.3— for aluminum of 61 percent. They do not take into account skin effect; this should be considered when the larger conductors are used, particularly at the higher frequencies. No allowance has been made for increased length due to sag when the conductors are suspended. The resistance values for the stranded conductors are two percent greater than for a solid rod of cross section equal to the total cross section of the wires of the cable. For stranded conductors D was taken as the diameter of a solid rod of equivalent cross sectional area. Actually D for stranded conductors is slightly greater, resulting in slightly less reactance than the table values. The table values are therefore conservative. In calculating the reactance values for the steel reinforced aluminum cable the presence of the steel strands was ignored.

TABLE VII—RATIO OF 60 CYCLE REACTANCE TO RESISTANCE AT 25° C

THE RESISTANCE VOLTS HAVING BEEN DETERMINED (AT 25° C) THE REACTANCE VOLTS MAY BE FOUND BY MULTIPLYING THE RESISTANCE VOLTS BY THE CONSTANTS GIVEN IN TABLE BELOW FOR THE SPACING AND SIZE OF CONDUCTORS CONTEMPLATED. THE RATIO FOR OTHER FREQUENCIES IS $\frac{F}{60}$ TIMES THE TABLE VALUES. FOR A TEMPERATURE OF 65° C (149° F) MULTIPLY TABLE VALUES BY .87

MATERIAL	TYPE	DIAMETER IN INCHES	B & S NO.	AREA IN CIRCULAR MILLS	RESISTANCE OF A SINGLE CONDUCTOR FOR IN OHMS AT 25° C (77° F) X X		DISTANCE (D) BETWEEN CENTERS OF CONDUCTORS X																										
					PER 1000 FEET	PER MILE	1'	2'	3'	4'	5'	6'	8'	12'	18'	2'	3'	4'	5'	6'	7'	8'	9'	11'	13'	15'	17'	19'	21'	23'	25'		
							1'	2'	3'	4'	5'	6'	8'	12'	18'	2'	3'	4'	5'	6'	7'	8'	9'	11'	13'	15'	17'	19'	21'	23'	25'		
COPPER	STRAINED	1/2	1000000	1000000	0.0557	0.321	1.10	0.55	0.36	0.24	0.17	0.13	0.10	0.08	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01			
		1/4	500000	500000	0.0652	0.377	1.22	0.61	0.40	0.27	0.19	0.14	0.11	0.08	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01		
		3/16	300000	300000	0.0739	0.418	1.31	0.66	0.43	0.29	0.21	0.15	0.12	0.09	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		1/8	150000	150000	0.0824	0.459	1.39	0.70	0.46	0.31	0.23	0.17	0.13	0.10	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		5/32	90000	90000	0.0908	0.499	1.46	0.74	0.49	0.33	0.25	0.19	0.14	0.11	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		3/16	45000	45000	0.1000	0.539	1.53	0.78	0.52	0.35	0.27	0.21	0.16	0.12	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		1/8	22500	22500	0.1100	0.579	1.60	0.82	0.55	0.37	0.29	0.23	0.18	0.14	0.11	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		5/32	11250	11250	0.1200	0.619	1.67	0.86	0.58	0.39	0.31	0.25	0.20	0.15	0.12	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		3/16	5625	5625	0.1300	0.659	1.74	0.90	0.61	0.41	0.33	0.27	0.22	0.17	0.13	0.11	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	
		1/8	2812	2812	0.1400	0.700	1.81	0.94	0.64	0.43	0.35	0.29	0.24	0.19	0.14	0.12	0.11	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	
	SOLID	1/2	100000	100000	0.0557	0.321	1.10	0.55	0.36	0.24	0.17	0.13	0.10	0.08	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01		
		1/4	50000	50000	0.0652	0.377	1.22	0.61	0.40	0.27	0.19	0.14	0.11	0.08	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		3/16	30000	30000	0.0739	0.418	1.31	0.66	0.43	0.29	0.21	0.15	0.12	0.09	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	
		1/8	15000	15000	0.0824	0.459	1.39	0.70	0.46	0.31	0.23	0.17	0.13	0.10	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		5/32	9000	9000	0.0908	0.499	1.46	0.74	0.49	0.33	0.25	0.19	0.14	0.11	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		3/16	4500	4500	0.1000	0.539	1.53	0.78	0.52	0.35	0.27	0.21	0.16	0.12	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		1/8	2250	2250	0.1100	0.579	1.60	0.82	0.55	0.37	0.29	0.23	0.18	0.14	0.11	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		5/32	1125	1125	0.1200	0.619	1.67	0.86	0.58	0.39	0.31	0.25	0.20	0.15	0.12	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		3/16	562	562	0.1300	0.659	1.74	0.90	0.61	0.41	0.33	0.27	0.22	0.17	0.13	0.11	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01	0.01
		1/8	281	281	0.1400	0.700	1.81	0.94	0.64	0.43	0.35	0.29	0.24	0.19	0.14	0.12	0.11	0.10	0.09	0.08	0.07	0.06	0.05	0.04	0.03	0.02	0.02	0.01	0.01	0.01	0.01	0.01	0.01

For three-phase regular flat spacing use $D = 1.26 A$. For three-phase irregular flat or triangular spacing use $D = \sqrt{ABC}$. For a two-phase line the spacing is the average distance between centers of conductors of the same phase.

xx At a temperature of 65° C (149° F) these resistance values would be increased by 15 percent. They are based upon a conductivity for copper of 97.3—for aluminum of 51 percent. They do not take into account skin effect; this should be considered when the larger conductors are used, particularly at the higher frequencies. No allowance has been made for increased length due to sag when the conductors are suspended. The resistance values for the stranded conductors are two percent greater than for a solid rod of cross section equal to the total cross section of the wires of the cable.

For stranded conductors D was taken as the diameter of a solid rod of equivalent cross sectional area. Actually D for stranded conductors is slightly greater, resulting in slightly less reactance than the table values. The table values are therefore conservative. In calculating the reactance values for the steel reinforced aluminum cable the presence of the steel strands was ignored.

TABLE VIII—CAPACITANCE TO NEUTRAL PER 1000 FEET OF SINGLE BARE CONDUCTOR

MATERIAL	TYPE	DIAMETER IN INCHES	R & S NO.	AREA IN CIRCULAR MILS	DISTANCE (D) BETWEEN CENTERS OF CONDUCTORS X																					
					1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19
COPPER					CAPACITANCE (C) TO NEUTRAL IN MICROFARADS PER 1000 FEET OF EACH CONDUCTOR OF A SINGLE-PHASE OR OF A SYMMETRICAL THREE-PHASE LINE. THE VALUES FOR CAPACITANCE WERE DERIVED FROM THE EQUATION $C = \frac{.007354}{\log_{10} \left[\frac{D}{2R} + \left(\frac{D}{2R} \right)^2 - 1 \right]^{1/2}}$ R BEING THE RADIUS OF THE CONDUCTOR EXPRESSED IN SAME TERMS AS D X X THE CAPACITANCE BETWEEN CONDUCTORS EQUALS ONE HALF THE TABLE VALUES																					
					1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19
SOLID	STRAINED	1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
		1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'	5'	6'	8'	12'	18'	2	3	4	5	6	7	8	9	11	13	15	17	19	21	23	25		
1/2	1'	2'	4'																							

TABLE IX—25 CYCLE CAPACITY SUSCEPTANCE TO NEUTRAL PER MILE OF SINGLE BARE CONDUCTOR

MICROMHOS PER MILE OF EACH CONDUCTOR OF A SINGLE-PHASE OR OF A SYMMETRICAL THREE-PHASE LINE—THE SUSCEPTANCE VALUES WERE DERIVED FROM THE EQUATION $b = 2\pi f C$ THE CHARGING CURRENT IN AMPERES PER MILE OF SINGLE CONDUCTOR TO NEUTRAL — THE (SUSCEPTANCE FROM TABLE X (VOLTS TO NEUTRAL) X 10^{-6}) THE SUSCEPTANCE BETWEEN CONDUCTORS EQUALS ONE HALF THE TABLE VALUES

MATERIAL	TYPE	DIAMETER IN INCHES	B & S NO	AREA IN CIRCULAR MILS	DISTANCE (D) BETWEEN CENTERS OF CONDUCTORS X																												
					1'	2'	3'	4'	5'	6'	8'	12'	18'	2	3	4	6	6	7	8	9	11	13	15	17	19	21	23	26				
COPPER	SOLID	1/4	102	100,000	213	115	820	790	710	618	523	433	345	270	214	166	128	98	75	58	44	33	25	19	14	11	8	6	5	4			
		1/8	104	100,000	137	70	878	763	692	603	513	423	333	258	192	145	107	80	60	45	34	26	19	14	10	7	5	4	3	2	2		
		3/16	106	100,000	177	107	862	760	687	597	507	417	327	252	186	139	101	74	54	40	30	22	16	12	8	6	4	3	2	2	2	2	
		1/4	108	100,000	148	105	825	723	640	550	460	370	280	205	139	101	74	54	40	30	22	16	12	8	6	4	3	2	2	2	2	2	
		3/16	110	100,000	151	98.6	805	710	627	537	447	357	267	192	126	90	63	43	31	23	17	12	8	6	4	3	2	2	2	2	2	2	
		1/4	112	100,000	136	92.8	772	683	626	534	442	350	260	185	119	83	56	39	28	20	14	10	7	5	3	2	2	2	2	2	2	2	2
		3/16	114	100,000	129	87.9	750	670	613	521	429	337	247	172	106	70	43	27	19	13	9	6	4	3	2	2	2	2	2	2	2	2	2
		1/4	116	100,000	122	83.0	732	652	595	503	411	319	229	154	88	52	35	24	17	12	8	5	3	2	2	2	2	2	2	2	2	2	2
		3/16	118	100,000	114	78.1	714	634	577	485	393	301	211	136	72	36	20	13	9	6	4	3	2	2	2	2	2	2	2	2	2	2	2
		1/4	120	100,000	107	73.2	700	620	563	471	379	287	197	122	56	20	11	7	5	3	2	2	2	2	2	2	2	2	2	2	2	2	2
		3/16	122	100,000	100	68.3	690	610	553	461	369	277	187	112	40	14	8	5	3	2	2	2	2	2	2	2	2	2	2	2	2	2	2
		1/4	124	100,000	93	63.4	680	600	543	451	359	267	177	102	24	8	4	2	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	126	100,000	86	58.5	670	590	533	441	349	257	167	92	8	3	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	128	100,000	79	53.6	660	580	523	431	339	249	157	82	2	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	130	100,000	72	48.7	650	570	513	421	329	239	147	72	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	132	100,000	65	43.8	640	560	503	411	321	231	137	62	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	134	100,000	58	38.9	630	550	493	401	311	221	127	52	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	136	100,000	51	34.0	620	540	483	391	301	211	117	42	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	138	100,000	44	29.1	610	530	473	381	291	201	107	32	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	140	100,000	37	24.2	600	520	463	371	281	191	97	22	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	142	100,000	30	19.3	590	510	453	361	271	181	87	12	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	144	100,000	23	14.4	580	500	443	351	261	171	77	2	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	146	100,000	16	9.5	570	490	433	341	251	161	67	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	148	100,000	9	4.6	560	480	423	331	241	151	57	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		3/16	150	100,000	2	0	550	470	413	321	231	141	47	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
		1/4	152	100,000	0	0	540	460	403	311	221	131	37	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
3/16	154	100,000	0	0	530	450	393	301	211	121	27	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	156	100,000	0	0	520	440	383	291	201	111	17	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	158	100,000	0	0	510	430	373	281	191	101	7	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	160	100,000	0	0	500	420	363	271	181	91	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	162	100,000	0	0	490	410	353	261	171	81	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	164	100,000	0	0	480	400	343	251	161	71	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	166	100,000	0	0	470	390	333	241	151	61	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	168	100,000	0	0	460	380	323	231	141	51	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	170	100,000	0	0	450	370	313	221	131	41	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	172	100,000	0	0	440	360	303	211	121	31	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	174	100,000	0	0	430	350	293	201	111	21	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	176	100,000	0	0	420	340	283	191	101	11	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	178	100,000	0	0	410	330	273	181	91	1	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	180	100,000	0	0	400	320	263	171	81	0	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	182	100,000	0	0	390	310	253	161	71	0	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	184	100,000	0	0	380	300	243	151	61	0	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	186	100,000	0	0	370	290	233	141	51	0	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	188	100,000	0	0	360	280	223	131	41	0	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
3/16	190	100,000	0	0	350	270	213	121	31	0	0	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1		
1/4	192	100,000	0	0	340	260	203	111	21	0	0	1	1</																				

TABLE XI—CHARGING K.V.A. IN THREE-PHASE CIRCUITS PER MILE OF THREE-BARE CONDUCTORS

CHARGING CURRENT PER MILE (EXPRESSED IN KVA 3 PHASE) FOR A SYMMETRICAL 3 PHASE CIRCUIT AT THE AVERAGE VOLTAGES AND SPACINGS OF CONDUCTORS STATED. FOR OTHER ARRANGEMENTS OF CONDUCTORS SEE FOOT NOTES. * For OTHER SPACINGS THESE VALUES WILL VARY DIRECTLY AS THE SUSCEPTANCE VALUES OF THE SPACINGS COMPARED. THE CHARGING K.V.A. 3 PHASE — (CHARGING CURRENT IN AMPERES TO NEUTRAL) X (VOLTS TO NEUTRAL) X 3. IN DETERMINING THE CHARGING K.V.A. FOR THE ENTIRE HIGH TENSION SYSTEM THE LENGTH OF ALL BRANCH CIRCUITS MUST BE INCLUDED AS WELL AS THAT OF THE MAIN CIRCUITS.

MATERIAL	TYPE	DIAMETER IN INCHES	B & S NO.	AREA IN CIRCULAR MILS	2 5 C Y C L E S										6 0 C Y C L E S																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																									
					20 KV					30 KV					40 KV					50 KV					60 KV					70 KV					80 KV					100 KV					120 KV					140 KV					160 KV					200 KV																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																												
					4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.	4 FT.	5 FT.	6 FT.	8 FT.	10 FT.																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																		
COPPER	STRAINED	1/31	33	280	1.38	1.78	2.25	2.82	3.40	3.96	4.50	5.03	5.56	6.08	6.59	7.10	7.60	8.09	8.57	9.04	9.50	9.95	10.40	10.84	11.27	11.70	12.12	12.54	12.95	13.36	13.76	14.16	14.55	14.94	15.32	15.70	16.07	16.44	16.81	17.17	17.54	17.90	18.26	18.61	18.97	19.32	19.67	20.02	20.37	20.71	21.06	21.40	21.74	22.08	22.42	22.76	23.10	23.44	23.78	24.11	24.45	24.78	25.11	25.44	25.77	26.10	26.43	26.76	27.09	27.42	27.75	28.08	28.41	28.74	29.07	29.40	29.73	30.06	30.39	30.72	31.05	31.38	31.71	32.04	32.37	32.70	33.03	33.36	33.69	34.02	34.35	34.68	35.01	35.34	35.67	36.00	36.33	36.66	36.99	37.32	37.65	37.98	38.31	38.64	38.97	39.30	39.63	39.96	40.29	40.62	40.95	41.28	41.61	41.94	42.27	42.60	42.93	43.26	43.59	43.92	44.25	44.58	44.91	45.24	45.57	45.90	46.23	46.56	46.89	47.22	47.55	47.88	48.21	48.54	48.87	49.20	49.53	49.86	50.19	50.52	50.85	51.18	51.51	51.84	52.17	52.50	52.83	53.16	53.49	53.82	54.15	54.48	54.81	55.14	55.47	55.80	56.13	56.46	56.79	57.12	57.45	57.78	58.11	58.44	58.77	59.10	59.43	59.76	60.09	60.42	60.75	61.08	61.41	61.74	62.07	62.40	62.73	63.06	63.39	63.72	64.05	64.38	64.71	65.04	65.37	65.70	66.03	66.36	66.69	67.02	67.35	67.68	68.01	68.34	68.67	69.00	69.33	69.66	69.99	70.32	70.65	70.98	71.31	71.64	71.97	72.30	72.63	72.96	73.29	73.62	73.95	74.28	74.61	74.94	75.27	75.60	75.93	76.26	76.59	76.92	77.25	77.58	77.91	78.24	78.57	78.90	79.23	79.56	79.89	80.22	80.55	80.88	81.21	81.54	81.87	82.20	82.53	82.86	83.19	83.52	83.85	84.18	84.51	84.84	85.17	85.50	85.83	86.16	86.49	86.82	87.15	87.48	87.81	88.14	88.47	88.80	89.13	89.46	89.79	90.12	90.45	90.78	91.11	91.44	91.77	92.10	92.43	92.76	93.09	93.42	93.75	94.08	94.41	94.74	95.07	95.40	95.73	96.06	96.39	96.72	97.05	97.38	97.71	98.04	98.37	98.70	99.03	99.36	99.69	100.02	100.35	100.68	101.01	101.34	101.67	102.00	102.33	102.66	102.99	103.32	103.65	103.98	104.31	104.64	104.97	105.30	105.63	105.96	106.29	106.62	106.95	107.28	107.61	107.94	108.27	108.60	108.93	109.26	109.59	109.92	110.25	110.58	110.91	111.24	111.57	111.90	112.23	112.56	112.89	113.22	113.55	113.88	114.21	114.54	114.87	115.20	115.53	115.86	116.19	116.52	116.85	117.18	117.51	117.84	118.17	118.50	118.83	119.16	119.49	119.82	120.15	120.48	120.81	121.14	121.47	121.80	122.13	122.46	122.79	123.12	123.45	123.78	124.11	124.44	124.77	125.10	125.43	125.76	126.09	126.42	126.75	127.08	127.41	127.74	128.07	128.40	128.73	129.06	129.39	129.72	130.05	130.38	130.71	131.04	131.37	131.70	132.03	132.36	132.69	133.02	133.35	133.68	134.01	134.34	134.67	135.00	135.33	135.66	135.99	136.32	136.65	136.98	137.31	137.64	137.97	138.30	138.63	138.96	139.29	139.62	139.95	140.28	140.61	140.94	141.27	141.60	141.93	142.26	142.59	142.92	143.25	143.58	143.91	144.24	144.57	144.90	145.23	145.56	145.89	146.22	146.55	146.88	147.21	147.54	147.87	148.20	148.53	148.86	149.19	149.52	149.85	150.18	150.51	150.84	151.17	151.50	151.83	152.16	152.49	152.82	153.15	153.48	153.81	154.14	154.47	154.80	155.13	155.46	155.79	156.12	156.45	156.78	157.11	157.44	157.77	158.10	158.43	158.76	159.09	159.42	159.75	160.08	160.41	160.74	161.07	161.40	161.73	162.06	162.39	162.72	163.05	163.38	163.71	164.04	164.37	164.70	165.03	165.36	165.69	166.02	166.35	166.68	167.01	167.34	167.67	168.00	168.33	168.66	168.99	169.32	169.65	169.98	170.31	170.64	170.97	171.30	171.63	171.96	172.29	172.62	172.95	173.28	173.61	173.94	174.27	174.60	174.93	175.26	175.59	175.92	176.25	176.58	176.91	177.24	177.57	177.90	178.23	178.56	178.89	179.22	179.55	179.88	180.21	180.54	180.87	181.20	181.53	181.86	182.19	182.52	182.85	183.18	183.51	183.84	184.17	184.50	184.83	185.16	185.49	185.82	186.15	186.48	186.81	187.14	187.47	187.80	188.13	188.46	188.79	189.12	189.45	189.78	190.11	190.44	190.77	191.10	191.43	191.76	192.09	192.42	192.75	193.08	193.41	193.74	194.07	194.40	194.73	195.06	195.39	195.72	196.05	196.38	196.71	197.04	197.37	197.70	198.03	198.36	198.69	199.02	199.35	199.68	200.01	200.34	200.67	201.00	201.33	201.66	201.99	202.32	202.65	202.98	203.31	203.64	203.97	204.30	204.63	204.96	205.29	205.62	205.95	206.28	206.61	206.94	207.27	207.60	207.93	208.26	208.59	208.92	209.25	209.58	209.91	210.24	210.57	210.90	211.23	211.56	211.89	212.22	212.55	212.88	213.21	213.54	213.87	214.20	214.53	214.86	215.19	215.52	215.85	216.18	216.51	216.84	217.17	217.50	217.83	218.16	218.49	218.82	219.15	219.48	219.81	220.14	220.47	220.80	221.13	221.46	221.79	222.12	222.45	222.78	223.11	223.44	223.77	224.10	224.43	224.76	225.09	225.42	225.75	226.08	226.41	226.74	227.07	227.40	227.73	228.06	228.39	228.72	229.05	229.38	229.71	230.04	230.37	230.70	231.03	231.36	231.69	232.02	232.35	232.68	233.01	233.34	233.67	234.00	234.33	234.66	234.99	235.32	235.65	235.98	236.31	236.64	236.97	237.30	237.63	237.96	238.29	238.62	238.95	239.28	239.61	239.94	240.27	240.60	240.93	241.26	241.59	241.92	242.25	242.58	242.91	243.24	243.57	243.90	244.23	244.56	244.89	245.22	245.55	245.88	246.21	246.54	246.87	247.20	247.53	247.86	248.19	248.52	248.85	249.18	249.51	249.84	250.17	250.50	250.83	251.16	251.49	251.82	252.15	252.48	252.81	253.14	253.47	253.80	254.13	254.46	254.79	255.12	255.45	255.78	256.11	256.44	256.77	257.10	257.43	257.76	258.09	258.42	258.75	259.08	259.41	259.74	260.07	260.40	260.73	261.06	261.39	261.72	262.05	262.38	262.71	263.04	263.37	263.70	264.03	264.36	264.69	265.02	265.35	265.68	266.01	266.34	266.67	267.00	267.33	267.66	267.99	268.32	268.65	268.98	269.31	269.64	269.97	270.30	270.63	270.96	271.29	271.62	271.95	272.28	272.61	272.94	273.27	273.60	273.93	274.26	274.59	274.92	275.25	275.58	275.91	276.24	276.57	276.90	277.23	277.56	277.89	278.22	278.55	278.88	279.21	279.54	279.87	280.20	280.53	280.86	281.19	281.52	281.85	282.18	282.51	282.84	283.17	283.50	283.83	284.16	284.49	284.82	285.15	285.48	285.81	286.14	286.47	286.80	287.13	287.46	287.79	288.12	288.45	288.78	289.11	289.44	289.77	290.10	290.43	290.76	291.09	291.42	291.75	292.08	292.41	292.74	293.07	293.40	293.73	294.06	294.39	294.72	295.05	295.38	295.71	296.04	296.37	296.70	297.03	297.36	297.69	298.02	298.35	298.68	299.01	299.34	299.67	300.00	300.33	300.66	300.99	301.32	301.65	301.98	302.31	302.64	302.97	303.30	303.63	303.96	304.29	304.62	304.95	305.28	305.61	305.94	306.27	306.60	306.93	307.26	307.59	307.92	308.25	308.58	308.91	309.24	309.57	309.90	310.23	310.56	310.89	311.22	311.55	311.88	312.21	312.54	312.87	313.20	313.53	313.86	314.19	314.52	314.85	315.18	315.51	315.84	316.17	316.50	316.83	317.16	317.49	317.82	318.15	318.48	318.81	319.14	319.47	319.80	320.13	320.46	320.79	321.12	321.45	321.78	322.11	322.44	322.77	323.10	323.43	323.7

Microfarads per 1000 feet of single conductor to neutral.

or

$$C = \frac{0.03883}{\log_{10} \frac{D}{R}} \dots\dots\dots (16)$$

Microfarads per mile of single conductor to neutral.

The above formulas are only applicable to ordinary overhead circuits when the distance from the conductor to other conductors, particularly the earth, is large compared to their distance apart. However, since the effect of the earth is usually small in most practical cases, the formulas give a very close approximation to the actual capacitance of overhead circuits.

The values of capacitance in Table VIII were derived by using formula (13). For calculating the capacitance for the stranded conductors, the actual overall diameter of the cable was taken. This introduces a small error which is negligible except for very close spacings not used in high tension transmission lines employing bare conductors.

CHARGING CURRENT

RELATION OF CHARGING CURRENTS OF SINGLE AND THREE-PHASE SYSTEMS

The diagrams (Fig. 11) may assist in forming a clear understanding of the relation of charging current

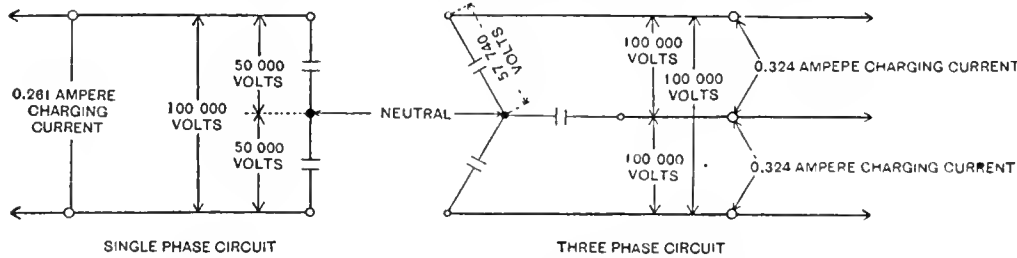


FIG. 11—CHARGING CURRENT IN SINGLE AND THREE-PHASE CIRCUITS

to susceptance for single and three-phase circuits. In the following consideration No. 0000 stranded copper conductors will be assumed as spaced nine feet between any two conductors, frequency 60 cycles, voltage 100 000 volts between conductors. Voltage to neutral will therefore be, for single phase circuit, 50 000 volts and for three-phase circuit 57 740 volts. Distance of transmission one mile. From Table VIII, a capacitance to neutral of 0.00282 microfarads per 1000 feet is obtained which is equivalent to 0.0149 microfarads per conductor to neutral for this one mile of circuit. The susceptance will therefore be as follows:—

Per conductor to neutral $2 \pi f C_n = 5.62$ microhms
 Between conductors $2 \pi f C_{12} = 2.81$ microhms

For Single-Phase Circuit—To neutral $5.62 \times 50\ 000 \times 10^6 = 0.281$ amperes or between conductors $2.81 \times 100\ 000 \times 10^6 = 0.281$ amperes therefore charging k.v.a. is $0.281 \times 50\ 000 \times 2 = 28.1$ k.v.a. single phase or $0.281 \times 100\ 000 = 28.1$ k.v.a. single phase.

For a Three-Phase Circuit—To neutral $5.62 \times 57\ 740 \times 10^6 = 0.324$ amperes. Therefore charging k.v.a. is $0.324 \times 57\ 740 \times 3 = 56.2$ k.v.a. three-phase.

It will be seen from the above that the charging current per conductor in the three-phase symmetrical

system is 15.5 percent greater than in the single-phase system, and the resulting charging k.v.a. is just double that of the single-phase system. The charge on any particular conductor is in phase with the voltage between that conductor and the neutral and the charging current for that conductor is 90 degrees ahead of the voltage drop from that conductor to neutral.

Grounding of the neutral point of a system has no effect upon the charging current when the system is in static balance. In determining the total charging current to be supplied by a given generating station, it should be remembered that in cases of duplicate transmission circuits, when both circuits are excited, the charging current will be approximately double what it would be if only one of the circuits were in use.

Tables IX and X contain values for capacitance susceptance to neutral in micromhos per mile of conductor. As indicated, the charging current in amperes per mile of single conductor to neutral = the (susceptance from table) \times (volts to neutral) $\times 10^{-6}$. Thus in a three-phase, 60 cycle, 100 000 volt, (57 740 volts to neutral), symmetrical circuit, the No. 0000 stranded conductors being arranged at the corners of an equilateral triangle spaced nine feet apart, the charging current per mile would be determined as follows:—

$$5.62 \times 57\ 740 \times 10^{-6} = 0.3245 \text{ amperes to neutral}$$

$$\text{or } 0.3245 \times 57\ 740 = 18.737 \text{ k.v.a. to neutral}$$

$$18.737 \times 3 = 56.2 \text{ K.v.a. total three phase}$$

Table XI is an extension of Tables IX and X from which values in k.v.a., three-phase for charging current have

been calculated for certain assumed spacings and average voltages. In the case cited above it was found that the charging current would be 56.2 k.v.a., three-phase per mile. Table XI gives this value directly for the conditions specified.

CHARGING CURRENT AT ZERO LOAD

The term charging current of a transmission circuit refers to the amount of current which flows into the circuit at the supply end with normal voltage held at the receiver end at zero load. If the circuit is long, its capacitance will be high and therefore the voltage at the supply end may be considerably less than at the receiver end. For instance a 60 cycle circuit 300 miles long, having certain constants will, with 100 000 volts maintained at the receiver end, have a voltage of only 80 000 volts at the supply end at zero load. This same circuit will at full load and 100 000 volts maintained at the receiver end, require 120 000 volts at the supply end. It is evident therefore that, since the charging current varies with the voltage, if the circuit has much capacitance the voltage along the circuit, and particularly near the supply end, will vary to a large extent

and consequently the charging current of the circuit will be different for different loads.

In case of the 300 mile circuit referred to above, the charging current at zero load will be very much less than it is at full load, because the average voltage at zero load is less than the average voltage at full load. At zero load the average voltage is less and at full load it is greater than the receiver end voltage.

It is customary to calculate the total charging current for the circuit by multiplying the total susceptance by the receiver end voltage. This would be correct if the voltage throughout the length of the circuit were held constant and of the same value as at the receiver end. This condition is approximately met within commercial lines and this method of determining the

susceptance by the receiver voltage. For a circuit 300 miles long the error in charging current is only two percent for 25 cycles and seven percent for 60 cycle circuits. The error in charging k.v.a. is four percent for 25 cycle and 32 percent for 60 cycle circuits.

RELATION OF INDUCTANCE TO CAPACITANCE

As conductors are brought closer together, the inductance decreases and the capacitance increases. These values change with changes in spacings between conductors in such a manner that their product $L \times C$ is practically a constant for all spacings (except very close spacings such as used in low-voltage service and lead-covered cables) and for all sizes of conductors. If there were no losses encountered by the electric

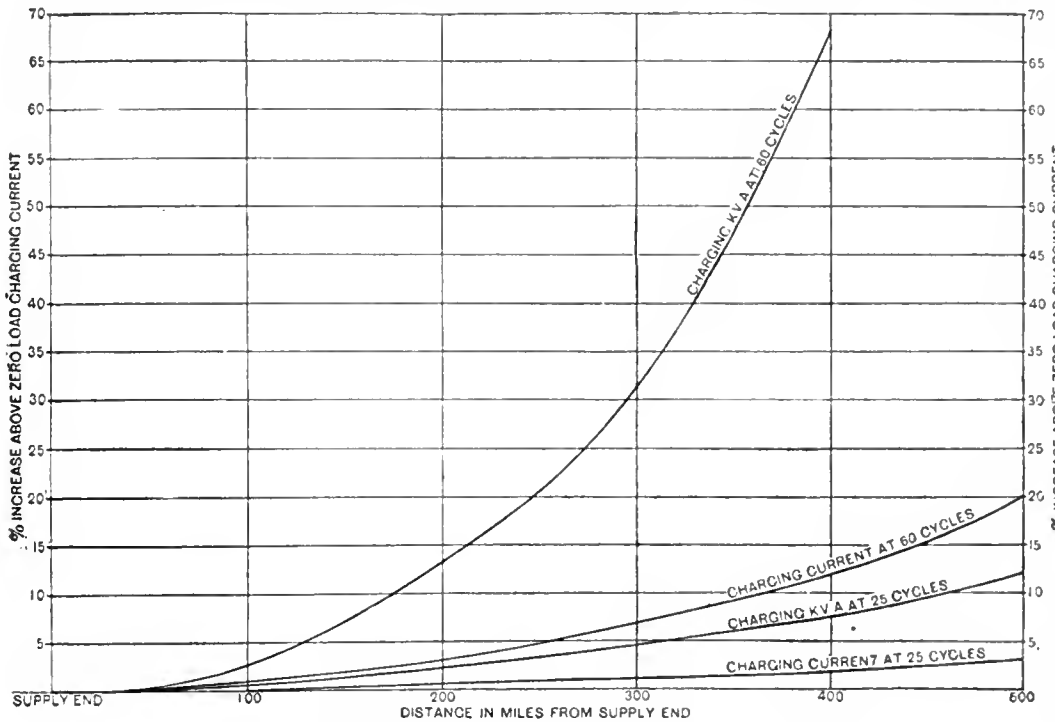


FIG. 12—CHARGING CURRENT AT ZERO LOAD FOR VARIOUS LENGTHS

At zero load the voltage (on account of the effect of capacitance) decreases as the supply end of the circuit is approached. The charging current at points along the circuit decreases directly as the voltage. If the charging current for zero load is estimated by the approximate method based upon the receiver voltage being maintained throughout the length of the circuit the result will be too high. The error will increase as the length of the circuit is increased; it will also increase rapidly as the frequency is raised. The error in the resulting K.V.A. required to charge the circuit will therefore increase very rapidly with an increase in distance or frequency. The curves below represent an approximation of this error.

total charging current is therefore sufficiently accurate for most practical purposes.

For the purpose of making exact calculation of the total current at the supply end of long circuits, the charging current must be calculated by mathematical formulas which accurately take into account the change in voltage along the circuit at zero load. This will be taken up in a later article. It may be interesting to note approximately, however, how the charging current and charging k.v.a., as determined by the above method, varies from what it would be if calculated by the rigorous formula. The curves in Fig. 12 represent an approximation to the error when calculating the charging current at zero load by multiplying the total

propagation in the conductors themselves the product of L and C would be a constant for all spacings and sizes of conductors.

In Table C is indicated the relation of the total inductance and capacitance, and their product, in two bare parallel conductors in air for a circuit one mile long. The values for L are in millihenries and for C in microfarads. Since the formulas by which L and C were calculated account for the flux within the conductors themselves, the product LC is not a constant, as will be seen by the tabulated values, although for the larger spacings such as used in high-

tension transmission the product is nearly a constant.

TABLE C—PRODUCT OF (TOTAL) L AND (TOTAL) C

Solid Conductors		Spacing Inches	Inductance L , Formula (4)	Capacitance C , Formula (14)	Product $L \cdot C$
Size	Diam. Inches				
1 000 000	1.00	2	1.053	0.03395	0.03575
1 000 000	1.00	24	2.653	0.01155	0.03064
1 000 000	1.00	300	4.279	0.00695	0.02974
0000	0.46	2	1.553	0.02079	0.03228
0000	0.46	24	3.153	0.00961	0.03030
0000	0.46	300	4.779	0.00623	0.02977

RELATION OF INDUCTANCE AND CAPACITANCE TO LIGHT VELOCITY

The propagation of the electric and the magnetic

fields in a dielectric, such as air, is the same as that of light. Along a transmission line it is retarded only slightly due to losses or the fact that the current is not confined to the surface of the conductors. If the inductance inside the conductors is negligible, then the velocity of the electric and the magnetic fields is the same as light, that is approximately 186 000 miles per second or approximately 3×10^{10} cm. per second. For high-tension transmission lines of large spacings, the inductance inside the conductor is relatively small, so that the speed of the electric field is practically that of light.

The following relation exists between inductance L in henries, capacity C in farads and velocity of light V per second:—

$$LC \text{ (in air)} = \frac{1}{V^2} \text{ or, } V = \frac{1}{\sqrt{LC}} \dots\dots\dots (17)$$

Thus it will be seen that if either L or C is known, the other may be determined since the velocity of light V is known. If values for L and C are taken which include the inductance inside the conductors, particularly if the conductors are very close together, it would be necessary to assume a velocity of electric propagation

somewhat less than that of light. If, on the other hand, the values for L and C external to the conductors are taken, then the above equation is rigidly correct.

In Table C, it was shown that for No. 0000 conductors, 300 inch spacing, the total values of L and C were for a single-phase line,—

$$L = 0.004\,779 \text{ henries per mile of circuit.}$$

$$C = 0.000\,000\,006\,23 \text{ farads per mile of circuit.}$$

therefore, $V = \frac{1}{\sqrt{0.004\,779 \times 0.000\,000\,006\,23}} =$
 183 000 miles per second (18)
 which is less than the speed of light.

If we take the inductance in the air space between the conductors, Formula (2); we arrive at the values,—

$$L = 0.004\,617\,9 \text{ henries per mile of circuit.}$$

$$C = 0.000\,000\,006\,23 \text{ farads per mile of circuit.}$$

therefore $V = \frac{1}{\sqrt{0.004\,617\,9 \times 0.000\,000\,006\,23}} =$
 186 000 miles per second (19)
 which is approximately the speed of light.

CHAPTER III

QUICK ESTIMATING TABLES

FOR every occasion where a complete calculation of a long distance transmission line is made, there are many where the size of wire needed to transmit a given amount of power economically is required quickly. This knowledge is, moreover, the basis for all transmission line calculations, as all methods of calculating regulation presuppose that the size of wire is known. To determine quickly and with the least possible calculation the approximate size of conductor corresponding to a given I^2R transmission loss for any ordinary voltage or distance, is the function of Tables XII to XXI inclusive. By including so many transmission voltages it is not intended to indicate that any of them might equally well be selected for a new installation. On the contrary it is very desirable in the consideration of a new installation, to eliminate consideration of some of the voltages now in use. This point will be considered later.

Since both the power-factor of the load, and the charging current of the circuit, as well as any change in the resistance of the conductors, will alter the I^2R loss, it is evident that it is impractical to present tables which will take into account the effect of all of these variables. The accompanying tables do, however, give the percentage I^2R loss corresponding to the two temperatures (25 and 65 degrees C) ordinarily encountered in practice and the usual load power-factors of unity and 80 percent lagging, upon which the k.v.a. values of the tables are based. The effect, however, of charging current, corona or leakage loss is not taken into account in these table values. The latter two (corona and leakage) are usually small and need not be considered here. The effect of charging current, may, however, with long circuits be material and will be discussed.

The values of k.v.a. in these tables are based upon the following percentage I^2R loss in transmission (neglecting the effect of charging current) :—

	Percent Loss At 25°C	Percent Loss At 65°C
Load at 100 percent P-F.	8.66	10.0
Load at 80 percent P-F.	10.8	12.5

These loss values are based upon the power delivered at the end of the circuit as 100 percent, and not upon the power at the supply end. If raising or lowering transformers are employed, the loss and voltage drop in them will, of course, be in addition to the above.

At first glance, some of these tables may appear to have been carried to extremes of k.v.a. values for the conductor sizes. This is because the tables are calculated for ten percent loss, (at 100 percent power-

factor and 65 degrees C) whereas the permissible loss is frequently much less than ten percent. As the loss is directly proportional to the load, the permissible loads for a given size wire and distance can be read almost directly for any loss. Thus for a two percent loss the permissible k.v.a. will be two-tenths the table values. Conversely, the size of wire to carry a given k.v.a. load at two percent loss will be the same as will carry five ($10 \div 2$) times the k.v.a. at ten percent loss. In other words to find the size of wire to carry a given k.v.a. load at any desired percent loss, find the ratio of the desired I^2R loss to the I^2R loss upon which the table values are based (corresponding of course to the temperature and the load power-factor). Divide this ratio into the k.v.a. to be transmitted. The result will be the table k.v.a. value corresponding to the desired I^2R loss.

For example:—Assume 400 k.v.a. is to be delivered a distance of 14 miles at 6000 volts, three-phase, and 80 percent power-factor lagging, at an assumed temperature of 25 degrees C. Table XV indicates that this condition will be met with an I^2R loss of 10.8 percent if No. 0 copper or 167 800 circ. mil aluminum conductors are used.

Now assume that the I^2R loss should not exceed 5.4 percent, in place of 10.8 percent (upon which the table values are based). $5.4 \div 10.8 = 0.5$ and $400 \div 0.5 = 800$ k.v.a. as the table value corresponding to an I^2R loss of 5.4 percent. The conductors corresponding to 800 k.v.a. table value (5.4 percent I^2R loss) will be seen to be No. 0000 copper or 336 420 circ. mil aluminum.

If conductors corresponding to 15 percent I^2R loss are desired the same procedure will be followed:— $15 \div 10.8 = 1.39$ and $400 \div 1.39 = 287$ k.v.a. table value. This table value corresponds to approximately No. 1 copper or 133 220 circ. mil aluminum conductors.

The table k.v.a. values have been tabulated for various distances. Should the actual distance be different from the table values and it is desired to obtain k.v.a. values corresponding to the losses upon which the table k.v.a. values have been calculated, the following procedure may be followed:—

For a given I^2R loss in a given conductor (effect of charging current neglected) the k.v.a. \times feet or the k.v.a. \times miles is a constant. Thus Table XII indicates that for 2 000 000 circ. mil cable, 756 000 k.v.a. \times feet is the constant; that is 755 k.v.a. may be transmitted 1000 feet; 378 k.v.a., 2000 feet, and so on. If the actual distance to be transmitted is 1300 feet the corresponding k.v.a. value will be $756\,000 \div 1300$ or 581 k.v.a. Usually the k.v.a. value can readily be approximated

for any distance with sufficient accuracy for the purpose for which these quick estimating tables are presented. One way of doing this would be as follows:— The k.v.a. value corresponding to 2500 ft. is 302 k.v.a.

TABLE XII—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED. BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	FOR LOAD POWER-FACTOR OF 100%—8.68% LOSS— AT 25° C FOR LOAD POWER-FACTOR OF 80%—10.8% LOSS— AT 65° C 10.0% LOSS 12.5 LOSS																
			220 VOLTS DELIVERED																
			50 FEET	100 FEET	150 FEET	200 FEET	250 FEET	300 FEET	400 FEET	500 FEET	600 FEET	750 FEET	1000 FEET	1500 FEET	2000 FEET	2500 FEET	3500 FEET	5000 FEET	1 MILE
2 000 000	1 800 000	1 700 000	15725	7542	5042	3781	3025	2521	1890	1512	1260	1008	754	504	378	302	216	151	143
1 600 000	1 500 000	1 400 000	12100	6050	4033	3025	2420	2016	1512	1210	1008	806	605	403	302	242	173	121	114
1 200 000	1 100 000	1 000 000	9047	4523	3015	2262	1808	1507	1131	905	753	603	452	301	226	181	129	90	85
950 000	900 000	850 000	7224	3612	2408	1804	1445	1204	903	722	602	482	361	241	181	144	103	72	68
800 000	750 000	700 000	6050	3025	2017	1512	1210	1008	754	605	504	403	302	202	151	113	84	60	57
650 000	600 000	550 000	4914	2457	1638	1228	983	819	614	491	409	328	244	164	123	98	70	49	45
500 000	450 000	400 000	3781	1890	1260	945	754	630	472	378	315	252	189	126	94	75	54	38	34
350 000	300 000	250 000	2645	1322	882	661	529	441	330	264	220	176	132	88	66	53	38	24	25
211 600	167 772	133 079	1600	800	533	400	320	244	200	160	133	107	80	63	40	32	23	16	15
1 2	105 560	83 694	800	400	260	200	160	133	100	80	66	53	40	27	20	16	11	8	8
3	52 624	41 738	396	198	132	99	79	63	46	39	33	26	20	13	10	8	6	4	4
4	33 088	26 244	251	125	84	62	50	42	31	25	21	16	12	8	6	5	4	3	3
5	26 244	20 822	198	99	66	50	40	33	25	20	16	13	10	8	6	5	4	3	2
6	20 822	16 512	151	75	50	37	31	25	19	15	12	10	8	6	5	4	3	2	2
			440 VOLTS DELIVERED																
			50 FEET	100 FEET	150 FEET	200 FEET	250 FEET	300 FEET	400 FEET	500 FEET	600 FEET	750 FEET	1000 FEET	1500 FEET	2000 FEET	2500 FEET	3500 FEET	5000 FEET	1 MILE
2 000 000	1 800 000	1 700 000	60500	30250	20166	15125	12100	10083	7562	6050	5042	4033	3025	2017	1512	1210	844	605	573
1 600 000	1 500 000	1 400 000	48400	24200	16133	12100	9680	8066	6050	4840	4033	3224	2420	1613	1210	968	691	484	458
1 200 000	1 100 000	1 000 000	36377	18188	12062	9047	7237	6031	4523	3618	3015	2412	1809	1206	904	724	517	362	343
950 000	900 000	850 000	28896	14448	9632	7224	5719	4816	3612	2889	2408	1926	1444	963	722	578	412	289	273
800 000	750 000	700 000	24300	12150	8064	6050	4940	4033	3025	2420	2016	1613	1210	807	605	484	346	242	226
650 000	600 000	550 000	19453	9827	6532	4913	3931	3276	2454	1945	1638	1310	982	755	491	393	281	194	186
500 000	450 000	400 000	15125	7562	5042	3781	3025	2521	1890	1512	1260	1008	756	504	378	302	216	151	143
350 000	300 000	250 000	10579	5289	3526	2644	2115	1763	1322	1058	881	705	529	353	264	211	151	106	100
211 600	167 772	133 079	6400	3200	2133	1600	1280	1067	800	640	533	426	320	213	160	128	91	64	61
1 2	105 560	83 694	3200	1600	1064	800	640	533	400	320	266	213	160	107	80	64	46	32	30
3	52 624	41 738	1584	792	529	396	317	264	198	158	132	105	79	53	40	32	22	16	15
4	33 088	26 244	1003	502	334	253	201	158	125	100	83	67	50	35	25	20	14	10	10
5	26 244	20 822	791	395	264	197	158	132	105	79	66	53	39	26	16	11	8	6	6
6	20 822	16 512	629	314	210	157	126	105	78	63	52	41	32	21	14	10	7	5	5

The heating limitations may, for the shorter distances, particularly if insulated or concealed conductors are employed, necessitate the use of larger conductors, resulting in a correspondingly less transmission loss. In the case of insulated or concealed conductors, should the k.v.a. values fall near or to the left of the heavy line, consult Table XXV for insulated or Table XXIII for bare conductors. The reactance for the larger conductors may be excessive, particularly for 60-cycle service, producing excessive voltage drop. This may be obviated by installing two or more parallel circuits or using three-conductor cables. For single-phase circuits the k.v.a. will be one-half the table values.

Hence the value corresponding to half this distance (1250 ft.) is 604 k.v.a., which is sufficiently accurate for practical purposes.

REACTANCE LIMITATIONS

The k.v.a. value of the tables naturally do not take into account the reactance of the circuit. It will be

TABLE XIII—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED, BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	FOR LOAD POWER-FACTOR OF 100%—8.66% LOSS— AT 25° C FOR LOAD POWER-FACTOR OF 80%—10.8% LOSS— AT 65° C 12.6 LOSS																
			550 VOLTS DELIVERED																
			50 FEET	100 FEET	150 FEET	200 FEET	250 FEET	300 FEET	400 FEET	500 FEET	600 FEET	750 FEET	1000 FEET	1500 FEET	2000 FEET	2500 FEET	3500 FEET	5000 FEET	1 MILE
2	2000 000		94.531	47246	31570	23363	18906	14745	11816	94.53	7877	6302	4727	3151	2363	1891	1350	94.5	864
1	1800 000		85815	42907	28605	21453	17163	14302	10727	85.81	7151	5721	4291	2860	2145	1716	1224	85.8	812
1	1700 000		80132	40064	26771	20033	16024	13355	10016	80.13	6677	5342	4007	2671	2003	1603	1144	80.1	758
1	1600 000		75625	37812	25208	18906	15135	12604	9453	75.62	6302	5042	3781	2521	1891	1512	1081	75.6	716
1	1500 000		70760	35380	23587	17690	14152	11793	8843	70.76	5896	4717	3538	2353	1769	1415	1011	70.7	670
1	1400 000		66120	33060	22040	16530	13224	11020	8245	66.12	5510	4408	3306	2204	1653	1322	945	66.1	625
1	1200 000		56542	28271	18847	14135	11308	9423	7067	56.54	4711	3769	2827	1885	1413	1131	808	56.5	535
1	1100 000		52155	26077	17385	13038	10431	8672	6519	52.15	4346	3477	2608	1738	1304	1043	745	52.2	492
1	1000 000	1590 000	47245	23632	15755	11816	9453	7877	5908	47.24	3938	3151	2363	1576	1182	945	675	47.3	448
9	950 000	1515 000	45149	22574	15050	11287	9030	7525	5613	45.15	3762	3010	2257	1505	1129	903	645	45.2	426
9	900 000	1431 000	42206	21303	14202	10652	8521	7101	5324	42.20	3550	2840	2130	1420	1065	852	608	42.2	403
9	850 000	1351 500	40064	20033	13355	10016	8013	6677	5008	40.06	3338	2671	2003	1336	1002	801	572	40.1	380
8	800 000	1272 000	37813	18907	12604	9453	7562	6302	4727	37.81	3151	2521	1891	1260	945	756	540	37.8	358
7	750 000	1192 500	35380	17385	11816	9453	7562	6302	4727	35.38	2957	2366	1739	1183	887	708	507	35.3	335
7	700 000	1113 000	33060	16530	11020	8245	6612	5510	4132	33.06	2755	2204	1653	1102	827	661	472	33.1	313
6	650 000	1033 500	30770	15355	10237	7677	6142	5118	3838	30.77	2559	2047	1535	1024	768	614	439	30.7	291
6	600 000	954 000	28271	14135	9424	7067	5614	4712	3538	28.27	2356	1888	1414	942	707	561	404	28.3	268
6	550 000	874 500	26077	13038	8693	6520	5215	4326	3260	26.07	2173	1738	1304	869	652	522	372	26.1	247
5	500 000	795 000	23633	11816	7878	5908	4726	3939	2954	23.63	1967	1575	1182	788	591	473	337	23.6	224
5	450 000	715 500	21228	10614	7076	5307	4246	3538	2653	21.22	1769	1388	1061	708	531	425	303	21.2	201
5	400 000	636 000	18965	9482	6322	4771	3793	3161	2370	18.96	1580	1243	948	632	474	379	271	18.9	180
4	350 000	556 500	16530	8245	5510	4132	3306	2755	2044	16.53	1377	1103	824	551	413	331	234	16.5	158
4	300 000	477 000	14152	7076	4723	3542	2823	2311	1771	14.15	1180	944	708	472	358	271	142	14.2	134
4	250 000	397 500	11863	5932	3954	2944	2372	1977	1483	11.86	988	791	593	395	297	237	170	11.9	112
3	211 600	336 420	10000	5000	3333	2500	2000	1666	1250	10.00	833	666	500	333	250	200	143	10.0	95
3	167 772	266 800	7940	3980	2633	1990	1582	1327	985	794	796	623	398	265	199	159	113	80	75
3	133 079	211 950	6302	3151	2101	1575	1260	1050	787	630	525	420	315	210	158	126	90	63	60
2	105 560	167 800	5000	2500	1666	1250	1000	833	625	500	416	333	250	166	125	100	71	50	47
2	83 694	133 220	3954	1977	1318	988	781	649	494	395	329	264	198	132	99	79	56	40	38
2	64 358	105 530	3134	1567	1045	783	627	522	372	313	261	209	157	105	78	63	45	31	30
3	52 624	83 640	2479	1239	826	620	496	413	310	248	207	165	124	83	62	50	35	25	24
4	41 738	66 370	1977	988	659	494	395	327	277	197	164	132	99	66	49	40	28	20	19
4	33 088	52 630	1477	783	522	392	313	261	196	156	130	104	78	52	39	31	22	16	15
6	26 244	41 740	1239	619	413	310	247	206	155	123	103	82	61	41	31	25	18	12	12
7	20 822		984	492	328	246	197	164	123	98	82	65	49	33	25	20	14	10	9
8	16 572		780	390	260	195	156	130	97	78	65	52	39	26	20	16	11	8	7
			1100 VOLTS DELIVERED																
			100 FEET	200 FEET	300 FEET	500 FEET	750 FEET	1000 FEET	2500 FEET	4000 FEET	1 MILE	1 1/4 MILES	1 1/2 MILES	2 MILES	2 1/2 MILES	3 MILES	3 1/2 MILES	4 MILES	5 MILES
2	2000 000		189062	94.531	63021	37812	25208	18906	7562	4726	3580	2865	2388	1790	1432	1194	1023	895	714
1	1800 000		171631	85815	57210	34326	22884	17163	6865	4291	3250	2600	2166	1625	1300	1083	928	812	650
1	1700 000		160265	80132	53424	32053	21368	16026	6410	4006	3035	2428	2023	1517	1214	1012	867	758	607
1	1600 000		151250	75625	50416	30250	20164	15125	6050	3781	2865	2291	1910	1432	1145	955	818	716	572
1	1500 000		141520	70760	47173	28304	18868	14152	5661	3538	2680	2144	1786	1340	1072	893	766	670	536
1	1400 000		132240	66120	44030	26948	17631	13224	5289	3306	2505	2003	1669	1252	1002	839	716	626	501
1	1200 000		113084	56542	37694	22617	15078	11308	4523	2827	2141	1713	1428	1070	856	714	612	535	428
1	1100 000		104310	52155	34770	20862	13908	10431	4172	2608	1974	1580	1317	988	790	668	562	494	395
1	1000 000	1590 000	94531	47265	31510	18906	12604	9453	3781	2363	1790	1432	1194	895	716	597	512	447	358
9	950 000	1515 000	90298	45149	30099	18059	12040	9030	3612	2257	1710	1368	1140	855	684	570	489	427	342
9	900 000	1431 000	85211	42206	28404	17042	11361	8521	3408	2130	1613	1291	1076	807	645	538	461	403	322
9	850 000	1351 500	80132	40064	26771	16026	10684	8013	3203	2003	1518	1214	1012	758	607	506	434	379	303
8	800 000	1272 000	75625	37814	25209	15126	10083	7563	3025	1891	1432	1144	955	716	573	477	409	358	286
7	750 000	1192 500	70968	35814	23634	14193	9462	7097	2839	1774	1344	1075	896	722	537	448	384	336	268
7	700 000	1113 000	66120	33060	22040	13224	8816	6612	2693	1669	1252	1002	831	624	601	477	358	313	250
6	650 000	1033 500	61421	30770	20474	12284	8189	6142	2457	1535	1163	930	775	582	466	388	332	291	232
6	600 000	954 000	56542	28271	18847	14135	11308	9453	2241	1374	1071	874	634	428	357	306	267	214	171
6	550 000	874 500	52155	26077	17385	13038	10431	8672	2086	1304	988	790	658	494	395	322	282	247	197
5	500 000	795 000	47265	23633	15755	11816	9453	7878	1890	1181	896	714	597	447	358	298	256	223	179
5	450 000	715 500	42907	21303	14202	10652	8521	7101	1498	1061	804	643	536	402	322	268	229	201	161
5	400 000	636 000	37931	18965	12644	9453	7586	6057	1517	948	718	575	479	359	287	239	205	179	143
4	350 000	556 500	33060	16530	11020	8245	6408	5000	1322	824	636	501	417	313	250	208	178	156	125
4	300 000	477 000	28337	14168	9446	7378	5833	4500	1133	708	538	430	358	268	215	179	153	134	107
4	250 000	397 500	23725	11862	7908	4745	3163	2372	949	593	444	359	300	224	179	150	128	112	89
3	211 600	336 420	20000	10000	6666	4000	2666	2000	800	500	378	303	252	189	152	126	109	94	76
3	167 772	266 800	15921	7940	5307	3184	2123	1592	637	399									

necessary in some cases of low voltage and single conductors (where the reactance is high) to use lower operating limits. This will be considered later by examples on voltage regulation.

TABLE XIV—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED, BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	FOR LOAD POWER-FACTOR OF 100%—8.66% LOSS— AT 25° C FOR LOAD POWER-FACTOR OF 80%—10.8% LOSS— AT 65° C 10.0% LOSS 12.5 LOSS																
			2200 VOLTS DELIVERED																
			100 FEET	200 FEET	300 FEET	500 FEET	750 FEET	1000 FEET	2500 FEET	4000 FEET	1 MILE	1 1/2 MILES	2 MILES	2 1/2 MILES	3 MILES	3 1/2 MILES	4 MILES	5 MILES	
2000 000			756 000	378 000	252 000	151 000	101 000	75 600	30 200	18 900	14 300	11 460	9 550	7 160	5 730	4 770	4 090	3 580	2 860
1800 000			686 000	343 000	229 000	137 000	91 500	68 600	27 400	17 100	13 000	10 400	8 670	6 500	5 200	4 330	3 710	3 250	2 600
1700 000			641 000	320 000	214 000	128 000	85 500	64 100	25 600	16 000	12 100	9 770	8 100	6 070	4 850	4 050	3 470	3 040	2 430
1600 000			605 000	302 000	202 000	121 000	80 700	60 500	24 200	15 100	11 400	9 170	7 640	5 730	4 580	3 820	3 270	2 860	2 290
1500 000			566 000	283 000	189 000	113 000	75 500	56 600	22 400	14 100	10 700	8 550	7 150	5 360	4 290	3 570	3 060	2 680	2 140
1400 000			529 000	264 000	176 000	106 000	70 500	52 900	21 100	13 200	10 000	7 900	6 680	5 010	4 010	3 340	2 860	2 500	2 000
1200 000			452 000	226 000	151 000	90 500	60 300	45 200	18 100	11 300	8 560	6 850	5 710	4 280	3 430	2 850	2 460	2 140	1 710
1100 000			417 000	208 000	139 000	83 400	55 400	41 700	16 700	10 400	7 900	6 320	5 160	3 830	3 030	2 580	2 200	1 840	1 410
1000 000	1 590 000		378 000	189 000	124 000	75 400	50 400	37 800	15 100	9 450	7 160	5 730	4 770	3 580	2 860	2 460	2 140	1 710	1 370
950 000	1 515 000		361 000	180 000	120 000	72 200	48 100	36 100	14 400	9 030	6 840	5 470	4 560	3 420	2 740	2 280	1 950	1 510	1 170
900 000	1 431 000		341 000	170 000	114 000	68 100	45 400	34 100	13 400	8 520	6 450	5 160	4 300	3 230	2 580	2 150	1 790	1 340	1 070
850 000	1 351 500		320 000	160 000	107 000	64 100	42 700	32 000	12 800	8 010	6 070	4 860	4 050	3 030	2 430	2 020	1 730	1 320	1 010
800 000	1 272 000		303 000	151 000	101 000	60 500	40 300	30 200	12 100	7 560	5 730	4 580	3 820	2 860	2 290	1 910	1 440	1 140	890
750 000	1 192 500		284 000	142 000	94 600	56 800	37 800	28 400	11 300	7 100	5 370	4 300	3 580	2 700	2 150	1 790	1 540	1 190	930
700 000	1 113 000		264 000	132 000	88 100	52 900	35 200	26 400	10 600	6 610	5 010	4 010	3 340	2 500	2 000	1 670	1 330	1 070	820
650 000	1 033 500		246 000	123 000	81 900	49 100	32 700	24 500	9 800	6 140	4 650	3 720	3 100	2 330	1 860	1 550	1 330	1 060	810
600 000	954 000		226 000	113 000	75 400	45 200	30 100	22 600	8 350	5 650	4 280	3 430	2 850	2 140	1 710	1 430	1 220	1 070	857
550 000	874 500		208 000	104 000	69 500	41 700	27 800	20 800	7 560	5 210	3 950	3 160	2 630	1 970	1 580	1 310	1 130	1 020	897
500 000	795 000		189 000	94 500	63 000	37 800	25 200	18 900	6 800	4 730	3 580	2 860	2 380	1 790	1 430	1 190	1 020	895	776
450 000	715 500		170 000	84 900	56 400	33 900	22 600	17 000	6 000	4 240	3 220	2 570	2 140	1 610	1 290	1 070	920	804	643
400 000	636 000		152 000	75 800	50 400	30 300	20 200	15 200	5 200	3 790	2 870	2 300	1 910	1 440	1 150	958	824	718	575
350 000	556 500		132 000	66 100	44 100	26 400	17 600	13 200	4 290	3 310	2 500	2 000	1 670	1 250	1 000	835	716	626	501
300 000	477 000		113 000	56 500	37 800	22 700	15 100	11 300	3 800	4 530	2 890	2 150	1 720	1 430	1 073	860	735	613	536
250 000	397 500		94 600	47 300	31 400	19 000	12 600	9 500	3 000	2 370	1 800	1 410	1 200	900	720	600	513	450	359
200 000	318 000		75 400	37 800	25 200	15 100	10 700	8 000	2 400	1 900	1 510	1 210	1 010	757	606	505	433	378	303
150 000	238 500		56 800	28 400	18 100	11 300	8 560	6 410	1 800	1 400	1 100	880	734	594	477	382	318	261	211
100 000	159 000		37 800	18 900	12 400	7 540	5 040	3 780	1 100	850	660	530	430	340	270	210	160	120	90
50 000	79 500		18 900	9 450	6 200	3 780	2 520	1 890	550	420	320	250	200	150	110	80	60	40	30
0	0		0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
1	105 560	167 800	40 000	20 000	13 300	8 000	5 330	4 000	1 600	1 000	757	604	505	380	303	252	217	189	151
2	83 694	133 220	31 600	15 800	10 500	6 330	4 220	3 160	1 260	790	600	480	400	300	239	200	170	140	110
3	66 358	103 530	25 100	12 500	8 360	5 010	3 340	2 510	1 000	630	475	380	317	237	190	158	136	118	94
4	52 624	83 640	19 800	9 900	6 610	3 970	2 640	1 980	790	500	375	300	250	188	150	125	107	94	75
5	41 738	66 370	15 800	7 910	5 270	3 160	2 110	1 580	630	395	300	240	200	150	120	100	86	75	60
6	33 088	52 630	12 500	6 270	4 180	2 510	1 670	1 250	500	313	237	190	159	118	95	79	68	59	47
7	26 244	41 740	9 900	4 960	3 310	1 980	1 320	990	400	250	188	150	125	94	75	63	54	47	37
8	20 822		7 900	3 930	2 620	1 570	1 050	790	310	197	149	119	99	79	65	54	42	37	30
9	16 572		6 250	3 120	2 080	1 250	830	620	250	156	118	95	79	65	54	47	37	30	24

4000 VOLTS DELIVERED																		
	1 MILE	1 1/2 MILES	2 MILES	2 1/2 MILES	3 MILES	3 1/2 MILES	4 MILES	5 MILES	6 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	
650 000	1 033 000	1 540 000	10 250	7 700	6 150	5 120	4 400	3 850	3 080	2 560	2 200	1 920	1 710	1 540	1 400	1 280	1 180	1 100
600 000	954 000	1 420 000	9 460	7 100	5 670	4 750	4 050	3 550	2 840	2 370	2 030	1 775	1 580	1 420	1 290	1 190	1 090	1 020
550 000	874 500	1 310 000	8 750	6 550	5 250	4 370	3 750	3 280	2 620	2 180	1 870	1 640	1 460	1 310	1 190	1 090	1 010	940
500 000	795 000	1 190 000	7 950	5 950	4 750	3 960	3 400	2 980	2 380	1 980	1 700	1 490	1 320	1 190	1 080	990	915	850
450 000	715 500	1 070 000	7 120	5 350	4 270	3 560	3 060	2 670	2 140	1 780	1 530	1 330	1 190	1 070	970	890	822	763
400 000	636 000	949 000	6 330	4 750	3 800	3 165	2 713	2 373	1 900	1 582	1 357	1 186	1 055	949	863	791	736	678
350 000	556 500	827 000	5 510	4 130	3 307	2 756	2 362	2 067	1 653	1 378	1 181	1 033	919	827	752	689	636	590
300 000	477 000	708 000	4 720	3 540	2 832	2 359	2 023	1 749	1 416	1 179	1 011	884	786	708	644	589	545	503
250 000	397 500	592 000	3 920	2 960	2 370	1 975	1 693	1 481	1 185	987	846	740	658	593	539	493	456	423
200 000	318 000	470 000	3 140	2 090	1 570	1 255	1 044	896	723	627	523	448	391	349	313	285	261	241
150 000	238 500	350 000	2 500	1 670	1 250	1 000	833	713	625	500	416	357	312	277	250	227	208	192
100 000	159 000	250 000	1 570	1 045	783	627	522	448	392	313	261	224	196	174	157	142	130	120
50 000	79 500	124 000	826	620	496	413	354	310	248	207	177	155	138	124	113	103	95	88
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

4400 VOLTS DELIVERED																	
	1 MILE	1 1/2 MILES	2 MILES	2 1/2 MILES	3 MILES	3 1/2 MILES	4 MILES	5 MILES	6 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES
650 000																	

TABLE XV—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED, BASED UPON THE FOLLOWING I^2R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	FOR LOAD POWER-FACTOR OF 100%— $\frac{AT 25^{\circ}C}{8.68\% \text{ LOSS}}$ — $\frac{AT 65^{\circ}C}{10.0\% \text{ LOSS}}$ FOR LOAD POWER-FACTOR OF 80%—10.8% LOSS—12.5 LOSS																
			6000 VOLTS DELIVERED																
			1 MILE	1½ MILES	2 MILES	2½ MILES	3 MILES	3½ MILES	4 MILES	5 MILES	6 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES
650000	1033000	34600	23100	17300	13800	11500	9900	8650	7630	6930	5780	4950	4320	3850	3460	3150	2890	2670	2470
600000	954000	31800	21200	15900	12700	10600	9100	7950	7000	6300	5300	4550	3920	3450	3150	2890	2670	2470	2270
550000	874500	29500	19700	14700	11800	9950	8430	7370	6500	5800	4920	4220	3690	3280	2950	2680	2460	2270	2110
500000	795000	26600	17700	13300	10600	8860	7400	6470	5720	5020	4250	3600	3130	2760	2460	2220	2000	1840	1700
450000	715500	23900	16000	11900	9550	8000	6880	5900	5200	4500	3770	3140	2700	2360	2070	1850	1690	1540	1400
400000	636000	21300	14200	10700	8540	7120	6100	5340	4700	4000	3300	2700	2300	1980	1730	1570	1420	1280	1150
350000	556500	18600	12400	9300	7440	6200	5310	4650	4000	3300	2660	2200	1870	1600	1400	1250	1100	970	850
300000	477000	15900	10600	7960	6370	5310	4550	3980	3390	2750	2250	1870	1570	1350	1170	1010	880	760	650
250000	397500	13300	8890	6670	5330	4440	3810	3330	2870	2370	1950	1600	1350	1160	990	850	730	620	520
200000	318000	10700	7200	5420	4300	3580	3080	2660	2240	1790	1470	1220	1010	860	730	610	500	410	330
150000	238500	8100	5800	4380	3520	2820	2350	2020	1760	1410	1180	970	800	680	580	490	410	340	280
100000	159000	5500	4400	3300	2620	2100	1780	1480	1200	970	790	660	550	460	390	320	260	210	170
75000	119500	3900	3100	2300	1800	1400	1100	890	710	580	480	390	320	260	210	170	140	110	80
50000	79500	2300	1800	1300	1000	750	590	480	390	320	260	210	170	140	110	80	60	40	30
25000	39750	1150	900	650	500	380	300	240	190	150	120	90	70	55	40	30	20	10	5

			6600 VOLTS DELIVERED																
			1 MILE	1½ MILES	2 MILES	2½ MILES	3 MILES	3½ MILES	4 MILES	5 MILES	6 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES
650000	1033000	41900	28000	20900	16800	14000	12000	10500	9400	8400	7000	6000	5250	4670	4190	3830	3500	3230	3000
600000	954000	38500	25700	19300	15400	12800	11000	9630	8500	7700	6400	5500	4810	4290	3870	3500	3210	2960	2750
550000	874500	35700	23800	17800	14300	11900	10200	8930	7900	7100	5950	5100	4460	3960	3570	3250	2980	2750	2550
500000	795000	32300	21500	16100	12900	10800	9250	8075	7160	6460	5400	4620	4030	3600	3230	2940	2690	2480	2310
450000	715500	29100	19400	14500	11600	9700	8200	7270	6520	5820	4850	4150	3640	3230	2910	2640	2420	2240	2080
400000	636000	25900	17300	12900	10300	8650	7400	6470	5780	5180	4320	3700	3240	2880	2590	2360	2160	1990	1830
350000	556500	22600	15000	11300	9040	7520	6450	5650	4920	4320	3500	3000	2620	2310	2050	1880	1740	1580	1460
300000	477000	19300	12900	9650	7720	6490	5620	4820	4160	3560	2900	2400	2050	1790	1560	1400	1260	1120	1000
250000	397500	16200	10800	8100	6480	5400	4620	4050	3440	2900	2310	2020	1800	1620	1470	1350	1240	1140	1040
200000	318000	13600	9050	6800	5430	4530	3890	3400	2920	2460	1940	1700	1510	1360	1230	1130	1040	960	890
150000	238500	10800	7200	5400	4320	3600	3080	2700	2160	1800	1540	1350	1200	1080	980	900	830	770	710
100000	159000	8350	5700	4270	3420	2850	2440	2140	1710	1420	1200	1070	950	855	778	713	657	610	570
75000	119500	6800	4520	3400	2720	2260	1940	1700	1360	1130	970	850	755	680	617	565	522	486	456
50000	79500	4200	3100	2290	1710	1350	1070	900	720	600	510	440	380	330	290	260	230	200	180
25000	39750	2100	1500	1100	860	712	585	495	425	355	295	252	221	196	177	161	146	136	126

			10 000 VOLTS DELIVERED																
			6 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	15 MILES	18 MILES	18 MILES	20 MILES	22 MILES	24 MILES	26 MILES	28 MILES
650000	1033000	16000	13800	12100	10700	9620	8750	8020	7400	6870	6420	6000	5350	4820	4380	4020	3680	3410	3180
600000	954000	14800	12700	11100	9820	8850	8050	7350	6800	6320	5900	5500	4810	4320	3900	3540	3240	2960	2750
550000	874500	13700	11700	10200	9100	8200	7450	6820	6300	5850	5450	5100	4360	3900	3500	3160	2860	2580	2390
500000	795000	12300	10600	9250	8250	7400	6720	6150	5700	5280	4920	4620	3800	3360	3000	2700	2400	2120	1940
450000	715500	11100	9550	8450	7420	6680	6080	5560	5130	4770	4450	4170	3300	2880	2520	2220	1940	1680	1500
400000	636000	9890	8480	7420	6594	5930	5390	4940	4560	4240	3950	3710	2800	2380	2020	1720	1460	1200	1040
350000	556500	8610	7380	6460	5742	5170	4700	4310	3970	3690	3400	3230	2270	1850	1500	1250	1000	840	710
300000	477000	7370	6320	5530	4910	4420	4020	3690	3400	3160	2950	2760	1800	1380	1040	810	650	510	410
250000	397500	6170	5290	4630	4110	3700	3370	3090	2850	2640	2470	2310	1350	1000	750	590	450	340	260
200000	318000	5000	4300	3700	3200	2840	2560	2300	2070	1910	1780	1660	1000	750	590	450	340	260	190
150000	238500	3900	3300	2800	2450	2140	1960	1780	1630	1510	1400	1310	800	600	460	340	260	190	140
100000	159000	2800	2400	2000	1750	1560	1420	1300	1200	1110	1040	976	600	450	340	260	190	140	100
75000	119500	2100	1800	1500	1300	1120	1030	949	882	823	771	721	450	330	250	180	130	90	60
50000	79500	1300	1100	900	780	690	620	560	510	470	430	390	250	180	130	90	60	40	30
25000	39750	650	550	450	390	340	300	270	240	210	190	170	100	70	50	30	20	10	5

			11 000 VOLTS DELIVERED																
			8 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	15 MILES	18 MILES	18 MILES	20 MILES	22 MILES	24 MILES	26 MILES	28 MILES
650000	1033000	19400	16600	14500	12900	11700	10600	9700	8920	8300	7750	7250	6450	5800	5270	4850	4450	4100	3820
600000	954000	17800	15300	13400	11900	10700	9750	8900	8230	7650	7100	6600	5750	5150	4620	4200	3800	3450	3180
550000	874500	16500	14200	12400	11100	9920	9000	8250	7620	7080	6500	6000	5100	4500	4000	3600	3200	2850	2580
500000	795000	14900	12800	11200	10000	8970	8170	7450	6800	6200	5600	5100	4150	3500	3000	2600	2200	1850	1600
450000	715500	13500	11500	10100	8950	8070	7350	6650	6000	5400	4800	4300	3300	2700	2200	1800	1400	1050	800
400000	636000	12000	10300	9000	8000	7200	6550	6000	5350	4750	4200	3700	2650	2100	1650	1250	850	500	350
350000	556500	10500	8950	7820	6960	6270	5680	5200	4550	4000	3500	3000	2000	1500	1100	700	350	200	100
300000	477000	9200	7850	6700	5900	5200	4620	4100	3450	2900	2400	1900	1000	700	500	300	150	100	50
250000	397500	8000	6800	5600	4800	4100	3500	3000	2350	1900	1400	1000	600	400	250	100	50	20	10
200000	318000	6800	5700	4500	3700	3000	2400	1900	1300	900	600	400	200	100	50	20	10	5	2
150000	238500	5600	4600	3400	2600	2000	1400	900	500	300	200	100	50	20	10	5	2	1	0
100000	159000	4400	3500	2300	1500	900	500	300	150	80	40	20	10	5	2	1	0	0	0
75000	119500	3200	2400	1600	900	500	250	150	70	30	15	5	2	1	0	0	0	0	0
50000	79500	2000	1400	900	500	250	100	50	20	10	5	2	1	0	0	0	0	0	0
25000	39750	1000	700	400	200</														

TABLE XVII—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED. BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																																			
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	AT 25° C												AT 85° C																							
			FOR LOAD POWER-FACTOR OF 100%—8.66% LOSS—												10.0% LOSS																							
			FOR LOAD POWER-FACTOR OF 80%—10.8% LOSS—12.5% LOSS																																			
20 000 VOLTS DELIVERED																																						
			7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	15 MILES	16 MILES	18 MILES	18 MILES	20 MILES	22 MILES	24 MILES	28 MILES	28 MILES	30 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	15 MILES	16 MILES	18 MILES	18 MILES	20 MILES	22 MILES	24 MILES	28 MILES	28 MILES	30 MILES
650 000	1033 000	55 000	47 200	42 700	38 500	34 900	32 000	29 500	27 200	25 500	23 600	21 300	19 200	17 400	16 000	14 800	13 400	12 800	12 800	47 200	42 700	38 500	34 900	32 000	29 500	27 200	25 500	23 600	21 300	19 200	17 400	16 000	14 800	13 400	12 800	12 800		
600 000	954 000	50 000	44 200	40 000	36 200	33 000	30 200	27 800	25 600	23 800	22 100	20 500	19 000	17 600	16 300	15 000	13 800	12 600	12 600	44 200	40 000	36 200	33 000	30 200	27 800	25 600	23 800	22 100	20 500	19 000	17 600	16 300	15 000	13 800	12 600	12 600		
550 000	874 500	46 800	41 000	37 500	34 500	31 800	29 200	27 000	25 000	23 400	21 800	20 400	19 000	17 800	16 600	15 400	14 300	13 200	13 200	41 000	37 500	34 500	31 800	29 200	27 000	25 000	23 400	21 800	20 400	19 000	17 800	16 600	15 400	14 300	13 200	13 200		
500 000	795 000	42 500	37 000	33 500	30 500	27 800	25 400	23 400	21 600	20 000	18 500	17 200	16 000	14 900	13 900	13 000	12 100	11 200	11 200	37 000	33 500	30 500	27 800	25 400	23 400	21 600	20 000	18 500	17 200	16 000	14 900	13 900	13 000	12 100	11 200	11 200		
450 000	715 000	38 200	33 200	29 600	26 600	24 300	22 200	20 500	19 000	17 600	16 300	15 100	14 000	13 000	12 100	11 200	10 400	9 600	9 600	33 200	29 600	26 600	24 300	22 200	20 500	19 000	17 600	16 300	15 100	14 000	13 000	12 100	11 200	10 400	9 600	9 600		
400 000	636 000	33 900	29 400	26 400	23 700	21 600	19 800	18 200	16 900	15 800	14 800	13 800	13 000	12 100	11 200	10 400	9 600	8 900	8 900	29 400	26 400	23 700	21 600	19 800	18 200	16 900	15 800	14 800	13 800	13 000	12 100	11 200	10 400	9 600	8 900	8 900		
350 000	556 500	29 500	25 800	23 000	20 700	18 800	17 200	15 900	14 700	13 600	12 600	11 800	11 000	10 300	9 400	8 600	7 900	7 300	7 300	25 800	23 000	20 700	18 800	17 200	15 900	14 700	13 600	12 600	11 800	11 000	10 300	9 400	8 600	7 900	7 300	7 300		
300 000	477 000	25 300	22 100	19 700	17 700	16 000	14 700	13 600	12 600	11 800	11 000	10 300	9 830	9 050	8 370	7 770	7 240	6 750	6 750	22 100	19 700	17 700	16 000	14 700	13 600	12 600	11 800	11 000	10 300	9 830	9 050	8 370	7 770	7 240	6 750	6 750		
250 000	397 500	21 100	18 500	16 400	14 800	13 400	12 300	11 400	10 600	9 880	9 260	8 730	8 230	7 760	7 320	6 910	6 520	6 160	6 160	18 500	16 400	14 800	13 400	12 300	11 400	10 600	9 880	9 260	8 730	8 230	7 760	7 320	6 910	6 520	6 160	6 160		
0000	211 600	17 800	15 300	13 500	12 000	10 900	10 000	9 300	8 700	8 150	7 700	7 280	6 880	6 500	6 140	5 800	5 480	5 180	5 180	15 300	13 500	12 000	10 900	10 000	9 300	8 700	8 150	7 700	7 280	6 880	6 500	6 140	5 800	5 480	5 180	5 180		
0000	167 772	2 66 800	14 200	12 400	11 000	9 950	9 040	8 280	7 650	7 110	6 630	6 220	5 830	5 460	5 110	4 780	4 460	4 160	4 160	12 400	11 000	9 950	9 040	8 280	7 650	7 110	6 630	6 220	5 830	5 460	5 110	4 780	4 460	4 160	4 160			
0	105 560	1 67 800	8 930	7 810	6 940	6 250	5 680	5 210	4 800	4 460	4 170	3 910	3 670	3 450	3 240	3 040	2 840	2 660	2 660	7 810	6 940	6 250	5 680	5 210	4 800	4 460	4 170	3 910	3 670	3 450	3 240	3 040	2 840	2 660	2 660			
1	83 694	1 33 220	7 050	6 170	5 490	4 920	4 450	4 040	3 700	3 410	3 150	2 910	2 700	2 500	2 300	2 100	1 920	1 760	1 760	6 170	5 490	4 920	4 450	4 040	3 700	3 410	3 150	2 910	2 700	2 500	2 300	2 100	1 920	1 760	1 760			
2	66 358	1 05 530	5 600	4 900	4 360	3 920	3 560	3 270	3 000	2 750	2 510	2 290	2 100	1 940	1 780	1 630	1 480	1 350	1 350	4 900	4 360	3 920	3 560	3 270	3 000	2 750	2 510	2 290	2 100	1 940	1 780	1 630	1 480	1 350	1 350			
3	52 624	83 640	4 430	3 870	3 440	3 100	2 820	2 580	2 380	2 210	2 070	1 940	1 820	1 700	1 580	1 470	1 370	1 280	1 280	3 870	3 440	3 100	2 820	2 580	2 380	2 210	2 070	1 940	1 820	1 700	1 580	1 470	1 370	1 280	1 280			
4	41 738	66 370	3 500	3 100	2 750	2 480	2 250	2 060	1 900	1 770	1 650	1 540	1 440	1 340	1 250	1 170	1 090	1 020	1 020	3 100	2 750	2 480	2 250	2 060	1 900	1 770	1 650	1 540	1 440	1 340	1 250	1 170	1 090	1 020	1 020			
5	33 088	52 630	2 810	2 460	2 180	1 960	1 790	1 640	1 510	1 400	1 310	1 230	1 150	1 080	1 010	940	880	820	820	2 460	2 180	1 960	1 790	1 640	1 510	1 400	1 310	1 230	1 150	1 080	1 010	940	880	820	820			
22 000 VOLTS DELIVERED																																						
			7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	15 MILES	16 MILES	18 MILES	18 MILES	20 MILES	22 MILES	24 MILES	28 MILES	28 MILES	30 MILES	7 MILES	8 MILES	9 MILES	10 MILES	11 MILES	12 MILES	13 MILES	14 MILES	15 MILES	16 MILES	18 MILES	18 MILES	20 MILES	22 MILES	24 MILES	28 MILES	28 MILES	30 MILES
650 000	1033 000	66 500	58 200	51 800	46 600	42 200	38 800	35 800	33 300	31 000	29 100	27 500	26 000	24 600	23 300	22 100	21 000	20 000	20 000	58 200	51 800	46 600	42 200	38 800	35 800	33 300	31 000	29 100	27 500	26 000	24 600	23 300	22 100	21 000	20 000	20 000		
600 000	954 000	61 200	53 500	47 600	42 800	38 900	35 600	32 900	30 600	28 500	26 500	24 700	23 000	21 400	20 000	19 000	18 000	17 000	17 000	53 500	47 600	42 800	38 900	35 600	32 900	30 600	28 500	26 500	24 700	23 000	21 400	20 000	19 000	18 000	17 000	17 000		
550 000	874 500	56 800	49 600	44 200	39 700	36 100	33 100	30 500	28 400	26 500	24 800	23 200	21 600	20 100	19 000	18 000	17 000	16 000	16 000	49 600	44 200	39 700	36 100	33 100	30 500	28 400	26 500	24 800	23 200	21 600	20 100	19 000	18 000	17 000	16 000	16 000		
500 000	795 000	51 300	44 800	40 000	35 900	32 600	29 900	27 600	25 700	23 900	22 400	20 900	19 500	18 200	17 000	16 000	15 000	14 000	14 000	44 800	40 000	35 900	32 600	29 900	27 600	25 700	23 900	22 400	20 900	19 500	18 200	17 000	16 000	15 000	14 000	14 000		
450 000	715 000	46 000	40 200	35 800	32 200	29 300	26 900	24 800	23 000	21 500	20 100	18 700	17 400	16 100	14 900	13 800	12 800	12 800	12 800	40 200	35 800	32 200	29 300	26 900	24 800	23 000	21 500	20 100	18 700	17 400	16 100	14 900	13 800	12 800	12 800			
400 000	636 000	41 200	36 000	32 000	28 800	26 200	24 000	22 100	20 400	19 000	17 600	16 300	15 100	14 000	13 000	12 100	11 200	10 400	10 400	36 000	32 000	28 800	26 200	24 000	22 100	20 400	19 000	17 600	16 300	15 100	14 000	13 000	12 100	11 200	10 400	10 400		
350 000	556 500	35 800	31 400	27 900	25 100	22 800	20 900	19 300	17 900	16 700	15 700	14 700	13 800	13 000	12 100	11 200	10 400	9 600	9 600	31 400	27 900	25 100	22 800	20 900	19 300	17 900	16 700	15 700	14 700	13 800	13 000	12 100	11 200	10 400	9 600	9 600		
300 000	477 000	30 200	26 700	23 800	21 200	19 000	17 100	15 500	14 200	13 000	12 000	11 100	10 300	9 500	8 700	8 000	7 300	6 600	6 600	26 700	23 800	21 200	19 000	17 100	15 500	14 200	13 000	12 000	11 100	10 300	9 500	8 700	8 000	7 300	6 600	6 600		
250 000	397 500	25 600	22 400	19 900	17 900	16 300	14 900	13 600	12 400	11 300	10 400	9 600	8 800	8 100	7 400	6 800	6 200	5 600	5 600	22 400	19 900	17 900	16 300	14 900	13 600	12 400	11 300	10 400	9 600	8 800	8 100	7 400	6 800	6 200	5 600	5 600		
0000	211 600	21 500	18 900	16 800	15 100	13 700	12 600	11 600	10 800	10 000	9 300	8 600	8 000	7 400	6 800	6 300	5 800	5 400	5 400	18 900	16 800	15 100	13 700	12 600	11 600	10 800	10 000	9 300	8 600	8 000	7 400	6 800	6 300	5 800	5 400	5 400		
0000	167 772	2 66 800	15 100	13 400	12 000	10 900	10 000	9 300	8 700	8 150	7 700	7 280	6 880	6 500	6 140	5 800	5 480	5 180	5 180	13 400	12 000	10 900	10 000	9 300	8 700	8 150	7 700	7 280	6 880	6 500	6 140	5 800	5 480	5 180	5 180			
0	105 560																																					

TABLE XVIII—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED, BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	40 000 VOLTS DELIVERED																
			14 MILES	16 MILES	18 MILES	20 MILES	22 MILES	24 MILES	26 MILES	28 MILES	30 MILES	32 MILES	36 MILES	40 MILES	44 MILES	48 MILES	52 MILES	56 MILES	60 MILES
650000	1033000	170000	96000	85300	76800	70000	64200	59000	55000	51300	48000	42600	38400	35000	32100	29500	27500	25600	23600
600000	954000	160000	88700	78800	70800	64500	59000	55000	51700	47300	44300	39400	35400	32200	29500	27200	25300	23400	21400
550000	874500	150000	83800	74000	66000	59200	54000	50000	46800	42300	39500	34700	30700	27600	25000	22800	21000	19100	17200
500000	795000	140000	78400	68600	60600	53800	48600	44600	41400	36900	34100	29300	25300	22200	19600	17400	15600	13700	11800
450000	715500	130000	73400	63600	55600	48800	43600	39600	36400	31900	29100	24300	20300	17200	14600	12400	10600	8700	6800
400000	636000	120000	68400	58600	50600	43800	38600	34600	31400	26900	24100	19300	15300	12200	9600	7400	5600	3700	1800
350000	556500	110000	63400	53600	45600	38800	33600	29600	26400	21900	19100	14300	10300	7200	4600	2400	500	0	0
300000	477000	100000	58400	48600	40600	33800	28600	24600	21400	16900	14100	9300	5300	2200	0	0	0	0	0
250000	397500	90000	53400	43600	35600	28800	23600	19600	16400	11900	9100	4300	0	0	0	0	0	0	0
0000	217600	336420	35700	31200	27800	25000	22700	20800	19200	17800	16700	15600	13900	12500	11400	10400	9410	8330	7330
0	167772	266800	28400	24800	22100	19900	18100	16600	15300	14200	13200	12400	10700	9400	8230	7650	7110	6330	5330
1	133079	211950	22200	19600	17400	15700	14200	13000	12000	11000	10200	9400	8700	7800	7130	6530	6030	5400	4330
2	105560	167800	17800	15200	13000	11300	10400	9600	8900	8300	7700	7100	6500	5900	5300	4700	4100	3500	2600
3	83694	133220	15600	13000	10800	9100	8200	7400	6700	6100	5500	4900	4300	3700	3100	2500	1900	1300	500
4	66358	105530	11200	9800	8700	7800	7100	6500	5900	5300	4700	4100	3500	2900	2300	1700	1100	500	0
5	52624	83640	8860	7750	6880	6200	5600	5000	4400	3800	3200	2600	2000	1400	800	200	0	0	0
6	41738	66370	7070	6190	5500	4900	4300	3700	3100	2500	1900	1300	700	100	0	0	0	0	0
7	33088	52630	5610	4910	4370	3830	3290	2750	2210	1670	1130	590	50	0	0	0	0	0	0

CONDUCTORS			44 000 VOLTS DELIVERED																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	44 000 VOLTS DELIVERED																
			14 MILES	16 MILES	18 MILES	20 MILES	22 MILES	24 MILES	26 MILES	28 MILES	30 MILES	32 MILES	36 MILES	40 MILES	44 MILES	48 MILES	52 MILES	56 MILES	60 MILES
650000	1033000	173000	106000	93000	84500	77500	71500	66500	62000	57000	53000	47500	43000	39000	35000	32100	29500	27500	25600
600000	954000	163000	98000	85000	76500	70500	65500	61000	56500	52000	48000	42500	38000	34000	31000	28500	26500	24600	22700
550000	874500	153000	93000	80000	71500	65500	60500	56000	52000	47500	43500	38000	33500	30000	27500	25500	23600	21700	19800
500000	795000	143000	88000	75000	66500	60500	55500	51000	47000	42500	38500	33000	28500	25000	22500	20500	18600	16700	14800
450000	715500	133000	83000	70000	61500	55500	50500	46000	42000	37500	33500	28000	23500	20000	17500	15500	13600	11700	9800
400000	636000	123000	78000	65000	56500	50500	45500	41000	37000	32500	28500	23000	18500	15000	12500	10500	8600	6700	4800
350000	556500	113000	73000	60000	51500	45500	40500	36000	32000	27500	23500	18000	13500	9000	6500	4500	2600	700	0
300000	477000	103000	68000	55000	46500	40500	35500	31000	27000	22500	18500	13000	8500	5000	3000	1000	0	0	0
250000	397500	93000	63000	50000	41500	35500	30500	26000	22000	17500	13500	8000	3500	1500	0	0	0	0	0
0000	217600	336420	43200	37900	33700	30300	27500	25200	23300	21600	20200	18900	16800	15000	13700	12600	11600	10600	9600
0	167772	266800	34400	30100	26800	24100	21900	20100	18400	16800	15400	14000	12000	10900	9900	9000	8200	7400	6600
1	133079	211950	27100	23700	20400	17700	15900	14100	12400	10800	9400	8000	6800	5800	4900	4100	3300	2500	1700
2	105560	167800	21700	18900	16200	14400	12600	10800	9200	7600	6200	4900	3800	3000	2200	1400	600	0	0
3	83694	133220	15600	13000	10800	9100	7400	5800	4200	2600	1000	0	0	0	0	0	0	0	0
4	66358	105530	11200	9800	8700	7800	7100	6500	5900	5300	4700	4100	3500	2900	2300	1700	1100	500	0
5	52624	83640	8860	7750	6880	6200	5600	5000	4400	3800	3200	2600	2000	1400	800	200	0	0	0
6	41738	66370	7070	6190	5500	4900	4300	3700	3100	2500	1900	1300	700	100	0	0	0	0	0
7	33088	52630	5610	4910	4370	3830	3290	2750	2210	1670	1130	590	50	0	0	0	0	0	0

CONDUCTORS			50 000 VOLTS DELIVERED																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	50 000 VOLTS DELIVERED																
			20 MILES	24 MILES	28 MILES	32 MILES	36 MILES	40 MILES	44 MILES	48 MILES	52 MILES	56 MILES	60 MILES	64 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES
650000	1033000	180000	100000	86000	75000	66800	60000	54800	50200	46200	43000	40000	37500	34000	30000	27400	25100	23100	21200
600000	954000	170000	92500	79200	69200	61500	55500	50300	46200	42500	39000	36000	33500	30000	27500	25200	23100	21200	19300
550000	874500	160000	85500	73500	64200	57000	51000	46000	42000	38500	35000	32000	29000	26000	23500	21400	19500	17600	15700
500000	795000	150000	80500	68500	59200	52000	46000	41000	37000	33500	30000	27000	24000	21000	18500	16600	14700	12800	10900
450000	715500	140000	75500	63500	54200	47000	41000	36000	32000	28500	25000	22000	19000	16000	13500	11600	9700	7800	5900
400000	636000	130000	70500	58500	49200	42000	36000	31000	27000	23500	20000	17000	14000	11000	8500	6600	4700	2800	900
350000	556500	120000	65500	53500	44200	37000	31000	26000	22000	18500	15000	12000	9000	6500	4600	2700	800	0	0
300000	477000	110000	60500	48500	39200	32000	26000	21000	17000	13500	10000	7000	4500	2600	700	0	0	0	0
250000	397500	100000	55500	43500	34200	27000	21000	16000	12000	8500	5000	2000	0	0	0	0	0	0	0
0000	217600	336420	39100	33500	29300	24400	21700	19800	18200	16800	15000	13900	12200	10800	9700	8800	8100	7500	6900
0	167772	266800	31100	25500	22200	19400	17300	15500	14000	12600	11000	9900	8900	8000	7200	6400	5600	4800	4100
1	133079	211950	24500	20400	17500	15300	13400	11700	10200	8900	7700	6600	5600	4700	3900	3100	2300	1500	700
2	105560	167800	19500	16300	13900	12200	10600	9200	7900	6800	5800	4900	4000	3200	2400	1600	800	0	0
3	83694	133220	15400	12200	10200	8600	7200	6000	5000	4100	3300	2500	1700	900	100	0	0	0	0
4	66358	105530	11200	9800	8700	7800	7100	6500	5900	5300	4700	4100	3500	2900	2300	1700	1100	500	0
5	52624	83640	8860	7750	6880	6200	5600	5000	4400	3800	3200	2600	2000	1400	800	200	0	0	0
6	41738	66370	7070	619															

TABLE XIX—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED, BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
			AT 26°C — 8.85% LOSS — 10.0% LOSS FOR LOAD POWER-FACTOR OF 100% — 80% — 10.8% LOSS — 12.5% LOSS																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	66 000 VOLTS DELIVERED																
			20 MILES	24 MILES	28 MILES	32 MILES	36 MILES	40 MILES	44 MILES	48 MILES	52 MILES	56 MILES	60 MILES	64 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES
650 000	1033 000	210 000	174 000	150 000	131 000	116 000	105 000	95 300	87 500	80 800	75 000	69 800	65 500	58 000	52 500	47 600	43 700	40 400	37 100
600 000	954 000	193 000	161 000	138 000	120 000	107 000	96 200	87 500	80 800	75 000	69 800	64 300	60 000	53 500	48 100	43 700	40 100	37 100	
550 000	874 500	178 000	148 000	127 000	111 000	99 000	89 200	81 000	74 200	68 700	63 800	59 500	55 500	49 500	44 600	40 500	37 100	34 000	
500 000	795 000	161 000	134 000	115 000	101 000	89 500	80 500	73 300	67 000	62 000	57 400	53 700	50 500	44 700	40 200	36 600	33 600	31 000	
450 000	715 000	145 000	121 000	104 000	91 000	80 700	72 700	66 000	60 500	56 000	51 800	48 500	45 500	40 300	36 300	33 500	30 700	28 000	
400 000	636 000	130 000	108 000	92 700	81 000	72 000	65 000	59 000	54 000	50 000	46 300	43 300	40 500	36 000	32 500	29 500	27 000	25 000	
350 000	556 500	113 000	94 000	80 500	70 500	62 500	56 500	51 500	47 200	43 500	40 500	37 700	35 200	31 200	28 200	25 700	23 600	21 700	
300 000	477 000	96 000	80 000	68 700	60 000	53 500	48 000	43 700	40 000	37 000	34 300	32 000	30 000	27 000	24 000	21 800	20 000	18 500	
250 000	397 500	80 500	67 200	57 600	50 500	44 700	40 300	36 700	33 600	30 000	28 800	26 900	25 200	22 300	20 100	18 300	16 800	15 500	
0000	211 600	336 420	68 000	56 700	48 700	42 500	37 700	34 000	31 000	28 300	26 100	24 300	22 700	21 200	18 800	17 000	15 500	14 100	13 000
000	167 772	266 800	45 000	38 600	32 600	28 300	25 000	22 100	20 000	18 100	16 900	15 300	14 000	12 700	11 500	10 500	9 500	8 600	8 000
00	133 079	211 950	42 500	35 500	30 400	26 600	23 700	21 300	19 300	17 700	16 300	15 200	14 200	13 300	11 800	10 600	9 650	8 850	8 150
0	105 560	167 800	34 000	28 400	24 300	21 300	18 900	17 000	15 500	14 200	13 100	12 200	11 300	10 600	9 450	8 500	7 700	7 100	6 550
1	83 694	133 220	27 000	22 500	19 300	16 800	15 000	13 500	12 200	11 200	10 400	9 600	8 900	8 400	7 500	6 700	6 100	5 600	5 200
2	66 358	105 530	21 300	17 800	15 200	13 300	11 800	10 600	9 700	8 900	8 200	7 600	7 100	6 500	5 900	5 300	4 850	4 450	4 100
3	52 624	83 640	16 800	14 000	12 000	10 500	9 300	8 400	7 600	7 000	6 400	6 000	5 600	5 200	4 700	4 200	3 820	3 500	3 240
4	41 738	66 370	13 500	11 200	9 600	8 400	7 500	6 700	6 100	5 600	5 200	4 800	4 500	4 200	3 750	3 375	3 070	2 810	2 600
			70 000 VOLTS DELIVERED																
			38 MILES	40 MILES	44 MILES	48 MILES	52 MILES	56 MILES	60 MILES	64 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES
650 000	1033 000	130 000	118 000	107 000	98 000	90 600	84 000	78 600	73 500	68 500	63 500	59 000	53 500	49 000	45 300	42 000	39 300	36 700	32 700
600 000	954 000	120 000	108 000	98 500	90 500	83 500	77 500	72 300	67 000	62 000	57 000	52 000	47 000	43 200	40 000	37 300	34 700	32 100	27 900
550 000	874 500	112 000	100 000	91 200	83 500	77 500	71 700	67 000	62 000	57 000	52 000	47 000	43 000	40 000	37 000	34 000	31 000	28 000	23 700
500 000	795 000	101 000	90 500	82 500	75 500	69 800	64 800	60 500	56 700	52 500	48 500	45 200	41 200	37 700	34 900	32 400	30 200	28 300	25 200
450 000	715 000	90 800	80 800	72 700	66 000	60 600	55 900	51 900	48 400	45 400	42 000	39 500	36 300	33 000	30 300	28 000	25 900	24 200	21 200
400 000	636 000	80 800	72 700	66 000	60 600	55 900	51 900	48 400	45 400	42 000	39 500	35 100	31 600	28 800	26 400	24 300	22 600	21 100	17 600
350 000	556 500	70 300	63 300	57 500	52 700	48 700	45 200	42 200	39 500	35 100	31 600	28 800	26 400	24 300	22 600	21 100	19 700	17 700	14 600
300 000	477 000	60 200	54 200	49 200	45 000	41 700	38 700	36 100	33 900	30 100	27 100	24 600	22 600	20 800	19 300	18 000	16 900	15 000	12 000
250 000	397 500	50 400	45 300	41 200	37 800	34 900	32 400	30 200	28 300	25 200	22 700	20 600	18 900	17 400	16 200	15 100	14 200	12 600	10 000
0000	211 600	336 420	42 500	38 300	34 800	31 900	29 400	27 300	25 500	23 900	22 200	20 900	19 100	17 400	15 900	14 700	13 600	12 600	11 000
000	167 772	266 800	33 800	30 400	27 700	25 400	23 400	21 700	20 300	19 000	17 600	16 300	15 200	13 800	12 700	11 700	10 900	9 520	8 460
00	133 079	211 950	26 700	24 000	21 800	20 000	18 400	17 100	16 000	15 000	13 300	12 200	11 300	10 900	10 000	9 240	8 580	7 500	6 670
0	105 560	167 800	21 200	19 100	17 400	15 900	14 700	13 600	12 700	12 000	11 000	10 600	9 570	8 700	7 970	7 360	6 830	6 380	5 310
1	83 694	133 220	16 800	15 100	13 700	12 600	11 600	10 800	10 100	9 450	8 400	7 560	6 970	6 300	5 820	5 400	5 040	4 720	4 200
2	66 358	105 530	13 300	12 000	10 900	10 000	9 240	8 580	8 000	7 600	6 670	6 000	5 460	5 000	4 620	4 290	4 000	3 750	3 330
3	52 624	83 640	10 500	9 500	8 630	7 920	7 300	6 780	6 330	5 900	5 270	4 750	4 320	3 960	3 670	3 390	3 160	2 970	2 640
			80 000 VOLTS DELIVERED																
			38 MILES	40 MILES	44 MILES	48 MILES	52 MILES	56 MILES	60 MILES	64 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES
650 000	1033 000	171 000	154 000	140 000	128 000	118 000	110 000	102 000	96 000	85 500	77 000	70 000	64 000	59 000	55 000	51 000	48 000	42 700	37 000
600 000	954 000	160 000	142 000	128 000	118 000	109 000	101 000	94 500	88 500	78 500	71 000	64 000	59 000	55 000	51 000	48 000	42 700	37 000	32 000
550 000	874 500	146 000	131 000	119 000	110 000	101 000	93 600	87 600	82 000	72 000	65 000	59 500	55 000	51 000	48 000	43 800	41 000	36 400	31 000
500 000	795 000	132 000	119 000	108 000	99 000	91 500	85 000	79 000	74 200	64 000	57 000	51 500	47 500	44 500	41 500	38 500	37 000	33 000	28 000
450 000	715 000	118 000	107 000	97 000	88 800	82 000	76 200	71 000	66 700	56 000	50 000	45 000	42 000	39 000	36 000	33 000	30 000	27 000	24 000
400 000	636 000	105 000	94 900	86 300	79 100	73 000	67 800	63 300	59 300	52 700	47 400	43 000	39 500	36 500	33 500	31 000	28 000	25 000	22 000
350 000	556 500	91 800	82 600	75 100	68 900	63 600	59 000	55 100	51 600	45 900	41 300	37 500	34 400	31 800	29 500	27 000	24 800	22 900	20 000
300 000	477 000	80 600	70 900	64 300	59 000	54 400	50 500	47 200	44 200	39 300	35 400	32 100	29 500	27 200	25 300	23 600	22 100	19 600	17 000
250 000	397 500	68 900	59 200	53 800	49 400	45 600	42 300	39 300	37 000	32 900	29 600	26 900	24 700	22 800	21 100	19 700	18 500	16 400	14 000
0000	211 600	336 420	55 500	50 300	45 400	41 700	38 400	35 700	33 300	31 200	27 900	25 200	22 700	20 800	19 000	17 800	16 700	15 000	13 000
000	167 772	266 800	44 200	39 800	36 200	33 100	30 600	28 400	26 500	24 800	22 100	19 900	18 100	16 600	15 300	14 200	13 200	12 400	11 000
00	133 079	211 950	34 800	31 300	28 500	26 100	24 100	22 400	20 900	19 600	17 400	15 600	14 200	13 100	12 100	11 200	10 400	9 800	8 700
0	105 560	167 800	27 800	25 000	22 700	20 800	19 200	17 800	16 700	15 600	13 900	12 500	11 300	10 400	9 610	8 930	8 330	7 810	6 940
1	83 694	133 220	21 900																

TABLE XX—QUICK ESTIMATING TABLE

CONDUCTORS			KILOVOLT- AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED. BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
B & S NO.	COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	100 000 VOLTS DELIVERED																
			52 MILES	68 MILES	80 MILES	84 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES	180 MILES	176 MILES	192 MILES	208 MILES
650 000	1 033 000	1 850 000	172 000	140 000	150 000	133 000	120 000	109 000	100 000	92 500	86 000	80 000	75 000	66 500	60 000	54 500	50 000	46 200	42 700
600 000	954 000	1 700 000	158 000	140 000	138 000	123 000	110 000	101 000	93 500	87 000	81 000	76 000	69 500	63 000	57 500	53 000	49 200	45 700	42 200
550 000	874 500	1 580 000	147 000	137 000	128 000	114 000	103 000	93 500	87 000	81 000	75 000	70 000	64 500	59 000	54 500	50 000	46 200	42 700	39 500
500 000	795 000	1 430 000	132 000	124 000	116 000	103 000	93 000	84 500	77 500	71 500	66 000	61 000	56 000	51 500	47 500	43 500	40 000	37 000	34 000
450 000	715 500	1 280 000	117 000	110 000	104 000	93 000	83 500	76 000	70 500	65 000	60 000	55 500	51 000	47 000	43 000	39 500	36 500	33 500	31 000
400 000	636 000	1 140 000	106 000	100 000	92 700	82 400	74 200	67 400	62 000	57 000	52 000	47 500	43 500	39 500	36 000	32 500	29 500	27 000	24 500
350 000	556 500	994 000	92 300	86 100	80 700	71 800	64 600	58 700	53 800	49 700	46 100	43 000	39 500	36 500	33 000	29 500	26 500	24 000	21 500
300 000	477 000	851 000	79 000	73 700	69 100	61 400	55 300	50 300	46 100	42 500	39 500	36 500	33 000	30 000	27 000	24 000	21 500	19 000	17 000
250 000	397 500	712 000	66 000	61 700	57 800	51 400	46 300	42 100	38 600	35 600	33 000	30 000	27 000	24 000	21 000	18 000	15 500	13 000	11 000
200 000	318 000	569 000	53 000	48 700	45 800	40 300	36 000	32 500	29 500	27 000	24 000	21 000	18 000	15 500	13 000	11 000	9 000	7 500	6 000
150 000	238 500	426 000	40 000	36 000	33 000	28 500	25 000	22 000	19 500	17 000	15 000	13 000	11 000	9 000	7 500	6 000	5 000	4 000	3 000
100 000	159 000	283 000	27 000	24 000	22 000	19 000	16 000	14 000	12 000	10 000	8 000	7 000	6 000	5 000	4 000	3 000	2 000	1 500	1 000
0000	77 500	143 000	14 000	13 000	12 000	11 000	10 000	9 000	8 000	7 000	6 000	5 000	4 000	3 000	2 000	1 500	1 000	7 000	5 000
0000	167 772	246 800	47 800	44 400	41 400	38 800	34 500	31 100	28 200	25 900	23 200	20 700	18 400	16 300	14 000	12 000	10 000	8 000	6 000
0000	133 079	211 950	37 700	35 000	32 700	30 600	27 200	24 500	22 300	20 400	18 800	17 500	15 300	13 600	12 000	10 200	8 400	7 000	5 000
0000	105 560	167 800	30 000	27 900	26 000	24 400	21 700	19 500	17 700	16 300	15 000	13 900	12 200	10 800	9 800	8 900	8 000	7 000	6 000
0000	77 500	133 220	23 000	21 000	20 000	19 300	17 100	15 400	14 000	12 900	11 900	11 000	10 300	9 600	8 600	7 700	7 000	6 400	5 900
0000	400 000	636 000	114 000	106 000	100 000	92 700	82 400	74 200	67 400	62 000	57 000	52 000	47 500	43 500	39 500	36 000	32 500	29 500	27 000

			110 000 VOLTS DELIVERED																
52 MILES	68 MILES	80 MILES	84 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES	180 MILES	176 MILES	192 MILES	208 MILES			
650 000	1 033 000	224 000	208 000	194 000	182 000	162 000	146 000	133 000	121 000	112 000	104 000	97 000	91 000	81 000	73 000	66 500	60 500	56 000	
600 000	954 000	206 000	190 000	176 000	167 000	148 000	134 000	122 000	111 000	103 000	95 500	89 000	83 500	74 000	67 000	61 000	56 500	52 000	
550 000	874 500	191 000	177 000	163 000	153 000	135 000	124 000	113 000	105 000	97 500	90 500	84 500	79 000	70 000	63 000	57 500	53 000	49 000	
500 000	795 000	173 000	160 000	150 000	140 000	125 000	112 000	102 000	93 500	86 000	80 000	74 000	68 500	60 000	54 000	49 000	45 000	41 000	
450 000	715 500	155 000	144 000	134 000	124 000	110 000	97 000	88 000	81 000	74 500	69 000	63 500	58 000	50 000	45 000	41 000	37 000	34 000	
400 000	636 000	139 000	128 000	118 000	108 000	94 000	81 000	73 000	66 000	60 500	55 000	49 500	44 000	36 000	31 000	27 000	24 000	21 500	
350 000	556 500	120 000	111 000	104 000	97 000	84 800	78 100	71 000	65 000	60 000	55 000	50 000	45 000	37 000	32 000	28 000	25 000	22 000	
300 000	477 000	103 000	95 600	87 000	80 000	69 000	62 900	56 800	51 500	47 000	42 500	38 000	33 500	25 000	21 000	18 000	15 500	13 000	
250 000	397 500	86 000	80 000	74 700	70 000	62 200	56 000	50 600	45 600	41 000	37 000	33 000	29 000	21 000	17 000	14 000	12 000	10 000	
200 000	318 000	72 000	67 500	63 000	59 100	52 500	47 200	42 900	39 400	36 300	33 000	29 500	26 000	18 000	14 000	11 000	9 000	7 500	
150 000	238 500	57 800	53 700	50 100	47 000	41 800	37 600	34 200	31 300	28 200	25 000	21 500	18 500	14 000	11 000	9 000	7 000	5 000	
100 000	159 000	43 000	40 200	37 500	34 900	32 900	29 600	26 900	24 700	22 800	21 000	19 000	16 400	12 000	9 000	7 000	5 000	3 000	
0000	77 500	34 300	33 700	31 500	29 500	26 200	23 600	21 400	19 700	18 100	16 900	15 700	14 700	13 100	11 800	10 700	9 800	8 900	
0000	167 772	246 800	57 800	53 700	50 100	47 000	41 800	37 600	34 200	31 300	28 200	25 000	21 500	18 500	14 000	11 000	9 000	7 000	
0000	133 079	211 950	45 600	42 300	39 500	37 000	32 900	29 600	26 900	24 700	22 800	21 000	19 000	16 400	12 000	9 000	7 000	5 000	
0000	105 560	167 800	37 000	34 300	31 500	28 500	25 500	22 500	20 000	18 000	16 300	14 700	13 100	11 800	10 700	9 800	8 900	8 000	
0000	77 500	133 220	28 700	26 600	24 900	23 000	20 700	18 600	16 900	15 500	14 300	13 000	11 600	10 300	9 300	8 500	7 800	7 200	

			120 000 VOLTS DELIVERED																
84 MILES	72 MILES	80 MILES	88 MILES	96 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES	160 MILES	176 MILES	192 MILES	208 MILES	224 MILES	240 MILES	256 MILES			
650 000	1 033 000	216 000	192 000	173 000	157 000	144 000	133 000	123 000	115 000	108 000	96 000	86 500	78 500	72 000	66 500	61 500	57 000	54 000	
600 000	954 000	199 000	177 000	160 000	145 000	133 000	123 000	114 000	106 000	100 000	88 500	80 500	72 500	66 500	61 500	57 000	53 000	50 000	
550 000	874 500	184 000	164 000	148 000	134 000	123 000	113 000	105 000	98 500	92 500	82 000	74 000	67 000	61 500	56 500	52 000	49 000	46 000	
500 000	795 000	167 000	148 000	133 000	121 000	111 000	102 000	95 000	89 000	83 500	74 000	66 500	60 500	55 000	51 000	47 500	44 500	41 700	
450 000	715 500	150 000	133 000	120 000	109 000	100 000	92 500	86 000	80 000	75 000	66 500	60 000	55 500	50 000	46 200	43 000	40 000	37 500	
400 000	636 000	133 000	119 000	107 000	97 000	89 000	82 100	76 000	70 500	66 000	57 000	51 000	47 000	43 000	39 500	36 500	33 500	31 000	
350 000	556 500	116 000	103 000	93 000	84 500	77 500	71 500	66 400	62 000	58 100	51 600	46 500	42 200	38 700	35 800	33 200	31 000	29 000	
300 000	477 000	99 000	88 500	79 600	72 400	66 400	61 200	56 900	53 100	49 800	44 200	39 800	36 200	33 100	30 600	28 400	26 500	24 800	
250 000	397 500	83 000	74 100	66 700	60 600	55 500	51 200	47 600	44 400	41 600	37 000	33 300	30 300	27 700	25 600	23 800	22 200	20 800	
200 000	318 000	70 300	62 500	56 200	51 100	46 800	43 200	40 000	37 500	35 100	31 200	28 100	25 500	23 400	21 600	20 000	18 700	17 500	
150 000	238 500	56 000	49 700	44 800	40 700	37 300	34 400	32 000	29 500	27 000	23 000	20 000	18 500	17 200	16 000	14 900	14 000	13 000	
100 000	159 000	44 100	39 200	35 300	32 100	29 400	27 100	25 200	23 300	21 800	19 600	17 600	16 000						

HEATING LIMITATIONS

The k.v.a. values given in these tables do not take into account the heating and consequently carrying capacity of the conductors. This may be ignored in the case of the longer overhead high-voltage transmission circuits. For very short circuits (especially for the lower voltages and particularly for insulated or concealed conductors) the carrying capacity (safe heating limits) of the conductors must be carefully considered.

approximately the point at which the carrying capacity of that particular conductor is reached if insulated and installed in a fully loaded four duct line. If the conductor is to be installed in a duct line having more than four ducts its capacity will be still further reduced. The position of this line is based upon the use of lead covered, paper insulated, three conductor, copper cables for sizes up to 700 000 circ. mils and of lead covered, paper insulated, single conductor, copper cables for the larger sizes. In other words, the position of this heavy

TABLE XXI—QUICK ESTIMATING TABLE

CONDUCTORS		KILOVOLT-AMPERES, 3 PHASE, WHICH MAY BE DELIVERED AT THE FOLLOWING VOLTAGES OVER THE VARIOUS CONDUCTORS FOR THE DISTANCES STATED, BASED UPON THE FOLLOWING I ² R LOSS (EFFECT OF CHARGING CURRENT NEGLECTED)																
		FOR LOAD POWER-FACTOR OF 100%—8.68% LOSS— AT 25°C FOR LOAD POWER-FACTOR OF 80%—10.8% LOSS— AT 65°C 12.5 LOSS																
COPPER AREA IN CIRCULAR MILS	ALUMINUM AREA IN CIRCULAR MILS	154,000 VOLTS DELIVERED																
		98 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES	160 MILES	178 MILES	192 MILES	208 MILES	224 MILES	240 MILES	256 MILES	288 MILES	320 MILES	352 MILES	384 MILES
650 000	1 033 000	237 500	219 500	203 500	190 000	178 500	158 500	142 500	129 500	118 500	109 500	102 000	95 000	89 000	79 200	71 200	64 700	59 300
600 000	954 000	219 000	202 000	187 500	175 000	164 000	145 000	131 500	119 500	109 500	101 000	93 500	87 500	82 000	73 000	65 700	59 700	54 700
550 000	874 500	202 000	187 000	173 000	161 500	151 500	133 500	121 300	110 200	101 000	93 300	86 600	80 800	75 700	67 300	60 700	55 200	50 500
500 000	795 000	183 500	169 000	157 000	146 500	137 500	122 000	109 500	99 500	91 500	84 500	78 500	73 200	68 700	61 000	55 000	50 000	45 700
450 000	715 500	164 500	152 000	141 000	131 600	123 500	109 500	98 800	89 800	82 300	76 000	70 600	65 800	61 700	54 800	49 300	44 900	41 200
400 000	636 000	147 000	136 000	126 500	117 500	110 500	98 000	88 300	80 200	73 500	67 800	63 000	58 800	55 200	49 000	44 200	40 200	36 800
350 000	556 500	128 200	118 500	109 500	102 500	96 000	85 300	76 800	69 800	64 000	59 200	54 800	51 200	48 000	42 700	38 500	35 000	32 100
300 000	477 000	109 500	102 000	94 000	87 500	82 000	73 000	65 700	59 700	54 700	50 500	46 800	43 700	41 000	36 500	32 900	29 800	27 300
250 000	397 500	91 500	84 500	78 500	73 200	68 600	61 000	54 800	49 800	45 700	42 200	39 200	36 600	34 300	30 500	27 400	24 900	22 900
	336 420	77 200	71 200	66 200	61 600	57 700	51 500	46 200	42 000	38 500	35 600	33 000	30 800	28 900	25 700	23 200	21 000	19 200
	266 800	61 500	56 800	52 700	49 200	46 100	41 000	36 900	33 600	30 700	28 400	26 400	24 600	23 100	20 500	18 400	16 800	15 400
		187,000 VOLTS DELIVERED																
		98 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES	160 MILES	178 MILES	192 MILES	208 MILES	224 MILES	240 MILES	256 MILES	288 MILES	320 MILES	352 MILES	384 MILES
650 000	1 033 000	350 000	323 000	300 000	280 000	263 000	234 000	210 000	191 000	175 000	161 500	150 000	140 000	131 500	116 500	105 000	95 200	87 500
600 000	954 000	323 000	297 000	277 000	258 000	242 000	215 000	193 500	176 000	161 500	149 000	138 000	129 000	121 000	107 500	97 000	88 000	80 700
550 000	874 500	299 000	275 000	256 000	239 000	224 000	199 000	179 000	162 500	149 000	137 500	128 000	119 500	112 000	99 500	89 500	81 500	74 600
500 000	795 000	270 000	250 000	232 000	216 000	203 000	180 000	162 000	147 500	135 000	125 000	115 500	108 000	101 000	90 000	81 000	73 700	67 600
450 000	715 500	243 000	225 000	209 000	194 500	182 000	162 000	145 500	132 500	121 500	112 000	104 000	97 200	91 000	81 000	73 000	66 200	60 700
400 000	636 000	217 000	200 000	185 500	173 500	162 500	144 500	130 000	118 500	108 500	100 000	93 000	86 700	81 300	72 200	65 000	59 000	54 200
350 000	556 500	189 000	174 500	162 000	151 000	141 500	126 000	113 500	103 000	94 500	87 000	81 000	75 500	70 800	63 000	56 700	51 500	47 200
300 000	477 000	161 000	149 000	138 000	129 000	121 000	107 500	96 500	88 000	80 500	74 300	69 000	64 500	60 500	53 700	48 300	44 000	40 300
250 000	397 500	134 500	124 500	115 500	107 500	101 000	89 800	80 800	73 500	67 300	62 200	57 800	53 800	50 500	44 800	40 400	36 800	33 700
	336 420	114 000	105 000	97 500	91 000	85 200	75 800	68 200	62 000	57 000	52 600	48 800	45 500	42 700	38 000	34 200	31 000	28 500
		220,000 VOLTS DELIVERED																
		98 MILES	104 MILES	112 MILES	120 MILES	128 MILES	144 MILES	160 MILES	178 MILES	192 MILES	208 MILES	224 MILES	240 MILES	256 MILES	288 MILES	320 MILES	352 MILES	384 MILES
650 000	1 033 000	485 000	447 000	417 000	388 000	364 000	323 000	291 000	265 000	243 000	224 000	208 000	194 000	182 000	162 000	145 500	132 500	121 500
	954 000	447 000	413 000	383 000	358 000	336 000	298 000	263 000	240 000	224 000	207 000	192 000	179 000	168 000	149 000	134 500	122 000	112 000
	874 500	413 000	382 000	354 000	331 000	310 000	276 000	249 000	226 000	207 000	191 000	177 500	165 500	155 000	138 000	124 000	112 800	103 500
	795 000	374 000	345 000	321 000	299 000	281 000	249 000	224 000	204 000	187 000	173 000	160 000	149 500	140 500	124 500	112 500	102 000	93 500
	715 500	336 000	310 000	288 000	269 000	252 000	224 000	202 000	183 000	168 000	155 000	144 000	134 500	126 000	112 000	101 000	92 000	84 200
	636 000	300 000	277 000	257 000	240 000	225 000	200 000	180 000	163 500	150 000	138 500	128 500	120 000	112 500	100 000	90 000	82 000	75 000

The loss due to corona will not be excessive with any of the above conductors used at sea level for the voltages stated. For elevations above sea level, check the values with Table XXII, especially for the smaller conductors. On long circuits of high voltage, the effect of charging current (also corona and leakage losses) will be to increase or decrease the I²R loss, depending on the amount of load and its power-factor. See Fig. 13

For circuits of short length the carrying capacity of conductors will frequently determine these sizes and not the economic transmission loss. The carrying capacity of bare copper conductors suspended in air and of insulated copper conductors in duct lines are given in tables XXIII and XXIV, both of which are to appear in subsequent articles.

Running diagonally across each table from XII to XVII inclusive, is a heavy line. The point at which this heavy line intersects the horizontal line containing the k.v.a. values for a given size of conductor indicates

line is based upon the k.v.a. values for carrying capacity given in Table XXIV and is placed upon the tables as a warning that the heating limit capacity of the conductors must be considered. To illustrate, suppose 220 volts is to be delivered, over 1 000 000 circ. mil, insulated, single conductor, copper cables in a fully loaded four duct conduit. Table XII indicates that 189 k.v.a. can be transmitted over these conductors a distance of 2000 ft. without overheating the cable. If it is desired to transmit 378 k.v.a. a distance of 1000 feet, the fact that this value occurs to the left of the heavy line, indicates that

it is beyond the safe carrying capacity for this size conductor in a four duct line. Reference to Table XXIV will show that 297 k.v.a. is the maximum capacity of this cable under the conditions stated. In this case, either a larger conductor, or two or more smaller conductors must be used to prevent overheating. This will result in a less loss than those upon which the table k.v.a. values are based, and in this case the heating of the cable will probably determine the size to use.

EFFECT OF CHARGING CURRENT IN ABOVE I²R LOSS VALUES

As stated previously, the percent I²R losses in the quick estimating tables are based upon the load current and therefore do not take into account the effect of the charging current which is of a distributed nature and superimposed upon the load current. The effect of the charging current is to increase or decrease the current in the circuit by an amount depending upon the relative

there will be a lagging component in the load current. The charging or leading current will be practically in opposition to the lagging component of the load current and will therefore tend to cancel or neutralize the lagging component of the load current. The result will be a reduction of the current in the circuit and consequently in the I²R loss. But if the circuit is very long, particularly if the frequency is 60 cycles and the load power-factor is near unity (lagging component in load current small) the comparatively large leading component (charging current) will not only neutralize the lagging component of the load current, but will produce a leading power-factor at points along the circuit. If the charging current is sufficiently high it will increase the current, causing an increase in the I²R loss. Thus the effect of charging current in circuits delivering a lagging load is to decrease the I²R loss up

to a certain amount and then, if the charging current is sufficiently large, to increase I²R loss.

The curves in Fig. 13 show this effect for 25 and 60 cycle circuits delivering loads of unity power-factor; also loads of 80 percent lagging power-factor for circuits up to 500 miles long. It will be seen that for circuits 300 miles long the effect of charging current will be to reduce the I²R loss by approximately 25 percent if the load is 80

percent lagging. If the load power-factor is unity the I²R loss will be increased approximately 10 percent for these particular problems if the frequency is 25, and 30 percent if the frequency is 60 cycles.

The curves in Fig. 13 show that for circuits 500 miles long, in which the entire charging current is furnished from one end of the circuit, the effect of this charging current is to increase the I²R loss by 300 percent if the frequency is 60 cycle and the load power-factor 100 percent. In other words a large part of the current in the circuit for such a long 60 cycle circuit is charging current so that the effect of the load current on the I²R loss is comparatively small. Of course such a long circuit, unless fed from two or more generating stations located at widely separated points along the transmission line, would not be commercially practical.

values of the lagging and leading quadrature components of the current in the circuit.

For instance assume that the power-factor of the load is unity. In such case there is no quadrature component in the load current. If, however, the circuit is of considerable length, and particularly if the frequency is 60 cycles, there will be an appreciable amount of charging current (quadrature leading component) added vectorially to the load current. The sum of these two currents in quadrature with each other will result in an increase of current in the circuit with a consequent increase in the I²R loss. Thus the effect of charging current in a circuit delivering a load of 100 percent power-factor will always be to increase the I²R loss.

If, however, the power-factor of the load is lagging,

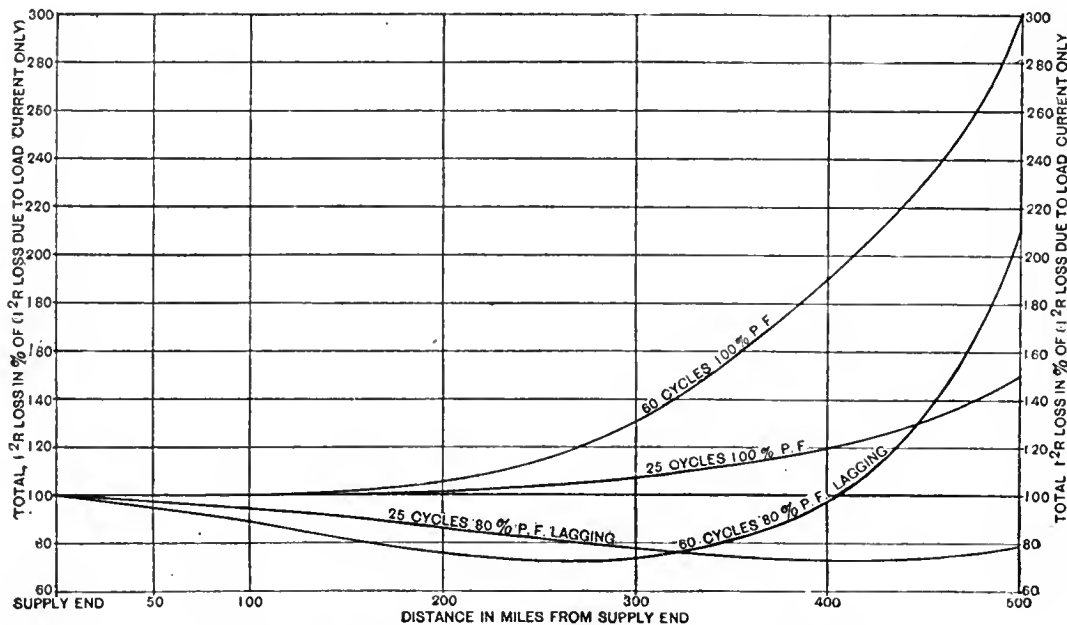


FIG. 13—EFFECT OF CHARGING CURRENT ON I²R TRANSMISSION LOSS

The curves represent (for certain circuits) an approximation of the resultant I²R loss, compared to what it would have been if there were no charging current present in the circuit. The effect of the charging current superimposed upon the receiver current is either to increase or to decrease the I²R loss of the circuit depending principally upon the relative amount of the leading and lagging components of the current in the circuit.

CHAPTER IV

CORONA EFFECT

In 1898 Dr. Chas. F. Scott presented a paper before the A.I.E.E. describing experimental tests (made during several years previous) relating to the energy loss between conductors due to corona effect. These investigations began at the Laboratory at Pittsburgh and were continued at Telluride, Colorado, in conjunction with the engineers of the Telluride Power Company. Preliminary observations were made by Mr. V. G. Converse and were continued in notable measurements by Mr. R. D. Mershon. These investigations were later followed by the work of Professor Ryan, by Mr. R. D. Mershon, Mr. F. W. Peek, Jr., Dr. J. B. Whitehead, Mr. G. Faccioli and others. The electrical profession is particularly indebted to Mr. Peek and Dr. Whitehead for the large amount of both practical and theoretical data which they have presented to the electrical profession on the subject. Mr. Peek developed and presented the empirical formulas which follow, for determining the disruptive critical voltage, the visual critical voltage and the power loss due to corona effect. The close accuracy of Mr. Peek's formulas has been confirmed by various investigators in different sections of the country. The following deductions concerning corona have to a large extent been previously presented by Mr. Peek.

CORONA, manifesting its presence usually by an electrostatic glow or luminous discharges, and audibly by a hissing sound, was clearly observed and studied in connection with electrostatic machines. It did not become a serious factor to be considered in connection with the design of commercial electrical apparatus until the increasing generator and transmission voltage emphasized its importance.

Although it is usual to think of corona effect only in connection with high-voltage transmission lines, it has received not a little thought of late by the designers of high-voltage generators and motors, notably large, high-voltage turbogenerators. By effectively insulating the portion of the conductor embedded in the iron of the armatures of alternating-current machines, particularly with mica, punctures to ground due to corona effect are not likely to occur. However, at the ends of the armature coils (where it is difficult to employ mica for insulating), where air is partially depended upon as an insulating medium between coils and ground, corona may appear. The presence of these corona stresses results in disintegrating and weakening some kinds of insulating materials, causing them to break down after a period of service. This deterioration of insulation may be due to local heating, mechanical vibration or chemical formations in the overstressed air, such as ozone, nitric acid, etc.

Higher voltages are being chosen as an economic means for reducing loss in transmission. These higher voltages may result in corona loss far in excess of the saving in transmission loss due to the adaptation of the higher voltages. It is, therefore, pertinent that particular consideration be given to the limitation of corona loss when the choice of conductors is made. This consideration will sometimes make it desirable to take advantage of the higher critical voltage limits of aluminum conductors (with steel reinforced centers) of an equivalent resistance, due to their greater diameter; or it may be desirable to obtain the necessary larger diameter by the use of copper conductors having some form of non-conducting centers or, for still larger diameters, of

aluminum conductors having such centers, in order to avoid skin effect. The use of copper conductors having hemp centers has in some instances given mechanical trouble.

The critical voltage at which corona becomes manifest, is not constant for a given line, but is somewhat dependent upon atmospheric conditions. Assuming a line employing conductors just within the critical voltage limitations for the conditions to be met, the corona loss in such a line would be almost negligible during fair weather, but during stormy weather, (particularly during snowstorms) this corona loss would be many times what it is during fair weather. On the other hand, since the storm will usually not appear over the whole length of lines at the same time and since storms occur only at intervals, it may often be economical to allow this loss to reach fairly high values during storms. Fog, sleet, rain and snowstorms lower the critical voltage and increase the losses. The effect of snow is greater than any other weather condition. Increase in temperature or decrease in barometric pressure lowers the voltage at which visual corona starts.

The critical voltage increases with both the diameter of conductors and their distance apart. This sometimes makes it desirable to use aluminum conductors as previously stated. It also increases with the horizontal or vertical arrangement of conductors, due to the fact that the two outside conductors considered as a pair are twice as far apart as are the other pairs. The same general rules apply to stranded conductors as to solid conductors, the actual diameter of the former being considered as the effective diameter of the conductor.

The losses due to corona effect increase very rapidly with increase in voltage after the critical voltage has been reached. A long transmission line having considerable capacitance may deliver a higher voltage than appears at the generator end of the line due to capacitance effect. The corona loss would in this case be greater per mile at the receiving end than at the sending end of the line.

The magnitude of the losses, as well as the critical voltage, is affected by atmospheric conditions;—hence they probably vary with the particular locality and the season of the year. Therefore, for a given locality, a voltage which is normally below the critical point, may at times be above the critical voltage, depending upon changes in the weather.

The material of the conductors does not seem to affect the losses. Sometimes the conductors of new transmission lines, when first placed in service will show visual corona, which may entirely disappear after a few hours or weeks of service. This may be due to scratches, particles of foreign substances, etc., on the conductors which are eliminated after the voltage stress has been kept on the conductors for a short time. Under such conditions the corona loss will also become less as the visual effect disappears.

The loss of power due to corona effect increases with frequency and increases as the square of the excess voltage above a certain critical voltage referred to as the "disruptive critical voltage" e_0 . This disruptive critical voltage is that voltage, at which a certain definite and constant potential gradient is reached at the conductor surface. This gradient g_0 is 30 kv maximum (21.1 kv effective) per centimeter, or 76.2 kv maximum (53.6 kv effective) per inch. These values are based upon an air density at sea level (25° C., 29.92 inches or 76 cm. barometer). This gradient is independent of the size of conductors and their distance apart, but is proportional to the air density, that is to the barometric pressure and the absolute temperatures. It may be considered as the dielectric strength of air. The presence of corona at a certain point of the system shows that a critical electric stress has been exceeded at that point. The corona loss is also proportional to the square root of the conductor radius r and inversely proportional to the square root of the conductor spacing.

The law by which corona losses increase with the voltage does not give a very steep curve, but a rather mild curve following the quadratic law at and above the critical limit. In other words there is no sharp elbow in the curve above which the losses increase very rapidly with the voltage and which could be adopted as the normal operating point of the circuit.

Table XXII, indicating the voltage limitations due to corona effect, has been worked up from Mr. F. W. Peek's formula as indicated at the bottom of the table. The values in this table are conservative and may in many cases be exceeded. They are the effective e_0 disruptive critical voltage between conductors for fair weather based upon δ values for 25 degrees C. (77 degrees F) and m_0 values of 0.87 for cable and 0.93 for wire. With these table values, corona loss should not be excessive during storms. If the values of Table XXII indicate that the conductors contemplated are close to the limit due to corona effect, a careful check should be made by the formula to determine definitely the corona loss for such conductors under storm operating conditions.

F. W. PEEK'S CORONA FORMULAE

Disruptive Critical Volts, Fair Weather (parallel wires)

$$e_0 = 2,302 m_0 g_0 \delta r \log_{10} \frac{s}{r} \dots\dots\dots (20)$$

effective kv to neutral,—

Visual Critical Volts—Fair Weather (parallel wires)

$$e_v = 2,302 m_v g_0 \delta r \left(1 + \frac{0.189}{\sqrt{r \delta}} \right) \log_{10} \frac{s}{r} \dots (21)$$

effective kv to neutral

Power Loss (fair weather)—

$$P = \frac{390}{\delta} (f + 25) \sqrt{\frac{r}{s}} (e - e_0)^2 10^{-5} \dots\dots\dots (22)$$

kw per mile of each conductor

Power Loss (Storm)—Storm power loss is higher and can generally be found with fair approximation by assuming $e_0 = 0.80$ times fair weather e_0 . It generally works out in practice that the e_0 voltage is the highest that should be used on transmission lines $\dots\dots\dots (22A)$

All of the above voltages are to neutral. To find voltages between lines multiply by 1.73 for three-phase, and by 2 for single phase.

Notation—

- e = Effective applied voltage in kv to neutral.
(This will vary at different points of the circuit and at different loads. At low loads and long lines of high voltage it may be higher at the receiving end than at the generator end due to inductive capacitance)
- e_0 = effective disruptive critical voltage in kv to neutral.
It is the voltage that gives a constant break down gradient for air of 76 kv maximum per inch, the "elastic limit" at which the air breaks down. Visual corona does not start at the disruptive critical voltage, but at a higher voltage e_v .
- e_v = effective visual critical kv to neutral (voltage at which visual corona starts)
- P = power loss in fair weather in kw per mile of single conductor,
- δ = $\frac{17.9b}{459 + t}$. This takes care of the effect of altitude and temperature, (air density). It is 1 at 25 degrees C. (77 degrees F.) and 29.92 inches (76 cm.), barometric pressure.
- g_0 = 53.6 kv per inch effective (disruptive gradient of air)
- b = barometric pressure in inches.
- t = maximum temperature in degrees F.
- f = frequency in cycles per second.
- m_0 = irregularity factor.
= 1 for polished wires.
= 0.98 to 0.93 for roughened or weathered wire.
= 0.87 to 0.83 for cables.
- m_v = m_0 for wires (1 to 0.93)
= 0.72 for local corona all along cables (7 strands)
= 0.82 for decided corona all along cables (7 strands)
- r = radius of conductor in inches.
- s = spacing in inches between conductor centers, based upon the assumption of a symmetrical triangular arrangement. For three-phase irregular flat or triangular spacing take $s = \frac{1}{\sqrt{3}} \overline{ABC}$. For three-phase regular flat spacing take $s = 1.26A$.

Theoretically, if the conductors were perfectly smooth, no loss would occur until the critical voltage, e_v is reached, when the loss should suddenly take a definite value, equal to that calculated by quadratic law, with e_v as the applied voltage and e_0 as the critical voltage in the equation. It should then follow the quadratic law for all higher voltages. On the weathered conductors used in practice, the quadratic law is followed over the whole range of voltage, starting at e_0 .

Example:—In order to show the variation in corona loss at different voltages and for different weather conditions, Table E has been calculated for No. 0 stranded copper conductors (105 560 circ. mils, 0.373 in. diameter) and for steel reinforced aluminum conductors (167 800 circ. mils, 0.501 in. diameter) having an equivalent resistance but greater diameter. F. W. Peek's formulas were used and the following assumptions were made:—

- f = 60 cycles.
- m_0 = 0.87
- m_v = 0.72
- g_0 = 53.6

r = 0.186 in. for copper = 0.250 in. for aluminum.
 s = 144 inches (delta arrangement of conductors).
 b = 28.9 corresponding to an altitude of 1000 feet.
 t = 77 degrees F. & therefore = 0.967.

$\frac{s}{r}$ = 774 for copper = 576 for aluminum
 $\log_{10} 774 = 2.89$ and $\log_{10} 576 = 2.76$

$\sqrt{\frac{r}{s}} = 0.036$ for copper and 0.0415 for aluminum.

DISRUPTIVE CRITICAL VOLTAGE—Fair Weather

$$e_0 = 2.302 m_0 g_0 \delta r \log_{10} \frac{s}{r} \quad (20)$$

effective kv to neutral

For the Copper Conductors

$$e_0 = 2.302 \times 0.87 \times 53.6 \times 0.967 \times 0.186 \times 2.89$$

= 55.8 kv to neutral (96 500 volts between conductors).

Table XXII gives, by interpolation, the limitation of e_0 for above conditions, as 96 500 volts between conductors. To find e_0 to neutral for any other altitude or temperatures insert the corresponding values of δ for the altitude and temperature in the formula.

TABLE D—WORKING TABLE— δ (DENSITY) VALUES

Altitude and Temperature Correction Factors

$\delta = \frac{17.9b}{459 + t}$ where b = barometric pressure in inches and t = temperature in degrees F.

Altitude in Feet	Barometer		δ Values for Different Temp.		
	In Inches	In Cm.	0° C. (32° F.)	25° C. (77° F.)	50° C. (122° F.)
Sea Level	30.0	76.2	1.09	1.00	0.925
500	29.45	74.8	1.07	0.985	0.910
1000	28.90	73.3	1.05	0.967	0.892
1500	28.30	71.8	1.03	0.947	0.873
2000	27.80	70.7	1.01	0.932	0.860
2500	27.25	69.2	0.955	0.912	0.841
3000	26.80	68.0	0.980	0.897	0.827
4000	25.75	65.3	0.940	0.860	0.793
5000	24.70	62.7	0.902	0.827	0.762
6000	23.90	60.7	0.875	0.800	0.738
7000	22.95	58.3	0.840	0.770	0.710
8000	22.05	56.0	0.805	0.738	0.682
9000	21.30	54.1	0.778	0.712	0.657
10 000	20.50	52.1	0.750	0.687	0.633
12 000	19.00	48.3	0.697	0.637	0.588
14 000	17.55	44.7	0.643	0.588	0.543
15 000	16.90	42.9	0.618	0.566	0.522

*This column contains the values for δ which were used in determining the values of e_0 in Table XXII. That is, the values for sea level in Table XXII multiplied by these δ values gives the e_0 values of the table for the higher altitudes.

For the Aluminum Conductors

$$e_0 = 2.302 \times 0.87 \times 53.6 \times 0.967 \times 0.25 \times 2.76$$

= 71.5 kv to neutral (123 500 volts between conductors).

Table XXII gives (by interpolation) the limitation for above conditions as 123 500 volts between conductors.

To find e_0 to neutral for any other altitude or temperature insert the corresponding value of δ for that altitude and temperature in the formula.

DISRUPTIVE CRITICAL VOLTAGE—Stormy Weather

e_0 during storm = approximately 80 percent e_0 during fair weather.

For the Copper Conductors

$$e_0 \text{ for storm} = 55.8 \times 0.80 = 44.6 \text{ kv to neutral or } 77000 \text{ volts between conductors.}$$

For the Aluminum Conductors

$$e_0 \text{ for storm} = 71.5 \times 0.80 = 57.2 \text{ kv to neutral or } 98800 \text{ volts between conductors.}$$

VISUAL CRITICAL VOLTAGE—Fair Weather

$$e_v = 2.302 m_v g_0 \delta r \left(1 + \frac{0.189}{\sqrt{r \delta}} \right) \log_{10} \frac{s}{r} \quad (21)$$

effective kv to neutral

For Copper Conductors

$$e_v = 2.302 \times 0.72 \times 53.6 \times 0.967 \times 0.186 \left(1 + \frac{0.189}{0.424} \right) 2.89$$

= 66.4 kv to neutral (115 000 volts between conductors).

To find e_v to neutral for any other altitude and temperature, insert the corresponding values of δ for that altitude and temperature in the formula above.

For the Aluminum Conductors

$$e_v = 2.302 \times 0.72 \times 53.6 \times 0.967 \times 0.25 \left(1 + \frac{0.189}{0.492} \right) 2.76$$

= 82 kv to neutral (141 500 volts between conductors).

To find e_v to neutral for any other altitude and temperature, insert the corresponding values of δ for that altitude and temperature in the formula above.

POWER LOSS

$$P = \frac{390}{\delta} (f + 25) \sqrt{\frac{r}{s}} (e - e_0)^2 10^{-5} \dots \dots \dots (22)$$

kw per mile of each conductor

The corona power loss corresponding to various conditions for the above circuit has been calculated by formulae (22) and (22A). They are given in Table E. However, in order to illustrate the application of the power loss formula the losses for the following conditions are determined below. Assuming that the No. 0 stranded copper conductors will be operated at 105 kv between conductors (60.7 kv to neutral).

For Fair Weather—Max. Temp. 50 degrees C. (122 degrees F.)— $E_0 = 51.3$ kv.

$$P = \frac{390}{\delta} (f + 25) \sqrt{\frac{r}{s}} (e - e_0)^2 10^{-5} \dots \dots \dots (22)$$

kw per mile of each conductor

$$P = \frac{390}{0.892} (60 + 25) \times 0.036 (60.7 - 51.3)^2 10^{-5}$$

= 1.2 kw per mile of each conductor or 3.6 kw per mile for three conductors.

For Stormy Weather—Max. Temp. 25 degrees C. (77 degrees F.)— $E_0 = 55.8 \times 0.8 = 44.6$ kv.

$$P = \frac{390}{0.967} (60 + 25) \times 0.036 (60.7 - 44.6)^2 10^{-5} \quad (22A)$$

= 3.2 kw per mile of each conductor or 9.6 kw per mile for three conductors.

By applying formula (20) to the above case it develops that the fair weather values of e_0 are for 25 degrees C. (77 degrees F.) 96 500 kv and for 50 degrees C. (122 degrees F.) 88 800 kv between conductors. Table XXII values for 25 degrees C. (77 degrees F.) confirm this.

Table E values for corona loss indicate that No. 0 copper conductors can, with 144 inch delta arrangement of conductors and 1000 ft. elevation be used at line voltages as high as 100 000 volts without excessive corona loss during stormy weather. At 100 000 volts and assuming a 25 degrees C. (77 degrees F) temperature during fair weather and storm conditions, the corona losses would be 0.1 kw per mile for fair weather and 6.5 kw per mile for stormy weather. If the transmission is single circuit 100 miles long and without branches, has an average altitude of 1000 feet and the storm condition existed throughout the length of the circuit, the power loss due to corona would be 6.5×100 or 650 kw. The capacity of such a circuit at 100 000 volts (see Table XX) would be roughly 15 000 kw at ten percent I^2R loss. The storm corona loss therefore would represent $\frac{650}{15000}$ or 4.3 percent. This, in addition to ten percent I^2R loss, would represent approximately 14 percent loss in transmission during the storm conditions.

In the above case it would probably be considered good engineering (so far as corona loss is concerned)

TABLE XXII—APPROXIMATE VOLTAGE LIMITATIONS RESULTING FROM CORONA

STRANDED COPPER CONDUCTORS

B & S NO AND CIRCULAR MILS	DIAMETER IN INCHES	ELEVATION IN FEET	LIMIT IN KILOVOLTS BETWEEN CONDUCTORS 3 PHASE FOR VARIOUS SPACINGS X													B & S NO. AND CIRCULAR MILS	DIAMETER IN INCHES	ELEVATION IN FEET	LIMIT IN KILOVOLTS BETWEEN CONDUCTORS 3 PHASE FOR VARIOUS SPACINGS X												
			3	4	6	7	8	9	11	13	15	18	25	3	4				6	7	8	9	11	13	15	18	25				
			FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.				FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.	FT.				
4	.232	SEA LEVEL	54	56	58	60	62	63	64	66	67	69	71	73	109	114	120	124	128	131	134	137	142	145	149	155	160				
		1000	52	54	56	58	60	61	62	64	65	67	69	71	73	107	112	118	122	126	130	133	137	141	145	150	155				
		2000	50	52	54	56	58	59	60	62	63	65	67	69	71	73	104	110	116	120	124	128	132	135	139	145	149				
		4000	44	46	48	50	51	53	54	55	57	58	59	61	63	96	102	107	111	115	118	122	125	129	133	138					
		8000	43	45	47	49	50	51	52	53	55	56	57	59	61	89	94	99	103	107	110	114	118	122	126	130	135				

SOLID COPPER CONDUCTORS

4	.204	SEA LEVEL	51	54	56	58	59	60	61	63	64	65	68	70	75	79	82	85	87	89	91	94	96	98	102	105	
		1000	49	52	54	56	57	58	59	60	62	63	65	68	70	73	77	80	83	85	87	89	91	93	96	99	
		2000	47	50	52	54	55	56	57	58	60	61	63	65	68	70	71	75	78	81	83	85	87	89	91	93	96
		4000	44	46	48	50	51	52	53	54	55	57	58	60	63	50	53	56	59	61	63	65	67	69	71	73	75
		8000	41	43	45	47	48	49	50	51	52	54	55	57	60	40	43	46	49	51	53	55	57	59	61	63	65

x For single phase or 2 phase multiply the 3 phase values by 1.16. The above are the disruptive critical voltage values for fair weather based upon a temperature of 25° C. (77° F.) and values for M₀ of 0.87 for stranded and 0.93 for solid conductors. Derived by Peek's formula: Kilovolts to neutral = 2.302 M₀ G₀^δ R Log₁₀ $\frac{S}{R}$, where G₀ = 53.6 Kilovolts per inch; S = Spacing in inches; R = Radius of conductor in inches; δ = $\frac{17.9 B}{459 + T}$; T = Temperature in degrees F.; B = Barometer pressure in inches.

to operate the No. 0 copper conductors at as high a line voltage as 100 000 volts. If, however, for other reasons, 120 000 is selected as the desirable operating voltage, then either a large diameter copper conductor or an aluminum conductor having a greater diameter but an equivalent conductivity to that of the No. 0 copper conductor should be selected.

TABLE E—COMPARISON OF CORONA LOSS

For No. 0 Stranded Copper Conductors 105 560 cir. mil (diameter 0.373 in.) and equivalent Aluminum Conductors 167 800 cir. mil (diameter 0.501 in.) Conductor Spacing (s) Delta = 144 in. Altitude 1000 feet—Barometer 28.9 inches. Calculated from formula (22)

Kilovolts		Corona Loss in Kw. per Mile for Three Conductors at 60 Cycles											
		Fair Weather—(Formula 22)						Stormy Weather—(Formula 22-A)					
		No. 0 Copper Radius 0.186 in.			Aluminum Radius 0.25 in.			No. 0 Copper Radius 0.186 in.			Aluminum Radius 0.25 in.		
Between Conductors	To Neutral	0° C 32° F δ=1.05 ε ₀ =60.5	25° C 77° F δ=0.967 ε ₀ =55.7	50° C 122° F δ=0.892 ε ₀ =51.3	0° C 32° F δ=1.05 ε ₀ =77.5	25° C 77° F δ=0.967 ε ₀ =71.5	50° C 122° F δ=0.892 ε ₀ =66.0	0° C 32° F δ=1.05 ε ₀ =48.4	25° C 77° F δ=0.967 ε ₀ =44.5	50° C 122° F δ=0.892 ε ₀ =41.0	0° C 32° F δ=1.05 ε ₀ =62.	25° C 77° F δ=0.967 ε ₀ =57.2	50° C 122° F δ=0.892 ε ₀ =52.7
		100	57.8	0.0	0.1	0.2	0	0	0	0.3	6.5	11.3	0
110	63.5	0.3	2.3	6.0	0	0	0	7.8	13.3	20.3	0	1.7	4.6
120	69.2	2.6	6.7	12.8	0	0	0.4	14.8	22.6	32.0	2.0	6.2	12.6
130	75.1	7.25	13.9	22.6	0.0	0.5	3.8	24.4	34.6	46.5	6.7	13.7	23.2
140	80.8	13.8	23.3	34.8	0.3	3.7	10.1	35.8	48.7	63.7	13.9	23.8	36.4
150	86.7	22.4	35.5	50.2	3.3	9.9	19.7	50.2	66.	84.	24.	37.2	53.3
160	92.4	35.0	49.8	67.7	8.7	18.7	32.2	66.	85.	106	36.	53.	73.
180	104.8	66.0	89.0	115.0	29.3	47.3	69.5	108.	135.	163.	72.	96.	125.

Note: At 25 cycles the losses would be $\frac{f_1 + 25}{f + 25} = \frac{25 + 25}{60 + 25} = \frac{50}{85}$ times the above table values. For conductors in a row (flat spacing) the corona loss would be reduced below the values for delta or triangular arrangement. For the higher voltages in the above table the conductor spacings would, in an actual installation, be greater than 144 in. (upon which basis the table values are given) thus giving actually less corona loss for the higher voltages than indicated by the table values.

The accompanying photograph illustrating corona on an experimental line is published with the kind permission of F. W. Peek, Jr.

Since the formulas pertaining to corona effect are to some extent worked up from test data they may be slightly changed from time to time. In case the problem at hand seems vitally near the critical point it will be well to consult the latest literature at that time as an additional check on the work.



CORONA AT 230 KV. 1.19 CM. DIAMETER, 0.47" CABLE. 310 CM. 10 FEET SPACING.

CHAPTER V

SPEED OF ELECTRIC PROPOGATION—RESONANCE PARALLELING TRANSMISSION CIRCUITS HEATING OF BARE CONDUCTORS

SPEED OF ELECTRIC PROPAGATION

ASTRONOMERS and investigators by various methods of determination have arrived at slightly different values for the speed of light. The Smithsonian Physical Tables give 186 347 miles per second as a close average estimate. In electrical engineering, the speed of light is usually stated as approximately 3×10^{10} centimeters per second. This is the equivalent of 186 451 miles per second. The speed of electrical propagation (assuming zero losses) is the same as that of light.

ELECTRIC WAVE LENGTH

Suppose a frequency of 60 cycles per second is impressed upon a circuit of infinite length. At the end of one sixtieth of a second the first impulse (neglecting retardation due to losses) will have traversed a distance of $186\,347 \div 60$ or 3106 miles. A section of such a circuit 3106 miles long would be designated as having a full wave length for a frequency of 60 cycles per second.

In Fig. 14, the dotted line or one cycle wave is shown as extending over a circuit 3106 miles long. In this case, when the first part of the wave arrives at a point 3106 miles distant, the end of the same wave is at the beginning of the circuit. For each half wave length the current is of equal value but flowing in opposite directions in the conductor. Such a circuit is designated as of full wave length. Since the velocity of the electric propagation is slightly less than that of light, being slightly retarded due to resistance and leakage losses, the actual wave length will be slightly less than 3106 miles. Thus for a 300 mile, 60 cycle, three-phase circuit consisting of No. 000 copper conductors having 10 ft. flat spacing, the wave length is calculated to be 2959 miles. The wave length of such a circuit is indicated by the heavy line on the accompanying sketch. In the case of this particular circuit the electric field has been retarded approximately five percent, due to the losses of the circuit, as indicated by the displacement of the dotted and full line curves.

QUARTER WAVE RESONANCE

If the end of a long trough filled with water is struck by a hammer, the impact will cause a wave in the water to start in front of the point of impact and travel to the far end of the tank. When this wave reaches the far end of the tank it will be reflected, traveling back toward the point of origin, but on account of resistance encountered it will be of diminishing height or amplitude. If, at the instant it gets back to the point of origin, the end of the tank is again struck by the hammer, the

resulting impulse will be that due to the second hammer blow plus that remaining from the first blow. The result will be that the second wave from the near end of the tank will be of greater amplitude than the first wave. If when the second wave arrives back at the near end, the end of the tank is struck again with the hammer the resulting third impulse will be of greater amplitude than the second impulse. If at the instant of the return of each succeeding impulse the end of the tank is struck, the result will be cumulative and each succeeding wave will be of greater magnitude than the one preceding until the point is reached where the losses due to resistance become sufficient to prevent a further increase in amplitude of the wave.

Under certain conditions a similar phenomenon may occur in electric circuits and this is known as "quarter wave resonance". If an electric impulse* is sent into a

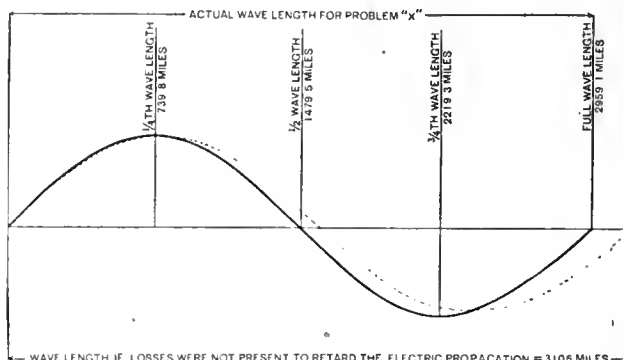


FIG. 14—WAVE LENGTH OF 60 CYCLE CIRCUIT

conductor, such as a transmission circuit, this impulse travels along the conductor at the velocity of light. If the circuit is open at the other end, the impulse is there reflected and returns at the same velocity. If at the moment when the impulse arrives at the starting point a second impulse is sent into the circuit, the returned first impulse adds itself to, and so increases the second impulse; the return of this second impulse adds itself to the third impulse, and so on; that is, if alternating impulses succeed each other at intervals equal to the time required by an impulse to travel over the circuit and back, the effects of successive impulses add themselves, and large currents and high e.m.f.'s may be produced by small impulses. This condition is known as quarter wave electric resonance. To produce this condition, it is necessary that the alternating impulses occur at time intervals equal to the time required for the impulses to travel the length of the line and back. For example, the time of one half wave or cycle of impressed e.m.f.

*For a complete study of this subject see "Transient Electric Phenomena and Oscillations" by C. P. Steinmetz, from which the above description of quarter wave resonance has largely been taken.

is the time required by light to travel twice the length of the line, or the time of one complete cycle is the time light requires to travel four times the length of the line. Stated another way, the number of cycles or frequency of the impressed alternating e.m.f.'s in resonance condition, is the velocity of light divided by four times the length of the line; or to have free oscillation or resonance condition, the length of the line is one quarter wave length of light. The cycles at which this condition is reached (if there were no losses present) would be determined as follows:—

$$\text{Frequency} = \frac{46587}{\text{Length in miles}} \dots\dots\dots (23)$$

or

$$\text{Length in miles} = \frac{46587}{\text{Frequency}} \dots\dots\dots (24)$$

RESONANCE LENGTHS OF CIRCUITS

Commercial frequencies are so low that to reach a quarter wave resonance condition with them the circuit would have to be of great length. The following values, for the sake of simplicity, are based upon the assumption that there are no losses in the circuit.

Fundamental Frequency	Resonance Length	Wave Length
15 cycles	3106 miles	12434 miles
25 cycles	1863 miles	7452 miles
40 cycles	1165 miles	4600 miles
60 cycles	776 miles	3106 miles

The above lengths are based upon the impressed or fundamental frequencies. If these impressed frequencies contain appreciable higher harmonics, some of the latter may approach resonance frequency and, if of sufficient magnitude, may cause trouble. Thus the length of circuit corresponding to resonance conditions of various harmonics of the fundamental is given below.

Cycles	Harmonics		
	3rd.	5th.	7th.
15	1035 miles	631 miles	444 miles
25	621 miles	372 miles	266 miles
40	388 miles	233 miles	166 miles
60	258 miles	155 miles	111 miles

Thus an impressed frequency of 60 cycles will not produce quarter wave electric resonance unless the circuit be approximately 776 miles long. If a third harmonic, however, is present in the impressed wave, this harmonic will develop quarter wave resonance in a circuit approximately 258 miles long, a 5th harmonic in a circuit approximately 155 miles long, and a 7th harmonic in a circuit approximately 111 miles long.

The above values are based upon no losses being encountered in transmission. Obviously this is an incorrect assumption, as electric propagation is always accompanied by more or less loss, depending upon the fundamental constants (resistance and leakage) of the circuit. The effect of such losses is to retard the velocity of the electric propagation, usually by an amount of five to ten percent below that of light. The above values of circuit lengths representing a condition for resonance may therefore be as much as ten percent above the actual lengths.

An investigation of the effects of higher harmonics

of the impressed wave is of importance in connection with very long distance transmission systems.

PARALLELING TRANSMISSION CIRCUITS

Transmission lines are frequently constructed with duplicate circuits which are normally operated in parallel. In other cases two circuits may lead from the generating station in divergent directions and at some distant point come together and be connected in parallel.

If the two circuits are fed from different generators, or sources of supply, the only condition necessary for paralleling the circuits is that the phase rotation of the two circuits be the same and that the regulation in speed of the prime movers of the generators feeding the two systems can be adjusted so as to bring the phases of the two circuits together for paralleling.

If, however, the two circuits which are to be connected in parallel are fed from the same source of supply, the case may become involved. There will be no trouble in obtaining the correct phase rotation, for should the circuits not rotate alike, it is only necessary to transpose any two of the connections of either of the circuits (assuming that the circuits are three-phase). The other condition to be met is that the phases of both circuits to be paralleled are the same, i. e., the voltages in the phases to be paralleled must pass through their zero and maximum values at the same instant.

If neither circuit has transformers between the points where they are to be connected in parallel, their phases will coincide and there will be no trouble about connecting them in parallel. If one circuit has no transformers and the other has transformers, the phase relations of the two circuits will depend upon the kind of transformer connections employed. Suppose it is assumed that the raising transformers are connected delta to star and the lowering transformers are connected delta to delta. With these connections the phases of the two circuits will be 30 electrical degrees apart and it will be impossible to parallel the circuits. In other words one delta-star or star-delta transformer connection produces a phase displacement of 30 degrees. It will be obvious that a second delta-star or star-delta connection will restore the original phase relation. A delta-delta connection or a star-star connection does not affect the phase relations. If both circuits have an even number of star and even number of delta windings, the equivalent resultant will be the same as if all the connections were either delta-delta or star-star; hence, there will be no resultant change in phase relations and the two circuits can be paralleled with each other or with a circuit having no transformations. If, however, both circuits have an odd number of delta and an odd number of star windings, any attempt to resolve them into the equivalent number of delta-delta and star-star connections will leave one star and one delta; the effect is the same as if there was one star-delta connection in the circuits. This will twist the phase relations of the terminals 30 degrees out of phase from the generators. Since both circuits will have an

equivalent phase displacement, they can be paralleled with one another, but since both are 30 degrees out of phase with the generators, they cannot be paralleled with a line having no transformations; nor with a line having an even number of star and delta connections.

When the phase angles of the two transmission circuits (receiving their power from a common source) are known to be such as to permit of parallel operation it is then necessary to phase them out before connecting the circuits together. The phase rotation can be checked most readily by means of a polyphase motor connected first to one circuit and then to the other, being careful to connect the leads in the same order in each case. If the motor runs in the same direction from both circuits, the phase rotation of the circuits will be the same. The phase angle can be readily tested by means of a single-phase synchroscope*. In case a polyphase motor and synchroscope are not available, the phasing out of the circuits may be accomplished by the use of a voltmeter and transformer.** As an illustration, assume that from a 4400 volt bus in a generating station a 4400 volt transmission circuit extends for some distance from the station. A second transmission circuit fed from the same bus but containing both raising and lowering transformers is to be paralleled at the farther end with the 4400 volt circuit which contains no transformers. The phase angles of the lines are assumed to be such as to permit paralleling the two circuits, with proper connections.

One of the transmission circuits is connected to one side of the paralleling switch as in Fig. 15 and the other circuit to the other side of the same switch. The three terminals on one side of the switch may be tagged 1-2-3. Likewise the three terminals on the other side of the switch may be tagged 4-5-6. Connect any two terminals together (1 and 4 in this case) by a jumper. Take voltage readings across the corresponding terminals 2 to 5, 3 to 6, and 3 to 5, 2 to 6. From these voltage readings it is a simple matter to indicate by a vector diagram the relative phase relations at the switch contacts of the two circuits to be paralleled. In the case illustrated, the readings indicate that the relative voltage relations on the two sides of the paralleling switches are as indicated by the full line delta 1-2-3, and the broken line delta 4-5-6. It will be seen that phase 1-3 will parallel with phase 4-5, that phase 1-2 will parallel with phase 6-5 and phase 2-3 will parallel with phase 4-6. In order to bring about this phase relation it will be necessary to change the transformer connections on the low-tension side of the lowering transformers, inside of the delta. That is the 6 end of the transformer windings 5-6 will be connected to the 4 end of transformer

4-5. The 4 end of transformer 4-6 will be connected to the 5 end of transformer, 5-6 and the 6 end of transformer 4-6 will be connected to the 5 end of transformer 4-5. These changes will shift the position of the delta 4-5-6 so that it will coincide with delta 1-2-3. A further test of voltage between switch terminals 2 to 5 and 3 to 6 should indicate zero voltage across the switch terminals to be connected together, in which case the paralleling switches may be closed. In order to measure the voltage across the paralleling switch contacts it will usually be necessary to employ a potential transformer. This transformer and voltmeter should be capable of withstanding 1.73 times the voltage of the circuit for, with the connections given in Fig. 15, one reading gave 7610 volts, whereas the voltage of the circuit was only 4400 volts.

In case there is a ground on both systems, the placing of a jumper across two of the switch contacts would result in a short-circuit. This jumper should not be placed across the switch until it has been shown by connecting a transformer across these two contacts that no potential exists between them.

HEATING OF BARE CONDUCTORS IN AIR

If the circuit is long, the voltage will probably be high and consequently the current to be transmitted

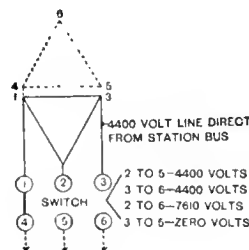


FIG. 15—TEST FOR PHASE SEQUENCE

small. In this case, the heating effect of the current will be small and unimportant. If, however, the circuit is short and an unusually large amount of power is to be transmitted, the current will be large. Since the I^2R loss varies as the square of the current and directly as the resistance, the heat generated, if the current is large, may be sufficient to overheat or anneal the material of the conductors. In some cases of unusually large amounts of power being transmitted short distances, the heating effect of the currents resulting may be sufficient to limit the amount of power that can be transmitted at a given voltage.

Table XXIII should be consulted in cases where the circuit is short and the amount of power to be transmitted large. In this table are columns containing current values which have been calculated corresponding to 10, 25 and 40 degrees C. rise in temperature for various sizes of bare copper conductors suspended in still air at a temperature of 25 degrees C. In other words these current values are based upon absolute temperatures of 35, 50 and 65 degrees C. The current values corresponding to a temperature rise of 40 degrees C.

*These tests are described in an article on "Phasing Out High Tension Lines" by E. C. Stone in the JOURNAL for Nov. 1917, p. 448.

**This method is described in an article on "Determination of Polarity of Transformers for Parallel Operation" by W. M. McConahey, in the JOURNAL for July 1912, p. 613. See also article on "Polarity of Transformers" by W. M. Dann in the JOURNAL for July 1916, p. 350.

ERRATUM

The formula used in calculating the values for table XXIII, page 43, embodied the only available information on this subject at the time the values were calculated. Recent exhaustive and carefully conducted tests, made by George E. Luke, indicate a wide difference in results from the table values, especially in the larger size conductors. The table values corresponding to 40° C rise should not, therefore, be used.

In the April, 1923 issue of the Electric Journal, page 127, appears an article entitled "Current Capacity of Wires and Coils" in which Mr. Luke gives the results of his tests and the empirical formula he developed as a result of the test.

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TABLE XXIII—HEATING CAPACITY FOR 40° C. RISE OF BARE COPPER CONDUCTORS SUSPENDED OUT OF DOORS

CONDUCTORS				AMPERES—BARE CONDUCTORS IN STILL AIR FOR TEMPERATURE RISES STATED.			APPROXIMATE CARRYING CAPACITY IN KVA 3 PHASE CORRESPONDING TO A TEMPERATURE RISE OF 40 C (BASED UPON AMPERES IN COLUMN MARKED "FOR 40°C RISE") FOR BARE COPPER CONDUCTORS SUSPENDED IN STILL AIR OUT OF DOORS.									
TYPE	B & S NO.	AREA IN CIRCULAR MILS	DIAMETER IN INCHES	FOR 10°C RISE	FOR 25°C RISE	FOR 40°C RISE	220 VOLTS	440 VOLTS	550 VOLTS	1100 VOLTS	2200 VOLTS	4000 VOLTS	4400 VOLTS	6000 VOLTS	6600 VOLTS	6900 VOLTS
				K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.	K.V.A.
STRAINED		2 000 000	.663	2140	3280	4050	1540	3080	3850	7700	15400	27900	30800	42000	46200	48300
		1 800 000	.6548	1980	3020	3760	1430	2860	3580	7150	14300	26000	28600	39100	42800	44800
		1 700 000	.6504	1890	2920	3600	1370	2740	3420	6850	13700	24800	27400	37100	40800	43000
		1 600 000	.6459	1810	2780	3440	1310	2620	3270	6550	13100	23800	26200	35800	39200	41000
		1 500 000	.6412	1720	2640	3300	1250	2500	3140	6280	12500	22800	25120	34300	37400	39200
		1 400 000	.6364	1635	2520	3100	1180	2360	2950	5990	11800	21400	23600	32200	35100	37000
		1 200 000	.6263	1460	2230	2760	1050	2100	2620	5250	10500	19100	21000	28700	31500	33000
		1 100 000	.6209	1360	2100	2580	980	1960	2460	4920	9840	17800	19680	26800	29400	30800
		1 000 000	.6152	1270	1950	2420	920	1840	2300	4600	9200	16700	18400	25100	27600	28800
		950 000	.6123	1215	1870	2320	880	1760	2200	4400	8800	16000	17600	24100	26400	27700
		900 000	.6093	1150	1780	2220	840	1680	2100	4200	8440	15300	16880	23000	25300	26500
		850 000	.6062	1120	1720	2130	800	1620	2030	4050	8100	14700	16200	22100	24300	25400
		800 000	.6031	1075	1640	2030	770	1540	1930	3870	7740	14000	15480	21100	23200	24200
		750 000	.6000	1025	1580	1940	740	1480	1840	3690	7380	13400	14760	20200	22100	23000
		700 000	.5969	980	1490	1830	710	1390	1740	3480	6960	12600	13920	19000	20900	21800
		650 000	.5929	920	1410	1740	660	1320	1660	3320	6640	12000	13280	18100	19800	20800
		600 000	.5893	870	1330	1630	620	1250	1560	3170	6240	11300	12480	17000	18700	19600
		550 000	.5853	810	1250	1530	580	1160	1450	2900	5800	10600	11600	15900	17500	18300
		500 000	.5815	755	1160	1430	540	1090	1360	2720	5440	9860	10880	14900	16300	17100
		450 000	.5772	700	1060	1320	500	1000	1250	2500	5000	9100	10000	13700	15100	15700
	400 000	.5728	640	980	1210	460	920	1150	2300	4600	8350	9200	12600	13800	14400	
	350 000	.5681	575	885	1090	415	830	1040	2080	4160	7500	8320	11300	12400	13000	
	300 000	.5630	515	785	970	370	740	930	1840	3680	6700	7360	10100	11100	11600	
	250 000	.5575	450	685	840	320	640	800	1600	3200	5800	6400	8700	9550	10000	
	0000	211 600	.528	385	605	750	285	570	715	1430	2860	5170	5720	7800	8550	8950
	000	167 772	.470	330	525	630	230	475	595	1190	2380	4320	4760	6500	7130	7450
	133 079	.418	280	425	527	500	400	500	1000	2000	3640	4000	5510	6000	6380	
	0	105 560	.373	235	360	444	170	336	423	846	1692	3060	3384	4600	5050	5300
	1	83 464	.332	195	300	370	141	282	352	704	1408	2550	2816	3840	4220	4400
	2	66 358	.292	162	250	307	111	233	292	584	1168	2120	2336	3180	3500	3660
	3	52 624	.260	136	210	258	98	196	245	490	980	1770	1960	2680	2940	3080
	4	41 738	.232	114	176	212	89	178	224	448	896	1620	1792	2440	2680	2800
	5	33 088	.206	96	147	182	69	138	173	346	692	1260	1384	1890	2080	2180
SOLID		2 000 000	.660	370	565	728	275	550	690	1380	2760	5030	5520	7550	8300	8700
		1 800 000	.650	340	475	588	245	488	560	1120	2240	4060	4480	6100	6700	7000
		1 700 000	.645	325	420	475	225	440	500	1000	1880	3760	4080	5550	6000	6300
		1 600 000	.640	305	385	415	158	316	395	790	1580	2860	3160	4300	4730	4950
		1 500 000	.635	285	355	385	132	264	330	660	1320	2400	2640	3620	3970	4150
		1 400 000	.630	265	325	355	112	224	280	560	1120	2040	2240	3060	3370	3520
		1 200 000	.620	229	280	307	93	186	233	465	930	1690	1860	2540	2790	2920
		1 100 000	.615	204	240	270	81	158	197	394	788	1430	1576	2150	2360	2470
		1 000 000	.610	182	200	240	71	132	165	330	660	1200	1320	1810	1980	2080
		950 000	.605	170	185	225	66	120	150	300	600	1100	1200	1600	1700	1780
		900 000	.600	158	170	210	60	110	140	270	540	1000	1100	1400	1500	1580
		850 000	.595	146	155	195	55	100	130	240	480	900	1000	1300	1400	1480
		800 000	.590	134	140	180	50	90	120	210	420	800	900	1200	1300	1380
		750 000	.585	122	125	165	45	80	110	180	360	700	800	1100	1200	1280
		700 000	.580	110	110	150	40	70	100	150	300	600	700	1000	1100	1180
		650 000	.575	98	98	135	35	60	90	120	240	500	600	900	1000	1080
		600 000	.570	86	86	120	30	50	80	100	200	400	500	800	900	980
		550 000	.565	74	74	105	25	40	70	90	160	300	400	700	800	880
		500 000	.560	62	62	90	20	30	60	80	120	200	300	600	700	780
		450 000	.555	50	50	75	15	20	50	70	100	150	200	500	600	680
	400 000	.550	38	38	60	10	10	40	60	80	120	150	400	500	580	
	350 000	.545	26	26	45	5	5	30	40	60	80	100	300	400	480	
	300 000	.540	14	14	30	0	0	20	30	40	60	80	200	300	380	
	250 000	.535	2	2	15	0	0	10	15	20	30	40	100	150	200	
	0000	211 600	.528	385	605	750	1430	1710	1950	2600	2860	3900	4270	6000	6500	6800
	000	167 772	.470	330	525	630	1190	1420	16200	21600	23800	32400	35600	54000	58000	61200
	133 079	.418	280	425	527	500	1000	1200	13700	18200	20000	27400	30000	45500	51500	
	0	105 560	.373	235	360	444	1150	1500	1750	2300	2600	3600	4000	5500	6000	6400
	1	83 464	.332	195	300	370	960	1280	1500	2000	2400	3200	3600	5000	5500	5900
	2	66 358	.292	162	250	307	720	960	1100	1400	1800	2400	2800	3800	4200	4500
	3	52 624	.260	136	210	258	490	670	820	1000	1300	1700	2100	2800	3100	3300
	4	41 738	.232	114	176	212	440	600	750	900	1200	1600	2000	2700	3000	3200
	5	33 088	.206	96	147	182	340	450	550	700	900	1200	1500	2000	2300	2500
	0000	211 600	.660	370	565	728	1380	1660	1890	2520	2760	3750	4150	6300	6800	7200
	000	167 772	.610	310	475	588	1120	1340	1500	2040	2240	3060	3350	5080	5500	5800
	133 079	.560	258	420	495	630	940	1130	1290	1710	1880	2580	2820	4260	4600	4900
	0	105 560	.625	218	335	415	7180	7900	9450	10800	14360	15800	21600	23600	30000	33200
	1	83 464	.589	192	280	348	6020	6620	7950	9050	12040	13600	18100	19800	30000	34400
	2	66 358	.558	154	238	295	5100	5600	6750	7650	10200	11200	16800	25500	30600	36000

(absolute temperature of 65 degrees C.) have also been expressed in the form of k.v.a., three-phase values corresponding to various transmission voltages. Thus No. 0000 stranded bare copper conductors suspended in still air out doors at 25 degrees C. will carry 750 amperes with a temperature rise of 40 degrees C. (absolute temperature 65 degrees C.). If the transmission voltage is 220 volts, the corresponding k.v.a. value will be 285 k.v.a. three-phase and if the transmission voltage is 10 000 volts, 13 000 k.v.a. may be transmitted with the same temperature rise.

As indicated by foot notes the values of the table were calculated by formulas from Foster's Handbook as follows:—

$$\text{Amperes} = 1100 \sqrt{\frac{TD^3}{R}} \text{ for stranded conductor....(25)}$$

$$\text{Amperes} = 1250 \sqrt{\frac{TD^3}{R}} \text{ for solid conductor.....(26)}$$

Where

T = Temperature rise in degrees C.

D = Diameter of conductors in inches.

R = Resistance of conductors in ohms per mil-foot at final temperature.

CHAPTER VI

DETERMINATION OF FREQUENCY & VOLTAGE

FREQUENCY DETERMINATION

Cost of Transformers—Sixty cycle transformers cost approximately 30 to 40 percent less than 25 cycle transformers; or stated another way, 25 cycle transformers cost approximately 40 to 66 percent more than 60 cycle transformers. The saving in first cost may vary between \$1.50 and \$2.50 per kv-a. in favor of 60 cycles. Assuming that the total kv-a. of transformer capacity connected to a transmission circuit is 2.5 times the kv-a. transmitted over the circuit, the saving in favor of 60 cycle transformers would be \$3.75 to \$6.25 or an average of \$5.00 per kv-a. transmitted. Assuming 20 000 kv-a. to be transmitted, the saving in cost at \$5.00 per kv-a. will be \$100 000 in favor of 60 cycle transformers. The actual difference in cost will depend upon the type of the transformers, that is, whether water or self-cooled and also upon their average capacity. The difference in cost will be greater for the self-cooled type and for the smaller capacities.

Weight and Space of Transformers—The less weight of 60 cycle transformers makes them easier to handle and they require less space for installation.

Higher Reactance—Inductive reactance at 60 cycles is 2.4 times its value at 25 cycles. This tends to produce poorer voltage regulation of the circuit. Higher reactance has one advantage for the larger systems in that it tends to limit short-circuit currents and thus assists the circuit opening devices to function properly. By virtue of the higher reactance it might be possible in some cases to obtain sufficient reactance in the transformers without the addition of current limiting reactance coils.

Efficiency—The efficiency of 60 cycle transformers is usually 0.25 to 0.50 percent higher than for 25 cycle transformers.

Charging Current—At 25 cycles both the charging current and the reactance are approximately 42 percent of their values for 60 cycles. This tends to give better regulation and usually higher efficiency in transmission. On the other hand, the higher transmission efficiency may be offset by the slightly lower efficiency of 25 cycle transformers. In cases of very long circuits (particularly if the circuits are in duplicate and both in service) or of transmission systems embracing many miles of high tension mains and feeders, the charging currents may be so great as to limit the choice in transmission voltage. On the other hand large charging currents may be permitted, provided under excited synchronous motors are used at various parts of the transmission

system for partially neutralizing this charging current and for maintaining constant voltage.

Inductive Disturbances—Lightning, switching and other phenomena cause disturbances on conductors of transmission circuits. The frequency of these disturbances is independent of that impressed on the system. After the removal of the disturbing influence they oscillate with the natural frequency of the line.

The natural frequency of the line is far above commercial frequencies but, if the transmission line is long, there may be some odd harmonic present in the fundamental impressed frequency which corresponds with the natural period of the line. This might tend to produce an unstable condition or resonance. This condition is somewhat less likely to occur at 25 cycles.

Summary—Although there are a number of large 25 cycle transmission systems in operation, they were mostly installed before the design of 60 cycle converting apparatus and electric light systems had reached their present state of perfection. Unless it is desirable to parallel with an existing 25 cycle system located in adjoining territory without the introduction of frequency changers, it is now quite general practice to choose the frequency of 60 cycles.*

VOLTAGE DETERMINATION

From a purely economic consideration of the conductors themselves, Kelvin's law for determining the most economical size of conductors would apply. Kelvin's law may be expressed as follows:—

"The most economical section of a conductor is that which makes the annual cost of the I^2R losses equal to the annual interest on the capital cost of the conducting material, plus the necessary annual allowance for depreciation". That is, the economical size of conductor for a given transmission will depend upon the cost of the conducting material and the cost of power wasted in transmission losses. The law of maximum economy may be stated as follows:—"The annual cost of the energy wasted per mile of the transmission circuit added to the annual allowance per mile for depreciation and interest on first cost, shall be a minimum".

Attempts have been made to determine by mathematical expression the most economical transmission voltage, all factors having been taken into account. There are so many diverse factors entering into such a

*For a complete discussion of this subject see a paper by D. B. Rushmore before the Schenectady section A. I. E. E., May 17, 1912, on "Frequency" and an article by B. G. Lamme on "The Technical Story of the Frequencies" in the JOURNAL for June, 1918, p. 230.

treatment as to make such an expression complicated, difficult and unsatisfactory. There are many points requiring careful investigation, not embraced by Kelvin's law, before the proper transmission voltage can be determined. Some of these points are given below.

Cost of Conductors—For a given percentage energy loss in transmission, the cross-section and consequently the weight of conductors required by the lower and medium voltage lines (up to approximately 30,000 volts) to transmit a given block of power varies inversely as the square of the transmission voltage. Thus if this voltage is doubled, the weight of the conductors will be reduced to one fourth with approximately a corresponding reduction in their cost. This saving in conducting material for a given energy loss in transmission becomes less as the higher voltages are reached, becoming

conductors, from which their cost may readily be calculated, is given in Table E-1. As an insurance against breakdown, important lines frequently are built with circuits in duplicate. In such cases the cost of conductors for two circuits should not be overlooked.

Table E-1 contains the weights of bare stranded copper cables per 1000 feet of circuit, also per mile of circuit. For the purpose of facilitating rapid calculation for any given case, the weights are given corresponding to one, two and three conductors for these two lengths of circuit.

Reduced Electric Surges—The better insulation necessitated by higher transmission voltages tends to make the circuit more secure against ordinary disturbances. Also the smaller currents resulting with the higher voltages cause less disturbance in the circuit in the case of grounds, short-circuits, switchings, lightning and other disturbances.

Less Reactance Volts Drop—Since the current corresponding to higher transmission voltages goes down as the voltage goes up, the voltage necessary to overcome the reactance of the circuit will be less, and the percentage reactance volts much less for higher volt-

TABLE E 1—WEIGHT OF BARE COPPER CONDUCTORS

B & S NO	AREA IN CIRCULAR MILS	WEIGHT IN POUNDS					
		PER 1000 FEET OF CIRCUIT			PER MILE OF CIRCUIT		
		NUMBER OF CONDUCTORS			NUMBER OF CONDUCTORS		
		ONE	TWO	THREE	ONE	TWO	THREE
2 000 000	6 180	12 360	18 540	32 630	65 260	97 890	
1 900 000	5 870	11 740	17 610	30 994	61 988	92 982	
1 800 000	5 560	11 120	16 800	29 357	58 714	88 071	
1 700 000	5 250	10 500	15 750	27 720	55 440	83 160	
1 600 000	4 940	9 880	14 820	26 083	52 166	78 249	
1 500 000	4 630	9 260	13 890	24 446	48 892	73 338	
1 400 000	4 320	8 640	12 960	22 810	45 620	68 430	
1 300 000	4 010	8 020	12 030	21 173	42 346	63 519	
1 200 000	3 710	7 420	11 130	19 539	39 178	58 767	
1 100 000	3 400	6 800	10 200	17 952	35 904	53 856	
1 000 000	3 090	6 180	9 270	16 315	32 630	48 945	
950 000	2 930	5 860	8 790	15 470	30 940	46 410	
900 000	2 780	5 560	8 340	14 678	29 356	44 034	
850 000	2 620	5 240	7 860	13 834	27 668	41 502	
800 000	2 470	4 940	7 410	13 042	26 084	39 126	
750 000	2 320	4 640	6 960	12 250	24 500	36 750	
700 000	2 160	4 320	6 480	11 405	22 810	34 215	
650 000	2 010	4 020	6 030	10 613	21 226	31 839	
600 000	1 850	3 700	5 550	9 768	19 536	29 304	
550 000	1 700	3 400	5 100	8 976	17 952	26 928	
500 000	1 540	3 080	4 620	8 173	16 262	24 393	
450 000	1 390	2 780	4 170	7 339	14 678	22 017	
400 000	1 240	2 480	3 720	6 547	13 094	19 641	
350 000	1 080	2 160	3 240	5 702	11 404	17 106	
300 000	926	1 852	2 778	4 889	9 778	14 667	
250 000	772	1 544	2 316	4 076	8 152	12 228	
200 000	653	1 306	1 959	3 448	6 896	10 344	
150 000	518	1 034	1 554	2 735	5 470	8 205	
100 000	411	822	1 233	2 170	4 340	6 510	
0000	326	652	978	1 721	3 442	5 163	
1	83 700	258	516	774	1 362	2 724	
2	62 400	205	410	615	1 082	2 164	
3	52 600	163	326	489	861	1 722	
4	41 700	129	258	387	681	1 362	
5	33 100	102	204	306	539	1 078	
6	26 300	81	162	243	428	856	
7	20 800	64	128	192	338	676	
8	16 300	51	102	153	269	538	

ing increasingly less as voltages go higher. This is for the reason that for the higher voltages at least two other sources of losses, leakage over insulators and the escape of energy through the air between the conductors (known as "corona") appear. In addition to these two losses, the charging current, which increases as the transmission voltage goes higher, may either increase or decrease the current in the circuit depending upon the power-factor of the load current and the relative amount of the leading and lagging components of the current in the circuit. Any change in the current of the circuit will consequently be accompanied by a corresponding change in the I²R loss. In fact, these sources of additional losses may, in some cases of long circuits or extensive systems, materially contribute toward limiting the transmission voltage. The weight of copper

TABLE F—PRESENT RELATIVE COSTS OF HIGH TENSION APPARATUS

Expressed in Percent (6600 Volt Costs Taken as 100%)

	6600 Volts	11000 Volts	13200 Volts	16500 Volts	22000 Volts	33000 Volts	44000 Volts	66000 Volts	88000 Volts	110000 Volts	176000 Volts
Transformers	100	102	104	106	108	115	125	150	175	200	225
Switches	100	100	100	100	100	110	115	155	255	420	
Electrolytic Arresters	100	151	160	195	205	320	430	640	1600	1900	2400
Insulators	100	135	185	365	430	650	1250	3500	5500	6500	7700

ages. Thus, if the transmission voltage is doubled, the current will be halved and for the same spacing of conductors the reactance volts drop will be one half, resulting in one fourth the percentage of the reactance-volts drop.

Cost of Transformers—If the transmission voltage exceeds 13 200 volts, banks of step-up transformers will be required of sufficient capacity to transform all of the kv-a. to be transmitted. A still greater capacity of step down transformers will be required to reduce the voltage to that suitable for operating motors and lights. In some cases two reductions from the transmission circuit voltage may be required, the first usually reducing to 22 000, 11 000 or 6600 volts for general distribution and the second reducing from the general distribution voltage to the proper voltage for motors and lights. The net result is that the total capacity in transformers connected to a transmission system employing both step up and step down transformers may vary from a minimum of two to a maximum of about four times the kv-a. transmitted over the high-tension circuits. The average condition we will assume as 2.5 times the kv-a. to be transmitted.

The cost of power transformers at the present time

for 66 000 volts service will vary between \$1.25 to \$3.00 for 60 cycle and \$2 to \$5 per kv-a. for 25 cycle service, depending upon their type and capacity. The total cost per kv-a. of transformers on a system would therefore be represented by approximately 2.5 times the above costs. The present relative costs of transformers for different voltages are given in Table F. For instance if the transmission voltage is increased from 33 000 to 66 000 volts the transformers will cost in the neighborhood of $150 \div 115$ or 31 percent more than they would cost for 33 000 volts. Knowing the amount of power to be transmitted, an approximate estimate may be made as to the additional cost of the necessary transformers for a higher voltage.

Cost of Insulators—Table F values indicate a wide difference in the cost of insulators for the higher volt-

Efficiency—The efficiency of transformers will be slightly higher for the lower voltages.

Small Customers—The furnishing of power to small customers at points along the transmission circuits should receive careful consideration. The cost of switching apparatus, lightning arresters and transformers required to permit service being given to such customers will be less for the lower voltage.

Charging Current—The amount of current required to charge the transmission circuits varies approximately as the transmission voltage. Therefore the charging current, expressed in kv-a. varies approximately as the square of the voltage. Thus the charging current required for a 33 000 volt circuit is approximately one half and the charging kv-a. one fourth that of a 66 000 volt circuit.

TABLE G—FORM OF TABULATION FOR DETERMINING VOLTAGES AND CONDUCTORS

BASED ON THE TRANSMISSION OF 10 000 KV-A. FOR TEN MILES AT 80 PERCENT POWER-FACTOR LAGGING, 60 CYCLES, THREE PHASE

VOLTAGE		AMPERES FOR 10,000 KVA	CONDUCTORS							VOLTAGE DROP AT FULL LOAD			FIRST COST					ANNUAL OPERATING COST				
BETWEEN CONDUCTORS	TO NEUTRAL		B & S OR CIRCULAR MILS	TOTAL WEIGHT IN POUNDS	RESISTANCE OHMS	TOTAL I ² R LOSS			RESISTANCE IR IN %	REACTANCE IX IN %	VOLTS DROP IN %	CONDUCTORS AT 25 CTS. PER POUND	TRANSFORMERS 25,000 KVA	HIGH TENSION SWITCHES	LIGHTNING ARRESTERS	INSULATORS	TOTAL	INTEREST ON FIRST COST AT 6 %	DEPRECIATION ON FIRST COST AT 10 %	I ² R LOSSES AT 1 CT PER KW-HOUR	TOTAL	
						KW FOR 10 HRS	LOSS IN %	2500 KVA														KW FOR 14 HRS
16,500	9 526	350	500 000	243 930	1.17	4 30	5.3	2.7	1 707 470	4.3	21.7	17.5	60 982	75 000	3 000	1 000	900	140 882	8 453	14 088	1 707.5	3 9616
			300 000	144 670	1.96	720	9.0	4.5	2 857 950	7.2	22.7	20	36 670	75 000	3 000	1 000	900	116 570	6 994	11 657	28 580	47 231
			*000	82 050	3.50	1286	16.1	8.0	5 102 700	12.9	24.2	25	20 512	75 000	3 000	1 000	900	100 412	6 025	10 041	51 027	67 093
22,000	12 702	262	300 000	146 670	1.96	403	5.0	2.5	1 598 700	4.0	12.8	11	36 670	76 500	3 000	1 050	1 200	118 420	7 105	11 842	15 987	34 934
			*000	82 050	3.50	720	9.0	4.5	2 857 950	7.2	13.6	14	20 512	76 500	3 000	1 050	1 200	102 262	6 136	10 226	28 580	44 942
			*0	51 630	5.55	1143	14.3	7.1	4 534 760	11.5	14.1	17.5	12 910	76 500	3 000	1 050	1 200	94 660	5 680	9 466	45 348	60 494
33 000	19 053	175	*00	65 100	4.42	406	5.1	2.5	1 609 650	4.0	6.5	7.0	16 275	82 500	3 300	1 600	1 980	105 655	6 340	10 565	16 097	33 002
			*2	32 460	8.83	811	10.1	5.0	3 215 650	8.0	6.8	10.5	8 117	82 500	3 300	1 600	1 980	97 497	5 850	9 749	32 156	47 753
			*4	20 430	14.1	1295	16.2	8.1	5 140 660	12.9	7.1	14.5	5 107	82 500	3 300	1 600	1 980	94 487	5 670	9 448	51 407	66 525
44 000	25 404	131	*2	32 460	8.83	454	5.7	2.9	1 805 290	4.6	3.9	6.0	8 117	90 000	3 450	2 200	3 960	107 727	6 463	10 772	18 053	35 288
			*5	16 170	17.8	916	11.4	5.8	3 639 780	9.1	4.0	9.5	4 040	90 000	3 450	2 200	3 960	103 650	6 219	10 365	36 378	52 982

ages; thus the increased cost of 66 000 volt insulators above the cost of 33 000 volt insulators is stated as $350 \div 650$ or 54 percent.

Cost of Other Apparatus—The cost of lightning arresters, high-tension circuit breakers and general insulation increase with the voltage. The increased cost of these items, however, may not have sufficient weight to materially influence the selection of the transmission voltage.

Cost of Buildings—Lower voltage transformers, switching equipment and lightning arresters require less space for insulation. If this apparatus is to be placed indoors, the cost of necessary buildings may be less. The amount of real estate required may also be less in case of the lower voltage.

Relative Cost Values—Table F contains relative cost values for different transmission voltages. They indicate approximately the variation, at the present time, in cost of the principal material which is affected by a change in transmission voltage. Cost values are very unstable at present but the table will serve in a general way to indicate comparative costs.

Summary—In deciding upon the transmission voltage, careful and full consideration should be given to the present (or probable future) voltage of any neighboring or adjacent systems. There is an increasing tendency to combine generating and transmission systems for purposes of economy, and insurance against breakdown in service. If a possible future consolidation is not kept in mind when selecting the transmission voltage, a voltage may be decided upon which would render it impossible to parallel with a neighboring system, except through connecting transformers. In this case the transformers of the two systems would probably not be interchangeable for service on either system.

If the contemplated transmission system is remote from any existing system, a study of the initial and operating costs should be made corresponding to various sizes of conductors and to various assumed transmission voltages. A suggested tabulation for such comparisons is shown in Table G. In this table, it is assumed that 10 000 kv-a. (8000 kw at 80 percent power-factor lagging), is to be transmitted a distance of ten miles at 60 cycles, three-phase for ten hours, followed by

2500 kv-a. (2000 kw at 80 percent power-factor lagging) for 14 hours. Delta spacing is assumed of three feet for the lower two and four feet for the higher two voltages. Raising and lowering transformers will be required of an assumed total capacity of $2.5 \times 10,000$ or 25 000 kv-a. Conductors of hard drawn stranded copper are employed, the resistance of the conductors being taken at a temperature of 25 degrees C. from Table II.

The cost of the pole or tower line, the right of way, buildings and real estate for buildings is not included in this tabulation. Neither is the difference in transformer efficiencies taken into account. The difference in these items will not be sufficient in this case greatly to influence the choice of the transmission voltage, because all of the voltages compared are relatively low. Because of the large amount of power to be transmitted a comparatively short distance, the approximate rule of 1000 volts per mile for short lines does not hold true for this problem.

Assuming for the sake of argument that the price values given in this form of tabulation are approximately correct for this problem and that there are no neighboring transmission systems, then the problem reduces to cost economics.

Since both the first and operating costs in Table G are higher for 16 500 volts than they are for 22 000 volts, it is evident that 16 500 volts is economically too low a voltage.

In the consideration of 22 000 volts it will be seen that, of the three sizes of conductors, the largest size (300 000 circ. mil.) will be the cheaper in the end. Thus, if No. 000 were selected, the first cost would be \$16 159 less than for 30 000 circ. mil conductors, but the operating cost (due to greater loss in transmission) will be approximately \$10 000 a year more. For a similar reason No. 0 conductors will be disqualified.

In the consideration of 33 000 volts, No. 00 conductors will be the choice and in the consideration of 44 000 volts, No. 2 conductors will be the choice. The choice then comes down to the following:—

Voltage Transmission	Conductors	Total Cost First	Annual Operating Cost
22 000	300 000 circ. mils	\$118 420	\$34 934
33 000	No. 00	105 655	33 002
44 000	No. 2	107 727	35 288

It will thus be seen that a voltage of 33 000 volts and No. 00 conductors are the most economical of those tabulated. The transmission loss will be 5.1 percent, the reactance 6.5 percent and the voltage drop seven percent at full load. The value assigned as the cost per

kw-hour for power lost in transmission will obviously have great influence in determining the proper economic size of conductors for any given transmission voltage. The cost of the copper will have a relatively greater importance on longer lines. As a matter of fact, a larger size than any of the conductors listed in Table G would be still more economical, under the conditions given. There have been numerous mistakes made in under-estimating the ultimate demand for electrical power and consequently adopting too low a transmission voltage. When in doubt the higher voltage will, in the course of time, most likely justify its adoption by reason of future growth not apparent at the time the choice is made.

The design and construction of transformers, circuit breakers, lightning arresters, etc. for a multiplicity of high-tension voltages is expensive. The manufacturers of such apparatus are endeavoring to standardize transmission voltages for the purpose of minimizing the number of designs of high-tension apparatus. This point could with mutual profit be taken up with the

TABLE H—COMMON TRANSMISSION VOLTAGES

Length of Line	Voltages
1 to 3 miles	550 or 2200 volts
3 to 5 miles	2200 or 6600 volts
5 to 10 miles	6600 or 13 200 volts
10 to 15 miles	13 200 or 22 000 volts
15 to 20 miles	22 000 or 33 000 volts
20 to 30 miles	33 000 or 44 000 volts
30 to 50 miles	44 000 or 66 000 volts
50 to 75 miles	66 000 or 88 000 volts
75 to 100 miles	88 000 or 110 000 volts
100 to 150 miles	110 000 or 132 000 volts
150 to 250 miles	132 000 or 154 000 volts
250 to 350 miles	154 000 or 220 000 volts

manufacturers before any particular voltage is decided upon.

The amount and cost of power to be transmitted is a very important factor in determining the economic transmission voltage. For average conditions isolated from existing transmission lines the voltages shown in Table H have been quite generally used. For exceptional cases, exceptional values will be used. For example if 40 000 kv-a. is to be transmitted 20 miles, 66 000 volts or higher might be used. On the other hand if a very small amount of power is to be transmitted, lower voltages would probably be selected.

At the present time the prospects seem bright for the standardization of the following "normal" system voltages.

44 000	132 000
66 000	154 000
88 000	*187 000
110 000	220 000

*The use of 187 000 volts is likely to occur only in case it is found necessary to have a voltage between 154 000 and 220 000 volts.

CHAPTER VII

PERFORMANCE OF SHORT TRANSMISSION LINES

(EFFECT OF CAPACITANCE NOT TAKEN INTO ACCOUNT)

THE PROBLEMS which come under the general heading of short transmission lines are those in which the capacitance of the circuit is so small that its effect upon the performance of the circuit may, for all practical purposes, be ignored. The effect of capacitance is to produce a current in leading quadrature with the voltage, usually designated as charging current. This leading component of current in the conductor does not appear in the load current at the receiving end of the circuit. It is zero at the receiving end of the circuit but increases at nearly a uniform rate as the sending end of the circuit is approached, at which point it ordinarily becomes a maximum.

The effect of this charging current flowing through the inductance of the circuit is to increase the receiving-end voltage and therefore to decrease the voltage drop under load. Since the charging current is 2.4 times greater for a frequency of 60 cycles than it is for a frequency of 25 cycles, its effect upon the voltage regulation will be considerably greater at 60 cycles than at 25 cycles. The effect of charging current upon the voltage regulation will also increase as the distance of transmission is increased.

If the circuit were without capacitance, there would be no charging current and consequently the mathematical and the two graphical solutions (impedance methods) which follow under the general heading of "short transmission lines" would all produce accurate results. All circuits, however, have some capacitance, and as the length or the frequency of the circuit increases, these three methods will therefore yield results of increasing inaccuracy. Some engineers consider these impedance methods sufficiently accurate for circuits 20 to 30 miles long while others use them for still longer circuits. To act as a guide, Table J indicates the error in the supply voltage as determined by these impedance methods, for circuits of different lengths corresponding to both 25 and 60 cycle frequencies. These three impedance methods produce practically the same results, and the sending end voltage, as determined by any of these methods, is always slightly high. In other words the effect of the charging current is to reduce the voltage necessary at the sending end, for maintaining a certain voltage at the receiving end of the circuit. The error referred to below for the three methods is expressed in percentage of the receiving end voltage. Thus, for a 30 mile, 25 cycle circuit, the error is 0.04 percent, and for a 30 mile, 60 cycle circuit the error is 0.2 percent. If an error of 0.5 percent is con-

sidered permissible, then the Dwight or the Mershon Chart methods, or the corresponding mathematical solution, may be used for 25 cycle circuits up to approximately 125 miles, and for 60 cycles circuits up to approximately 50 miles. Of course these impedance methods may be used for still longer circuits by making proper allowance to compensate for the fundamental error.

DIAGRAM ILLUSTRATING A SHORT TRANSMISSION CIRCUIT

Fig. 16 illustrates the relation between the various elements in short transmission circuits, when the effect of capacitance and leakage is not taken into account. The current flowing in such a circuit meets two opposing e.m.f.'s.; i.e. of resistance in phase with the current and reactance in lagging quadrature with the current.

The upper part of Fig. 16 illustrates such a circuit schematically and the lower part vectorially. The volt-

TABLE J

Length of Circuit (Miles)	Error in Percentage of Receiver Voltage	
	25 cycles	60 cycles
20	+0.02	+0.10
30	+0.04	+0.2
50	+0.1	+0.5
100	+0.4	+1.9
200	+1.4	+8.0
300	+3.3	+18.0

age component required at the sending end to overcome the resistance IR of the circuit is indicated in the vector diagram by a short line parallel with the base line I , representing the phase of the current. These lines are drawn parallel, since the resistance voltage drop is in phase with the current. The voltage component required at the sending end to overcome the reactance IX of the circuit is indicated by a line in quadrature or at right angles, to the phase of the current. The reactance is in quadrature with the current for the reason that the rate of change in the magnetic field (consequently the e.m.f. of self-induction or reactance) surrounding the conductor is greatest when the current is passing through zero. The hypotenuse IZ of this small right angle impedance triangle represents the impedance voltage of the circuit. It represents the direction and value of the resulting voltage necessary to overcome the combined effect of the resistance and the reactance of the circuit.

The relative values and phases of the receiving and

sending end voltages, and their phase relations with the current I , are also indicated on the vector diagram. This diagram is plotted for a receiving end load based upon 80 percent power-factor lagging. E_s represents the value of the voltage required at the sending end of the circuit to maintain the voltage E_r at the receiving end, when the impedance of the circuit is IZ and the receiving end power-factor is 80 percent lagging. The phase angle θ_s indicates the amount by which the current lags behind the voltage at the sending end; $\cos \theta_s$ being the power-factor of the load as measured at the sending end. Likewise $\cos \theta_r$ is the power-factor of the load at the receiving end.

TAPS TAKEN OFF CIRCUIT

Usually the main transmission circuit is tapped and power taken off at one or more points along the circuit. The performance of such a circuit must be calculated by steps thus:—Assume a circuit 200 miles long with 10 000 kw taken off at the middle and 10 000 kw at the receiving end. From the conditions known or assumed at the receiving end, calculate the corresponding send-

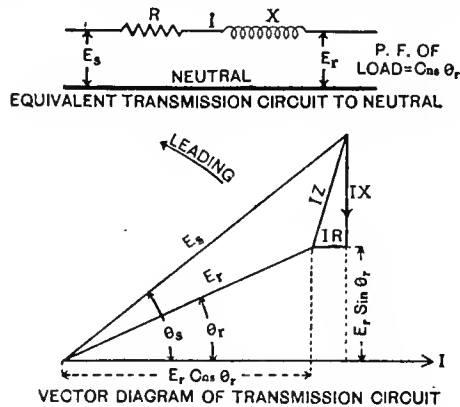


FIG. 16—DIAGRAMS FOR SHORT TRANSMISSION LINES

Impedance method, capacitance effect not taken into account.

ing end conditions, that is the voltage, power and power-factor at the substation in the middle of the circuit. To the calculated value of the actual power in kilowatts add the losses at the substation in the middle of the circuit. Any leading or lagging component in the substation load current must also be added algebraically, in order to determine the power-factor at the sending side of the substation. This will then be the receiving end conditions at the substation in the middle of the circuit, from which the corresponding conditions at the sending end of the circuit may be calculated. If the sending end conditions are fixed, and the receiving end conditions are to be determined, the substation losses will in such case be subtracted in place of added.

CABLE AND AERIAL LINES IN SERIES—COMPOSITE LINES

In some cases it is necessary to place part of a transmission circuit underground, and in other cases it may be desirable to use two or more sizes of conductors in series. The result will be that the circuit constants will be different for the various sections. If the effect of capacitance be neglected, the combined circuit may

be treated as a single circuit having a certain total resistance R and a total reactance X .

PROBLEMS

Later a table will be presented listing a large number of transmission circuits from 20 to 500 miles long, at both 25 and 60 cycles operating at from 10 000 to 200 000 volts. These problems are numbered from 1 to 64. When a reference is made in the following to some problem number it will refer to one of this list of problems.

SYMBOLS

The symbols which will be employed in the following treatment are given below:—

FOR LOAD CONDITIONS

- $Kv-a_r$ = (total) at receiving end.
- $Kv-a_{rn}$ = (one conductor to neutral) at receiving end.
- $Kv-a_s$ = (total) at sending end.
- $Kv-a_{sn}$ = (one conductor to neutral) at sending end.
- Kw_r = Kw (total) at receiving end.
- Kw_{rn} = Kw (one conductor to neutral) at receiving end.
- Kw_s = Kw (total) at sending end.
- Kw_{sn} = Kw (one conductor to neutral) at sending end.
- E_r = Voltage between conductors at receiving end.
- E_{rn} = Voltage from conductors to neutral at receiving end.
- E_s = Voltage between conductors at sending end.
- E_{sn} = Voltage from conductors to neutral at sending end.
- I_r = Current in amperes per conductor at receiving end.
- I_s = Current in amperes per conductor at sending end.
- $\cos \theta_r$ = Power-factor at receiving end.
- $\cos \theta_s$ = Power-factor at sending end.

FOR ZERO LOAD CONDITIONS

The symbols corresponding to zero load conditions are as indicated above for load conditions with the addition of a sub zero.

THE FUNDAMENTAL OR LINEAR CONSTANTS

The fundamental, or "linear constants" of the circuit for each conductor per unit length are represented as follows:—

- r = Linear resistance in ohms per conductor mile (taken from Table II)
- x = Linear reactance in ohms per conductor mile (taken from Table IV or V)
- b = Linear capacitance susceptance to neutral in mhos per conductor mile (taken from Table IX or X)
- g = Linear leakage conductance to neutral in mhos per conductor mile. (This represents the direct escape of active power through the air between conductors and of active power leakage over the insulators. These losses must be estimated for conditions similar to those of the circuit under consideration. For all lines except those of great length and high voltage it is common practice to disregard the effects of leakage or corona loss and to take g as equal to zero.)
- z = Linear impedance = $\sqrt{r^2 + x^2}$
- y = Linear admittance = $\sqrt{g^2 + b^2}$

If the length of each conductor of the circuit in unit length is designated as l we have

- rl = Total resistance in ohms per conductor = R
- xl = Total reactance in ohms per conductor = X
- bl = Total susceptance in mhos per conductor to neutral = B
- gl = Total conductance in mhos per conductor to neutral = G

then,

$$Z = \sqrt{R^2 + X^2} \text{ ohms}$$

$$\text{and, } Y = \sqrt{G^2 + B^2} \text{ mhos}$$

IR = Voltage necessary to overcome the resistance.

IX = Voltage necessary to overcome the reactance.

IZ = Voltage necessary to overcome the impedance.

METHODS FOR DETERMINING THE CONSTANTS OF THE CIRCUIT

Several different methods for determining the fundamental constants of the circuit are in use. These methods are illustrated below.

Problem—Find the resistance volts IR and the reactance volts IX in percent of delivered volts E_r for the following conditions:—100 kw active power to be delivered at 1000 volts, three-phase, 60 cycles, over three No. 0000 stranded, hard drawn, copper conductors, circuit one mile long, with a symmetrical delta arrangement of conductors, two foot spacing, the temperature being taken as 25 degrees C.

Resistance of one mile of single conductor = 0.277 ohm (from Table II)

Reactance of one mile of single conductor = 0.595 ohm (from Table V)

Method No. 1—When three-phase circuits first came into use, it was customary (and correct), in determining the loss and voltage regulation, to consider them equivalent to two single-phase circuits, each single-phase circuit transmitting one-half the power of the three-phase system. This practice is still followed by some engineers; thus:—

$$\frac{50,000}{1000} = 50 \text{ amp. per conductor for each single-phase circuit.}$$

$$\frac{0.277 \times 2 \times 50}{1000} \times 100 = 2.77\% \text{ resistance volts drop of single-phase circuit.}$$

$$\frac{0.595 \times 2 \times 50}{1000} \times 100 = 5.95\% \text{ reactance volts drop of single-phase circuit.}$$

Method No. 2 consists of treating the case as a straight three-phase problem. Thus:

$$\frac{100,000}{1000 \times 1.732} = 57.73 \text{ amperes per conductor of three-phase circuit.}$$

$$\frac{0.277 \times 1.732 \times 57.73}{1000} \times 100 = 2.77\% \text{ resistance volts drop of three-phase circuit.}$$

$$\frac{0.595 \times 1.732 \times 57.73}{1000} \times 100 = 5.95\% \text{ reactance volts drop of three-phase circuit.}$$

Method No. 3 consists in assuming one-third the total power transmitted over one conductor with neutral or ground return (resistance and reactance of return being taken as zero). Such an equivalent circuit is shown by diagram in the upper part of Fig. 16. Thus the circuit constants for the above problem would be determined as follows:—

$$\text{Watts per phase} = \frac{100,000}{3} = 33,333 \text{ watts.}$$

$$\text{Volts to neutral} = 1000 \times 0.5774 \text{ or } 577.4 \text{ volts.}$$

$$\frac{33,333}{577.4} = 57.74 \text{ amperes per conductor; (same as for method No. 2)}$$

$$\frac{0.277 \times 57.74}{577.4} \times 100 = 2.77\% \text{ resistance volts drop of three-phase circuit.}$$

$$\frac{0.595 \times 57.74}{577.4} \times 100 = 5.95\% \text{ reactance volts drop of three-phase circuit.}$$

It will be seen that all three methods produce the same results. *Method No. 3* seems the most readily adaptable to various kinds of transmission systems and will be used exclusively in the treatment of the problems which will follow.

APPLICATION OF THE TABLES

Numerous tables of constants, charts, etc., have been presented, and a few more will follow. Chart II plainly indicates the application of these tables, etc. to the calculation of transmission circuits and the sequence in which they should be consulted.

GRAPHICAL VS. MATHEMATICAL SOLUTIONS

At the time of the design of a transmission circuit the actual maximum load or power-factor of the load that the circuit will be called upon to transmit is seldom known. An unforeseen development leading to an increased demand for electrical energy may result in a greatly increased load to be transmitted. The actual length of a circuit (especially when located in a hilly or rolling country) is never known with mathematical accuracy. Moreover, the actual resistance of the conductors varies to a large extent with temperature variations along the circuit.

When it is considered that there are so many indeterminate variables which vitally affect the performance of a transmission circuit, it would seem that a comparatively long and highly mathematical solution for determining the exact performance, necessarily based upon rigid assumptions, is hardly justified. In many cases the economic loss in transmission will determine the size of conductors and, if the circuit is very long, synchronous machinery is likely to be employed for controlling the voltage.

Mathematical solutions have one very important virtue, in that they provide an entirely different but parallel route in the solution of such problems, and therefore are valuable as a check against serious errors in the results obtained by the more simple graphical solutions.

In the following treatment, simple but highly accurate graphical solutions will be first presented, for determining the performance not only of short transmission lines, but also for long lines. For short lines the Dwight and the Mershon charts will be used. For long lines, where the effect of capacitance must be accurately accounted for, the Wilkinson Charts, supplemented with vector diagrams will be used. These three forms of graphical solutions will, when correctly applied to any power transmission problem, produce results in which the error will be much less than that due to irregularities in line construction and inaccurate assumptions of circuit constants. These three graphical solutions will in each case be followed by mathematical solutions. In the case of short lines the usual formulas employing trigonometric functions will be employed, and in the case of long lines the convergent series, and two different forms of hyperbolic solutions will be employed.

GRAPHICAL SOLUTION

When the receiving end load conditions, that is, the voltage, the load and the power-factor are known, the IR volts required to overcome the resistance and the IX volts required to overcome the reactance of the circuit, may be readily calculated.

On a piece of plain paper or cross-section paper divided into tenths, a vector diagram of the current and of the various voltage drops of the circuit may be laid out to a convenient scale. Whichever kind of paper is used, the procedure will be as in the following example.

Single-Phase Problem—Find the voltage at the sending end of a single-phase circuit 16 miles long, consisting of two stranded, hard drawn No. 0000 copper conductors spaced three feet apart. Temperatures taken as 25 degrees C. Load conditions at receiving end assumed as 4000 kv-a (3200 kw at 80 percent power-factor lagging) 20 000 volts, single-phase, 60 cycles.

$$Kv-a_{rn} = \frac{4000}{2} = 2000 \text{ kv-a to neutral.}$$

$$E_{rn} = \frac{20\,000}{2} = 10\,000 \text{ volts to neutral.}$$

$$I_r = \frac{2\,000\,000}{10\,000} = 200 \text{ amperes per conductor.}$$

The fundamental constants per conductor are:—

$$R = 16 \times 0.277 \text{ (from Table II)} = 4.432 \text{ ohms}$$

$$X = 16 \times 0.644 \text{ (from Table V)} = 10.304 \text{ ohms}$$

$$\text{and } IR = 200 \times 4.432 = 886 \text{ volts resistance drop}$$

$$= \frac{886}{10\,000} \times 100 = 8.86 \text{ percent}$$

$$IX = 200 \times 10.304 = 2061 \text{ volts reactance drop}$$

$$= \frac{2061}{10\,000} \times 100 = 20.61 \text{ percent}$$

Having determined the above values a vector diagram may be made as follows:—

Draw an arc quadrant having a radius of 10 000 (the receiving end voltage to neutral) to some convenient scale, as shown in Fig. 17. The radius which represents the base, or horizontal line will be assumed as representing the phase of the current at the receiving end of the circuit. Divide this base line into ten equal parts. These ten divisions will then correspond to loads of corresponding power-factors. Since a load has been assumed having a power-factor of 80 percent lagging, draw a vertical line from the 0.8 division on the base line, until it intersects the arc of the circle. From this point of intersection draw a line to the right and parallel with the base line. To the same scale as that plotted for the receiver voltage (10 000) measure off to the right 886 volts to D . This is the voltage which, as determined above is required to overcome the resistance of one conductor of the circuit. It is sometimes stated as the voltage consumed by the line resistance. It will be noted that this voltage drop is in phase with the current at the receiving end. From this point lay off vertically, and to the same scale, 2061 volts which is, as determined above, the volts necessary to overcome the reactance of one conductor of the circuit. This is sometimes stated as the voltage consumed by the line reactance. Connect this last point by a straight

CHART II.—APPLICATION OF TABLES TO SHORT TRANSMISSION LINES (EFFECT OF CAPACITANCE NOT TAKEN INTO ACCOUNT) OVER HEAD BARE CONDUCTORS

Starting with the kv-a., voltage and power-factor at the receiving end known.

QUICK ESTIMATING TABLES XII TO XXI INC.

From the quick estimating table corresponding to the voltage to be delivered, determine the size of the conductors corresponding to the permissible transmission loss.

HEATING LIMITATION—TABLE XXIII

If the distance of transmission is short and the amount of power transmitted very large there is a possibility of overheating the conductors—to guard against such overheating the carrying capacity of the conductors contemplated should be checked by this table.

CORONA LIMITATION—TABLE XXII

If the transmission is at 30 000 volts, or higher, this table should be consulted to avoid the employment of conductors having diameters so small as to result in excessive corona loss.

RESISTANCE—TABLES I AND II

From one of these tables obtain the resistance per unit length of single conductor corresponding to the maximum operating temperature—calculate the total resistance for one conductor of the circuit—if the conductor is large (250 000 circ. mils or more) the increase in resistance due to skin effect should be added.

IR TRANSMISSION LOSS

Calculate the IR loss of one conductor by multiplying its total resistance by the square of the current—to obtain the total loss multiply this result by the number of conductors of the circuit.

REACTANCE—TABLES IV AND V

From one of these tables obtain the reactance per unit length of single conductor. Calculate the total reactance for one conductor of the circuit. If the reactance is excessive (20 to 30 percent reactance volts will in many cases be considered excessive) consult Table VI or VII. Having decided upon the maximum permissible reactance the corresponding resistance may be found by dividing this reactance by the ratio value in Table VI or VII. When the reactance is excessive, it may be reduced by installing two or more circuits and connecting them in parallel, or by the employment of three conductor cables. Using larger conductors will not materially reduce the reactance. The substitution of a higher transmission voltage, with its correspondingly less current, will also result in less reactance.

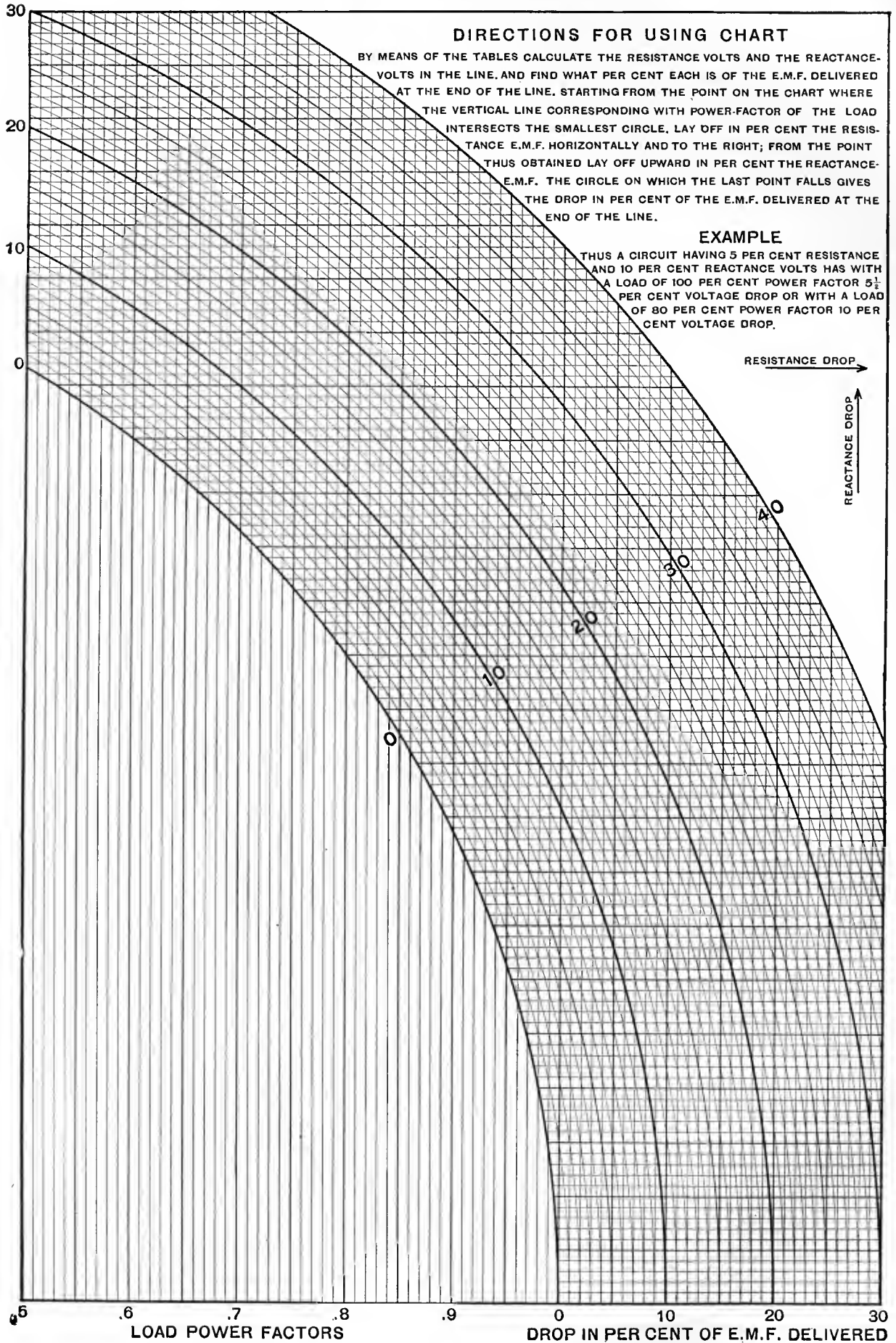
GRAPHICAL SOLUTION

A simple graphical solution, as described in the text, may be made by which the kv-a., the voltage and the power-factor at the sending end of the circuit may be determined graphically. Or the voltage at the sending end may be determined graphically by the use of either the Dwight or the Mershon chart. With the Mershon chart the power-factor at the sending end may be read directly from the chart.

MATHEMATICAL SOLUTION

As a precaution against errors the results obtained graphically should be checked by a mathematical solution, in cases where accuracy is essential.

CHART III-MERSHON CHART



cent of E_r , the impedance triangle is traced upon the chart and the voltage drop in percentage of E_r is read directly as indicated by the directions. All values on the chart are expressed in percent of the receiving end voltage.

Single-Phase Problem—Taking the resistance voltage as 8.86 percent and the reactance voltages 20.61 percent of the receiving end voltage, for the above single-phase problem, (Fig. 17) and tracing these values upon the Mershon Chart for a receiving end load of 80 percent power-factor lagging, the voltage drop is determined as 19.9 percent. The calculated value being 19.98 percent, the error by the chart is seen to be negligible.

WHEN THE SENDING END CONDITIONS ARE FIXED

When the conditions at the sending end are fixed and those at the receiving end are to be determined, the solving of the problem by the Mershon Chart is more complicated. In such cases, it is usual to estimate what the probable receiving end condition will be. From these estimated receiving end conditions, determine by the chart the corresponding sending end conditions. If the conditions as determined by this assumption are materially different from the known conditions, another assumption should be made. The corresponding sending end conditions should then be checked with the known conditions. Several such trials will usually be necessary to solve such problems.

GRAPHICAL SOLUTION BY THE DWIGHT CHART

Mr. H. B. Dwight has worked up a straight line chart, shown as Chart IV, in which the resistance and the reactance of the circuit have been taken into account through the medium of spacing lines marked for various sizes of conductors.* The use of this chart does not, therefore, require the calculation of the resistance and reactance or the use of tables of such constants. The Dwight Chart is also constructed so as to be applicable to loads of leading as well as to loads of lagging power-factors, whereas the Mershon chart, as generally constructed, is applicable to loads of lagging power-factor only. However the Mershon Chart can be made applicable for the solving of problems of leading as well as lagging power-factor loads by extending it through the lower right-hand quadrant. The application of synchronous condensers frequently gives rise to loads of leading power-factor. The Dwight Chart is well adapted to the solution of such circuits. Still another feature of this chart is that formulas are given which take capacitance effect into account with sufficient accuracy for circuits with a length up to approximately 100 miles.

Single-Phase Problem—Find the voltage at the sending end of a single-phase circuit 16 miles long, consisting of two stranded, hard-drawn, No. 0000 copper conductors, spaced three feet apart. Temperature

taken as 25 degrees C. Load condition at receiving end assumed as 4000 kv-a (3200 kw at 80 percent power-factor lagging) 20 000 volts single-phase, 60 cycles.

From Table II the resistance of No. 0000 stranded, hard-drawn, copper conductors at 25 degrees C. is found to be 0.277 ohm per wire per mile. Lay a straight edge across the Dwight Chart from the resistance value per mile 0.277 (as read on the lower half of the vertical line to the extreme right) to the spacing of three feet for copper conductors and 60 cycles at the extreme left. Along this straight edge read factor $V = 0.62$, corresponding to a lagging power-factor of 80 percent. This factor V is equivalent to the change in receiving end voltage per total ampere per mile of circuit, due to the line impedance.

It will be noted that opposite the resistance values (extreme right vertical line) is placed the corresponding sizes of copper and aluminum conductors on the basis of a temperature of 20 degrees C. If the temperature is assumed to be 20 degrees C. it will not be necessary to consult a table of resistance values. In such a case, the straight edge would simply be placed over the division of the vertical resistance line corresponding to the size and material of conductors. Marking a resistance value on this vertical line makes the chart adaptable to resistance values corresponding to conductors at any temperature. Had the power factor been leading, in place of lagging, the corresponding resistance point would have been located on the upper half of the vertical resistance line.

Continuing following the directions on the chart for short lines, we obtain the following. Since the circuit is single-phase, use $2V = 1.24$

$$\text{Voltage drop in percent of } E_r = \frac{100\,000 \times 4000 \times 16 \times 1.24}{20\,000^2} = 19.84 \text{ percent}$$

The voltage drop, as calculated mathematically, is 19.98 percent representing an error of 0.14 percent by the chart.

Three-Phase Problem (No. 33)—Find the voltage at the sending end of a three-phase circuit, 20 miles long, consisting of three No. 0000 stranded, hard-drawn, copper conductors, spaced three feet apart in a delta arrangement. Temperature taken as 25 degrees C. Load conditions at receiving end assumed as 1300 kv-a (1040 kw at 80 percent power-factor lagging) 10 000 volts, three-phase, 60 cycles.

From Table II, the resistance per wire per mile is again found to be 0.277 ohm and since the spacing and frequency are both the same as in the case of the above single-phase problem, we again obtain $V = 0.62$. The voltage drop in percent of E_r is therefore

$$\frac{100\,000 \times 1300 \times 20 \times 0.62}{10\,000^2} = 16.12 \text{ percent}$$

The voltage drop as calculated mathematically is 16.16 percent, representing an error of 0.04 percent.

CAPACITANCE

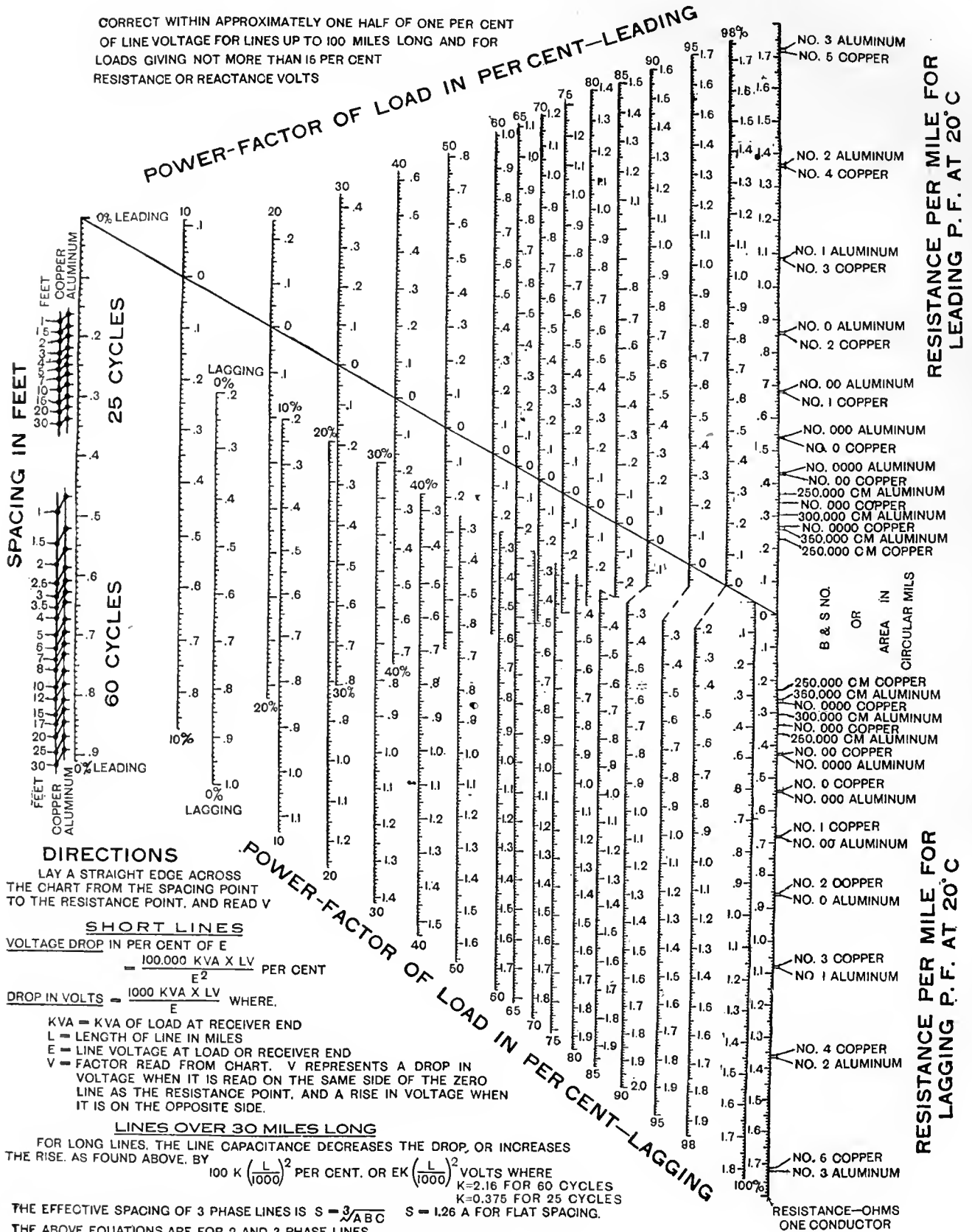
In long circuits the effect of capacitance is to decrease the voltage drop, or increase the voltage rise, as

*The basis of the construction of this chart is described in the JOURNAL for July, 1915, p. 306.

CHART—IV DWIGHT CHART

FOR DETERMINING THE VOLTAGE REGULATION OF TRANSMISSION CIRCUITS CONTAINING CAPACITANCE

CORRECT WITHIN APPROXIMATELY ONE HALF OF ONE PER CENT OF LINE VOLTAGE FOR LINES UP TO 100 MILES LONG AND FOR LOADS GIVING NOT MORE THAN 16 PER CENT RESISTANCE OR REACTANCE VOLTS



will be explained later. The Dwight and Mershon charts do not recognize the effect which capacitance has upon the voltage drop. In the lower left hand corner of the Dwight Chart, however, there is placed a formula by which a correction may be applied to the voltage drop as given by the chart. This correction accounts for the effect of the charging current (resulting from capacitance) quite accurately, provided the circuit is not too long or the frequency too high. The application of this corrective factor will be evident from the following problem.

TABLE K—COSINES, SINES AND TANGENTS

ANGLE	COS θ (P F)	SIN θ	TAN θ
0° 00'	1.000	0.0000	0.0000
8° 06'	0.990	0.1409	0.1423
11° 28'	0.980	0.1988	0.2028
14° 04'	0.970	0.2430	0.2506
16° 15'	0.960	0.2798	0.2915
18° 11'	0.950	0.3120	0.3285
19° 56'	0.940	0.3410	0.3627
21° 33'	0.930	0.3673	0.3949
23° 04'	0.920	0.3918	0.4258
24° 29'	0.910	0.4144	0.4554
25° 50'	0.900	0.4357	0.4841
27° 07'	0.890	0.4558	0.5121
28° 21'	0.880	0.4748	0.5396
29° 32'	0.870	0.4929	0.5665
30° 41'	0.860	0.5103	0.5934
31° 47'	0.850	0.5267	0.6196
32° 51'	0.840	0.5424	0.6457
33° 54'	0.830	0.5577	0.6720
34° 54'	0.820	0.5721	0.6976
35° 54'	0.810	0.5864	0.7239
36° 52'	0.800	0.6000	0.7499
37° 48'	0.790	0.6129	0.7757
38° 44'	0.780	0.6257	0.8021
39° 38'	0.770	0.6379	0.8283
40° 32'	0.760	0.6499	0.8551
41° 24'	0.750	0.6613	0.8816
42° 16'	0.740	0.6726	0.9089
43° 06'	0.730	0.6833	0.9358
43° 56'	0.720	0.6938	0.9634
44° 45'	0.710	0.7040	0.9913
45° 34'	0.700	0.7141	1.0199
46° 22'	0.690	0.7238	1.0489
47° 09'	0.680	0.7331	1.0780
47° 55'	0.670	0.7422	1.1074
48° 42'	0.660	0.7513	1.1383
49° 27'	0.650	0.7598	1.1688
50° 12'	0.640	0.7683	1.2002
50° 57'	0.630	0.7766	1.2327
51° 41'	0.620	0.7846	1.2655
52° 24'	0.610	0.7923	1.2985
53° 07'	0.600	0.8000	1.3327
53° 50'	0.590	0.8073	1.3680
54° 32'	0.580	0.8145	1.4037
55° 14'	0.570	0.8215	1.4406
55° 56'	0.560	0.8284	1.4788
56° 37'	0.550	0.8350	1.5175
57° 18'	0.540	0.8415	1.5577
57° 59'	0.530	0.8479	1.5993
58° 40'	0.520	0.8542	1.6426
59° 20'	0.510	0.8601	1.6864
60° 00'	0.500	0.8660	1.7320
60° 39'	0.490	0.8716	1.7783
61° 18'	0.480	0.8771	1.8265
61° 57'	0.470	0.8825	1.8768

Three-Phase Problem (No. 45)—Find the voltage at the sending end of a three-phase circuit, 100 miles long, consisting of three No. 0000, stranded, hard-drawn copper conductors, spaced nine feet apart in a delta arrangement. Temperature assumed as 25 degrees C. Load conditions at receiving end assumed as 22 000 kv-a, 80 percent power-factor lagging, 88 000 volts, 60 cycles.

From Table II the resistance is found to be 0.277 ohm per mile. From Dwight Chart read $V = 0.70$. Then, the voltage drop in percent of E_r , if the line were short, would be,

$$\frac{100\,000 \times 22\,000 \times 100 \times 0.70}{88\,000^2} = 19.89 \text{ percent}$$

From directions on the Dwight chart for circuits over 30 miles long, the charging current of this circuit is found to be such as to decrease the voltage drop under load conditions or to increase the voltage at zero load by the amount of $100 \times 2.16 \left(\frac{100}{1000}\right)^2 = 2.16$ percent. Hence the voltage at the sending end, under load conditions, will be $19.89 - 2.16 = 17.73$ percent. The actual result as calculated rigorously is 17.94 percent. Thus the error by the Dwight graphical solution is approximately 0.21 percent.

If the power-factor of the load is assumed as 100 percent (problem 46) in place of 80 percent lagging, we get $V = 0.33$ and find the error for the Dwight graphical solution of this 100 mile, 60 cycle circuit to be approximately 0.75 percent. It should be noted, however, that the reactance volts are in this case 22 percent of the receiving end voltage.

SENDING END CONDITIONS FIXED

When the sending end conditions are fixed, a different form of solution must be employed to determine the size of conductors corresponding to a given voltage drop. In such cases, the Dwight Chart is particularly applicable. To use the chart for the solution of such problems proceed as follows. First V is calculated by means of the formulas on the chart, and then a straight edge is placed through V (on the line corresponding to the power-factor of the load) and the point for the spacing and frequency to be used, and the required size of conductor can be seen at a glance on the resistance scale at the right. To make this application of the chart clear, the following is given,—

$$\text{Voltage drop in percent of } E_r = \frac{100\,000 K_{v-a} \times L \cdot I'}{E_r^2} \quad (28)$$

Hence

$$V = \frac{\text{Voltage drop in percent of } E_r \times E_r^2}{100\,000 K_{v-a} \times L} \dots\dots\dots (29)$$

Applying (29) to the above problem No. 33 we get

$$V = \frac{16.12 \times 10\,000^2}{100\,000 \times 1300 \times 20} = 0.62$$

Following the above directions, the resistance per mile is found to be 0.277 ohm and the corresponding size of conductor No. 0000 copper.

MATHEMATICAL SOLUTION

In order to check any one, or all of the above described graphical methods, a complete mathematical solution may be made by applying the various trigonometrical formulas, Fig. 18, to the values of the problem under consideration. These formulas have been arranged to meet the conditions of loads of either lagging or leading power-factors, and for conditions fixed at either the receiving or the sending ends.

There are numerous problems requiring a solution

where the voltage at the sending end, and the kilowatts and the power-factor of the load at the receiving end are fixed. In such cases it is required to determine the corresponding receiving end voltage. This determination can be made mathematically, but such a solution is tedious, since the formulas applying to such cases are cumbersome. Formulas are given at the bottom of Fig. 18 which may be applied to such problems. Time and labor may, however, be saved in solving such problems by the employment of a cut-and-try method usually used in such cases, as follows:—

Assume what the voltage drop will be, corresponding to the size of conductors likely to be used. On the basis of this assumption the receiving end voltage is fixed; thus, all of the receiving end conditions are assumed to be fixed. The corresponding sending end voltage is then readily determined by one of the graphical methods described. If the sending end voltage thus determined is found to be materially different from the fixed sending end voltage, another trial, based upon a different receiving end voltage, will probably suffice.

Single-Phase Problem—Find the characteristics of the load at the sending end of a single-phase circuit, 16 miles long, consisting of two stranded, hard drawn, copper conductors, spaced three feet apart; temperature taken as 25 degrees C.; load conditions at receiving end assumed as 4000 kv-a (3200 kw at 80 percent power-factor lagging) 20 000 volts, 60 cycles; transmission loss to be approximately ten percent.

Following the procedure given in Chart II, consult Quick Estimating Table XVII for a delivered voltage of 20 000. Since the conditions of the above problem are a power-factor of 80 percent, and a temperature 25 degrees C, the corresponding kv-a values are as indicated at the head of the table on the basis of 10.8 percent loss in transmission for a three-phase circuit. For a single-phase circuit the corresponding values will be one-half the table values. Thus the 4 000 kv-a single phase circuit of the problem is equivalent to 8000 kv-a, three-phase on the table. From the table, it is seen that for a distance of 16 miles 7810 kv-a, three-phase can be transmitted over No. 0000 conductors with a loss of 10.8 percent. 7810 kv-a is near enough to 8000 kv-a, and the loss of 10.8 percent is near enough to an assumed loss of ten percent, so we decide that No. 0000 copper conductors come nearest to the proper size to meet the conditions of the problem. The loss with No. 0000 conductors will be $\frac{8000}{7810} \times 10.8 = 11.06$ percent, as will be shown later.

Table XXIII indicates that there will be no overheating of this size of conductor.

Table XXII indicates that 20 000 volts is too low to result in corona loss with No. 0000 conductors, at any reasonable altitude. Then,—

$$Kv-a_{rn} = \frac{4000}{2} = 2000 \text{ kv-a to neutral.}$$

$$Kw_{rn} = \frac{3200}{2} = 1600 \text{ kw to neutral.}$$

$$E_{rn} = \frac{20\,000}{2} = 10\,000 \text{ volts to neutral.}$$

$$I_r = \frac{2\,000\,000}{10\,000} = 200 \text{ amperes per conductor.}$$

The resistance per conductor is

$$R = 16 \times 0.277 \text{ (from Table II)} = 4.432 \text{ ohms.}$$

The reactance per conductor is

$$X = 16 \times 0.644 \text{ (from Table V)} = 10.304 \text{ ohms.}$$

$$\text{and } IR = 200 \times 4.432 = 866 \text{ volts, resistance drop}$$

$$= \frac{866}{10\,000} \times 100 = 8.86 \text{ percent}$$

$$IX = 200 \times 10.304 = 2061 \text{ volts, reactance drop}$$

$$= \frac{2061}{10\,000} \times 100 = 20.61 \text{ percent}$$

$$E_{sn} = \sqrt{(10\,000 \times 0.8 + 866)^2 + (10\,000 \times 0.6 + 2061)^2} = 11\,998 \text{ volts to neutral} \dots\dots\dots (30)$$

$$\theta_s = \tan^{-1} \left(\frac{10\,000 \times 0.6 + 2061}{10\,000 \times 0.8 + 866} \right) = 42^\circ 13' \dots\dots (31)$$

$$\text{Percent } PF_s = (\text{Cos. } 42^\circ 13') \times 100 = 74.06 \text{ percent} \dots\dots (32)$$

$$Kv-a_{sn} = \frac{200 \times 11\,998}{1000} = 2399.6 \text{ kv-a per conductor} \dots\dots (33)$$

$$Kw_{sn} = 2399.6 \times 0.7406 = 1777.1 \text{ kw per conductor} \dots\dots (34)$$

$$\text{Percent voltage drop} = \frac{11\,998 - 10\,000}{10\,000} \times 100 = 19.98 \text{ percent} \dots\dots\dots (46)$$

$$\text{Transmission loss} = \frac{(200)^2 \times 4.432}{1000} = 177.28 \text{ kw per conductor} \dots\dots\dots (47)$$

$$\text{Percent transmission loss} = \frac{177.28 \times 2}{3200} \times 100 = 11.08 \text{ percent} \dots\dots\dots (48)$$

Three-Phase Problem (No. 33)—Find the characteristics of the load at the sending end of a three-phase circuit 20 miles long, consisting of three stranded, hard-drawn, copper conductors, spaced in a three foot delta. Temperature taken as 25 degrees C. Load conditions at receiving end assumed as 1300 kv-a. (1040 kw at 80 percent power-factor lagging) 10 000 volts, 60 cycles; transmission loss not to exceed ten percent.

Following the procedure given in Chart II, the following results are obtained:—

Consult Table XV for a delivered voltage of 10 000 volts. Since the conditions of the above problems are, power-factor of load 80 percent, temperature 25 degrees C. the corresponding three-phase kv-a values of the table are on the basis of 10.8 percent loss in transmission. From Table XV it is seen that 1240 kv-a, three-phase can be transmitted over No. 000 conductors, or 1560 kv-a., three-phase over No. 0000 conductors at 10.8 percent loss. Since the loss for the problem is not to exceed ten percent and 1300 kv-a is to be transmitted, we will select No. 0000 conductors. The loss for these conductors will therefore be $\frac{1300}{1560}$ of 10.8, or nine percent as will be shown later.

Table XXIII indicates that there will be no overheating of this size of conductor when carrying 1300 kv-a, three-phase.

Table XXII indicates that 10 000 volts is too low to result in corona loss with No. 0000 conductors at any reasonable altitude. Then:—

$$Kv-a_{rn} = \frac{1300}{3} = 433.33 \text{ kv-a to neutral.}$$

$$Kw_{rn} = \frac{1040}{3} = 346.6 \text{ kw to neutral.}$$

$$E_{ra} = \frac{10000}{1.732} = 5774 \text{ volts to neutral.}$$

$$I_r = \frac{433333}{5774} = 75.05 \text{ amperes per conductor.}$$

The resistance per conductor is,—

$$R = 20 \times 0.277 \text{ (from Table II)} = 5.54 \text{ ohms.}$$

The reactance per conductor is,—

$$X = 20 \times 0.644 \text{ (from Table V)} = 12.88 \text{ ohms.}$$

and

$$IR = 75.05 \times 5.54 = 415.8 \text{ volts, resistance drop.}$$

$$= \frac{415.8}{5774} \times 100 = 7.20 \text{ percent.}$$

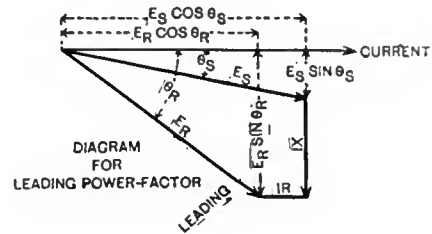
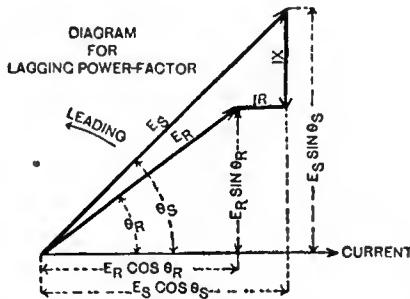
$$\text{Transmission loss} = \frac{(75.05^2) \times 5.54}{1000} = 31.20 \text{ kw per conductor}$$

..... (47)

$$\text{Percent transmission loss} = \frac{31.20 \times 3}{1040} \times 100 = 9.00 \text{ percent}$$

MIXED SENDING AND RECEIVING END CONDITIONS FIXED

Branch circuits are frequently run from the main transmission trunk circuit to the center of some local distribution. In such cases the voltage at the sending end and the current or the power and power-factor at



LOADS OF LAGGING POWER-FACTOR

LOADS OF LEADING POWER-FACTOR

WHEN RECEIVING-END CONDITIONS ARE FIXED

$$E_S = \sqrt{(E_R \cos \theta_R + IR)^2 + (E_R \sin \theta_R + IX)^2} \quad (30)$$

$$\theta_S = \tan^{-1} \left(\frac{E_R \sin \theta_R + IX}{E_R \cos \theta_R + IR} \right) \quad (31)$$

$$\% \text{ PF}_S = \cos \theta_S \times 100 \quad (32)$$

$$\text{KV-A}_{SN} = \frac{I \times E_{SN}}{1000} \text{ PER CONDUCTOR} \quad (33)$$

$$\text{KW}_{SN} = \text{KV-A}_{SN} \times \cos \theta_S \text{ PER CONDUCTOR} \quad (34)$$

WHEN RECEIVING-END CONDITIONS ARE FIXED

$$E_S = \sqrt{(E_R \cos \theta_R + IR)^2 + (E_R \sin \theta_R - IX)^2} \quad (40)$$

$$\theta_S = \tan^{-1} \left(\frac{E_R \sin \theta_R - IX}{E_R \cos \theta_R + IR} \right) \quad (41)$$

$$\% \text{ PF}_S = \cos \theta_S \times 100 \quad (32)$$

$$\text{KV-A}_{SN} = \frac{I \times E_{SN}}{1000} \text{ PER CONDUCTOR} \quad (33)$$

$$\text{KW}_{SN} = \text{KV-A}_{SN} \times \cos \theta_S \text{ PER CONDUCTOR} \quad (34)$$

WHEN SENDING-END CONDITIONS ARE FIXED

$$E_R = \sqrt{(E_S \cos \theta_S - IR)^2 + (E_S \sin \theta_S - IX)^2} \quad (35)$$

$$\theta_R = \tan^{-1} \left(\frac{E_S \sin \theta_S - IX}{E_S \cos \theta_S - IR} \right) \quad (36)$$

$$\% \text{ PF}_R = \cos \theta_R \times 100 \quad (37)$$

$$\text{KV-A}_{RN} = \frac{I \times E_{RN}}{1000} \text{ PER CONDUCTOR} \quad (38)$$

$$\text{KW}_{RN} = \text{KV-A}_{RN} \times \cos \theta_R \text{ PER CONDUCTOR} \quad (39)$$

WHEN SENDING-END CONDITIONS ARE FIXED

$$E_R = \sqrt{(E_S \cos \theta_S - IR)^2 + (E_S \sin \theta_S + IX)^2} \quad (42)$$

$$\theta_R = \tan^{-1} \left(\frac{E_S \sin \theta_S + IX}{E_S \cos \theta_S - IR} \right) \quad (43)$$

$$\% \text{ PF}_R = \cos \theta_R \times 100 \quad (37)$$

$$\text{KV-A}_{RN} = \frac{I \times E_{RN}}{1000} \text{ PER CONDUCTOR} \quad (38)$$

$$\text{KW}_{RN} = \text{KV-A}_{RN} \times \cos \theta_R \text{ PER CONDUCTOR} \quad (39)$$

GENERAL FORMULAS

WHEN THE VOLTAGE AT SENDING END AND THE AMPERES AND POWER-FACTOR AT RECEIVING END ARE FIXED

$$E_R = -I(R \cos \theta_R \pm X \sin \theta_R) + \sqrt{E_S^2 - I^2(R^2 \sin^2 \theta_R + X^2 \cos^2 \theta_R)} \pm 2I^2 R X \cos \theta_R \sin \theta_R \quad (44)$$

★ USE + WHEN THE POWER-FACTOR OF THE LOAD IS LAGGING AND - WHEN THE POWER-FACTOR IS LEADING

WHEN THE VOLTAGE AT SENDING END AND THE POWER AND POWER-FACTOR AT RECEIVING END ARE FIXED (POWER FACTOR LAGGING)

$$E_{RN} = A \sqrt{1 \pm \sqrt{1 - \frac{(R^2 + X^2) \text{KW}_{RN}^2 \times 10^8}{A^4 \cos^2 \theta_R}}} \quad \text{WHERE } A = E_{SN} \sqrt{\frac{1}{2} - \frac{1000 \text{KW}_{RN} (R \cos \theta_R + X \sin \theta_R)}{E_{SN}^2 \cos \theta_R}} \quad (45)$$

% VOLTAGE DROP = $\frac{E_S - E_R}{E_R} \times 100 \quad (46)$

TRANSMISSION LOSS = $\frac{I^2 R}{1000}$ KW PER CONDUCTOR (47)

% TRANSMISSION LOSS = $\frac{\text{TOTAL } I^2 R \text{ (IN KW)}}{\text{TOTAL KW}_R} \times 100 \quad (48)$

FIG. 18—TRIGONOMETRICAL FORMULAS FOR SHORT TRANSMISSION LINES
Capacitance effect not taken into account.

$$IX = 75.05 \times 12.88 = 966.6 \text{ volts, reactance drop.}$$

$$= \frac{966.6}{5774} \times 100 = 16.74 \text{ percent.}$$

$$E_{ra} = \sqrt{(5774 \times 0.8 + 415.8)^2 + (5774 \times 0.6 + 966.6)^2} = 6707 \text{ volts to neutral} \quad (30)$$

$$\theta_s = \tan^{-1} \left(\frac{5774 \times 0.6 + 966.6}{5774 \times 0.8 + 415.8} \right) = 41^\circ 22' \quad (31)$$

$$\text{PF}_s = (\cos 41^\circ 22') \times 100 = 75.05 \text{ percent} \quad (32)$$

$$\text{Kv-a}_{sa} = \frac{75.05 \times 6707}{1000} = 503.4 \text{ kv-a per conductor.} \quad (33)$$

$$\text{Kw}_{sa} = 503.4 \times 0.7505 = 377.8 \text{ kw per conductor} \quad (34)$$

$$\text{Percent voltage drop} = \frac{6707 - 5774}{5774} \times 100 = 16.16 \text{ percent} \quad (46)$$

the receiving end are approximately fixed. In such cases the calculation for the voltage at the receiving end requires more arithmetical work than is required when all the conditions at one end of the circuit are fixed. Such problems can be more readily solved graphically, as previously explained, but may be solved mathematically by applying formula (44) or (45), Fig. 18.

To illustrate the application of formula (44) we will apply the values of Problem 33 to formula (44) and calculate the receiving end voltage. Thus we have as fixed conditions:—

$$\begin{aligned}
 E_{sn} &= 6707 \text{ volts} \\
 I_r &= 75.05 \text{ amperes} \\
 \cos \theta_r &= 0.8 \\
 \sin \theta_r &= 0.6 \\
 R &= 5.54 \text{ ohms} \\
 X &= 12.88 \text{ ohms} \\
 IR &= 415.8 \text{ volts}
 \end{aligned}$$

Then

$$\begin{aligned}
 E_r &= -75.05 (5.54 \times 0.8 + 12.88 \times 0.6) + \\
 &\frac{\sqrt{6707^2 - 75.05^2 (5.54^2 \times 0.6^2 + 12.88^2 \times 0.8^2) + 2 \times}}{75.05^2 \times 5.54 \times 12.88 \times 0.8 \times 0.6 \dots \dots \dots (44)} \\
 &= -913 + \sqrt{44983849 - 660242 + 385831} \\
 &= -913 + 6637 = 5774 \text{ volts.}
 \end{aligned}$$

To illustrate the application of formula (45) we will apply the values of Problem 33 to formula (45)

TABLE L
ILLUSTRATING VARIATION IN REACTANCE

Resulting from Changes in the Conductors and Transmission Voltages

CONDUCTORS	Total I ² R Loss (KW)	IR		IX		Approximate Voltage Regulation at	
		Volts	Per Cent.	Volts	Per Cent.	100 Per Cent. Power Factor	80 Per Cent. Power Factor (Lag.)
RECEIVING END VOLTAGE — 6600							
Single Circuit of three 500,000 circ. mil bare overhead conductors	129	123	3.22	622	16.32	4.5	12.8
Two circuits each of three 250,000 circ. mil bare overhead conductors.	129	123	8.22	833	8.73	3.6	7.7
One Circuit of 600,000 circ. mil three-conductor cable. Insulation thickness $\frac{11}{16}$ by $\frac{11}{16}$ inches.	129	1.23	3.22	172	4.52	3.2	5.0
RECEIVING END VOLTAGE — 13 200							
Single circuit of three 125,000 circ. mil bare overhead conductors.	129	247	3.22	354	4.64	3.2	5.1

and calculate the receiving end voltage. Thus we have as fixed conditions:—

$$\begin{aligned}
 E_{sn} &= 6707 \text{ volts} \\
 Kw_{rn} &= 346.6 \text{ kw} \\
 R &= 5.54 \text{ ohms} \\
 X &= 12.88 \text{ ohms} \\
 \cos \theta_r &= 0.8 \\
 \sin \theta_r &= 0.6
 \end{aligned}$$

$$A = 6707 \sqrt{0.5 - \frac{1000 \times 346.6 (5.54 \times 0.8 + 12.88 \times 0.6)}{6707^2 \times 0.8}}$$

$$\begin{aligned}
 E_{rn} &= A \sqrt{1 + \sqrt{1 - \frac{(5.54^2 + 12.88^2) 346.6^2 \times 10^6}{A^4 \times 0.8^2}}} \dots (45) \\
 A &= 6707 \sqrt{0.5 - 0.1172} = 4152 \\
 E_{rn} &= 4152 \sqrt{1 + 0.936} = 5774 \text{ volts}
 \end{aligned}$$

Alternative to (44) and (45)—The following formulas have been proposed by Mr. H. B. Dwight to meet the mixed conditions referred to,—

$$\begin{aligned}
 E_{sn} &= 6707 \text{ volts} \\
 1000 \times Kw_{rn} &= 346\,600 \text{ watts} \\
 1000 \times \text{reactive } Kv\text{-}a_{rn} &= 346\,600 \times \frac{0.6}{0.8} = 260\,000 \text{ v-a} \\
 R &= 5.54 \text{ ohms} \\
 X &= 12.88 \text{ ohms} \\
 L &= 346\,600 \times 5.54 + 260\,000 \times 12.88 = 5\,270\,000 \\
 M &= 346\,600 \times 12.88 - 260\,000 \times 5.54 = 3\,025\,000
 \end{aligned}$$

$$\begin{aligned}
 E^2 &= 0.5 E_s^2 - L + 0.5 \sqrt{E_s^4 - 4 E_s^2 I_r - 4 M^2} \\
 E &= 5774 \text{ volts}
 \end{aligned}$$

or

$$\begin{aligned}
 E &= E_s - \frac{L}{E_s} - \frac{L^2}{E_s^3} - \frac{M^2}{2 E_s^3} - \frac{2L^3}{E_s^5} - \frac{3}{2} \frac{LM^2}{E_s^5} - \frac{5L^4}{E_s^7} \\
 &\quad - \frac{5L^2 M^2}{E_s^7} - \frac{5}{8} \frac{M^4}{E_s^7} \\
 E &= 5779 \text{ volts}
 \end{aligned}$$

CIRCUITS OF EXCESSIVE REACTANCE

If a large amount of power is to be transmitted at comparatively low voltage, particularly if the frequency is high, the reactance of the circuit will be high compared with its resistance. If the reactance is excessive (20 to 30 percent reactance volts may in some cases be considered excessive), the voltage regulation of the circuit may be seriously impaired.

As will be seen by consulting Tables VI and VII, there is a fixed relation between the resistance and the reactance of a circuit for a given frequency, size and spacing of conductors. This ratio is 2.4 times greater for 60 cycle than it is for 25 cycle circuits. For a given size of conductor the reactance can be varied only slightly by changing the spacing of overhead bare conductors. Substituting a larger or smaller conductor may change the resistance materially, but this will have little effect upon the reactance.

The reactance may be reduced by either or all of the following methods. The circuit may be split up into two or more circuits employing smaller conductors and these circuits connected in parallel. The voltage may be raised, if the installation is new, and smaller conductors employed; or the overhead conductors may be replaced by three conductor cables. To illustrate the above methods, the following problem has been assumed and the results tabulated.

A HIGH REACTANCE PROBLEM

Table L refers to the following problem—4000 kv-a, three-phase, 60 cycles, is to be delivered a distance of three miles over hard-drawn, stranded copper conductors. The I²R loss is to remain at 129 kw. The spacing of the overhead conductors assumed as 3 by 3 by 3 ft. Temperature 25 degrees C.

It is evident from Table L that if two three-phase circuits, each consisting of three 250 000 circ. mil. conductors are installed in place of one three-phase circuit, consisting of three 500 000 circ. mil. conductors, the reactance will be reduced by nearly one half, and a corresponding improvement in the voltage drop or regulation will occur, particularly if the load power-factor is 80 percent lagging. A further improvement along this line will be obtained if a single three-conductor cable is employed. Doubling the voltage for the overhead circuit and employing three 125 000 circ. mil. conductors results in practically as good performance in voltage regulation as for the 6600 volt three-conductor cable.

*See article by Mr. H. B. Dwight on "Effect of a Tie Line between Two Substations" in the *Electrical Review*, Dec. 21, 1918, p. 966. The formulas given in this article make complete allowance for the effect of capacitance and are very similar to the above.

CHAPTER VIII

PERFORMANCE OF LONG TRANSMISSION LINES

(GRAPHICAL SOLUTION)

THE E.M.F. of self-induction in a transmission circuit may either add to or subtract from the impressed voltage at the sending end, depending upon the relative phase relations between the current and the voltage at the receiving end of the circuit. This is illustrated by means of voltage vectors in Fig. 20, in which the phase of the current is assumed to be constant in the horizontal direction indicated by the arrow on the end of the current vector. The voltage at the receiving end is also assumed as constant at 100 volts. The vector representing the receiving end voltage ($E_r = 100$ volts) is shown in two positions corresponding to leading current, two positions corresponding to lagging current and in one position corresponding to unity power-factor. The components IR and IX of the supply voltage necessary to overcome the resistance R and the reactance X (e.m.f. of self-induction) of the circuit are assumed to be 10 volts and 20 volts respectively. Since the current is assumed as constant, IX and IR are also constant. The impedance triangle of the voltage components required to overcome the combined effect of the resistance and the reactance of this circuit is therefore constant. It is shown in five different positions about the semicircle, corresponding to five different load power-factors. The voltage E_s at the sending-end required to maintain 100 volts at the receiving-end is indicated for each of the five positions of the impedance triangle.

Counter-clockwise rotation of the vectors will be considered as positive. This means that when the current is lagging behind the impressed e.m.f., the voltage vector will be in the forward or leading direction from the current vector as indicated by the arrow. When the current leads the impressed voltage, the voltage vector will be in the opposite, or clockwise direction from the current vector. In other words, assuming the vectors all rotating at the same speed about the point O in a counter-clockwise direction, the current vector will be behind the voltage vector when the current is lagging and ahead of it when the current is leading.

The alternating magnetic flux surrounding the conductors, resulting from current flowing through them, generates in them a counter e.m.f. of self-induction. This e.m.f. of self-induction has its maximum value when the current is passing through zero and is therefore in lagging quadrature with the current. On the diagrams an arrow in the line IX , indicates the direction of the e.m.f. of self-induction. It will be seen that since the direction of the current is assumed constant, the e.m.f. of self-induction acts downward in all

five impedance diagrams. The sending-end voltage is therefore opposed or favored by this self-induced voltage (see arrows) to a greater or less extent depending upon the power-factor of the load. Thus at lagging loads of high power-factor, the self-induced voltage acts approximately at right angles to the sending-end voltage, and therefore requires a small component of the sending-end voltage to balance or neutralize its effect. As the power-factor of the receiving-end load decreases in the lagging direction (upper quadrant of diagram) the sending-end voltage swings around more nearly in line with the direction of the induced voltage, thus requiring a greater component of the sending-end voltage to counter-balance its effect. At zero power-factor lagging, the direction of the sending-end voltage and that of the induced e.m.f. are practically in opposition, (as indicated by the arrows), so that the component of the sending-end voltage required to overcome the induced voltage is a maximum, or nearly as much as the e.m.f. of self-induction. It is interesting to note that at zero lagging power-factor, when the effect of self-induction on line voltage drop reaches a maximum, the sending-end voltage component IR necessary to overcome the resistance of the circuit, (now nearly at right angles to the supply voltage), is a minimum. The reverse of these conditions is true for receiving-end loads of power-factors near unity.

Now consider receiving-end loads of leading power-factors, (lower quadrant of diagram). It will be seen that the e.m.f. of self-induction does not now oppose the sending-end voltage (indicated by direction of the arrows) but has a direction more or less parallel to that of the sending-end voltage. At high leading power-factors, the e.m.f. of self-induction has little effect on the sending-end voltage, but as zero leading power-factor is approached these two e.m.f.'s more nearly come in phase with each other. At zero power-factor leading, the e.m.f. of self-induction adds almost directly to the sending-end voltage.

It will be seen, therefore, that for receiving-end loads of lagging power-factor, the sending-end voltage is greater than the receiving-end voltage, by an amount necessary to overcome the resistance and self-induction of the circuit. For receiving-end loads of leading power-factor, the sending-end voltage is less than the receiving-end voltage, for the reason that the e.m.f. of self-induction is in such a position as to assist the sending-end voltage.

The following values from Fig. 20 illustrate these conditions:

Power-Factor of Receiving End Load	Supply Voltage
0 percent lagging	120.4
80 percent lagging	120.4
100 percent	111.8
80 percent leading	98.5
0 percent leading	80.6

The condition of leading power-factor at the receiving-end would be unusual in practice, since the power-factor of receiving-end loads is usually lagging. In cases, however, where condensers are used for voltage or power-factor control, the power-factor at the receiving-end may be leading. If the circuit were without inductance, there could be no rise in voltage at the

degrees behind it and the other as the result of the line charging current and lagging 90 degrees behind it. These two combine at an angle, with each other and with the impressed e.m.f. at the sending-end.

CHARGING CURRENT

Conductors of a circuit, being separated by a dielectric (such as air, in overhead circuits, or insulation in cables), form a condenser. When alternating-current flows through such a circuit, current (known as charging current) virtually passes from one conductor through the dielectric to the other conductors, which are at a different potential. This current is in shunt with the circuit, and differs from the current which passes between conductors over the insulators etc. (leakage current) or through the air (corona effect) only in that the charging current leads the voltage by 90 degrees, whereas the leakage current is in phase with the voltage.

For a given spacing of conductors, the charging current increases with the voltage, the frequency and the length of the circuit. For long high-voltage circuits, particularly at 60 cycles per second, the charging current may be as much as the full-load current of the circuit, or more. In some cases of long 60 cycle circuits, where a comparatively small amount of power is to be transmitted, it is necessary to limit the voltage of transmission, in order that the charging current may not be so great as to overload the generators. This charging current, being in leading quadrature with the voltage, represents nearly all reactive power, but it is just as effective in heating the generator windings as if it represented active power. On the other hand, it combines with the receiving-end current at an angle (depending upon the power-factor of the receiver load) in such a manner that the addition of the full-load receiving-end current, in extreme cases, may not greatly increase the sending-end current. In other words (if the charging current is near full-load current) the current at the generator end may not increase much when full load at the receiver end is added, over what it is when no load is taken off at the receiving-end.

Since the e.m.f. of self-induction due to the charging component is proportional to the charging current, its effect upon the voltage regulation of the circuit will also be proportional to the charging current. For a short low-voltage circuit, the charging current is so small that its effect on voltage regulation may be ignored. On the longer circuits, especially long 60 cycle circuits, such as will be considered later, its effect must be given careful consideration.

VARIATION IN CURRENT AND VOLTAGE ALONG THE CIRCUIT

It was explained above and illustrated in Fig. 20 that with a receiving-end load of leading power-factor,

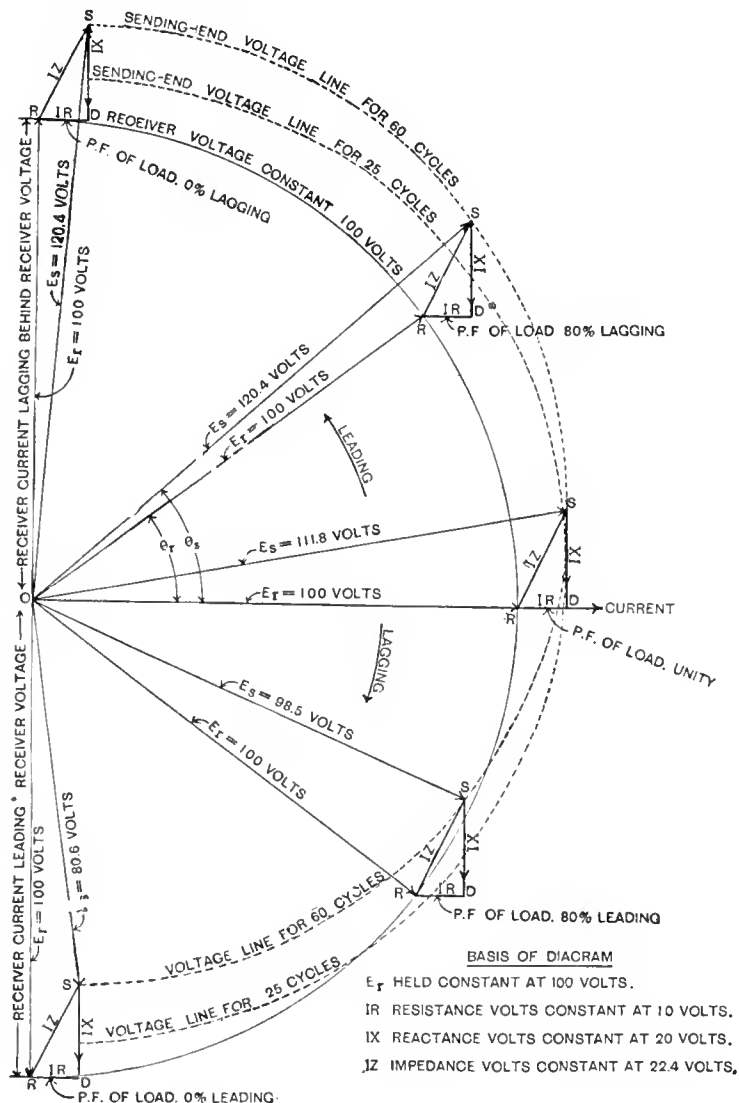


FIG. 20—EFFECT OF SELF INDUCTION ON REGULATION

receiving-end, for in such a case, IX of the diagram would disappear, and the voltage drop would be the same as with direct current. All alternating-current circuits are inductive, and the greater their inductance, the greater will be the voltage drop, or the voltage rise along the circuit.

Any alternating-current circuit may be looked upon as containing three active e.m.f.'s out of phase with each other. In addition to the impressed e.m.f. at the sending-end, there are two e.m.f.'s of self-induction, one as the result of the receiving-end current and lagging 90

the voltage at the sending-end of the circuit might be less than that at the receiving-end. It was shown that the e.m.f. of self-induction, resulting from the leading current, tends to raise the voltage along the circuit. This boosting effect of the voltage is entirely due to the leading component of the load current.

If, now, it is assumed that the power-factor of the receiving-end load is 100 percent, there will be no leading component in the load current, and therefore there can be no boosting of the voltage due to the load current. Since, however, all circuits have capacitance, and since the current is alternating, charging current will flow into the line and this being a leading current, the same tendency to raise the voltage along the circuit will take place as is illustrated by Fig. 20.

The upper part of Fig. 21 is intended to give a physical conception of what takes place in an alternating-current circuit. As the load current starts out from the sending-end, and travels along the conductor, it meets with ohmic resistance. This is represented by r in Fig. 21. It also meets with reactance in quadrature to the current. This is represented by jx in the diagram. Superimposed upon this load current is a current flowing from one conductor to the others, in phase with the voltage at that point and representing true power. This current is the result of leakage over insulators and of corona effect between the conductors. It is represented by the letter g in the diagrams. Then there is the charging current in leading quadrature with the voltage. This current does not consume any active power except that necessary to overcome the resistance to its flow.

In Fig. 21 the four linear constants of the alternating-current circuit, r representing the resistance, jx representing the reactance, g representing the leakage and b representing the susceptance, are shown as located, or lumped, at six different points along the circuit. This is as they would appear in an artificial circuit divided into six units. In any actual line, these four constants are distributed quite evenly throughout the length of the circuit.

VOLTAGE AND CURRENT DISTRIBUTION FOR PROBLEM X

The effect of the charging current flowing through the inductance of the circuit gives rise to a very interesting phenomenon. In order to illustrate this effect, the current and voltage distribution for a 60 cycle, 1000 volt, three-phase circuit, 300 miles long, is plotted in Fig. 21. This circuit will be referred to as problem X. In such a long 60 cycle circuit, this phenomenon is quite pronounced; so that such a problem serves well as an illustration. The voltage and the current have been determined for points 50 miles apart along the circuit. Values for both the current and the voltage under zero load, also under load conditions have been plotted. The load conditions refer to a receiving-end load of 18000 kv-a, at 90 percent power-factor, lagging, 60 cycle three-phase. The voltage is assumed as being held constant 104000 volts at the receiving-end, for both zero and full-load conditions.

Zero-Load Conditions—Without any load being taken from the circuit, it will be seen that the charging current at the sending-end approaches in value that established when under full load; i.e., 94.75 amperes. The charging current drops down to approximately 50 amperes at the middle, and to zero at the receiving-end of the unloaded circuit. The lower full line curve shows how this current is distributed along the circuit. Starting at zero, at the receiving-end of the circuit, it increases as the sending-end of the circuit is approached, at which point it reaches its maximum value of 87.89 amperes. The voltage distribution under zero-load conditions is somewhat opposite to that of the current distribution. That is the voltage (104000 volts at the receiving-end) keeps falling lower until it reached a value of 84676 at the sending-end. It should be noted that the voltage curve for zero load condition drops down rapidly as the sending-end is approached. The reason for this is the large charging current flowing through the inductance of the circuit at this end of the circuit. The larger the charging current the greater the resultant boosting of the receiving-end voltage.

Load Conditions—When 16000 kv-a at 90 percent power-factor lagging is taken from the circuit at the receiving-end, the current at this end goes up to 99.92 amperes. As the supply end is approached the current becomes less, reaching its lowest value (approximately 83 amperes) in the middle of the circuit. At the supply end it is 94.75 amperes, which is less than it is at the receiver end. Thus the full line representing the current in amperes along the circuit assumes the form of an arc, bending downward in the middle of the circuit. The shape of this current curve is dependent upon the relative values of the leading and lagging components of the current at points along the circuit. The reason that the current is a minimum near the middle of the circuit, is because this is the point where the lagging current of the load and the leading charging current of the circuit balance or neutralize each other, and the power-factor is therefore unity. Starting at the receiving-end, the power-factor is 90 percent lagging. As the middle of the circuit is approached, the increasing charging current neutralizes an increasing portion of the lagging component of the load current. Near the middle of the circuit, this lagging component is entirely neutralized, and the power-factor therefore rises to unity. Passing the middle and approaching the sending-end there is no more lagging component to be neutralized, and the increasing charging current causes a decreasing leading power-factor which, when the sending-end is reached, becomes 93.42 percent leading. It will, therefore, be seen that the power-factor as well as the current and voltage varies throughout the length of the circuit.

The voltage distribution under load condition is indicated by the top broken line. In order that the receiving-end voltage may be maintained constant at 104000 volts, the voltage at the sending-end will vary

from 84 676 volts at zero load to 122 370 volts at the assumed load.

THE AUXILIARY CONSTANTS

With the impedance methods considered under the general heading of "Short Transmission Lines" the current was considered as of the same value throughout the circuit, and the voltage drop along the circuit was considered as proportional to the distance. These assumptions, which are permissible in case of short lines, are satisfied by simple trigonometric formulas.

The rigorous solution for circuits of great electrical length accurately takes into account the effect produced by the non-uniform distribution of the current and the voltage throughout the length of the circuit. This effect will hereafter be referred to as the *distribution effect* of the circuit, and may be taken into account

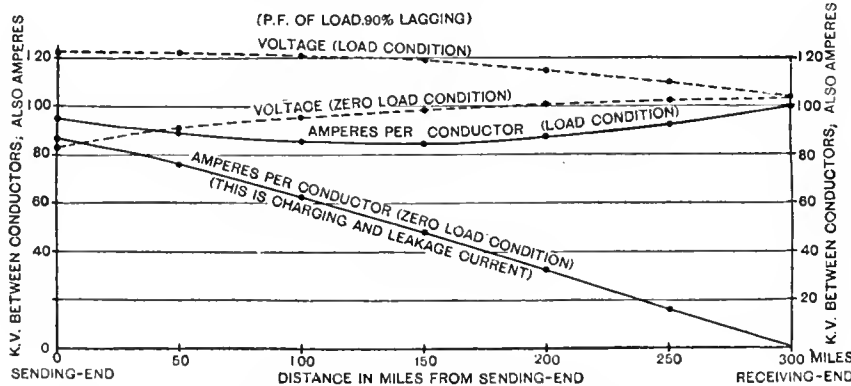
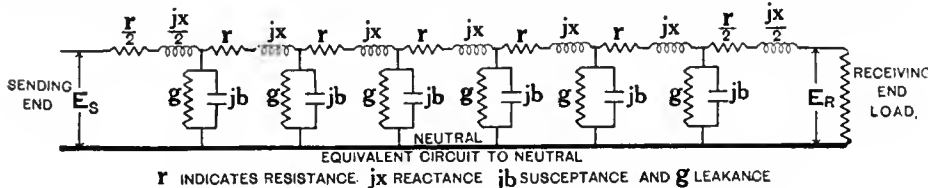


FIG. 21—DIAGRAMS OF TRANSMISSION CIRCUIT—PROBLEM X

300 miles long, 104,000 volts delivered, 60 cycle. The upper diagram gives a physical conception of the conditions along the line. The curves show the variation in current and voltage along the circuit.

through the application of the so called auxiliary constants of the circuit.

The auxiliary constants *A*, *B* and *C* of the circuit are functions of its physical properties, and of the frequency only. They are entirely independent of the voltage or current of the circuit. The various solutions for long transmission circuits are in effect schemes for determining the values of these three auxiliary constants. Mathematically they may be calculated, by hyperbolic functions or by their equivalent convergent series. Graphically they may be obtained to a high degree of accuracy from the accompanying Wilkinson Charts for overhead circuits not exceeding 300 miles in length. Having determined the values for these three constants for a given circuit, the remainder of the solution is just as simple as for short lines. It is only necessary to apply any desired load conditions to these constants and plot the results by vector diagrams.

DIAGRAM OF THE AUXILIARY CONSTANTS

In Fig. 22 are shown voltage and current diagrams representing the application of the auxiliary constants to the solution of transmission circuit problems. To construct the voltage vector diagram, the two auxiliary constants *A* and *B* are required, and to construct the current vector diagram, constants *A* and *C* are required.

Since these diagrams are based upon one volt and one ampere at the receiving-end, it is necessary to multiply the values of the auxiliary constants by the volts or the amperes at the receiving-end, in order to apply the auxiliary constants to a specific problem. Since the diagrams are shown corresponding to unity power-factor, it will also be necessary to change the position of the impedance and charging current triangles in case the power-factor differs from unity. This will be explained later.

Constants a₁ and a₂—Referring to the voltage diagram, Fig. 22, if the line is electrically short the charging current, and consequently its effect upon the voltage regulation is small. In such a case the auxiliary constant *a₁* would be unity, and the auxiliary constant *a₂* would be zero. In other words, the impedance diagram would (for a power-factor of 100 percent) be built upon the end of the vector *ER*, the point *O* coinciding with the point *R*. In such a case, the voltage at the sending end, at zero load, would be the same as that at the receiving-end. If the circuit contains appreciable capacitance, the e.m.f. of self-induction, resulting from the charging currents which will

flow, will result in a lower voltage at zero load at the sending-end than at the receiving-end of the line, as previously explained. Obviously, the load impedance triangle must be attached to the end of the vector representing the voltage at the sending-end of the circuit at zero load. This is the vector *EO* of the voltage diagram, Fig. 22. This voltage diagram corresponds to that of a 60 cycle circuit, 300 miles in length. In such a circuit, the effect of the charging current is sufficiently great to cause the shifting of the point *O* from *R* (in a short line) to the position shown in Fig. 22. In other words, the voltage at zero load at the sending-end has shifted from *ER* for circuits of short electrical length, to *EO* for this long 60 cycle circuit. The auxiliary constants *a₁* and *a₂*, therefore, determine the length and position of the vector representing the sending-end voltage at zero load. Actually, the constant *a₂* represents the volts resistance drop due to the charging current, for each volt at the

receiving-end of the circuit. That is, the line *OF* equals approximately one-half the charging current times the resistance *R*, taking into account, of course, the distributed nature of the circuit. If the circuit is short, it would be sufficiently accurate to assume that the total charging current flows through one-half of the resistance of the circuit. To make this clear, it will be shown later that, for problem *X*, the resistance per conductor $R = 105$ ohms and the auxiliary constant $C_2 = 0.001463$. Thus, this line will take 0.001463 ampere charging current, at zero load, for each volt maintained at the receiving-end, and since $OF =$ approximately $I_c \times \frac{R}{2}$ we have $OF (a_2) = 0.001463 \times \frac{105}{2} = 0.0768075$. The exact value of a_2 as calculated rigorously, taking into account the distributed nature of the circuit, is 0.076831. Since the charging current is in

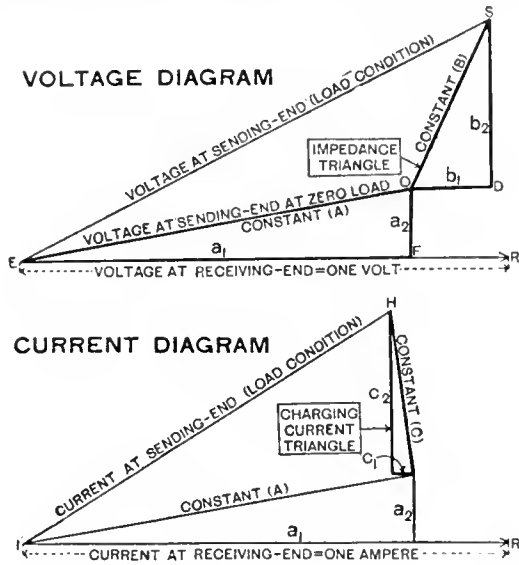


FIG. 22—DIAGRAMMATIC REPRESENTATION OF AUXILIARY CONSTANTS OF A TRANSMISSION CIRCUIT

The vectors are based upon one volt and one ampere being delivered to the receiving end at unity power-factor. These diagrams correspond to those of a long circuit. leading quadrature with the voltage *ER*, the resistance drop *OF* due to the charging current is also at right angles to *ER*, as in Fig. 22.

The length of the line *FR* or $(I - a_1)$, represents the voltage consumed by the charging current flowing through the inductance of the circuit. This may also be expressed with small error if the circuit is not of great electrical length as $I_c \times \frac{X}{2}$. The reactance per conductor for problem *X* is 249 ohms. Therefore $FR = 0.001463 \times \frac{249}{2} = 0.182143$ and $a_1 = 1.000000 - 0.182143 = 0.817857$. The exact value for a_1 as calculated rigorously, taking into account the distributed nature of the circuit, is 0.810558. The vector *FR*, representing the voltage consumed by the charging current flowing through the inductance, is naturally in quadrature with the vector *OF*, representing the voltage consumed by the charging current flowing through the resistance of the circuit.

Constants b_1 and b_2 represent respectively the resistance and the reactance in ohms, as modified by the distributed nature of the circuit. The values for these constants, multiplied by the current in amperes at the receiver-end of the circuit, give the *IR* and *IX* volts

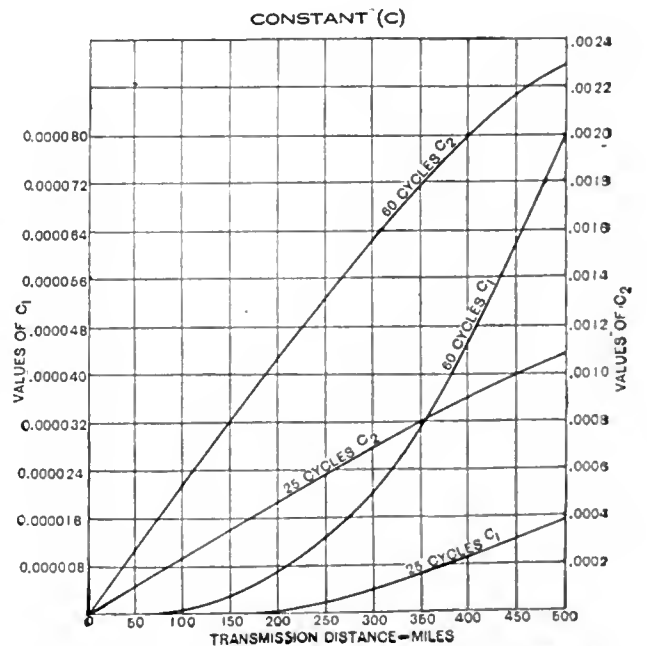
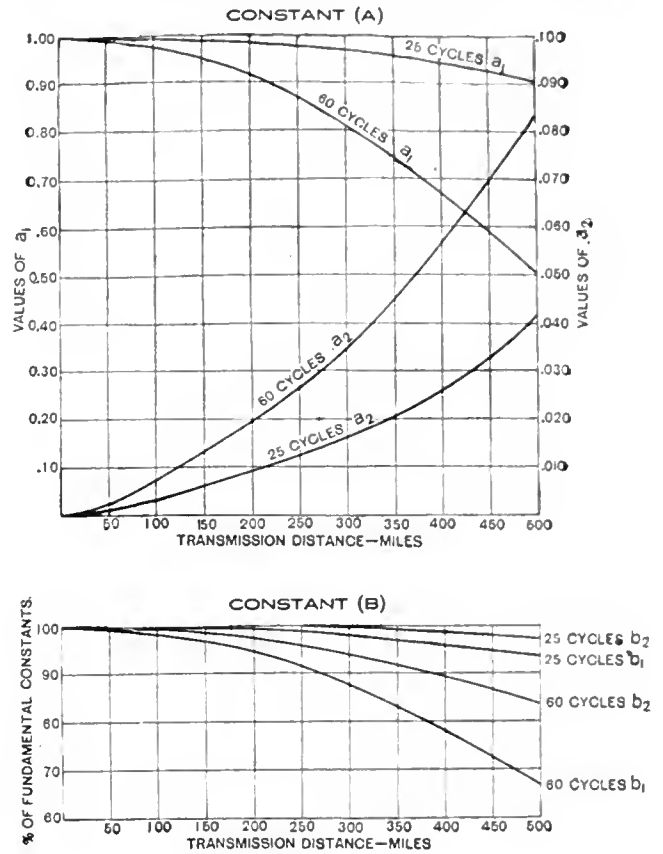


FIG. 23—VARIATION OF THE AUXILIARY CONSTANTS FOR CIRCUITS OF DIFFERENT LENGTHS

drop consumed respectively by the resistance and the reactance of the circuit. To illustrate this, the values of *R* and *X* for problem *X* are $R = 105$ ohms and $X = 249$ ohms per conductor. The distribution effect of the circuit modifies these linear values of *R* and *X* so that

their effective values are $b_1 = 91.7486$ and $b_2 = 235.868$ ohms. The impedance triangle, as modified so as to take into exact account the distributed nature of the circuit, is therefore smaller than it would be if the circuit were without capacitance.

Constants c_1 and c_2 represent respectively conductance and susceptance in mhos as modified by the distributed nature of the circuit. The values for these constants, multiplied by the volts at the receiving-end of the circuit, give the current consumed respectively by the conductance and the susceptance of the circuit. To illustrate, the value of B for problem X is 0.001563 mho per conductor. The distribution effect of the circuit modifies this fundamental value so that its effective

values are $b_1 = 91.7486$ and $b_2 = 235.868$ ohms. In other words these curves have been plotted from calculated values for these constants for certain circuits.

When the circuit is short, these constants do not vary materially from the linear constants of the circuit, but when the circuit becomes long, they depart rapidly, particularly if the frequency is high.

AUXILIARY CONSTANTS	WAVE LENGTH OF THE CIRCUIT AND TRANSMISSION DISTANCE—MILES							
	1/8TH	1/4	3/8TH	1/2	5/8TH	3/4	7/8TH	FULL
	389.8 MILES	739.8 MILES	1109.7 MILES	1479.5 MILES	1849.4 MILES	2219.3 MILES	2589.2 MILES	2959.1 MILES
a_1	+0.716	0	-0.789	-1.209	-0.942	0	+1.191	+1.922
a_2	+0.113	+0.323	+0.350	0	-0.622	-1.104	-0.958	0
b_1	+105	+87	-77.5	-276	-330	-122	+292	+670
b_2	+281	+428	+350	+55.5	-330	-605	-560	-135
c_1	-0.00075	-0.00059	-0.0012	-0.016	-0.0101	+0.00071	+0.0028	+0.0039
c_2	+0.00174	+0.00247	+0.00169	-0.000322	-0.00250	-0.0035	-0.00233	+0.00078
(A)	.725 /89°58'	.323 /90°00'	.843 /156°05'	1.209 /180°00'	1.129 /213°24'	1.104 /270°00'	1.528 /321°11'	1.922 /360°00'
(B)	301.4 /69°37'	437 /78°34'	358.8 /102°29'	282.3 /168°34'	469.5 /225°07'	619.3 /258°34'	635.7 /297°23'	682.4 /348°34'
(C)	.001743 /92°27'	.002527 /101°26'	.002075 /125°21'	.001633 /191°26'	.002715 /247°39'	.003582 /281°26'	.003677 /320°15'	.003947 /371°24'

FIG. 25—VARIATION OF THE AUXILIARY CONSTANTS For problem X, up to full wave length.

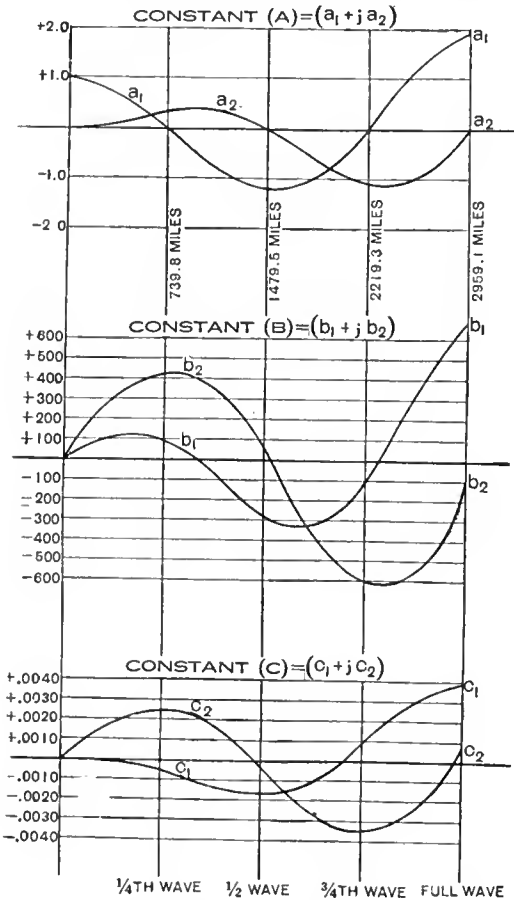


FIG. 24—VARIATION OF THE AUXILIARY CONSTANTS

For a 60 cycle circuit (problem X) up to full wave length.

For a 60 cycle circuit (problem X) up to full wave length. The value of c_1 is so small that its effect is negligible for all except very long circuits. For power circuits it will usually be sufficiently accurate to neglect c_1 . The value c_2 will in such cases represent the charging current at zero load per volt at the receiving-end. Thus c_2 , multiplied by the receiving-end voltage, gives the charging current at zero load for the circuit. For problem X, $c_2 = 0.001463$, and this, multiplied by the receiving-end voltage to neutral $60 \times 0.44 = 87.85$ amperes charging current per conductor.

VARIATION IN THE AUXILIARY CONSTANTS

The curves, Fig. 23, will serve to illustrate in a general way how the auxiliary constants vary for both

The auxiliary constants have been calculated for problem X up to and including a full wave length, namely 2959 miles. Calculations were made only for distances representing each 1/8th wave, that is each 370 miles. The results are tabulated in Fig. 25, and are plotted graphically in Fig. 24. It is interesting to note how these auxiliary constants vary with increasing negative and positive values as the circuit increases in length. A polar diagram is plotted in Fig. 26, indicating the manner in which the auxiliary constant A and its rectangular co-ordinates vary. Although these extreme variations are instructive and interesting, they are not encountered in power transmission circuits, although they will be in long distance telephone practice.

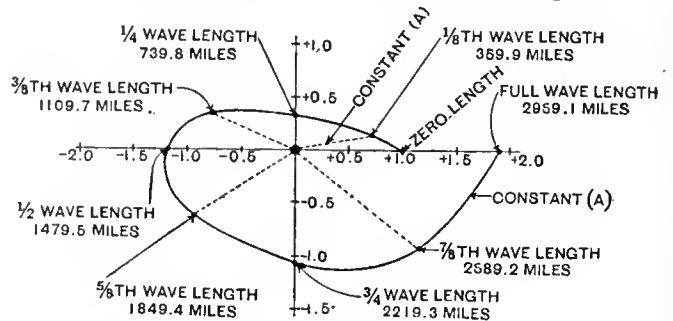


FIG. 26—POLAR DIAGRAM

Showing the variation of the auxiliary constant A for problem X, up to full wave length.

THE WILKINSON CHARTS

Mr. T. A. Wilkinson has prepared charts from which the auxiliary constants may be read directly, thus abridging a great amount of tedious mathematical calculation. These charts, are plotted for circuits of lengths up to and including 300 miles.*

*Similar Charts by Mr. Wilkinson were published in the *Electrical World* for Mar. 16, 1918.

CHART V—WILKINSON CHART A

(FOR DETERMINING AUXILIARY CONSTANTS—ZERO LOAD VOLTAGE)

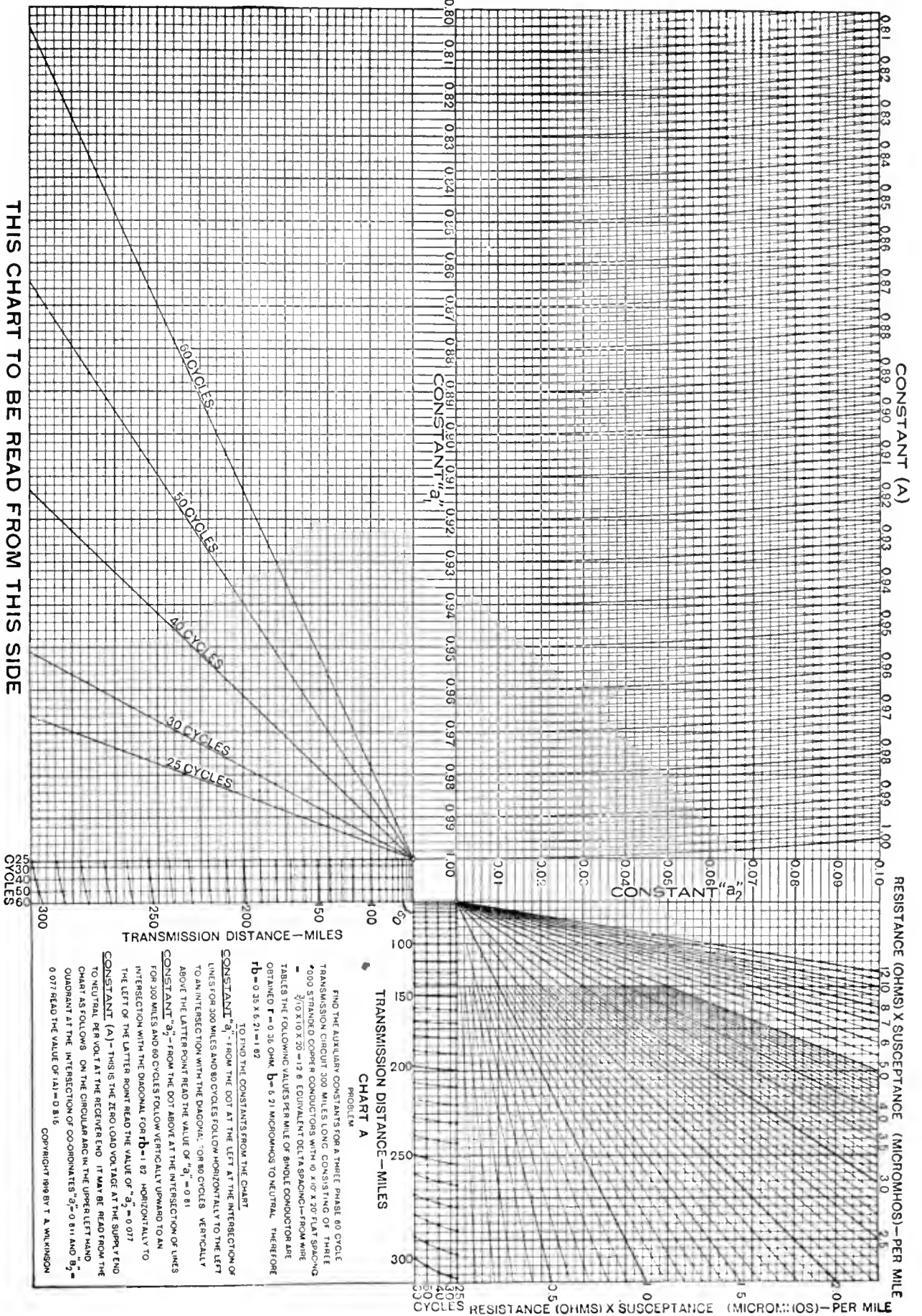


CHART A

PROBLEM
 FIND THE AUXILIARY CONSTANTS FOR A THREE PHASE 60 CYCLE TRANSMISSION CIRCUIT 300 MILES LONG CONSISTING OF THREE 4000 STRANDED COPPER CONDUCTORS WITH 10 X 10" X 20" FLAT SPACING $\frac{3}{10} \times 10 \times 20 = 12$ ft. EQUIVALENT DELTA SPACING—FROM WIRE TABLES THE FOLLOWING VALUES PER MILE OF SINGLE CONDUCTOR ARE OBTAINED $r = 0.35$ OHM $b = 5.21$ MICROMHOS TO NEUTRAL. THEREFORE $rb = 0.35 \times 5.21 = 1.82$

TO FIND THE CONSTANTS FROM THE CHART
CONSTANT "a₁"—FROM THE DOT AT THE INTERSECTION OF LINES FOR 300 MILES AND 60 CYCLES FOLLOW HORIZONTALLY TO THE LEFT TO AN INTERSECTION WITH THE DIAGONAL. FOR 60 CYCLES VERTICALLY ABOVE THE LATTER POINT READ THE VALUE OF "a₁" = 0.81
CONSTANT "a₂"—FROM THE DOT ABOVE AT THE INTERSECTION OF LINES FOR 300 MILES AND 60 CYCLES FOLLOW VERTICALLY UPWARD TO AN INTERSECTION WITH THE DIAGONAL FOR $rb = 1.82$ HORIZONTALLY TO THE LEFT OF THE LATTER POINT READ THE VALUE OF "a₂" = 0.077
CONSTANT (A)—THIS IS THE ZERO LOAD VOLTAGE AT THE SUPPLY END TO NEUTRAL PER VOLT AT THE RECEIVER END. IT MAY BE READ FROM THE CHART AS FOLLOWS ON THE CIRCULAR ARC IN THE UPPER LEFT HAND QUADRANT AT THE INTERSECTION OF COORDINATES "a₁" = 0.81 AND "a₂" = 0.077 READ THE VALUE OF (A) = 0.816

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their effective values are $b_1 = 91.7486$ and $b_2 = 235.868$ ohms. The impedance triangle, as modified so as to take into exact account the distributed nature of the circuit, is therefore smaller than it would be if the circuit were without capacitance.

Constants c_1 and c_2 represent respectively conductance and susceptance in mhos as modified by the distributed nature of the circuit. The values for these constants, multiplied by the volts at the receiving-end of the circuit, give the current consumed respectively by the conductance and the susceptance of the circuit. To illustrate, the value of B for problem X is 0.001563 mho per conductor. The distribution effect of the circuit modifies this fundamental value so that its effective

values are $b_1 = 91.7486$ and $b_2 = 235.868$ ohms. In other words these curves have been plotted from calculated values for these constants for certain circuits.

When the circuit is short, these constants do not vary materially from the linear constants of the circuit, but when the circuit becomes long, they depart rapidly, particularly if the frequency is high.

AUXILIARY CONSTANTS	WAVE LENGTH OF THE CIRCUIT AND TRANSMISSION DISTANCE—MILES							
	1/8TH	1/4	3/8TH	1/2	5/8TH	3/4	7/8TH	FULL
	389.9	739.8	1109.7	1479.5	1849.4	2219.3	2589.2	2959.1
	MILES	MILES	MILES	MILES	MILES	MILES	MILES	MILES
a_1	+0.716	0	-0.789	-1.209	-0.942	0	+1.191	+1.922
a_2	+1.113	+3.223	+3.550	0	-6.222	-1.104	-9.558	0
b_1	+105	+87	-77.5	-276	-330	-122	+292	+670
b_2	+281	+428	+350	+55.5	-330	-605	-560	-735
c_1	-0.00075	-0.00350	-0.0012	-0.016	-0.0101	+0.0071	+0.0028	+0.0037
c_2	+0.00174	+0.00247	+0.00169	-0.00322	-0.00250	-0.0035	-0.00233	+0.0078
(A)	72.5 /89°58'	323 /90°00'	843 /156°05'	1209 /180°00'	1129 /213°26'	1104 /270°00'	1528 /32°11'	1922 /360°00'
(B)	301.4 /69°37'	437 /78°34'	358.8 /102°29'	282.3 /148°34'	469.3 /225°07'	619.3 /268°34'	635.7 /297°23'	682.4 /348°34'
(C)	0.01743 /92°29'	0.02527 /101°26'	0.02075 /125°21'	0.01633 /149°26'	0.02715 /247°57'	0.03582 /281°26'	0.03677 /320°15'	0.03947 /371°26'

FIG. 25—VARIATION OF THE AUXILIARY CONSTANTS For problem X up to full wave length.

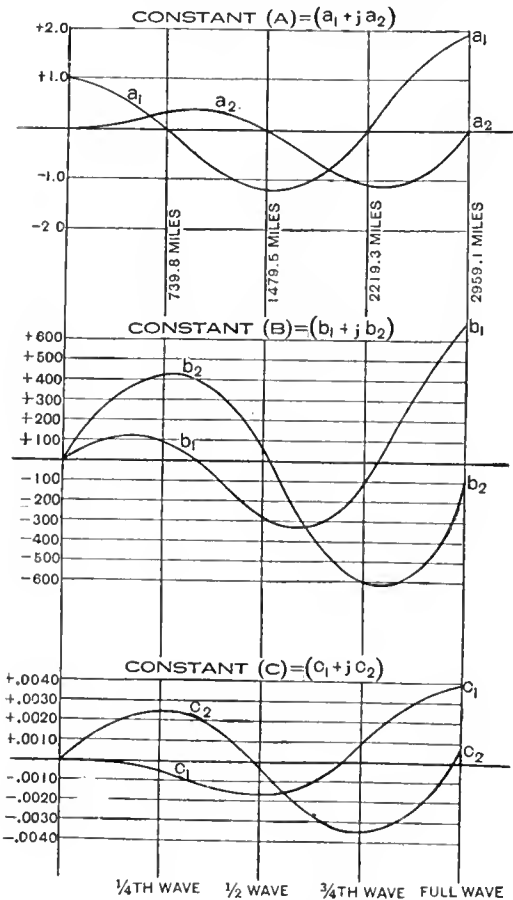


FIG. 24—VARIATION OF THE AUXILIARY CONSTANTS For a 60 cycle circuit (problem X) up to full wave length.

The auxiliary constants have been calculated for problem X up to and including a full wave length, namely 2959 miles. Calculations were made only for distances representing each 1/8th wave, that is each 370 miles. The results are tabulated in Fig. 25, and are plotted graphically in Fig. 24. It is interesting to note how these auxiliary constants vary with increasing negative and positive values as the circuit increases in length. A polar diagram is plotted in Fig. 26, indicating the manner in which the auxiliary constant A and its rectangular co-ordinates vary. Although these extreme variations are instructive and interesting, they are not encountered in power transmission circuits, although they will be in long distance telephone practice.

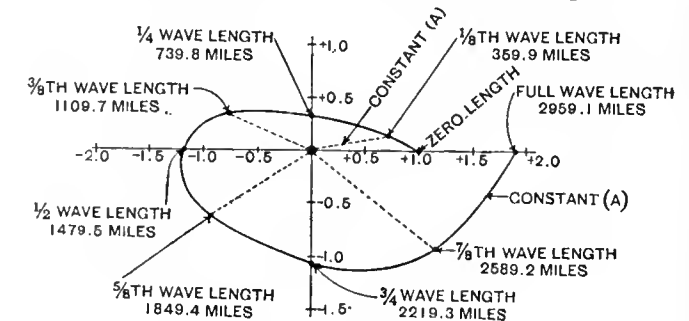


FIG. 26—POLAR DIAGRAM Showing the variation of the auxiliary constant A for problem X, up to full wave length.

value $c_2 = 0.001463$. The value of c_1 is so small that its effect is negligible for all except very long circuits. For power circuits it will usually be sufficiently accurate to neglect c_1 . The value c_2 will in such cases represent the charging current at zero load per volt at the receiving-end. Thus c_2 , multiplied by the receiving-end voltage, gives the charging current at zero load for the circuit. For problem X, $c_2 = 0.001463$, and this, multiplied by the receiving-end voltage to neutral $60,044 = 87.85$ amperes charging current per conductor.

VARIATION IN THE AUXILIARY CONSTANTS

The curves, Fig. 23, will serve to illustrate in a general way how the auxiliary constants vary for both

THE WILKINSON CHARTS

Mr. T. A. Wilkinson has prepared charts from which the auxiliary constants may be read directly, thus abridging a great amount of tedious mathematical calculation. These charts, are plotted for circuits of lengths up to and including 300 miles.*

*Similar Charts by Mr. Wilkinson were published in the *Electrical World* for Mar. 16, 1918.

CHART V—WILKINSON CHART A

(FOR DETERMINING AUXILIARY CONSTANTS—ZERO LOAD VOLTAGE)

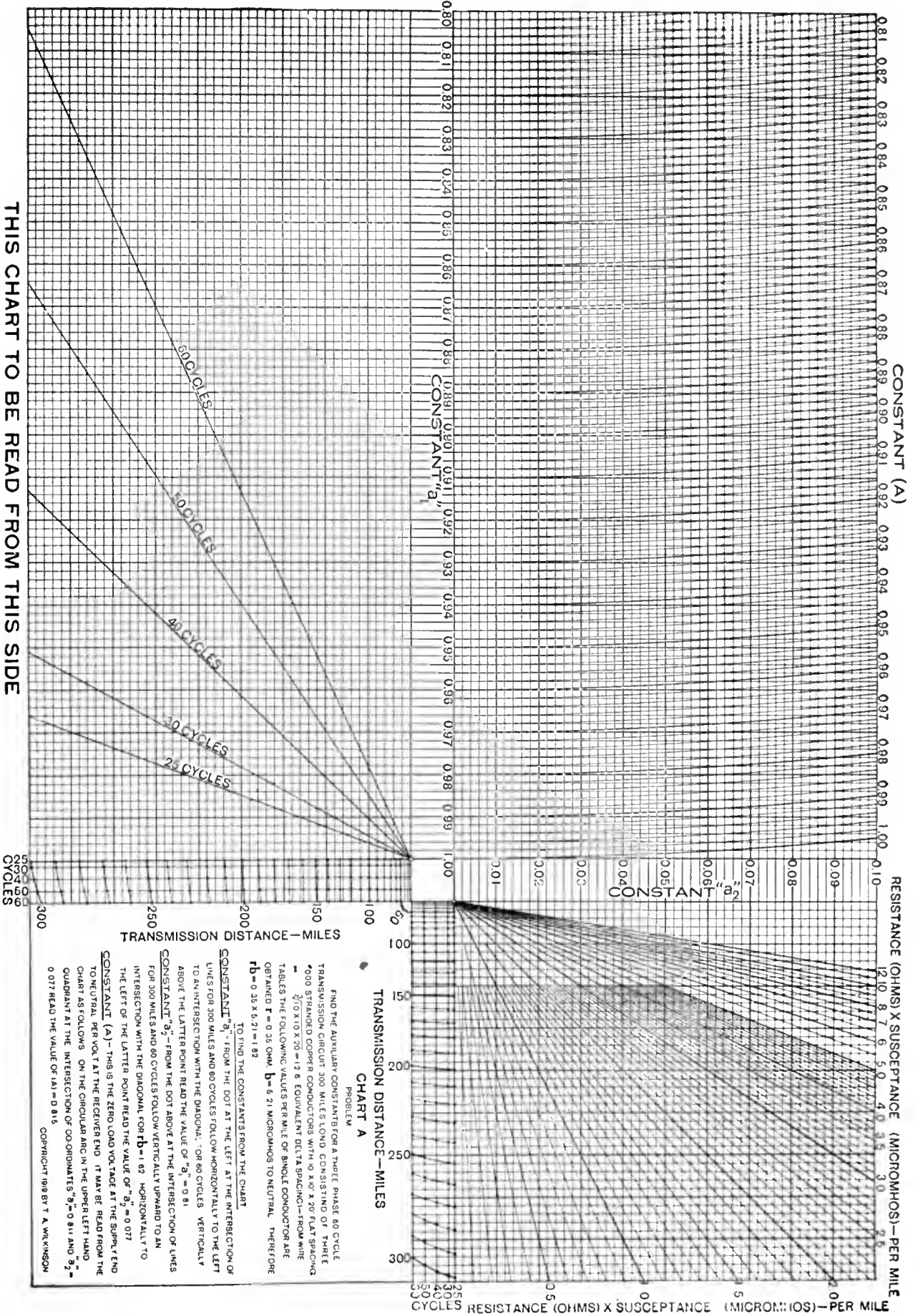


CHART VI—WILKINSON CHART B

(FOR DETERMINING AUXILIARY CONSTANTS—LINE IMPEDANCE)

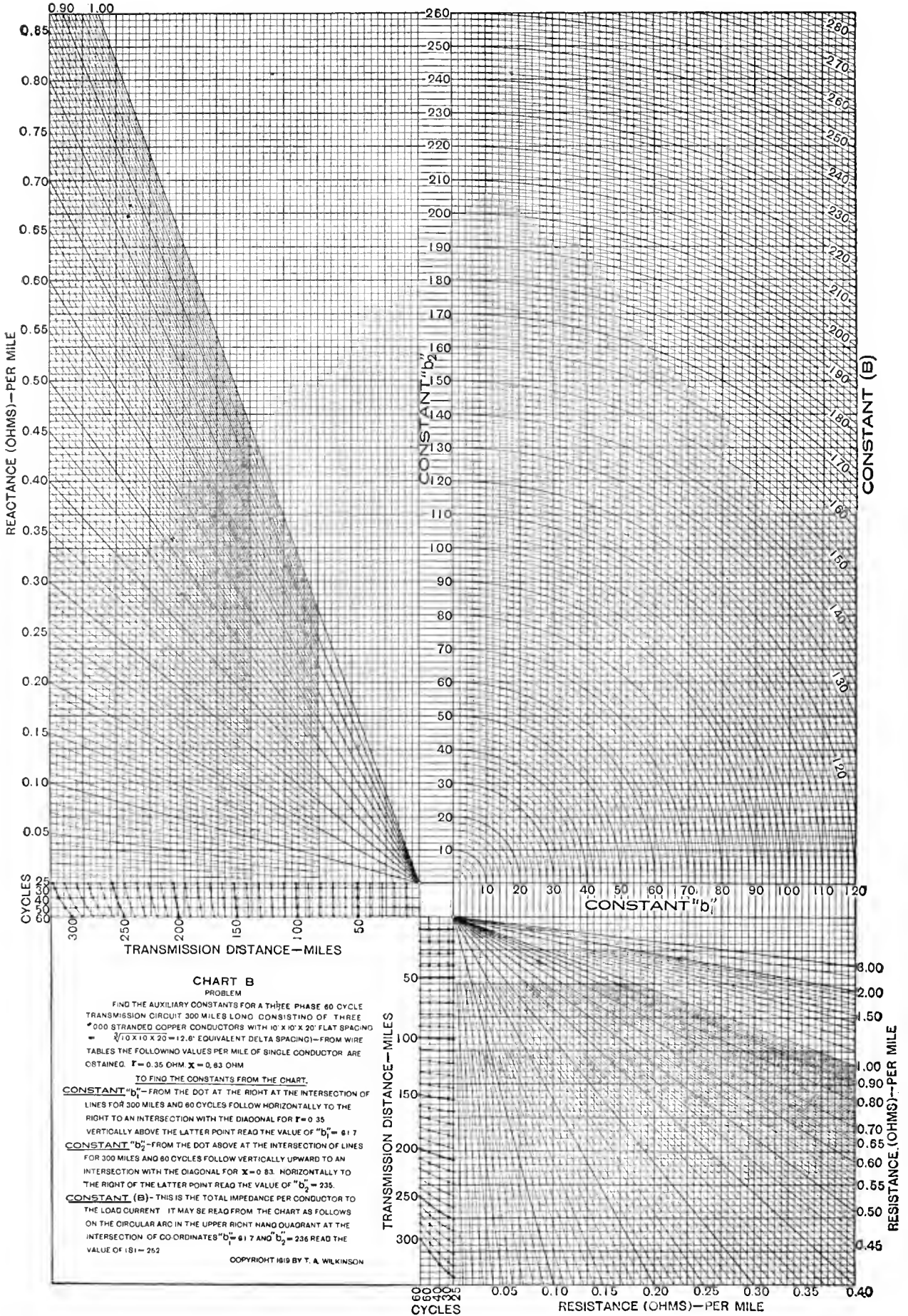


CHART B
PROBLEM

FIND THE AUXILIARY CONSTANTS FOR A THREE PHASE 60 CYCLE TRANSMISSION CIRCUIT 300 MILES LONG CONSISTING OF THREE 000 STRANDED COPPER CONDUCTORS WITH 10' X 10' X 20' FLAT SPACING = $\sqrt{3}/10 \times 10 \times 20 = 12.6'$ EQUIVALENT DELTA SPACING) FROM WIRE TABLES THE FOLLOWING VALUES PER MILE OF SINGLE CONDUCTOR ARE OBTAINED. $r = 0.35$ OHM $x = 0.83$ OHM

TO FIND THE CONSTANTS FROM THE CHART.

CONSTANT "b₁"—FROM THE DOT AT THE RIGHT AT THE INTERSECTION OF LINES FOR 300 MILES AND 60 CYCLES FOLLOW HORIZONTALLY TO THE RIGHT TO AN INTERSECTION WITH THE DIAGONAL FOR $r = 0.35$ VERTICALLY ABOVE THE LATTER POINT READ THE VALUE OF "b₁" = 61.7

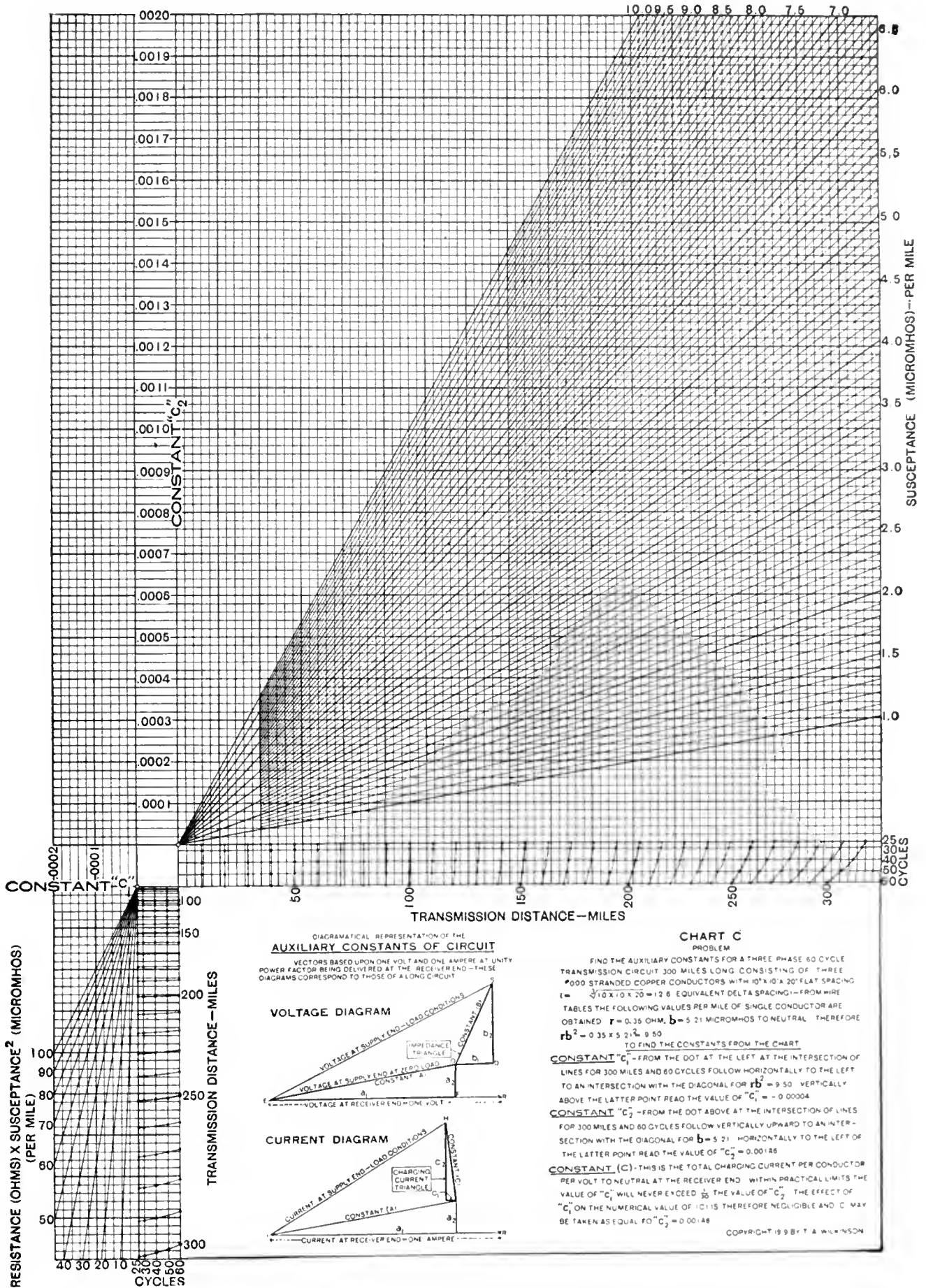
CONSTANT "b₂"—FROM THE DOT ABOVE AT THE INTERSECTION OF LINES FOR 300 MILES AND 60 CYCLES FOLLOW VERTICALLY UPWARD TO AN INTERSECTION WITH THE DIAGONAL FOR $x = 0.83$. HORIZONTALLY TO THE RIGHT OF THE LATTER POINT READ THE VALUE OF "b₂" = 235.

CONSTANT (B)—THIS IS THE TOTAL IMPEDANCE PER CONDUCTOR TO THE LOAD CURRENT IT MAY BE READ FROM THE CHART AS FOLLOWS ON THE CIRCULAR ARC IN THE UPPER RIGHT HAND QUADRANT AT THE INTERSECTION OF CO-ORDINATES "b₁" = 61.7 AND "b₂" = 235 READ THE VALUE OF B = 181 = 252

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CHART VII—WILKINSON CHART C

(FOR DETERMINING AUXILIARY CONSTANTS—CHARGING CURRENT)



The reading of these charts is simplified by reason of the fact that all three charts are somewhat similar. In following any of them, the start is made from the intersection of the short arc representing length of circuit and the straight line representing the frequency. From this intersection a straight line is followed to a diagonal line and thence at right angles to the constant required. Thus in a few minutes the auxiliary constants of the circuit may be obtained directly from the chart, whereas by a mathematical solution from 15 minutes to an hour might be consumed in obtaining them. It is not, however, the time saved in obtaining these constants which is most important. The greatest advantage in this graphical solution for the auxiliary constants is that it not only abridges the use of a form of mathematics which the average engineer is inefficient in using, but it tends to prevent serious mistakes being made. In calculating these auxiliary constants by either convergent series or hyperbolic methods, an incorrect algebraic sign assigned to a number may cause a very serious error. Errors of magnitude are less likely to occur when using a comparatively simple graphical solution.

In order to determine the accuracy obtainable by a complete graphical solution, using the Wilkinson Charts for obtaining the auxiliary constants and vector diagrams for the remainder of the solutions, 48 problems were solved both graphically and mathematically. These problems consisted of circuits varying between 20 and 300 miles in length, and voltages varying between 10 000 and 200 000 volts. Twenty-four problems were for 25 cycle, and the same number for 60 cycle circuits. The maximum error in supply end voltage by the graphical solution employing a four times magnifying glass was one-fourth of one percent. A tabulation of the results as determined by various methods for these circuits will follow later.

APPLICATION OF TABLES

The application of the tables to long transmission lines follows, in general, the same plan as for short lines, published as Chart II, with such modifications as are produced by the effects of distributed capacitance and reactance. The procedure best suited for long transmission lines is shown in Chart VIII.

GRAPHICAL SOLUTION OF PROBLEM X

Problem X—Length of circuit 300 miles, conductors three No. 000 stranded copper spaced 10 by 10 by 20 feet (equivalent delta 12.6 feet) Temperature taken as 25 degrees C. Load conditions at receiving end 18 000 kv-a, (16 200 kw at 90 percent power-factor lagging) 104 000 volts, three-phase, 60 cycles.

$$E_{r_3} = \frac{104\,000}{1732} = 60.046 \text{ volts.}$$

$$I_r = \frac{6000 \times 1000}{60.046} = 99.92 \text{ amperes.}$$

CHART VIII.—APPLICATION OF TABLES TO LONG TRANSMISSION LINES

(EFFECT OF DISTRIBUTED CAPACITANCE TAKEN INTO ACCOUNT) OVERHEAD BARE CONDUCTORS

Starting with the kv-a., voltage and power-factor at the receiving end known.

QUICK ESTIMATING TABLES XII TO XXI INC.

From the quick estimating table corresponding to the voltage to be delivered, determine the size of the conductors corresponding to the permissible transmission loss.

CORONA LIMITATION—TABLE XXII

If the transmission is at 30 000 volts, or higher, this table should be consulted to avoid the employment of conductors having diameters so small as to result in excessive corona loss.

RESISTANCE—TABLE II

From this table obtain the resistance per unit length of single conductor corresponding to the maximum operating temperature—calculate the total resistance for one conductor of the circuit—if the conductor is large (250 000 circ. mils or more) the increase in resistance due to skin effect should be added.

REACTANCE—TABLES IV AND V

From one of these tables obtain the reactance per unit length of single conductor. Calculate the total reactance for one conductor of the circuit. If the reactance is excessive (20 to 30 percent reactance volts will in many cases be considered excessive) consult Table VI or VII. Having decided upon the maximum permissible reactance the corresponding resistance may be found by dividing this reactance by the ratio value in Table VI or VII. When the reactance is excessive, it may be reduced by installing two or more circuits and connecting them in parallel, or by the employment of three conductor cables. Using larger conductors will not materially reduce the reactance. The substitution of a higher transmission voltage, with its correspondingly less current, will also result in less reactance.

CAPACITANCE SUSCEPTANCE—TABLES IX AND X

From one of these tables obtain the capacitance susceptance to neutral, per unit length of single conductor. Calculate the total susceptance for one conductor of the circuit to neutral.

GRAPHICAL SOLUTION

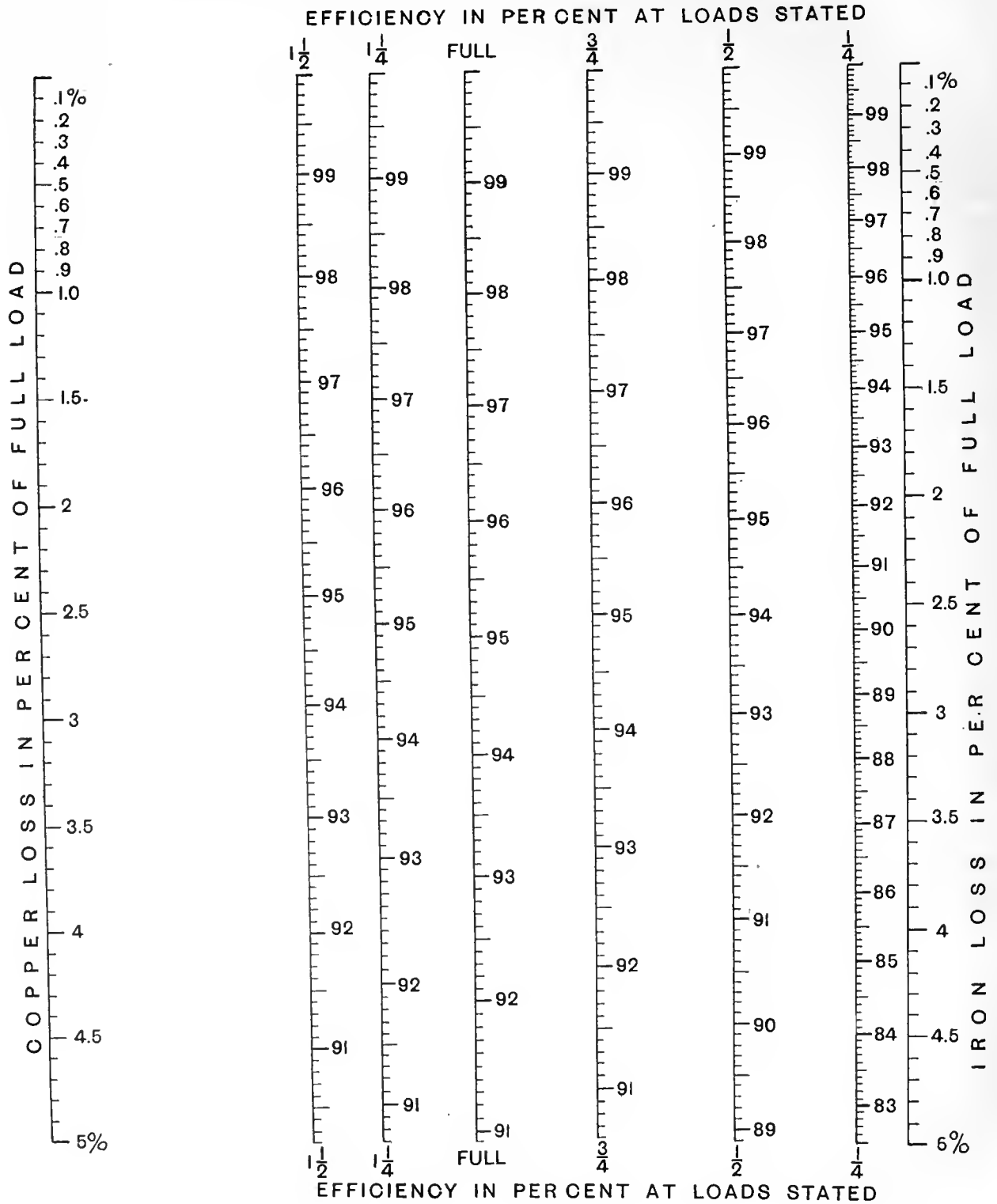
From the Wilkinson charts obtain the auxiliary constants. Applying these auxiliary constants to the load conditions of the problems, make a complete graphical solution as explained in the text. Vector diagrams of the voltage and the current at both ends of the circuit are then constructed, from which the complete performance can be readily obtained graphically.

MATHEMATICAL SOLUTION

As a precaution against errors in those cases where accuracy is essential, the result obtained graphically should be checked by the convergent series or the hyperbolic method.

CHART IX—PETER'S EFFICIENCY CHART

FOR DETERMINING TRANSFORMER LOSSES AND EFFICIENCIES

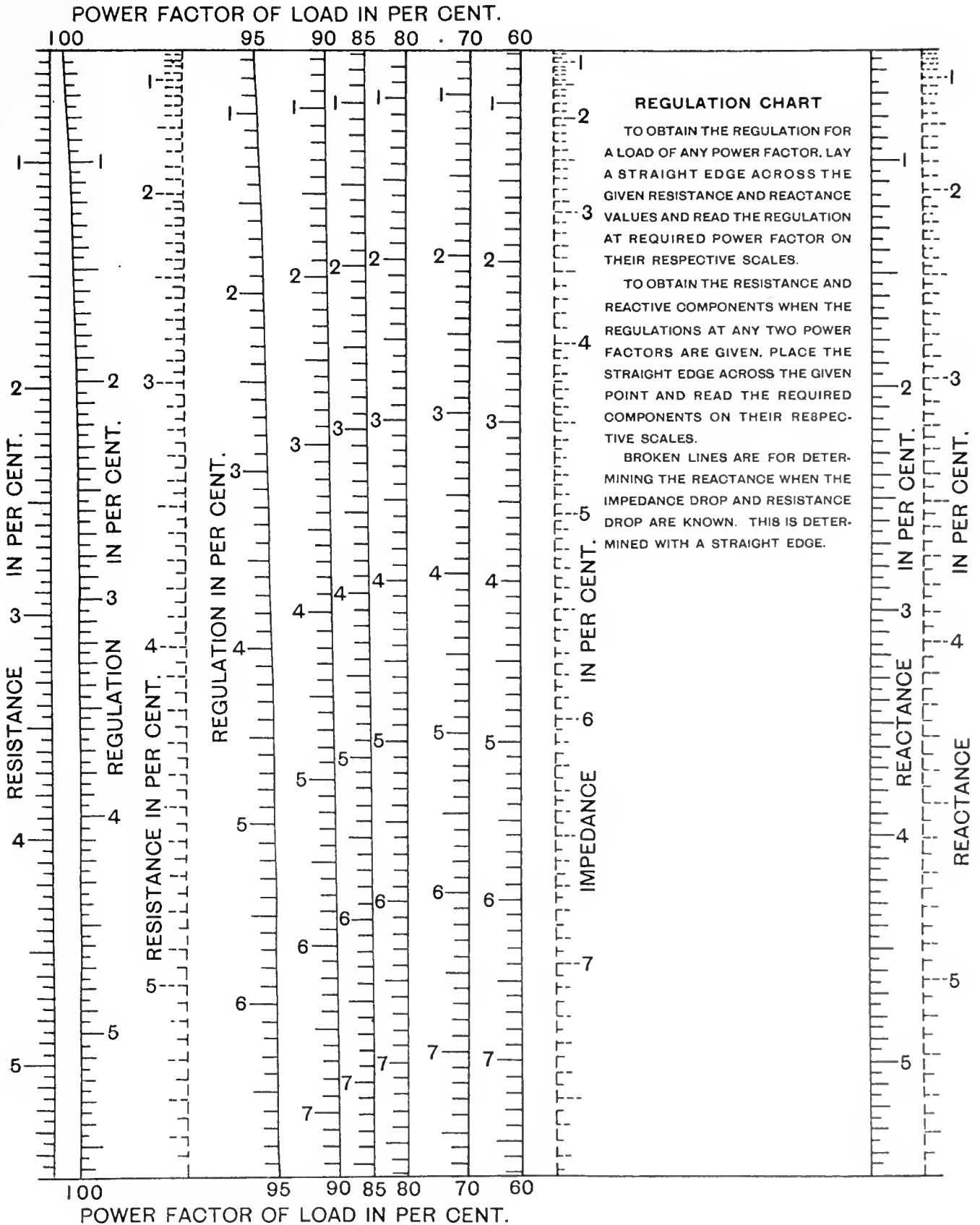


TO OBTAIN EFFICIENCY AT ANY LOAD LAY STRAIGHT EDGE AT GIVEN IRON AND COPPER LOSS POINTS AND READ THE EFFICIENCY AT REQUIRED LOAD ON THEIR RESPECTIVE SCALES WHERE STRAIGHT EDGE CROSSES THEM.

VICE VERSA. TO OBTAIN LOSSES, PLACE STRAIGHT EDGE ACROSS ANY TWO GIVEN EFFICIENCY POINTS AND READ PER CENT IRON AND COPPER LOSS ON THEIR RESPECTIVE SCALES.

CHART X--PETER'S REGULATION CHART

FOR DETERMINING TRANSFORMER REGULATION



the resistance per conductor including an equivalent value to correspond to the resistance in the high and low tension windings of two transformers will be,—

$$R + R_t = 105 + 6.25 + 6.25 = 117.5 \text{ ohms.}$$

The percent reactance volts of a transformer having 3.74 percent regulation at 80 percent lagging power-factor and 1.04 percent resistance volts may be read directly from Peter's Regulation Chart (Chart X) by laying a straight edge along the points corresponding to 1.04 percent resistance and 3.74 on the 80 percent power-factor line. The intersection of the straight edge with the last solid line at the right will give the percent reactance, = 4.85 percent.

The percent reactance volts can also be read directly from the Mershon Chart. To do this, follow

TABLE M—APPROXIMATION OF RESISTANCE AND REACTANCE VOLTS FOR TRANSFORMERS OF VARIOUS CAPACITIES

Transformer Capacity in Kv-a	Voltage Drop in Percent			
	Resistance		Reactance	
	25 cycles	60 cycles	25 cycles	60 cycles
300	2.15	1.3	4.0	5.6
500	1.4	1.2	4.1	6.0
750	1.2	1.1	4.2	6.3
1000	1.7	1.1	6.0	6.5
1500	1.4	0.9	6.2	7.0
2000	1.3	0.8	6.4	7.0
3000	1.2	0.75	6.8	7.0
5000	1.1	0.65	7.2	7.0
7500	1.0	0.6	7.8	8.0
10 000	1.0	0.6	8.0	8.0
15 000	0.95	0.55	8.0	8.5
25 000	0.9	0.5	8.0	9.0

upward the vertical line in the Mershon Chart corresponding to 80 percent power-factor until it intersects the first arc. From this point of intersection follow the horizontal line to the right a distance corresponding to 1.04 percent resistance volts. From this point thus obtained follow the vertical line until the arc representing 3.74 percent voltage drops is reached. The length of this vertical line will be the percentage reactance volts of the transformer, in this case 4.8 percent. Of course the reactance may, if desired, be calculated by following the general construction traced out as above described upon the Mershon chart, but the chart will give sufficiently accurate values for practical purposes.

The volts necessary to overcome the reactance of the windings of one of these transformers is therefore found to be $60 \times 0.46 \times 0.048 = 2882$ volts to neutral. The

ohms reactance will therefore be $\frac{2882}{99.92} = 28.84$ ohms to neutral for each transformer. Since the reactance of each line conductor is 249 ohms, the reactance per conductor, including an equivalent value to correspond to the reactance in the high and low tension windings of two transformers will be,—

$$X + X_t = 249 + 28.84 + 28.84 = 306.68 \text{ ohms.}$$

The impedance of one conductor of the circuit of problem X including the raising and lowering transformers will be,—

$$Z = 117.5 + j 306.68 \text{ ohms}$$

and $Y =$ (assumed to be the same as without the transformers).

With the assumed values for the impedance, the performance of the combined circuit may be calculated as though there were no transformers in the circuit.

VOLTAGE AND CURRENT AT INTERMEDIATE POINTS ALONG THE CIRCUIT

Thus far we have considered the electrical condition at the two ends of a transmission circuit only. Occasionally it may be desired to determine the voltage or the current at a point, or at various points along the circuit. In Fig. 21, graphs of the voltage and of the current are shown for points between the terminals of a circuit corresponding to the condition of zero load, and also of rated load. The graphs were plotted by determining graphically the voltage and the current for points at 50 mile intervals along this 300 mile circuit, as follows:—

To determine the conditions 250 miles from the sending-end, (50 miles from the receiving-end) the three auxiliary constants were obtained from the Wilkinson charts corresponding to a circuit 50 miles long. In other words, it was assumed that the circuit was only 50 miles long. By multiplying these auxiliary constants by the known voltage and current at the receiving-end of the circuit, voltage and current diagrams were constructed as in Fig. 27 and on these, the corresponding values of voltage and current at the sending-end of the 50 mile section were scaled off. This gives the conditions, for the load assumed, at a point 250 miles from the sending-end. In a similar manner the voltage and current at this point, corresponding to zero load at the receiving-end, may be obtained. A similar procedure will determine the electrical conditions for a point 100 miles from the receiving-end (200 miles from the sending-end). The auxiliary constants will this time be read from the charts, corresponding to a 100 mile circuit, but the same receiving-end conditions will be used, as before. The electrical condition for any intermediate points along any smooth line, may thus be readily determined.

CHAPTER IX

PERFORMANCE OF LONG TRANSMISSION LINES

(RIGOROUS CONVERGENT SERIES SOLUTION)

THE APPROXIMATE electrical performance of overhead circuits having a length not exceeding 300 miles, may readily be determined by the use of the Wilkinson Charts for determining the values of the auxiliary constants, supplemented by vector diagrams representing the current and voltages of the circuits. In important cases, as a final check upon the values obtained by the simple graphical solution, a mathematical solution yielding rigorous results should be made. If the circuit is more than 300 miles long, a mathematical solution yielding rigorous values will be required for determining the correct values of at least the auxiliary constants.

FORMS OF RIGOROUS SOLUTIONS

The most direct method for determining mathematically the exact performance of circuits of great electrical length is by the employment of hyperbolic functions, and the fundamental equations are usually expressed in such terms. Many engineers have a general aversion to the use of mathematical expressions employing hyperbolic functions. One reason for this is that the older engineers attended college before the hyperbolic theory as applied to transmission circuits had been developed, and tables of such functions were not at that time available.

In 1893 Dr. A. E. Kennelly introduced vector arithmetic into alternating-current computation for the first time.* Although real hyperbolic functions had well recognized uses in applied science, it was in 1894** that he, for the first time, suggested and illustrated the application of vector hyperbolic functions to the determinations of the electrical performance of transmission circuits. Since that time Dr. Kennelly has been a most persistent advocate of the employment of these functions in electrical engineering problems. To advance their use, he has calculated and published numerous tables and charts of such functions. Such tables were, until recently, incomplete and the result was that it was necessary, in using these tables, to interpolate values, thus introducing complications and inaccuracies into the calculations.

Tables of hyperbolic functions and charts are now sufficiently extensive and complete for accurate work. The universities quite generally are encouraging instruction of students in the hyperbolic theory. It is there-

fore to be expected that, in the future, the employment of hyperbolic functions for the solution of long transmission lines will come into general use.

The fundamental hyperbolic equations expressing the electrical behavior of transmission circuits may be expressed in the form of convergent series and, in such form have, in some cases, certain advantages over the hyperbolic form. The convergent series form of solution does not require the employment of tables or charts of hyperbolic functions, whereas hyperbolic forms of solutions do require such tables or charts. If, therefore, such tables or charts are not available, hyperbolic solutions cannot be employed.

While the amount of arithmetical work involved is considerable, any degree of accuracy may readily be obtained by the convergent series solution by working out the terms for the auxiliary constants until they become too small to have any effect upon the results. This can also be done with hyperbolic functions, but exact interpolation of such functions from tabular values, may be considered more difficult than the working out of an extra term or two in the convergent series form of solution. The above remarks apply to cases where an unusual degree of accuracy is required. Later will be included a tabulation of the performance of 64 different electrical circuits, as determined by a rigorous, and also by eight different approximate methods of calculation. As the rigorous values are taken as 100 percent correct, in determining the percent error by the approximate methods, it was important that the so called "rigorous" values be exact. To make them so, it was found convenient to employ the convergent series form of solution for these particular problems, covering circuits up to 500 miles long and potentials up to 200 000 volts. For the calculation of the performance of practical power transmission circuits, tables of hyperbolic functions are now sufficiently complete to yield results well within the errors due to variation in the assumed linear constants of the circuits from their actual values.

The employment of convergent series requires a working knowledge of complex quantities only, whereas the employment of hyperbolic functions in addition leads into hyperbolic trigonometry. As literature pertaining to the hyperbolic theory becomes more generally available, and as the younger engineers take up active engineering work, the hyperbolic theory will become more generally used.

For the purpose of providing a choice of rigorous methods, both convergent series and two forms of hy-

**Trans. Am. Inst. Elec. Engrs.*, Vol. X, page 175 "Impedance."

***"Electrical World"*, Vol. XXIII, No. 1, page 17, January 1894, "The Fall of Pressure in Long-Distance Alternating-Current Conductors."

perbolic solutions are given. The numerical values employed in these solutions have been carried to what may appear as an unnecessary degree of precision. The reason for this is to demonstrate the fact that all of these rigorous solutions yield the same results. For practical problems less accuracy would be essential, thus reducing the amount of arithmetical work.

Before taking up the rigorous solutions, it has been thought desirable to review the rules regarding the use of complex quantities and vector operations.

COMPLEX QUANTITIES

The calculation of the auxiliary constants of the circuit by convergent series, and the further calculation of the electrical performance of the circuit, involve the use of complex numbers, that is, numbers containing j terms. Thus $A = a_1 + ja_2$ is a complex quantity. To the beginner, expressions containing j terms may seem difficult to understand. It cannot be made too emphatic that the rules governing the use of such terms are so simple (embodying only the simple rules of algebra) that the beginner will shortly be surprised with the ease at which complex quantities are handled.

j Terms—In the complex notation $Z = X + jY$, the prefix j indicates that the value Y is measured along the axis perpendicular to that of X , or what is called the imaginary axis. There need be no significance attached to the symbol j other than that of a mere distinguishing mark, to designate a distance above or below the reference axis in the vector diagram. However, great use is made of a further assigned significance. It has a numerical significance in the form of $j = \sqrt{-1}$ which enables all formal algebraic operations, multiplication, addition, extraction of roots, etc. incident to computation involving complex quantities, to be carried out rigorously. This numerical designation for j does not prevent its use as a designating symbol for the vertical direction in the vector diagram.*

PLANE VECTORS

Alternating voltages and currents which vary according to the sine or cosine law, may be represented graphically by directed straight lines, called plane vectors. The length of the vector represents the effective value of the alternating quantity, while the position of the vector with respect to a selected reference vector, base or axis, gives the phase displacement. The line OP , of Fig. 29, represents a plane vector inclined at an angle of $33^\circ 41'$ with the base OS (the axis of reference). The length of the line OP is a measure, to some assumed scale, of the effective value of the voltage or current, while the angle SOP gives the phase displacement.

Counter-clockwise rotation is considered positive. Thus, in Fig. 29, if the line OS represents the instantaneous direction of the current and the line OP that of the voltage at the same instant, the current is represented

as lagging behind the voltage by the angle $33^\circ 41'$. By means of vectors the relative phase position and value of either currents or e.m.f.'s can be represented in the same manner as forces in mechanics.

The position of P , with respect to O , is usually defined in terms of rectangular or polar co-ordinates. In rectangular co-ordinates there are two fixed mutually perpendicular axes, $-XOX$ and $-YOY$ (Fig. 31) in the plane of reference. The former, $-XOX$, is called the real axis, or axis of real quantities. The latter, $-YOY$, is called the imaginary axis, or axis of imaginary quantities. The qualifying adjective "imaginary" does not mean that there is anything indeterminate or fictitious about this axis. The perpendicular projections of P (Fig. 31) on the X and Y axes are respectively the real component X , and the imaginary component Y .

The magnitude and sign of the rectangular components X and Y completely determine the position of the vector OP . Positive is indicated to right and upward, negative to the left and downward as indicated in Fig. 30. Thus, if X and Y are both positive, OP lies in the first quadrant. If X and Y are both negative, OP lies in the third quadrant. If X is $-$ and Y is $+$, OP lies in the second quadrant. If X is $+$ and Y is $-$, OP lies in the fourth quadrant. Any plane vector may be completely specified by its real and imaginary components X and Y . Thus, beneath Fig. 31, is a table in which the point P is located in the plane by co-ordinates for all quadrants.

From Fig. 30 it is evident that, mathematically, the quadrature numbers are just as real as the others. The quadrature numbers represent the vertical, and the ordinary numbers the horizontal directions.

VECTOR OPERATIONS

In general, in the handling of complex numbers involving j terms, the simple rules of algebra are followed. In Fig. 32 two vector quantities are shown. Vector A has a magnitude of 5 units and is inclined in the positive or leading direction at an angle of $36^\circ 52'$ with the horizontal reference vector, and vector B has a magnitude of 4.47 units, and is inclined in the positive or leading direction at an angle of $63^\circ 26'$ with the reference vector. These vector quantities are expressed in rectangular co-ordinate as $A = +4 + j3$, $B = +2 + j4$ or in polar co-ordinates as $A = 5 / 36^\circ 52'$, $B = 4.47 / 63^\circ 26'$. The prefix j simply means that the number following it is measured along the vertical or Y axis. The dot under the vector designation indicates that A is expressed as a complex number, so that the absolute value of A would be $\sqrt{(4)^2 + (3)^2} = 5$ and of $B = \sqrt{(2)^2 + (4)^2} = 4.47$. The absolute value of a complex number is called its "size"; while the angle is called its "slope".

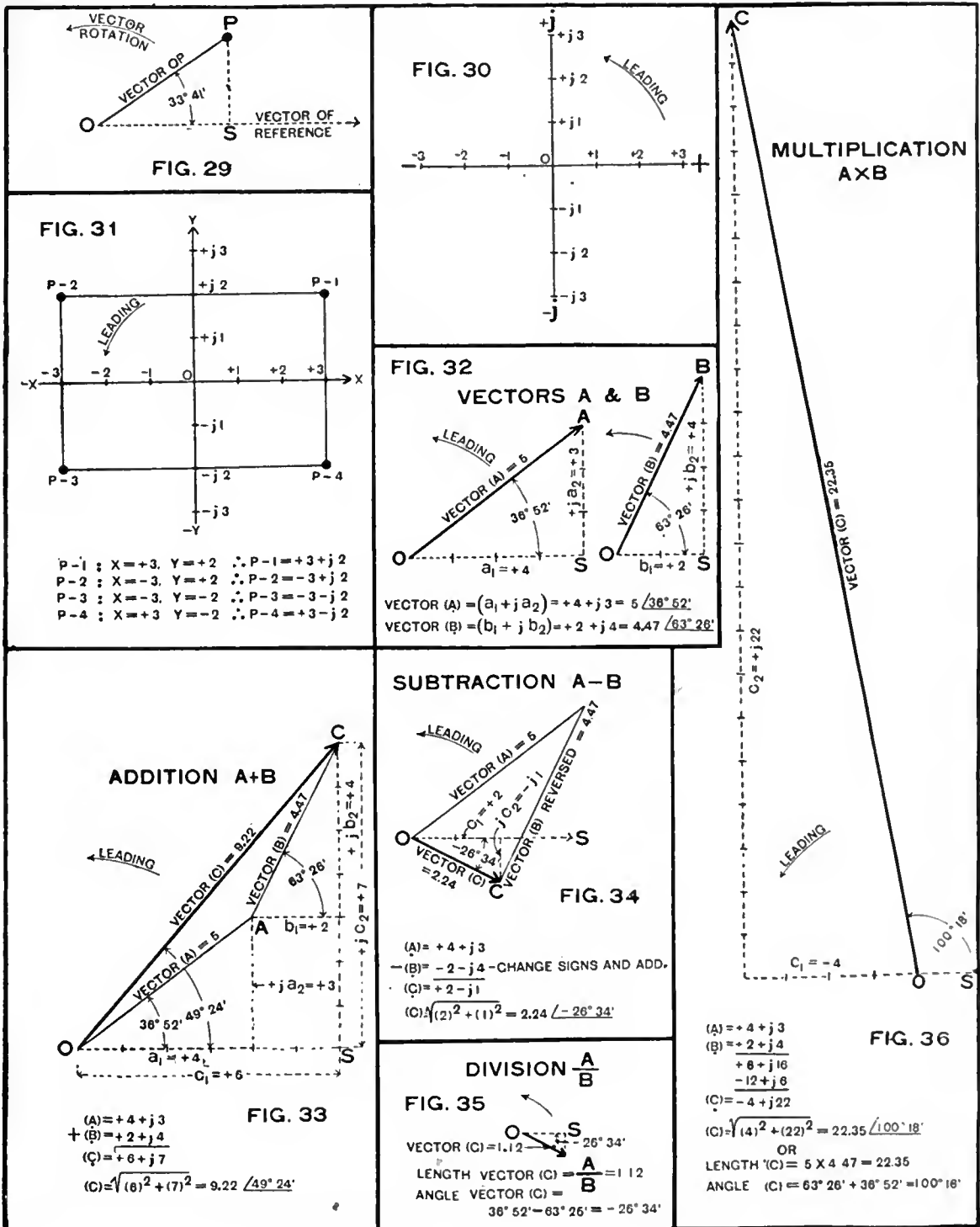
In order to illustrate the handling of complex quantities, the various operations of addition, subtraction, multiplication, division, evolution and involution of the vectors A and B in Fig. 32, will be performed.

*For an extended explanation of j terms, reference is made to Dr. Charles P. Steinmetz's "Engineering Mathematics", and Dr. A. E. Kennelly's "Artificial Electric Lines."

Addition—Fig. 33 illustrates the addition of these vectors expressed in rectangular co-ordinates. The resulting vector will have as its real component, the algebraic

sum of the reals, and as its imaginary component, the algebraic sum of the imaginaries. Thus,—

Subtraction—Fig. 34 illustrates the subtraction $A - B$ —



FIGS. 29 TO 36—EXAMPLES OF VECTOR SOLUTIONS

braic sum of the reals, and as its imaginary component, the algebraic sum of the imaginaries. Thus:

$$\begin{aligned} A &= +4 + j3 \\ + B &= +2 + j4 \\ \hline A + B &= C = +6 + j7 \\ C &= \sqrt{(6)^2 + (7)^2} = 9.22 \text{ absolute.} \end{aligned}$$

The resulting vector has, therefore, a size of 9.22

B. This is simply addition after the signs of both of the components of the vector to be subtracted have been reversed. Thus,—

$$\begin{aligned} A &= +4 + j3 \\ - B &= -2 - j4 \\ \hline A - B &= C = +2 - j1 \\ C &= \sqrt{(2)^2 + (1)^2} = 2.24 \text{ absolute.} \end{aligned}$$

The resulting vector C has therefore a size of 2.24 units and a slope of $-26^\circ 34'$. In polar co-ordinates, $C = 2.24 \angle 260^\circ 34'$.

Division—To divide one plane vector by another, divide their sizes and subtract their slopes, Fig. 35. Thus,—

$$\text{Absolute value of } C = \frac{5}{4.47} = 1.12$$

Angle of inclination of $C = 360^\circ 52' - 63^\circ 26' = -26^\circ 34'$ in the negative direction. In polar co-ordinates $C = 1.12 \angle 260^\circ 34'$.

Multiplication—Fig. 36 illustrates the multiplication of the vectors A and B . Here the rules of algebra also apply, except that when two j terms are multiplied signs are assigned opposite to those which would be used in the ordinary solution of an algebraic problem. This is for the reason that,—

$$j = \sqrt{-1}$$

hence, $j^2 = -1$

Hence where j^2 occurs it is replaced by its value -1 and therefore,—

$$\begin{aligned} -j \times j &= +1 \\ j^2 &= -1 \\ j^4 &= +1 \\ j^6 &= +j, \text{ etc.} \end{aligned}$$

Thus, to get the product of A and B :—

$$\begin{array}{r} A = +4 + j3 \\ B = +2 + j4 \\ \hline +8 + j6 \\ -12 + j16 \end{array}$$

$$A \times B = C = -4 + j22 = 22.35 \text{ absolute}$$

The resulting vector C has therefore a size of 22.35 units and is inclined in the positive direction at an angle of $100^\circ 18'$ to the vector of reference. The polar expression is $C = 22.35 \angle 100^\circ 18'$

The magnitude and position of the product may be also determined by multiplying the sizes of the vectors and adding their slopes. Thus :—

$$\begin{aligned} \text{Size of } C &= 5 \times 4.47 = 22.35 \text{ (as above)} \\ \text{Slope of } C &= 63^\circ 26' + 36^\circ 52' = 100^\circ 18'. \end{aligned}$$

Involution—Involution is multiple multiplication. To obtain the power of a plane vector, find the power of the polar value and multiply the angle by the power to which the vector is to be raised. Thus,—vector $A = 5 \angle 36^\circ 52'$; and $(5 \angle 36^\circ 52')^2 = 5^2 \angle 73^\circ 44' = 25 \angle 73^\circ 44'$.

Evolution—To find the root of a polar plane vector, find the root of the polar value and then divide the slope by the root desired. Thus vector $A = 5 \angle 36^\circ 52'$; and $\sqrt[3]{5 \angle 36^\circ 52'} = 2.236 \angle 18^\circ 26'$.

SOLUTION BY CONVERGENT SERIES

The hyperbolic formula for determining the operating characteristics of a transmission circuit in which exact account is taken of all the electric properties of the circuit is frequently expressed in the following form,—

$$E_s = E_r \cosh \sqrt{ZY} + I_r \sqrt{\frac{Z}{Y}} \sinh \sqrt{ZY} \dots (51)$$

$$I_s = I_r \cosh \sqrt{ZY} + E_r \frac{1}{\sqrt{\frac{Z}{Y}}} \sinh \sqrt{ZY} \dots (52)$$

Since \sqrt{ZY} is complex, the hyperbolic functions of complex quantities are required in solving these equations.

In above formula, expressed in hyperbolic language, the three auxiliary constants A , B and C which take into account the "distributed" nature of the circuit are represented by the quantities—

$$A = \cosh \sqrt{ZY} \dots (53)$$

$$B = \sqrt{\frac{Z}{Y}} \sinh \sqrt{ZY} \dots (54)$$

$$C = \frac{1}{\sqrt{\frac{Z}{Y}}} \sinh \sqrt{ZY} \dots (55)$$

Equations (51) and (52) above may therefore be expressed in terms of the auxiliary constants, A , B and C , as follows :—

$$E_s = E_r A + I_r B \dots (56)$$

$$I_s = I_r A + E_r C \dots (57)$$

$$\text{or } E_r = E_s A - I_s B \dots (58)$$

$$I_r = I_s A - E_s C \dots (59)$$

These three auxiliary constants may be calculated by convergent series as follows :—

$$A = \left[1 + \frac{YZ}{2} + \frac{Y^2 Z^2}{24} + \frac{Y^3 Z^3}{720} + \frac{Y^4 Z^4}{40320} + \text{etc.} \right] \dots (60)$$

$$B = Z \left[1 + \frac{YZ}{6} + \frac{Y^2 Z^2}{120} + \frac{Y^3 Z^3}{5040} + \frac{Y^4 Z^4}{362880} + \text{etc.} \right] \dots (61)$$

$$C = Y \left[1 + \frac{YZ}{6} + \frac{Y^2 Z^2}{120} + \frac{Y^3 Z^3}{5040} + \frac{Y^4 Z^4}{362880} + \text{etc.} \right] \dots (62)$$

The above series are simply expressions for the auxiliary constants as previously stated. These constants are functions of the physical properties of the circuit and of the frequency only, and not of the voltage or the current. After the values for the auxiliary constants have been calculated for a given circuit and frequency their numerical values may be applied directly to any numerical values of E and I for which a solution is desired. From this point on, the performance of the circuit may be determined either by the graphical method previously described or by mathematical calculation.

Any degree of accuracy may be obtained by the use of convergent series for determining the auxiliary constants, by simply using a sufficient number of terms in the series. The rapidity of convergence of these series is dependent upon the value of the argument ZY and thus upon the square of the length of the circuit and frequency, and also, to a lesser extent upon the product of total circuit conductance and total circuit resistance.

As far as calculations based upon the more or less uncertain values of the fundamental constants of the circuit are concerned, the use of three terms in the series expression yields results in a 300 mile circuit which are sufficiently close to the exact values as given by the use of hyperbolic functions (infinite number of terms). In the case of shorter circuits two terms will give a high degree of accuracy. The number of terms necessary will be determined while doing the work, for it is usual to figure out the terms of the series until they become too small to be considered when added to $\frac{YZ}{2}$ or $\frac{Y Z}{6}$.

In Table N are given values for the auxiliary constants (expressed in rectangular co-ordinates) illustrating the convergence of the series for a 300 mile, 60 cycle circuit (Problem X), the complete calculation of which will follow.

Table N shows that even for a 60 cycle, 300 mile circuit, three terms give sufficiently accurate results for determining constant A, whereas two terms are sufficient for determining constants B and C. This is on account of the slower convergence of the hyperbolic cosine series.

TABLE N—CONVERGENT SERIES TERMS FOR PROBLEM X.

No. of Terms	Constant A	Constant B	Constant C
1	1.000000 + j 0.000000	105 + j 249	0 + j 0.001563
2	+ 0.805407 + j 0.082057	+ 91.3788 + j 235.7211	- 0.000043 + j 0.001462
3	+ 0.810596 + j 0.076735	+ 91.7527 + j 235.8678	- 0.000041 + j 0.001463
4	+ 0.810558 + j 0.076832	+ 91.7486 + j 235.8680	- 0.000041 + j 0.001463
Infinite	+ 0.810558 + j 0.076831	+ 91.7486 + j 235.8680	- 0.000041 + j 0.001463

CALCULATION FOR THE AUXILIARY CONSTANTS BY CONVERGENT SERIES

The form of solution and procedure indicated in Chart XI for the calculation of the auxiliary constants by convergent series is suggested as being complete and easy to follow.

First the physical characteristics of the circuit and the frequency are stated. These are the only features having any bearing upon the value of the auxiliary constants for a given circuit. The voltage and current to be transmitted do not affect these constants. The resistance, reactance, conductance, and susceptance to neutral per mile are ascertained from the tables for one conductor of the circuit. These values are then multiplied by the length of the circuit in miles and set down as total per conductor.

The values of Y and Z must now be set down for the problem in the form of complex quantities. Thus $Z = R + jX = 105 + j249$ and $Y = G + jB = 0 + j0.001563$ since zero leakage conductance has been assumed for this case. Conductance G represents the true power loss in the form of leakage over insulators and of corona loss through the air between conductors. Corona loss corresponding to the assumed atmospheric conditions may be estimated by applying Peek's formula (See Chapter IV on Corona). Insulator

leakage may be approximated from the most suitable test data available. It is general practice in the solution of all but the very longest high-voltage circuits to ignore the effect of the losses due to leakage and corona effect. These losses will be ignored in this case, so that G becomes zero. After Z and Y have been written down in the form of complex quantities the product YZ should be found as previously described for the multiplication of complex quantities. The second, third and fourth power of YZ may then be found, if desired. Chart XI shows the fourth power, but on all but the longest circuits a total of four terms will be sufficient, and for most problems three terms will give sufficient accuracy. The range of accuracy has been previously indicated for a 300 mile circuit on the basis of any number of terms being used up to and including infinity. The values in Chart XI are carried out to six decimal places whereas four places will usually give sufficient accuracy for calculating the values of the constants A and B. The smallness of the value of constant C may make six places desirable when calculating its value.

After the values of YZ, Y² Z², Y³ Z³ etc., have been calculated they are divided by 2, 24, 720 etc., respectively, set down and added to 1. This gives the value of the auxiliary constant A, as + 0.810558 + j 0.076831 which is also referred to as a₁ + ja₂. The absolute value of the constant A = 0.8142 is simply the square root of the sum of the square of a₁ and a₂. The polar value of A is thus 0.8142 / 5° 24' 53".

The solution for the constant B is of the same general form as the solution for the constant A, except that the values of YZ, Y² Z², and Y³ Z³ etc., are divided by 6, 120 and 5040 respectively. After these results are added to 1 they are multiplied by Z, the product being the value of the auxiliary constant B or b₁ + jb₂. The absolute value of B is obtained in the same manner as the absolute value of A.

The solution for C is the same as for B except that in place of the constant B series being multiplied by Z it is multiplied by Y and the values of C or c₁ + jc₂ obtained.

AUXILIARY CONSTANTS OF VARIOUS CIRCUITS

In Chart XII are tabulated exact values for the auxiliary constants for the 64 problems to which frequent reference will be made. These auxiliary constants have been calculated by convergent series, the results having been checked through the medium of three separate calculations made at different times. They are therefore believed exact to at least five significant digits. The results have been expressed in both rectangular and polar co-ordinates.

CALCULATIONS OF PERFORMANCE

In Chart XIII is given the complete calculation of the electrical performance for problem X, starting with

CHART XI—EXAMPLE ILLUSTRATING RIGOROUS SOLUTION FOR THE AUXILIARY CONSTANTS BY CONVERGENT SERIES FOR PROBLEM X.

PHYSICAL CHARACTERISTICS OF CIRCUIT - FREQUENCY

LENGTH, 300 MILES. CYCLES, 60.
 CONDUCTORS - #000 STRANDED COPPER.
 SPACING OF CONDUCTORS 10 X 10 X 20 FEET.
 EQUIVALENT DELTA SPACING = 12.6 FEET.

LINEAR LINE CONSTANTS

FROM TABLES - PER MILE

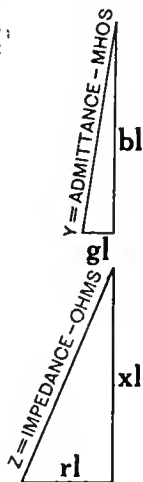
TABLE NO. 2, $r = .350$ OHM AT 25° C,
 TABLE NO. 5, $x = .830$ OHM (BY INTERPOLATION).
 TABLE NO. 10 $b = 6.21 \times 10^{-6}$ MHO (BY INTERPOLATION).
 $g =$ (IN THIS CASE TAKEN AS ZERO).

TOTAL PER CONDUCTOR

$R = rl = .350 \times 300 = 105$ OHMS TOTAL RESISTANCE.
 $X = xl = .830 \times 300 = 249$ OHMS TOTAL REACTANCE.
 $B = bl = 6.21 \times 300 \times 10^{-6} = .001563$ MHO TOTAL SUSCEPTANCE.
 $G = gl = 0 \times 300 = 0$ MHO TOTAL CONDUCTANCE.

MULTIPLICATION OF YZ

$$\begin{array}{r} Y = 0 + j .001563 \\ Z = 105 + j 249 \\ \hline = 0 \\ = -.389187 + j .164115 \\ \hline YZ = -.389187 + j .164115 \\ YZ = -.389187 + j .164115 \\ = + .151466 - j .063871 \\ = -.026934 - j .063871 \\ \hline Y^2 Z^2 = + .124532 - j .127742 \\ YZ = -.389187 + j .164115 \\ = -.048466 + j .020437 \\ = + .020964 + j .049715 \\ \hline Y^3 Z^3 = -.027502 + j .070152 \\ YZ = -.389187 + j .164115 \\ = + .010703 - j .004513 \\ = -.011513 - j .027302 \\ \hline Y^4 Z^4 = .000810 - j .031815 \end{array}$$



NOTE

THE AUXILIARY CONSTANTS OF THE CIRCUIT (A) (B) & (C) MAY BE OBTAINED GRAPHICALLY FROM THE WILKINSON CHARTS

SOLUTION FOR (A)

$$(A) = \left[1 + \frac{YZ}{2} + \frac{Y^2 Z^2}{24} + \frac{Y^3 Z^3}{720} + \frac{Y^4 Z^4}{40,320} \right]$$

1.000000

$$\frac{YZ}{2} = -.194593 + j .082057$$

$$\frac{Y^2 Z^2}{24} = + .005189 - j .005322$$

$$\frac{Y^3 Z^3}{720} = -.000038 + j .000097$$

$$\frac{Y^4 Z^4}{40,320} = -.000000 - j .000001$$

$$(A) = + .810568 + j .076831$$

$$= (a_1 + j a_2)$$

$$= 0.8142 \angle 5^\circ 24' 53''$$

SOLUTION FOR (B)

$$(B) = Z \left[1 + \frac{YZ}{6} + \frac{Y^2 Z^2}{120} + \frac{Y^3 Z^3}{5,040} + \frac{Y^4 Z^4}{362,880} \right]$$

1.000000

$$\frac{YZ}{6} = -.064883 + j .027352$$

$$\frac{Y^2 Z^2}{120} = + .001038 - j .001064$$

$$\frac{Y^3 Z^3}{5,040} = -.000005 + j .000014$$

$$\frac{Y^4 Z^4}{362,880} = -.000000 - j .000000$$

$$(B) = Z (+ .93617 + j .026302)$$

$$Z = 105 + j 249$$

$$+ 98.2978 + j 233.1063$$

$$- 6.5492 + j 2.7617$$

$$(B) = + 91.7486 + j 235.8680$$

$$= (b_1 + j b_2)$$

$$= 253.083 \angle 68^\circ 44' 41'' \text{ OHMS}$$

SOLUTION FOR (C)

$$(C) = Y \left[1 + \frac{YZ}{6} + \frac{Y^2 Z^2}{120} + \frac{Y^3 Z^3}{5,040} + \frac{Y^4 Z^4}{362,880} \right]$$

$$(C) = Y (+ 93617 + j 026302)$$

$$Y = 0 + j 001563$$

$$(C) = -.000041 + j 001463$$

$$= (c_1 + j c_2)$$

$$= .001464 \angle 91^\circ 36' 18'' \text{ MHO}$$

the values for the auxiliary constants and the receiving end load conditions known. The calculations are carried out by the employment of complex numbers, the complete performance being calculated for both load and zero load conditions. In order to give a more clear understanding of these mathematical operations the reader is referred to the vector diagrams of Fig. 37.

In Chart XIII are given the formulas for determining the E_s and I_s values under load conditions. On Fig. 37 these two same formulas are given, but in the form of vector diagrams, upon which vectors the numerical values corresponding to problem X are stated. With the numerical values of the vectors and angles stated, it should be a comparatively simple manner to

follow graphically (Fig. 37) the mathematical calculations shown in Chart XIII.

The formulas for E_s and I_s which are stated in Chart XIII and in Fig. 37 contain a complex number ($\cos \theta_r \pm j \sin \theta_r$) not previously stated in connection with the fundamental hyperbolic formulas for long circuits. The formulas previously given were based upon unity power-factor. The introduction of this new complex number is made necessary in order that the effect of the power-factor of the load current may be included in the calculations. The function of this new complex number is to rotate the current vector through an angle corresponding to the power-factor of the load current. It will be referred to as the rotating triangle. If the

CHART XII—AUXILIARY CONSTANTS OF VARIOUS CIRCUITS

PROBLEM NO.	LENGTH OF CIRCUIT—(MILES)	CONDUCTORS	SPACING—DELTA	LINEAR CONSTANTS				AUXILIARY CONSTANTS OF CIRCUIT					
				TOTAL PER CONDUCTOR ★				THESE AUXILIARY CONSTANTS TAKE INTO ACCOUNT THE EFFECT OF DISTRIBUTED CAPACITANCE. THEY HAVE BEEN CALCULATED RIGOROUSLY BY CONVERGENT SERIES					
				r _l	x _l	b _l	g _l	CONSTANT (A)		CONSTANT (B)		CONSTANT (C)	
				FROM TABLE NO 2	FROM TABLE NOS 4 & 5	FROM TABLE NOS 9 & 10		a ₁	a ₂	b ₁	b ₂	c ₁	c ₂
25 CYCLES													
1	20	0000 COPPER	3	5.54	5.36	57.2	0	.999847 + j.000158	5.5394 + j.53600	0	+ j.000057		
2	"	"	"	"	"	"	0	= .999847 / 0° 0' 32"	= 7.7081 / 44° 3' 27"	= .000057 / 90° 0' 0"			
3	"	"	3	5.54	5.36	57.2	0	.999847 + j.000158	5.5394 + j.53600	0	+ j.000057		
4	"	"	"	"	"	"	0	= .999847 / 0° 0' 32"	= 7.7081 / 44° 3' 27"	= .000057 / 90° 0' 0"			
5	30	0000 COPPER	4	8.31	8.5	81.0	0	.999656 + j.000336	8.3082 + j.84999	0	+ j.000081		
6	"	"	"	"	"	"	0	= .999656 / 0° 1' 10"	= 11.886 / 45° 39' 12"	= .000081 / 90° 0' 0"			
7	"	"	4	8.31	8.5	81.0	0	.999656 + j.000336	8.3082 + j.84999	0	+ j.000081		
8	"	"	"	"	"	"	0	= .999656 / 0° 1' 10"	= 11.886 / 45° 39' 12"	= .000081 / 90° 0' 0"			
9	50	0000 COPPER	4	13.85	14.1	135	0	.999048 + j.000935	13.841 + j.140996	0	+ j.000135		
10	"	"	"	"	"	"	0	= .999048 / 0° 3' 12"	= 19.757 / 45° 31' 44"	= .000135 / 90° 0' 0"			
11	"	"	6	13.85	15.1	125	0	.999056 + j.000866	13.8413 + j.150991	0	+ j.000125		
12	"	"	"	"	"	"	0	= .999056 / 0° 2' 58"	= 20.4833 / 47° 29' 20"	= .000125 / 90° 0' 0"			
13	100	0000 COPPER	9	27.7	32.2	233	0	.996248 + j.003224	27.6307 + j.321894	0	+ j.000233		
14	"	"	"	"	"	"	0	= .996253 / 0° 11' 7"	= 42.4218 / 49° 21' 28"	= .000233 / 90° 0' 0"			
15	"	"	11	27.7	33.2	226	0	.996249 + j.003126	27.6308 + j.331874	0	+ j.000226		
16	"	"	"	"	"	"	0	= .996254 / 0° 10' 47"	= 43.1841 / 50° 13' 13"	= .000226 / 90° 0' 0"			
17	200	300M COPPER	11	39.2	64.8	464	0	.984991 + j.009049	38.808 + j.64594	-0.00001 + j.000462			
18	"	"	"	"	"	"	0	= .985033 / 0° 31' 35"	= 75.356 / 59° 0' 10"	= .000462 / 90° 7' 27"			
19	"	"	17	39.2	69.2	434	0	.985009 + j.008464	38.8084 + j.68965	-0.00001 + j.000432			
20	"	"	"	"	"	"	0	= .985050 / 0° 29' 31"	= 79.134 / 60° 37' 58"	= .000432 / 90° 7' 54"			
21	300	636M ALUM.	11	44.1	91.2	747	0	.966085 + j.016285	43.1033 + j.90408	-0.00004 + j.000739			
22	"	"	"	"	"	"	0	= .966222 / 0° 57' 1"	= 100.157 / 64° 30' 36"	= .000739 / 90° 17' 10"			
23	"	"	21	44.1	101	672	0	.966219 + j.014650	43.1070 + j.100077	-0.00003 + j.000664			
24	"	"	"	"	"	"	0	= .966330 / 0° 52' 6"	= 108.966 / 66° 41' 48"	= .000664 / 90° 15' 24"			
25	400	636M ALUM.	17	58.8	130	928	0	.940161 + j.026738	56.4555 + j.127927	-0.00008 + j.000909			
26	"	"	"	"	"	"	0	= .940541 / 1° 37' 45"	= 139.83 / 66° 11' 16"	= .000909 / 90° 30' 14"			
27	"	"	21	58.8	134	896	0	.940452 + j.025819	56.4664 + j.131842	-0.00008 + j.000878			
28	"	"	"	"	"	"	0	= .940801 / 1° 34' 20"	= 143.425 / 66° 48' 54"	= .000878 / 90° 31' 18"			
29	500	636M ALUM.	17	73.5	163	1160	0	.906642 + j.041299	68.928 + j.158928	-0.00016 + j.001124			
30	"	"	"	"	"	"	0	= .907583 / 2° 36' 14"	= 173.23 / 66° 33' 13"	= .001124 / 90° 48' 56"			
31	"	"	21	73.5	168	1120	0	.907109 + j.039880	68.9507 + j.16376	-0.00015 + j.001085			
32	"	"	"	"	"	"	0	= .907985 / 2° 31' 2"	= 177.684 / 67° 10' 0"	= .001085 / 90° 47' 33"			
60 CYCLES													
33	20	0000 COPPER	3	5.54	12.88	137	0	.999118 + j.000379	5.53675 + j.128769	0	+ j.000137		
34	"	"	"	"	"	"	0	= .999118 / 0° 1' 18"	= 14.0167 / 66° 44' 0"	= .000137 / 90° 0' 0"			
35	"	"	3	5.54	12.88	137	0	.999118 + j.000379	5.53675 + j.128769	0	+ j.000137		
36	"	"	"	"	"	"	0	= .999118 / 0° 1' 18"	= 14.0167 / 66° 44' 0"	= .000137 / 90° 0' 0"			
37	30	0000 COPPER	4	8.31	20.4	195	0	.998011 + j.00081	8.299 + j.203887	0	+ j.000195		
38	"	"	"	"	"	"	0	= .998011 / 0° 2' 47"	= 22.014 / 67° 51' 6"	= .000195 / 90° 0' 0"			
39	"	"	4	8.31	20.4	195	0	.998011 + j.00081	8.299 + j.203887	0	+ j.000195		
40	"	"	"	"	"	"	0	= .998011 / 0° 2' 47"	= 22.014 / 67° 51' 6"	= .000195 / 90° 0' 0"			
41	50	0000 COPPER	4	13.85	34.0	324	0	.994496 + j.002239	13.7992 + j.339479	0	+ j.000323		
42	"	"	"	"	"	"	0	= .994498 / 0° 7' 33"	= 36.645 / 67° 52' 45"	= .000323 / 90° 0' 0"			
43	"	"	6	13.85	36.4	301	0	.994526 + j.002081	13.7994 + j.363432	0	+ j.000300		
44	"	"	"	"	"	"	0	= .994528 / 0° 7' 10"	= 38.874 / 67° 12' 30"	= .000300 / 90° 0' 0"			
45	100	0000 COPPER	9	27.7	77.4	562	0	.97832 + j.007728	27.2996 + j.769116	-0.00001 + j.000558			
46	"	"	"	"	"	"	0	= .97835 / 0° 27' 10"	= 81.6129 / 70° 27' 36"	= .000558 / 90° 6' 11"			
47	"	"	11	27.7	79.7	542	0	.97847 + j.007452	27.302 + j.791963	-0.00001 + j.000538			
48	"	"	"	"	"	"	0	= .978498 / 0° 26' 14"	= 83.77 / 70° 58' 30"	= .000538 / 90° 6' 27"			
49	200	300M COPPER	11	39.2	156	1116	0	.914128 + j.021243	36.9541 + j.151791	-0.00008 + j.001084			
50	"	"	"	"	"	"	0	= .914375 / 1° 19' 31"	= 156.224 / 76° 19' 2"	= .001084 / 90° 25' 23"			
51	"	"	17	39.2	166	1044	0	.914524 + j.019876	36.9641 + j.161507	-0.00007 + j.001014			
52	"	"	"	"	"	"	0	= .914740 / 1° 14' 40"	= 165.69 / 77° 6' 31"	= .001014 / 90° 23' 43"			
53	300	636M ALUM.	11	44.1	220	1794	0	.808816 + j.037006	38.4655 + j.206359	-0.00023 + j.001678			
54	"	"	"	"	"	"	0	= .809662 / 2° 37' 0"	= 209.913 / 79° 26' 28"	= .001678 / 90° 47' 8"			
55	"	"	21	44.1	243	1614	0	.810022 + j.033307	38.5002 + j.227918	-0.00018 + j.001510			
56	"	"	"	"	"	"	0	= .810701 / 2° 21' 14"	= 231.147 / 80° 24' 43"	= .001510 / 90° 4' 6"			
57	400	636M ALUM.	17	58.8	314	2212	0	.671701 + j.057759	45.8726 + j.28004	-0.00044 + j.001958			
58	"	"	"	"	"	"	0	= .674179 / 4° 54' 54"	= 283.77 / 80° 41' 50"	= .001958 / 91° 18' 0"			
59	"	"	21	58.8	322	2152	0	.672455 + j.056208	45.9013 + j.287194	-0.00042 + j.001912			
60	"	"	"	"	"	"	0	= .674800 / 4° 46' 39"	= 290.839 / 80° 55' 10"	= .001912 / 91° 15' 2"			
61	500	636M ALUM.	17	73.5	390	2785	0	.502772 + j.084790	48.9614 + j.325247	-0.00085 + j.002307			
62	"	"	"	"	"	"	0	= .509871 / 9° 34' 20"	= 328.912 / 81° 26' 21"	= .002307 / 92° 6' 32"			
63	"	"	21	73.5	402	2690	0	.504852 + j.081969	49.061 + j.335414	-0.00079 - j.002230			
64	"	"	"	"	"	"	0	= .511463 / 9° 13' 12"	= 338.98 / 81° 40' 43"	= .002230 / 92° 11' 45"			

*r_l is the resistance in ohms at 25° C (77° F), x_l the reactance in ohms, b_l the susceptance in micromhos to neutral (multiply by 10⁻⁶ to convert to mhos). The x and b values for the 63600 circ. mil aluminum cable were taken as those of 70000 circ. mil copper on the assumption that these two conductors would have approximately the same diameter. g_l the loss resulting from leakage over insulators and from corona has, for simplicity, been assumed as zero.

CHART XIII—RIGOROUS CALCULATION OF PERFORMANCE WHEN RECEIVING END CONDITIONS ARE FIXED

$KV-A_R = 18\ 000.$ $KW_R = 18\ 200.$ $E_R = 104\ 000$ VOLTS 3 PHASE. $PF_R = 90.00\%$ LAGGING.

PER PHASE TO NEUTRAL

$KV-A_{RN} = \frac{18\ 000}{3} = 6\ 000.$ $KW_{RN} = \frac{16\ 200}{3} = 5\ 400.$ $E_{RN} = \frac{104\ 000}{1.732} = 60\ 048.$ $I_R = \frac{6\ 000 \times 1\ 000}{60\ 046} = 99.92$ AMPERES.

AUXILIARY CONSTANTS OF CIRCUIT

(A) = $+ .810558 + j .076831$ (B) = $+91.7486 + j\ 235.868$ (C) = $- .000041 + j .001463$
 = $(a_1 + j a_2)$ = $(b_1 + j b_2)$ = $(C_1 + j C_2)$
 = $.8142 / 5^\circ 24' 53''$ = $253.083 / 68^\circ 44' 41''$ OHMS = $.001464 / 91^\circ 36' 18''$ MHO

SOLUTION FOR E_S LOAD CONDITIONS

SOLUTION FOR I_S

$E_S = E_R(a_1 + j a_2) + I_R(\cos \theta_R \pm j \sin \theta_R)(b_1 + j b_2) \star$

$I_S = I_R(\cos \theta_R \pm j \sin \theta_R)(a_1 + j a_2) + E_R(C_1 + j C_2) \star$

$\star \pm$ THIS SIGN IS MINUS WHEN THE P. F. IS LAGGING AND PLUS WHEN THE P. F. IS LEADING

$(a_1 + j a_2) = + .810558 + j .076831$
 $\times E_{RN} = \frac{60046}{}$
 $E_{RN}(a_1 + j a_2) = + 48871 + j 4613$
 $(\cos \theta_R - j \sin \theta_R) = + .9 - j .438$
 $\times I_R = \frac{99.92}{}$
 $I_R(\cos \theta_R - j \sin \theta_R) = + 89.93 - j 43.56$
 $\times (b_1 + j b_2) = + 91.75 + j 235.87$
 $+ 8251 + j 21212$
 $+ 10274 - j 3997$
 $I_R(\cos \theta_R - j \sin \theta_R)(b_1 + j b_2) = + 18525 + j 17215$
 $+ E_{RN}(a_1 + j a_2) = + 48671 + j 4613$
 $E_{SN} = + 67196 + j 21828$
 $= \sqrt{(67196)^2 + (21828)^2}$
 $E_{SN} = 70\ 652$ VOLTS TO NEUTRAL.

$I_R(\cos \theta_R - j \sin \theta_R) = + 89.93 - j 43.56$
 $\times (a_1 + j a_2) = + .810558 + j .076831$
 $+ 72.993 + j 6.909$
 $+ 3.347 - j 35.308$
 $I_R(\cos \theta_R - j \sin \theta_R)(a_1 + j a_2) = + 76.240 - j 28.399$
 $(C_1 + j C_2) = - .000041 + j .001463$
 $\times E_{RN} = \frac{60046}{}$
 $E_{RN}(C_1 + j C_2) = - 2.462 + j 87.85$
 $+ I_R(\cos \theta_R - j \sin \theta_R)(a_1 + j a_2) = + 76.240 - j 28.399$
 $I_S = + 73.778 + j 59.451$
 $= \sqrt{(73.778)^2 + (59.451)^2}$
 $I_S = 94.75$ AMPERES.

$KW_{SN} = (87.188 \times 73.778) + (21.828 \times 59.451) = 6,255$ KW PER PHASE.

EFFICIENCY = $\frac{5\ 400 \times 100}{6\ 255} = 86.33\%$.

$KV-A_{SN} = (70.662 \times 94.75) = 6\ 694$ KV-A PER PHASE.

$PF_S = \frac{6\ 255 \times 100}{6\ 694} = 93.42\%$ LEADING.

LOSS = $6255 - 5400 = 855$ KW PER PHASE.

PHASE ANGLES— AT FULL LOAD THE VOLTAGE AT THE SENDING END LEADS THE VOLTAGE AT THE RECEIVER END BY THE ANGLE

$\tan^{-1} \frac{21\ 828}{67\ 196} = \tan^{-1} 1.325 = 53^\circ 00'$, AND THE CURRENT AT THE SENDING-END LEADS THE VOLTAGE AT THE RECEIVING-END BY THE ANGLE

$\tan^{-1} \frac{59.451}{73.778} = \tan^{-1} 1.806 = 38^\circ 52'$, HENCE THE CURRENT AT THE SENDING-END LEADS THE VOLTAGE AT THE SENDING-END BY THE ANGLE $38^\circ 52'$

— ANGLE $18^\circ 00' = 20^\circ 52'$. THE POWER-FACTOR AT THE SENDING-END IS THEREFORE $\cos 20^\circ 52' = 93.42\%$ LEADING AT LOAD SPECIFIED.

ZERO LOAD CONDITIONS

$E_{SNO} = 48671 + j 4613$
 $= \sqrt{(48671)^2 + (4613)^2}$

$I_{SNO} = - 2.462 + j 87.85$
 $= \sqrt{(- 2.462)^2 + (87.85)^2}$

$KW_{SNO} = (48.671 \times - 2.462) + (46.13 \times 87.85) = 285.43$ KW PER PHASE.

$KV-A_{SNO} = 48.889 \times 87.89 = 4\ 297$ KV A PER PHASE.

$E_{SNO} = 48\ 889$ VOLTS.

$I_{SNO} = 87.89$ AMPERES.

$PF_{SO} = \frac{285.43 \times 100}{4.297} = 6.64\%$ LEADING.

REGULATION

A RISE IN VOLTAGE AT THE SENDING-END OCCURS OF $70\ 652 - 48\ 889 = 21\ 763$ VOLTS TO NEUTRAL WHEN THE LOAD IS INCREASED FROM ZERO TO 99.92 AMPERES AT 90% POWER FACTOR LAGGING AT THE RECEIVER END WITH CONSTANT VOLTAGE AT THE RECEIVING END.

PHASE ANGLES

AT ZERO LOAD THE VOLTAGE AT THE SENDING-END LEADS THE VOLTAGE AT THE RECEIVER END BY THE ANGLE $\tan^{-1} \frac{4613}{48\ 671} =$

$\tan^{-1} .0947 = 5^\circ 25'$ AND THE CURRENT AT THE SUPPLY END LEADS THE VOLTAGE AT THE RECEIVER END BY THE ANGLE $\tan^{-1} \frac{87.85}{- 2.462} =$

$\tan^{-1} (-35.71) = 91^\circ 36'$ —HENCE THE CURRENT AT THE SUPPLY END LEADS THE VOLTAGE AT THE SUPPLY END BY THE ANGLE $(91^\circ 36') - (5^\circ 25') =$

$86^\circ 11'$. THE POWER FACTOR AT THE SENDING-END IS THEREFORE $\cos 86^\circ 11' = 6.64\%$ LEADING AT ZERO LOAD.

load power-factor is 100 percent, this rotating triangle will equal $r \pm j 0$, hence it has no effect or power to rotate. If the power-factor of the load is 80 percent the rotating triangle would have a numerical value of $0.8 \pm j 0.6$.

The various phase angles given in Chart XIII show whether the power-factor at the supply end is leading or lagging. These various phase angles are given to make the discussion complete. Actually, in order to determine whether the power-factor at the supply end is leading or lagging, it is only necessary to note if the supply end

current vector leads or lags behind the supply end voltage vector. At the lower end of Fig. 37 combined current and voltage vectors are shown for this problem, corresponding to both load and zero load conditions.

In Chart XIV is given a complete calculation of the electrical performance of problem X, starting with the values for the auxiliary constants and the sending end load condition known. In other words the supply end conditions which were derived by calculation in Chart XIII have in this case been assumed as fixed, and the receiver end conditions calculated. The reason that

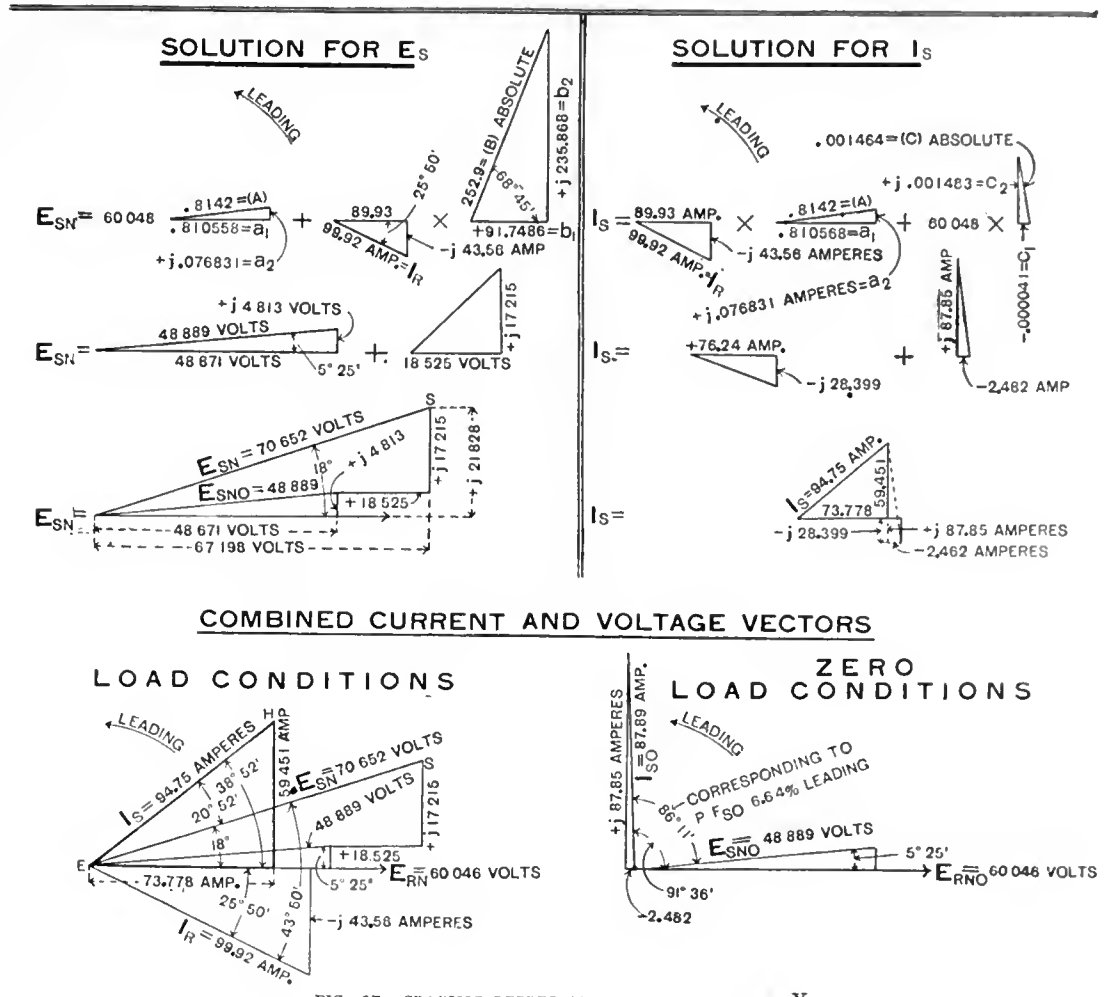


FIG. 37—GRAPHIC REPRESENTATION OF PROBLEM X
 Illustrating rigorous calculations of performance when receiving end conditions are fixed.

there is a slight difference between the receiving end conditions as calculated on Chart XIV and the known receiving end conditions is that the value for the sine in the rotating triangle (0.436) in chart XIII was carried out to only three places, whereas in Chart XIV it was carried out to four places. If the values for the rotating triangles had been carried out to five or six places in the calculations in both charts, the receiving end conditions would have checked exactly.

TERMINAL VOLTAGES AT ZERO LOAD

For a given circuit and frequency, the relation of the voltage at the two ends of the circuit is fixed. The ratio of sending end to the receiving end voltage is expressed by the constant A . The ratio of receiving to sending end voltage is expressed by $\frac{1}{A}$. For example, problem X, the sending end voltage under load is 70 652 volts. If the load is thrown off, and this sending end voltage is maintained constant at 70 652 volts, the receiving end voltage will rise to a value of $\frac{70\ 652}{0.8142} = 86\ 775$ volts to neutral. The rise in percent of sending end voltage is therefore $\frac{100 \times 86\ 775 - 70\ 652}{70\ 652} = 22.82$ percent.

PERFORMANCE OF VARIOUS CIRCUITS

In Chart XV is tabulated the complete performance of the 64 problems for which the auxiliary constants are tabulated in Chart XII. The auxiliary constants in Chart XII were applied to the fixed load conditions as stated in Chart XV for the receiving end, and both load and zero load conditions at the sending end were calculated and tabulated.

The object of calculating and tabulating the values for the 64 problems was two fold. First to obtain data on 25 and 60 cycle problems covering a wide range which would provide a basis for constructing curves, illustrating the effect that distance in transmission has upon the performance of circuits and upon the auxiliary constants of the circuit. Second, to give the student a wide range of problems from which he could choose, and from which he could start with the tabulated values as fixed at either end and calculate the conditions at the other end. It is believed that such problems will furnish very profitable practice for the student and will also serve as a general guide when making calculations on problems of similar length and fundamental or lineal constants. It is not intended that the figures given for longer circuits, included in these tabulations, shall coincide with ordinary conditions encountered in practice.

CHART XIV—RIGOROUS CALCULATION OF PERFORMANCE WHEN SENDING END CONDITIONS ARE FIXED

$KV-A_S = 20\ 082.$ $KW_S = 18\ 785.$ $E_S = 122\ 369$ VOLTS 3 PHASE. $PF_S = 93.42\%$ LEADING.

PER PHASE TO NEUTRAL

$KV-A_{SN} = \frac{20\ 082}{3} = 6\ 694.$ $KW_{SN} = \frac{18\ 785}{3} = 6\ 265.$ $E_{SN} = \frac{122\ 369}{1.732} = 70\ 852.$ $I_S = \frac{6\ 694 \times 1000}{70\ 852} = 94.75$ AMPERES.

AUXILIARY CONSTANTS OF CIRCUIT

(A) = +.810558 + j.076831 (B) = +91.7486 + j 236.868 (C) = -.000041 + j.001463
 = (a₁ + j a₂) = (b₁ + j b₂) = (C₁ + j C₂)
 = .8142 / 5° 24' 53" = 253.083 / 68° 44' 41" OHMS = .001464 \ 91° 36' 18" MHO

SOLUTION FOR E_R LOAD CONDITIONS SOLUTION FOR I_R

$E_R = E_S(a_1 + j a_2) - I_S(\cos \theta_S \pm j \sin \theta_S)(b_1 + j b_2) \star$

$I_R = I_S(\cos \theta_S \pm j \sin \theta_S)(a_1 + j a_2) - E_S(C_1 + j C_2) \star$

★ ± THIS SIGN IS MINUS WHEN THE P. F. IS LAGGING AND PLUS WHEN THE P. F. IS LEADING

(a₁ + j a₂) = +.810558 + j.076831
 x E_{SN} = + 70652

 E_{SN}(a₁ + j a₂) = + 57268 + j 5428

 (cos θ_S + j sin θ_S) = + .9342 + j .3567
 x I_S = + 94.75

 I_S(cos θ_S + j sin θ_S) = + 88.52 + j 33.8
 x (b₁ + j b₂) = + 91.75 + j 235.9

 + 8122 + 20882
 - 7973 + j 3101

 I_S(cos θ_S + j sin θ_S)(b₁ + j b₂) = + 149 + j 23983

I_S(cos θ_S + j sin θ_S) = + 88.52 + j 33.8
 x (a₁ + j a₂) = + .810558 + j .076831

 + 71.751 + j 6.801
 - 2.597 + j 27.397

 I_S(cos θ_S + j sin θ_S)(a₁ + j a₂) = + 69.154 + j 34.198

 (C₁ + j C₂) = -.000041 + j .001463
 x E_{SN} = 70652

 E_{SN}(C₁ + j C₂) = -2.897 + j 103.36

E_{SN}(a₁ + j a₂) = + 57268 + j 5428
 - I_S(cos θ_S + j sin θ_S)(b₁ + j b₂) = - 149 - j 23983

I_S(cos θ_S + j sin θ_S)(a₁ + j a₂) = + 69.154 + j 34.20
 - E_{SN}(C₁ + j C₂) = + 2.897 - j 103.36

CHANGE SIGNS AND ADD

E_{RN} = + 67119 - j 18555
 = $\sqrt{(67119)^2 + (18555)^2}$
 E_{RN} = 60 057 VOLTS TO NEUTRAL.

I_R = 72.051 - j 69.16
 = $\sqrt{(72.051)^2 + (69.16)^2}$
 I_R = 99.87 AMPERES.

$KW_{RN} = (67119 \times 72.051) + (18555 \times 69.16) = 6\ 399$ KW PER PHASE.

$PF_R = \frac{6\ 399 \times 100}{6\ 998} = 90.01\%$ LAGGING.

$KV-A_{RN} = (60.057 \times 99.87) = 5\ 998$ KV-A PER PHASE.

$EFFICIENCY = \frac{6\ 399 \times 100}{6\ 265} = 86.32\%$.

LOSS = 6 255 - 5 399 = 856 KW PER PHASE.

PHASE ANGLES AT FULL LOAD THE VOLTAGE AT THE RECEIVER END LAGS BEHIND THE VOLTAGE AT THE SENDING-END BY THE ANGLE $TAN^{-1} \frac{18\ 555}{67\ 119} = TAN^{-1}.325 = 18^\circ 0'$; AND THE CURRENT AT THE RECEIVER END LAGS BEHIND THE VOLTAGE AT THE SENDING-END BY THE ANGLE $TAN^{-1} \frac{69.16}{72.051} = TAN^{-1}.959 = 43^\circ 50'$. HENCE THE CURRENT AT THE RECEIVER END LAGS BEHIND THE VOLTAGE AT THE RECEIVER END BY THE ANGLE $43^\circ 50' - 18^\circ 0' = 25^\circ 50'$. THE POWER-FACTOR AT THE RECEIVER END IS THEREFORE $COS\ 25^\circ 50' = 90\%$ LAGGING.

ZERO LOAD CONDITIONS

$E_{RNO} = \frac{E_{SNO}(a_1 - j a_2)}{(a_1^2 + a_2^2)} = \frac{48\ 898(.81056 - j.076831)}{(.81056)^2 + (.076831)^2} = \frac{39635 - j 3757}{.6629} = 59\ 790 - j 5667 = 60\ 058$ VOLTS.

$I_{SO} = E_{SNO} \frac{(C_1 a_1 + C_2 a_2) + j(C_2 a_1 - C_1 a_2)}{(a_1^2 + a_2^2)} = 48\ 898 \frac{(-.000041 \times .81056) + (.001463 \times .076831) + j((.001463 \times .81056) - (-.000041 \times .076831))}{.6629}$

$I_{SO} = 48\ 898 \frac{(+.0000792 + j.001189)}{.6629} = 48\ 898(.000119 + j.001794) = 48\ 898 \times .001798 = 87.92$ AMPERES.

REGULATION

A RISE IN VOLTAGE AT THE SENDING-END OCCURS OF 70 852 - 48 898 = 21 754 VOLTS TO NEUTRAL WHEN THE LOAD IS INCREASED FROM ZERO TO 99.87 AMPERES AT 90.01% POWER FACTOR LAGGING AT THE RECEIVER END WITH CONSTANT VOLTAGE AT THE RECEIVING END.

PHASE ANGLES

AT ZERO LOAD THE VOLTAGE AT THE RECEIVER END LAGS BEHIND THE VOLTAGE AT THE SENDING-END BY THE ANGLE $TAN^{-1} \frac{5\ 667}{69\ 790} = TAN^{-1}.0948 = 5^\circ 25'$; AND THE CURRENT AT THE SENDING-END LEADS THE VOLTAGE AT THE SENDING-END BY THE ANGLE $TAN^{-1} \frac{0.001794}{.000119} = TAN^{-1} 15.08 = 86^\circ 11'$. THE POWER-FACTOR AT THE SENDING-END IS THEREFORE $COS\ 86^\circ 11' = 6.64\%$ LEADING AT ZERO LOAD.

CHART XV—CALCULATED PERFORMANCE OF VARIOUS CIRCUITS

PROBLEM	RECEIVING-END CONDITIONS FIXED							SENDING-END CONDITIONS—CALCULATED *											
	LOAD CONDITIONS							LOAD CONDITIONS					ZERO LOAD						
	KV-A _R	E _R 3 PHASE	TO NEUTRAL				TO NEUTRAL					TO NEUTRAL							
			KV-A _{RN}	KW _{RN}	E _{RN}	I _R	PF _R %	KV-A _{SN}	KW _{SN}	E _{SN}	I _S	PF _S %	★ ★	LINE DROP IN % OF E _{RN}	LINE LOSS IN % OF KW _R	KV-A _{SNO}	KW _{SNO}	E _{SNO}	I _{SO}
2 5 C Y C L E S																			
1	1300	10 000	433.3	346.6	5 774	75	80 LAG	474.63	377.52	6 347	74.78	79.53	-9.92	8.92	1.963		5 773	.34	
2	"	"	"	433.3	"	"	100	465.09	464.21	6 202	74.99	99.81	-7.41	7.13	"		"	"	"
3	5000	20 000	1666.6	1333.3	11 550	144.4	80 LAG	1821.9	1449.5	12 653	143.99	79.56	-9.55	8.71	7.622		11 548	.66	
4	"	"	"	1666.6	"	"	100	1786.6	1783.33	12 372	144.38	99.80	-7.12	7.00	"		"	"	"
5	3500	20 000	1167	933	11 550	101	80 LAG	1278.6	1017.45	12 733	100.42	79.58	-10.24	9.05	10.85		11 546	.94	
6	"	"	"	1167	"	"	100	1253.5	1254.22	12 415	100.97	99.82	-7.49	7.22	"		"	"	"
7	8000	30 000	2667	2133	17 320	15.4	80 LAG	2 928.2	2 329.8	19 125	153.11	79.56	-10.42	9.23	24.29		17 313	1403	
8	"	"	"	2667	"	"	100	2 849.8	2 860.5	18 640	153.96	99.48	-7.62	7.26	"		"	"	"
9	5000	30 000	1667	1333	17 320	96.2	80 LAG	1 817.3	1 459.2	19 184	94.73	80.29	-10.76	9.47	10.32		17 304	2.33	
10	"	"	"	1667	"	"	100	1 796.4	1 794.2	18 685	96.14	99.88	-7.89	7.63	"		"	"	"
11	20 000	60 000	6667	5 333	34 640	192.5	80 LAG	7 303.9	5 841.0	38 490	189.26	79.97	-11.11	9.53	149.8	.13	34 607	4.33	.09
12	"	"	"	6 667	"	"	100	7 192.1	7 181.2	37 387	192.37	99.85	-7.93	7.71	"		"	"	"
13	22 000	88 000	7333	5 867	50 810	144.4	80 LAG	7 762.5	6 419.6	56 619	137.1	82.70	-11.43	9.42	599.3	1.94	50 620	11.84	.32
14	"	"	"	7 333	"	"	100	7 915.4	7 915.2	54 820	144.39	100.00	-7.89	7.94	"		"	"	"
15	40 000	120 000	13 333	10 667	69 290	192.5	80 LAG	14 106	11 648	77 147	182.85	82.58	-14.34	9.19	1081	3.39	69 030	15.66	.31
16	"	"	"	13 333	"	"	100	14 366	14 366	74 642	192.47	100.00	-7.73	7.75	"		"	"	"
17	25 000	120 000	8 333	6 667	69 290	120.3	80 LAG	7 886.5	7 156.1	76 754	102.75	90.74	-10.77	7.34	2185	15.29	68 253	32.01	.70
18	"	"	"	8 333	"	"	100	9 025.4	8 913.0	73 401	122.76	98.75x	-5.73	6.26	"		"	"	"
19	40 000	140 000	13 333	10 667	80 830	165	80 LAG	13 270	11 610	91 761	144.52	87.49	-13.52	8.84	2780	17.44	79 622	34.92	.63
20	"	"	"	13 333	"	"	100	14 459	14 412	86 863	166.66	99.68x	-7.46	8.09	"		"	"	"
21	20 000	120 000	6 667	5 333	69 290	96.2	80 LAG	5 683.7	5 642.4	75 682	75.08	99.29	-9.22	5.81	3428	39.22	66 950	51.21	1.14
22	"	"	"	6 667	"	"	100	7 652.7	7 105.4	71 762	106.64	92.85x	-3.57	6.57	"		"	"	"
23	60 000	200 000	20 000	16 000	115 500	173.2	80 LAG	17 576	17 048	128 450	136.83	96.99	-11.21	6.55	8559	91.03	111 611	76.69	1.06
24	"	"	"	20 000	"	"	100	22 287	21 381	120 574	184.84	95.98x	-4.39	6.90	"		"	"	"
25	20 000	140 000	6 667	5 333	80 830	82.5	80 LAG	5 959.3	5 621.1	86 404	68.97	94.33x	-6.89	5.40	5585	109.6	76 024	73.47	1.96
26	"	"	"	6 667	"	"	100	8 808.9	7 166.1	81 647	107.89	84.34x	-4.01	7.47	"		"	"	"
27	50 000	200 000	16 667	13 333	115 500	144.4	80 LAG	14 295	14 222	127 267	112.32	99.49x	-10.19	6.67	11 018	202.04	108 663	101.4	1.84
28	"	"	"	16 667	"	"	100	20 322	18 066	118 833	171.01	89.70x	-2.89	8.40	"		"	"	"
29	15 000	140 000	5 000	4 000	80 830	618.6	80 LAG	6 183.5	4 237	83 045	74.46	68.52x	-2.74	5.92	6 665	208.54	73 360	90.85	3.13
30	"	"	"	5 000	"	"	100	8 518.7	5 479	78 655	108.30	64.32x	+2.69	9.58	"		"	"	"
31	40 000	200 000	13 333	10 667	115 500	115.5	80 LAG	13 277	11 383	123 401	107.59	85.74x	-6.85	6.71	13 140	395.8	104 378	75.3	3.01
32	"	"	"	13 333	"	"	100	19 096	14 672	115 162	165.82	76.83x	+0.29	10.05	"		"	"	"
6 0 C Y C L E S																			
33	1300	10 000	433.3	346.6	5 774	75	80 LAG	499.03	377.44	6 702	74.46	75.63	-16.07	8.90	4.558		5 769	.79	
34	"	"	"	433.3	"	"	100	469.05	464.18	6 259	74.94	98.96	-8.40	7.13	"		"	"	"
35	5000	20 000	1667	1333	11 550	144.4	80 LAG	1 911.02	1 448.95	13 333	143.33	75.82	-15.44	8.70	18.23		11 540	4.58	
36	"	"	"	1667	"	"	100	1 800.6	1 783.3	12 480	144.28	99.04	-8.05	6.98	"		"	"	"
37	3500	20 000	1167	933	11 550	101	80 LAG	1 341.0	1 016.8	13 482	99.47	75.82	-16.73	8.98	25.93		11 527	2.25	
38	"	"	"	1167	"	"	100	1 264.0	1 251.2	12 537	100.82	98.99	-8.55	7.22	"		"	"	"
39	8000	30 000	2667	2133	17 320	15.4	80 LAG	3 073.6	2 327.9	20 268	151.65	75.74	-17.02	9.13	58.43		17 286	3.38	
40	"	"	"	2667	"	"	100	2 894.7	2 864.1	18 830	153.73	98.94	-8.72	7.39	"		"	"	"
41	5000	30 000	1667	1333	17 320	96.2	80 LAG	1 879.2	1 456.2	20 331	92.43	77.40	-17.38	9.24	96.29	.22	17 225	5.59	.22
42	"	"	"	1667	"	"	100	1 806.1	1 794.1	18 845	95.84	99.33	-8.80	7.62	"		"	"	"
43	20 000	60 000	6 667	5 333	34 640	192.5	80 LAG	7 592.8	5 830.1	40 976	185.42	76.73	-18.24	9.32	357.9	.75	34 450	10.39	.21
44	"	"	"	6 667	"	"	100	7 243.0	7 180.2	37 773	191.75	99.15	-9.05	7.70	"		"	"	"
45	22 000	88 000	7 333	5 867	50 810	144.4	80 LAG	7 578.7	6 380.0	59 925	126.97	84.18	-17.94	8.74	1 409	8.62	49 710	28.35	.61
46	"	"	"	7 333	"	"	100	7 915.5	7 915.3	54 869	144.26	100.00	-7.99	7.94	"		"	"	"
47	44 000	120 000	13 333	10 667	69 290	192.5	80 LAG	13 796	11 579	81 710	168.84	83.93	-17.92	8.55	2 528	144.9	67 800	37.28	.57
48	"	"	"	13 333	"	"	100	14 366	14 365	74 735	192.22	100.00	-7.86	7.74	"		"	"	"
49	25 000	120 000	8 333	6 667	69 290	120.3	80 LAG	7 082.3	7 075.1	79 000	82.65	99.89	-14.01	6.12	475.9	75.47	63 357	75.11	1.59
50	"	"	"	8 333	"	"	100	9 473.0	8 949.6	70 599	134.18	94.47x	-1.89	7.40	"		"	"	"
51	40 000	140 000	13 333	10 667	80 830	165	80 LAG	11 827	11 461	96 727	122.27	96.90	-19.67	7.44	6 060	89.78	73 938	81.96	1.48
52	"	"	"	13 333	"	"	100	14 666	14 438	84 862	172.82	98.44x	-4.99	8.29	"		"	"	"
53	20 000	120 000	6 667	5 333	69 290	96.2	80 LAG	6 972.8	5 626.6	72 747	93.85	80.69x	-4.99	5.50	6 523	208.8	56 101	116.27	3.20
54	"	"	"	6 667	"	"	100	9 061.7	7 239.4	63 810	142.01	79.89x	+7.91	8.59	"		"	"	"
55	60 000	200 000	20 000	16 000	115 500	173.2	80 LAG	18 728	16 908	126 541	148.00	90.28x	-9.56	5.68	16 330	476.4	93 636	174.4	2.92
56	"	"	"	20 000	"	"	100	24 796	21 658	109 189	227.09	87.34x	+5.47	8.29	"		"	"	"
57	20 000	140 000	6 667	5 333	80 830	82.5	80 LAG	10 089	5 794.2	74 182	136.01	57.45x	+8.22	8.69	8 626	545.8	54 494	158.3	6.33
58	"	"	"	6 667	"	"	100	11 014	7 539.6	64 377	171.08	68.45x	+20.35	13.09	"		"	"	"
59	50 000	200 000	16 667	13 333	115 500	144.4	80 LAG	21 139	14 343	113 606	186.07	67.85x	+16.9	7.58	17 217	1 057	77 934	220.4	6.14
60	"	"	"	16 667	"	"	100	23 946	18 757	96 987	246.84	78.39x	+16.03	12.54	"		"	"	"
61	15 000	140 000	5 000	4 000	80 830	618.6	80 LAG	10 233	4 801.2	59 046	173.30	46.92x	+26.95	20.03	7 690	998.8	41 213	186.6	12.94
62	"	"	"	5 000	"	"	100	9 918.4	6 230.9	51 327	193.24	62.84x	+36.50	24.60	"		"	"	"
63	40 000	200 000	13 333	10 667	115 500	115.5	80 LAG	21 948	12 248	93 725	234.17	55.80x	+18.81	14.82	5 223	1 907	59 074	257.7	12.52
64	"	"	"	13 333	"	"	100	21 750	16 020	80 106	271.52	73.66x	+30.44	20.5	"		"	"	"

The above performances are based upon values for the auxiliary constants as given on Chart XII

CHAPTER X

HYPERBOLIC FUNCTIONS

In the consideration of the hyperbolic theory as applied to transmission circuits, the writer desires to express his high appreciation of the excellent literature already existing. Dr. A. E. Kennelly's pioneer work and advocacy of the application of hyperbolic functions to the solution of transmission circuits has been too extensive and well known to warrant a complete list of his contributions. His most important treatises are "Hyperbolic Functions Applied to Electrical Engineering", 1916; "Tables of Complex Hyperbolic and Circular Functions", 1914; "Chart Atlas of Hyperbolic Functions", 1914, which provides a ready means of obtaining values for complex functions, thus materially shortening and simplifying calculations, and "Artificial Electric Lines", 1917.

"Electrical Phenomena in Parallel Conductors" by Dr. Frederick Eugene Pernot, 1918, is an excellent treatise on the subject and contains valuable tables of logarithms of real hyperbolic functions from $x = 0$ to $x = 2.00$ in steps of 0.001.

An article "Long-Line Phenomena and Vector Locus Diagrams" in the *Electrical World* of Feb. 1, 1919, p. 212, by Prof. Edy Velandar is an excellent and valuable contribution on the subject, because of its simplicity in explaining complicated phenomena.

To employ hyperbolic functions successfully in the solution of transmission circuits it is not necessary for the worker to have a thorough understanding of how they have been derived. On the other hand it is quite desirable to understand the basis upon which they have been computed. A brief review of hyperbolic trigonometry is therefore given before taking up the solution of circuits.

CIRCULAR angles derive their name from the fact that they are functions of the circle, whose equation is $x^2 + y^2 = r$. Tabulated values of such functions are based upon a radius of unit length. The geometrical construction illustrating three of the functions, the sine, cosine and tangent of circular angles is indicated in Fig. 38. The angle AOP , indicated by full lines in the positive or counter-clockwise direction, has been drawn to correspond to one radian. The radian is an angular unit of such magnitude that the length of the arc which subtends the radian is numerically equal to that of the radius of the circle. Thus, the number of radians in a complete circle is 2π . Expressed in degrees the radian is equal approximately to $57^\circ 17' 44.8''$. The segment AOP of any angle AOP of one radian has an area equal to one-half the area of a unit square. Therefore the angle may be expressed in radians as,—

$$\frac{\text{Length of arc}}{\text{radius}} \quad \text{or} \quad \frac{2 \times \text{area}}{(\text{radius})^2}$$

Circular functions are obtained as follows,—

$$\text{Circular angle} = \frac{2 \times \text{area}}{(\text{radius})^2} \text{ radians}$$

$$\text{Sine } \theta = \frac{Y}{R}$$

$$\text{Cosine } \theta = \frac{X}{R}$$

$$\text{Tangent } \theta = \frac{Y}{X}$$

The variations in the circular functions, sine, cosine and tangent are indicated graphically in Fig. 39 for a complete revolution of 360 degrees. Since for the second and each succeeding revolution these graphs would simply be repeated, circular functions are said to have a period equal to 2π radians. In other words, adding 2π to a circular angle expressed in radians does not change the value of a circular function.

REAL HYPERBOLIC ANGLES

Real hyperbolic angles derive their name because they are functions of an equilateral hyperbola. A hyperbola is a plane curve, such that the difference between the distances from any point on the curve to two fixed points called the foci is constant. In an equilateral hyperbola, Fig. 40, the asymptotes OS and OS' are straight lines at right angles to each other and make equal angles with the X-axis. The hyperbola continually approaches the asymptotes, and meets them at infinity. The equation of such a hyperbola is $x^2 - y^2 = r$.

The hyperbolic angle AOP of Fig. 40, called for convenience θ^* , has been drawn so as to correspond to an angle of one hyperbolic radian, or one "hyp" as it is usually designated. Hyperbolic angles are determined by the area of the sector they enclose. Thus the hyperbolic angle of one hyp AOP , encloses an area AOP of one-half, or the same as the area AOP of the corresponding circular angle of Fig. 38. It should be observed here that although one circular radian subtends an angle AOP of $57^\circ 17' 44.8''$, one hyperbolic radian subtends a circular angle AOP of $37^\circ 17' 33.67''$ (0.65087 circular radian).

In the same way as for the circle the hyperbolic angle may be expressed in radians as,—

$$\frac{\text{Length of arc}}{\rho} \quad \text{or} \quad \frac{2 \times \text{area}}{(\text{radius})^2}$$

where ρ = the integrated mean radius from O to AP . As an illustration, the length of the arc AP , Fig. 40

*A "hyperbolic angle", in the sense above described, is not the opening between two lines intersecting in a plane, but a quantity otherwise analogous to a circular angle and the argument x of the function $\sinh x$, $\cosh x$, $\tanh x$, etc. The use of the term hyperbolic angle can only be justified by its convenience of analogy.

is 1.3167 and the mean integrated radius to arc *AP* is 1.3167.

Hyperbolic functions, distinguished from circular functions by the letter *h* affixed, are obtained as follows:—

The variations in hyperbolic functions are indicated graphically in Fig. 41 for hyperbolic angles up to approximately 2.0 hyps for the sine and cosine and up to 3.0 hyps for the tangent.

Hyperbolic functions have no true period, but add-

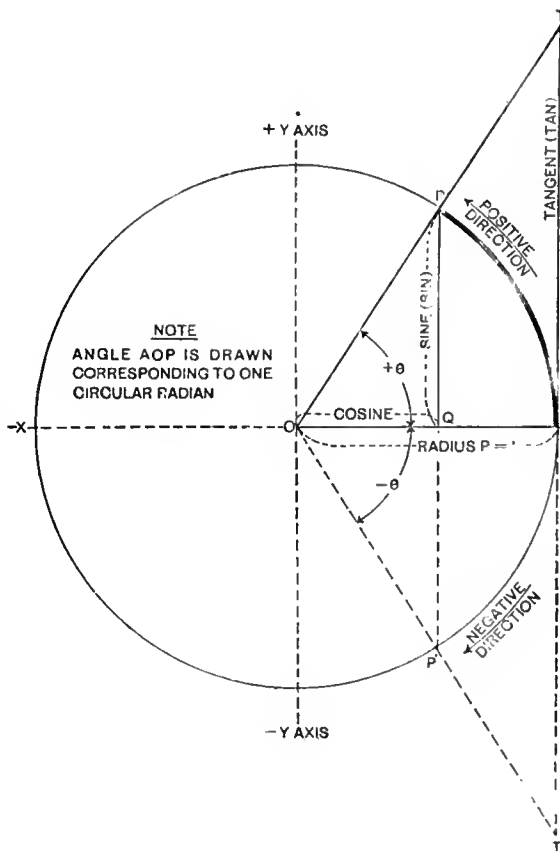


FIG. 38—REAL CIRCULAR ANGLES
 $X^2 + Y^2 = r^2$

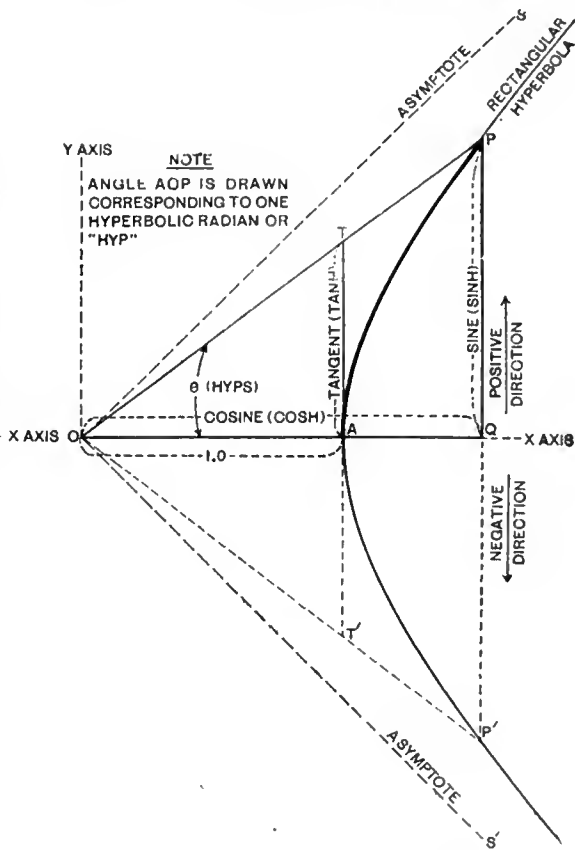


FIG. 40—REAL HYPERBOLIC ANGLES
 $X^2 - Y^2 = r^2$

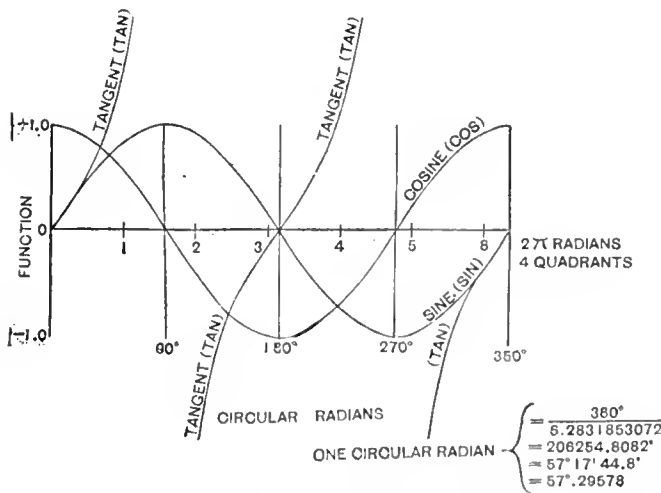


FIG. 39—GRAPHS OF CIRCULAR FUNCTIONS

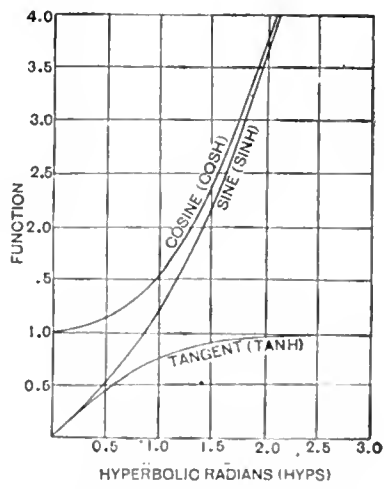


FIG. 41—GRAPHS OF HYPERBOLIC FUNCTIONS

$$\text{Hyperbolic angle } \theta = \frac{\text{Length of arc } AP}{\text{Length of mean radius}} \text{ radians.}$$

$$\text{Cosh } \theta = \frac{X}{OA}$$

$$\text{Sinh } \theta = \frac{Y}{OA}$$

$$\text{Tanh } \theta = \frac{Y}{X}$$

ing a $2\pi j$ to the hyperbolic angle does not change the values of the functions, hence these functions have an imaginary period of $2\pi j$.

Circular functions can be used to express the phase relations of current and voltage, but not the magnitude, or size, whereas hyperbolic functions, continually in-

creasing or decreasing, can be used to express the magnitude of current in a long circuit.

In Fig. 42 is shown a circular angle corresponding to one circular radian divided into five equal parts, each of 0.2 radian. Assuming unity radius, each of the arcs will have a constant length of 0.2 and a constant mean radius of 1.0. In Fig. 42 is shown a hyperbolic angle corresponding to one hyperbolic radian divided into five equal hyperbolic angles each of 0.2 hyperbolic radian. In this case the length of the arcs corresponding to each subdivision increases as the hyperbolic angle increases. The lengths of the corresponding integrated mean radii vectors also increase with the angle. By dividing the length of the arc of any of the five subdivisions by the length of the mean radius for that subdivision it will be seen that each subdivision represents 0.2 hyps.

From the above it will be evident that in radian measure, the magnitudes of circular and hyperbolic

plex angle takes, the construction for the cosine of a hyperbolic complex angle is illustrated by Fig. 43.

CONSTRUCTION FOR COSH θ

The construction, Fig. 43, assumes that the real part, that is the hyperbolic sector subtends an angle of one hyperbolic radian and the imaginary part, that is the circular sector, subtends an angle of one circular radian. This hyperbolic complex angle has therefore a numerical value of $r + j r$ hyperbolic radian. These numerical values embrace sectors sufficiently large for the purpose of clear illustration. The actual construction for obtaining the complex function $\cosh(\theta_1 + j\theta_2) = \cosh(r + j r \text{ hyperbolic radians})$ may be carried out as follows:—

On a piece of stiff card board lay out to a suitable scale the hyperbolic sector $\theta_1 = EOC$, equal to one hyp as shown in the upper left hand corner of Fig. 43. This may readily be plotted by the aid of a table of real hyperbolic functions for say each one tenth of a hyp up to and including one hyp. These are then plotted on the cardboard and joined with a curved line thus forming the arc EC of Fig. 43. The ends of the arc are then joined with O by straight lines. The real part of this hyperbolic complex angle is then cut out of the cardboard.

The circular part $j\theta_2$ of this complex angle is traced upon the cardboard as follows:— With radius equal to $\cosh \theta_1$ (to the same scale as used when tracing

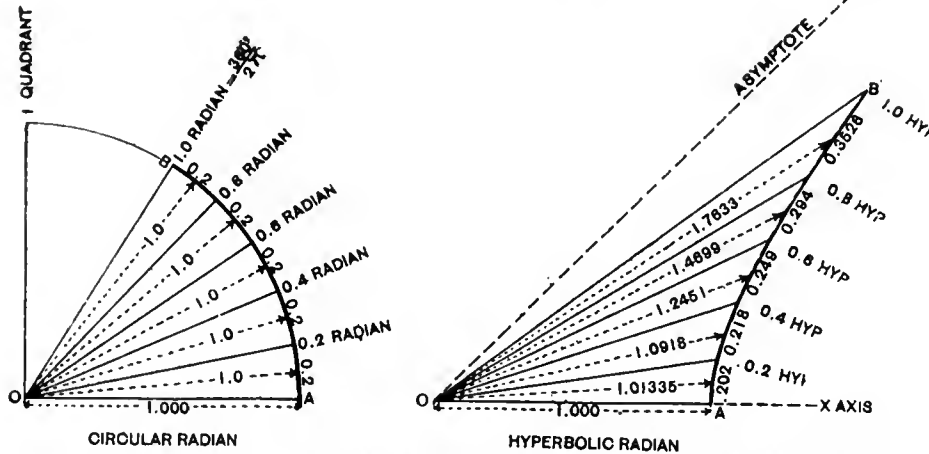


FIG. 42—SUBDIVISION OF A CIRCULAR AND A HYPERBOLIC RADIAN INTO FIVE SECTORS OF 0.2 RADIAN EACH

angles are similarly defined with reference to the area of circular and hyperbolic sectors.

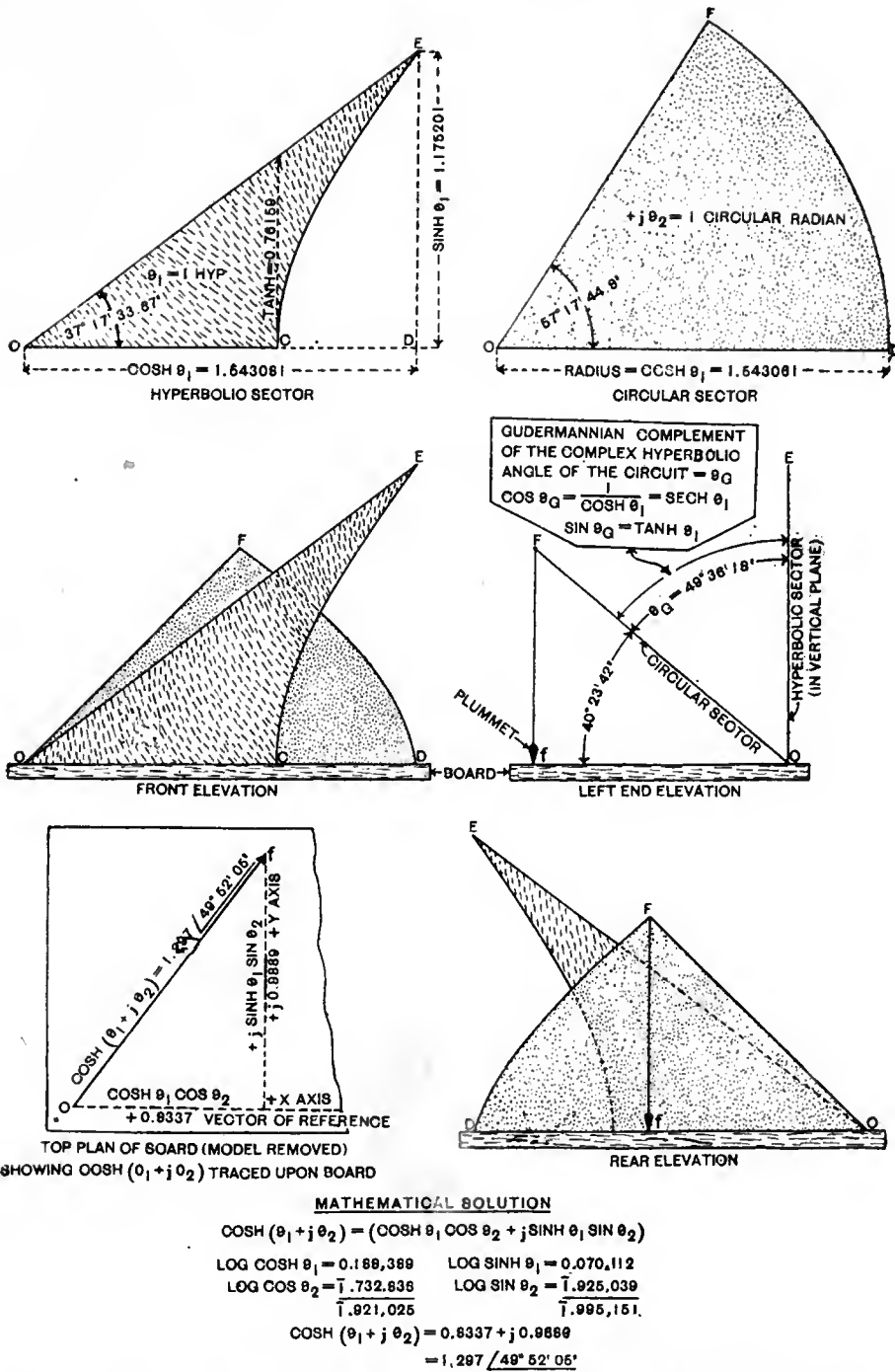
COMPLEX ANGLES AND THEIR FUNCTIONS

A complex angle is one which is associated with both a hyperbolic and a circular sector. If the complex angle is hyperbolic, its real part relates to a hyperbolic and its imaginary to a circular sector. On the other hand, if the complex angle is circular, its real part relates to a circular and its imaginary part to a hyperbolic sector. Complex hyperbolic trigonometry and complex circular trigonometry thus unite in a common geometrical relationship.

In the following treatment for the solution of transmission circuits by hyperbolic functions, only hyperbolic complex angles will enter into the solution. Such a complex angle will then consist of a combination of a "real" hyperbolic sector and a so-called "imaginary" or circular sector. The circular sector will occupy a plane inclined at an angle to the plane of the hyperbolic sector. In other words, the complex angle will be of the three-dimensional order. The construction of such a complex angle may be difficult to follow if viewed only from one direction. In order to illustrate the form that a com-

plex angle is now cut out of the piece of cardboard. This gives models of the two parts of the complex angle which may be arranged to form the complex angle $r + j r$ hyps. These two models are shown at the top of Fig. 43.

The two parts of the complex angle are arranged as follows:—Upon a drawing board or any flat surface occupying a horizontal plane, place the hyperbolic sector θ_1 in a vertical position. The plane of this hyperbolic sector will then be at right angles to the plane of the drawing board. The circular sector $j\theta_2$ is now placed in a vertical position just back of the hyperbolic sector. The toes O of each sector will then coincide, as well as the line OD of the circular sector with the line OC of the hyperbolic sector. The top of the circular sector is now turned back so that the plane of the circular sector lies at an angle with the vertical plane occupied by the hyperbolic sector. This displacement angle between the planes of the two sectors is



circular sector of this complex angle is moved in the forward direction through an angle of $49^\circ 36' 18''$ so that the plane of the circular sector assumes an angle of $90^\circ 00' 00'' - 49^\circ 36' 18'' = 40^\circ 23' 42''$ with the horizontal plane of the drawing board. From the end of the circular sector (point F) thus inclined, a plummet may be suspended until it meets the horizontal plane of the drawing board at the point f of the illustration. In other words, the point F is projected orthogonally onto the horizontal plane of the drawing board.

A top view of the drawing board, with the model removed, is illustrated in the lower left hand corner of Fig. 43. The line OF ($1.297 / 49^\circ 52' 05''$) traced upon the horizontal drawing board, is a vector representing the complex cosine of the complex angle $\theta_1 + j\theta_2 = 1 + j1$ hyperbolic radians. This complex cosine has rectangular coordinates of $+0.8337$ and $+j0.9889$.

At the bottom of Fig. 43 is given the mathematical expression for the exact solution for the cosine of a complex hyperbolic angle following the construction illustrated. There are numerous other mathematical equations with their equivalent geometrical constructions which will produce the same values for the cosine, but the above is probably as easy to follow as any, and will therefore be used exclusively hereafter.

CONSTRUCTION FOR $\operatorname{SINH} \theta$

The construction for the sine of the complex hyperbolic angle $1 + j1$ is indicated in Fig. 44. In this case the same construction may be used for obtaining the sinh as for determining the cosh of the complex angle with the following two exceptions.

The circular sector is made one quadrant (90°) larger. In other words the angle DOF' is $90^\circ + 57^\circ 17' 44.8''$ or $147^\circ 17' 44.8''$ as indicated by Fig. 44. It occupies the same plane as when determining the cosh of the angle but is simply extended in the forward direction through one quadrant, as indicated by the dotted lines of Fig. 44. The plummet is again suspended, this time from point F' upon the horizontal board, which it

known as the "gudermannian complement" of the hyperbolic angle θ . It will be referred to as θ_g . The front elevation of Fig. 43 illustrates how these two sectors would appear when viewed from the front. To the right of this illustration is shown how these two sectors would appear when viewed from the left hand end of the model. The displacement angle θ_g has a value for this particular complex angle of $49^\circ 36' 18''$. This numerical value is determined by virtue of the fact that this displacement angle has a cosine of

$$\frac{1}{\operatorname{cosh} \theta_1} = \frac{1}{1.543081} = 0.64805 \text{ or cosine of } \theta_g = \operatorname{sech} \theta_1 = 0.64805.$$

It has a sine of $\operatorname{tanh} \theta_1 = 0.76159$.

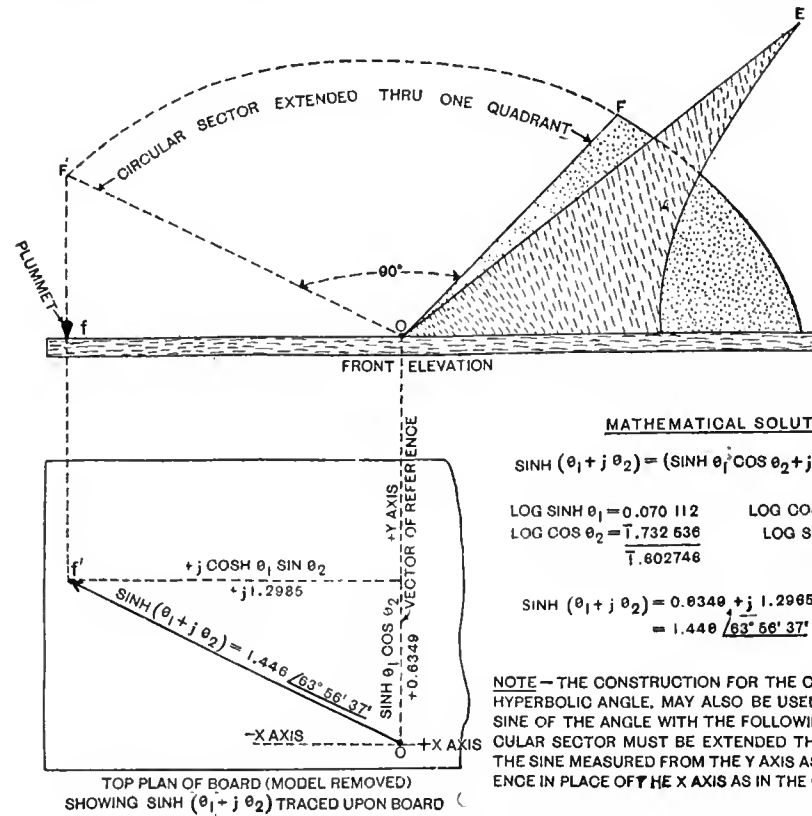
The angle whose cosine is 0.64805 and whose sine is 0.76159 is $49^\circ 36' 18''$. Thus the top part of the

meets at point f' . The other difference is that the sine OF' is read off from the Y axis as the vector of reference in place of the X axis as in the case of the cosine. Thus the circular sector has been carried forward through an angle of 90 degrees in the circular angle plane and the vector of reference has been advanced 90 degrees in the horizontal plane of reference. The sine of this angle is $1.446 / 63^\circ 56' 37''$ and has rectangular components of $0.6349 + j1.2985$. The mathematical

that Dr. Kennelly's description of the model and its application in determining the cosh and sinh of complex angles may be followed as given in the following paragraphs.

DESCRIPTION OF MODEL

In this model, the cosine or sine of a complex angle, either hyperbolic or circular, can be produced, by two successive orthogonal projections onto the XY plane, one projection being made from a rectangular hyperbola, and the other projection being then made from a particular circle definitely selected from among a theoretically infinite number of such circles, all concentric at the origin O , which circles, however, are not coplanar. The selection of the particular circle is determined by the foot of the projection from the hyperbola. This effects a geometrical process which is easily apprehended and visualized; so that once it has been realized by the student, the three-dimensional artifice is rendered superfluous, and he can roughly trace out a complex sine or cosine on an imaginary drawing board, with his eyes closed. The model, however, possesses certain interesting geometrical properties as a three-dimensional structure.



MATHEMATICAL SOLUTION

$$\text{SINH}(\theta_1 + j\theta_2) = (\text{SINH} \theta_1 \cos \theta_2 + j \text{COSH} \theta_1 \sin \theta_2)$$

$\text{LOG SINH} \theta_1 = 0.070112$	$\text{LOG COSH} \theta_1 = 0.168369$
$\text{LOG COS} \theta_2 = \overline{1.732536}$	$\text{LOG SIN} \theta_2 = \overline{1.925039}$
$\overline{1.802746}$	0.113428

$$\text{SINH}(\theta_1 + j\theta_2) = 0.6349 + j1.2985 = 1.446 / 63^\circ 56' 37''$$

NOTE—THE CONSTRUCTION FOR THE COSINE OF THE COMPLEX HYPERBOLIC ANGLE, MAY ALSO BE USED FOR DETERMINING THE SINE OF THE ANGLE WITH THE FOLLOWING CHANGES:—THE CIRCULAR SECTOR MUST BE EXTENDED THRU ONE QUADRANT AND THE SINE MEASURED FROM THE Y AXIS AS THE VECTOR OF REFERENCE IN PLACE OF THE X AXIS AS IN THE CASE OF THE COSINE.

FIG. 44—GRAPHICAL CONSTRUCTION FOR THE HYPERBOLIC SINE OF THE COMPLEX ANGLE $\theta_1 + j\theta_2 = 1 + j1$ hyperbolic radians.

expression for exact solution for the sine of a complex angle likewise accompanies the illustrated geometrical construction.

MODEL FOR ILLUSTRATING THE FUNCTIONS OF A COMPLEX ANGLE

Dr. Kennelly has recently constructed a model* for illustrating complex angles and for obtaining approximate values for the functions of such angles. Drawings made from photographs of this model are shown in Figs. 45, 46 and 47. The construction of a complex angle as above described is that employed by Dr. Kennelly in building his model. Since the model is applicable to tracing out numerous complex angles, it may seem a little difficult at the start. It was therefore thought desirable to precede the description of the model which is applicable to the solution of so many angles with a similar solution of a single definite complex angle. With the procedure for the solution, as given above, for cosh and sinh of $1 + j1$ hyperbolic radians in mind, it is believed

*This model was described in a paper read by him at a meeting of the American Academy of Arts and Sciences in April 1919.

The eight wire semicircles are formed with the following respective radii, in decimeters: 1.0, 1.020..., 1.081..., 1.185..., 1.337..., 1.543..., 1.810..., and 2.150..., which are the respective cosines of 0, 0.2, 0.4, 0.6, 0.8, 1.0, 1.2, and 1.4 hyperbolic radians, according to ordinary tables of real hyperbolic functions. These successive semi-circles therefore have radii equal to the cosines of successively increasing real hyperbolic angles θ_1 , by steps of 0.2, from 0 to 1.4 hyperbolic radians, inclusive. All of these semicircles have their common center at the origin O , in the plane $X O Y$, of the drawing board. The planes of the semicircles arc, however, displaced. The smallest circle of unit radius (1 decimeter), occupies the vertical plane $X O Z$, the

the eight wire semicircles are formed with the following respective radii, in decimeters: 1.0, 1.020..., 1.081..., 1.185..., 1.337..., 1.543..., 1.810..., and 2.150..., which are the respective cosines of 0, 0.2, 0.4, 0.6, 0.8, 1.0, 1.2, and 1.4 hyperbolic radians, according to ordinary tables of real hyperbolic functions. These successive semi-circles therefore have radii equal to the cosines of successively increasing real hyperbolic angles θ_1 , by steps of 0.2, from 0 to 1.4 hyperbolic radians, inclusive. All of these semicircles have their common center at the origin O , in the plane $X O Y$, of the drawing board. The planes of the semicircles arc, however, displaced. The smallest circle of unit radius (1 decimeter), occupies the vertical plane $X O Z$,

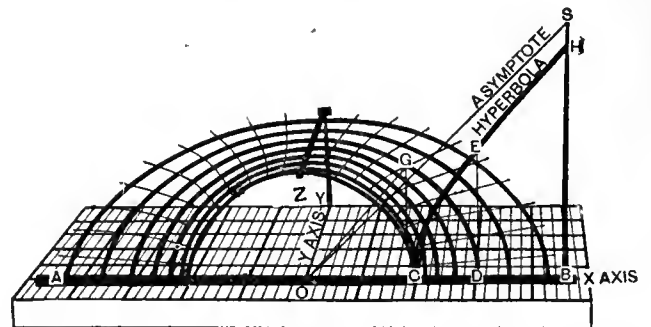


FIG. 45—DRAWING FROM A PHOTOGRAPH OF A GEOMETRICAL MODEL FOR THE ORTHOGONAL PROJECTION OF THE SINES AND COSINES OF COMPLEX ANGLES. THIS MODEL WAS DEVELOPED BY A. E. KENNELLY.

in which also lies the rectangular semi-hyperbola $X O H$. Angular distances corresponding to 0.2, 0.4, . . . 1.4 hyperbolic radians, are marked off along this hyperbola at successive corresponding intervals of 0.2. The cosines of these angles, as obtainable projectively on the $O X$ axis are marked off between C and B along the brass supporting bar, and at each mark, a semicircle rises from the $X Y$ plane, at a certain angle θ_0 with the vertical $X O Z$ plane. This displacement angle is determined by the relation,—

$$\cos \theta_0 = \frac{1}{\cosh \theta_1} = \operatorname{sech} \theta_1$$

Where θ_1 is the particular hyperbolic angle selected. This means, as is well known, that the displacement angle θ_0 between the plane of any semicircle and the vertical plane $Z O X$ is equal to the gudermannian of the hyperbolic angle θ_1 .

The model is, of course, only a skeleton structure of eight stages. If it could be completely developed, the number of semicircles would become infinite, and they would form a smooth continuous surface in three dimensions. Along the midplane $Z O Y$, all of these circles would have the same level, raised one decimeter above the horizontal drawing board plane of reference $X O Y$. The circles would increase in radius without limit, and would cover the entire $X O Y$ plane to infinity, the hyperbola extending likewise to infinity towards its asymptote $O S$, in the $X O Z$ plane. The actual model is thus the skeleton of the upper central sheet of the entire theoretical surface, near the origin.

The semicircles are also marked off in uniform steps of circular angle. Each step is taken, for convenience, as nine degrees, or one tenth of a quadrant. Corresponding angular steps on all of the eight semicircles are connected by thin wires, as shown in the illustrations.

A front elevation of the model, taken from a point on the $O Y$ axis—15 units from O , is given in Fig. 46. It will be seen that any tie wire, connecting corresponding circular angular

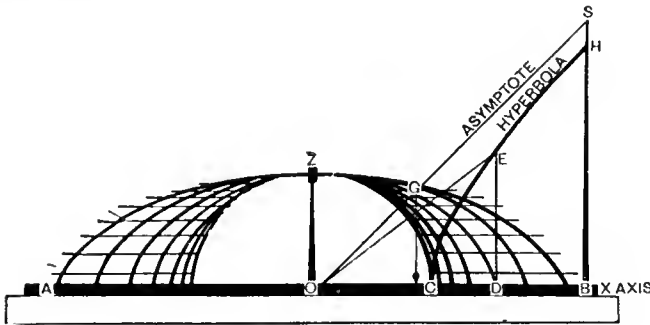


FIG. 46—FRONT ELEVATION OF MODEL
From a point on the $O Y$ axis, — 15 units from O .

points on the semicircles, is level, and lies at a constant height $\sin \theta_2$ decimeters above the drawing board. That is, the tie wire that connects all points of circular angle θ_2 , measured from $O X$ positively towards $O Y$, lies at the uniform height $\sin \theta_2$ decimeters above the drawing board.

A plan view of the model, taken from a point on the $O Z$ axis, + 15 units above O , is given in Fig. 47. It will be seen that each semicircle forms an ellipse, when projected on the base plane $X O Y$. The semi-major axis of this ellipse has length $\cosh \theta_1$, where θ_1 is the hyperbolic angle corresponding to that semicircle. The semi-minor axis is,—

$$\cosh \theta_1 \sin \theta_0 = \cosh \theta_1 \tanh \theta_1 = \sinh \theta_1$$

from the well known relation that exists between a hyperbolic angle and its gudermannian circular angle; namely,—

$$\sin \theta_0 = \tanh \theta_1$$

All of these ellipses have the same center of reference O . Any such system, having semi-major axes $\cosh \theta_1$, and semi-minor axes $\sinh \theta_1$, are well known to be confocal, and the foci must lie at the points +1 and -1 in the $X O Y$ plane, or the points in which the innermost circle cuts that plane.

PROCEDURE FOR PROJECTING $\cosh (\pm \theta_1 \pm j\theta_2)$

Thus premised, the process of finding the cosine of a complex hyperbolic angle $\theta_1 + j\theta_2$; that is, the process of finding $\cosh (\theta_1 + j\theta_2)$ is as follows:

Find the arc $C E$, Fig. 45, from $C = +1$ along the rectangular hyperbola $C E H$, which subtends θ_1 radians. The hyperbolic sector comprised between the radius, $O C$, the hyper-

bolic arc, and the radius vector $O E$, on this arc from the origin O , will then include $\frac{\theta_1}{2}$ sq. dm. of area. Drop a vertical perpendicular from E onto $O X$. It will mark off a horizontal distance $O D$ equal to $\cosh \theta_1$. Proceed along the circle which rises at D , in a positive or counterclockwise direction, through θ_2 circular radians, thus reaching on that circle a point G whose elevation above the drawing board is $\sin \theta_2$ decimeters. The area enclosed by a radius vector from the origin O on the circle, followed between the axis $O C$ and the circular curve, will be $\frac{\theta_2}{2} \cosh^2 \theta_1$ sq. dms.

From G , drop a vertical plummet, as in Fig. 46, on to the drawing board. In other words, project G orthogonally on the plane $X O Y$. Let g be the point on the drawing board at which the plummet from G touches the surface. Then it is easily seen that Og on the drawing board is the required magnitude and direction of $\cosh (\theta_1 + j\theta_2)$, in decimeters, with reference to $O X$ as the initial line in the plane $X O Y$. It may be read off either in rectangular coordinates along axes $O X$ and $O Y$ on a tracing cloth surface as shown in Fig. 47, or in polar coordinates printed on a sheet seen through the tracing cloth.

If the circular angle θ_2 , i. e., the imaginary hyperbolic angle $j\theta_2$, lies between π and 2π radians, (in quadrants 3 and 4), the point G will lie on the under side of the plane $X O Y$, and the projection onto g in that plane must be made upwards, instead of downwards.

If the hyperbolic angle whose cosine is required has a negative imaginary component, according to the expression $\cosh (\theta_1 - j\theta_2)$, then starting from the projected point D , we must trace out the circular angle in the negative or clockwise direction, as viewed from the front of the model.

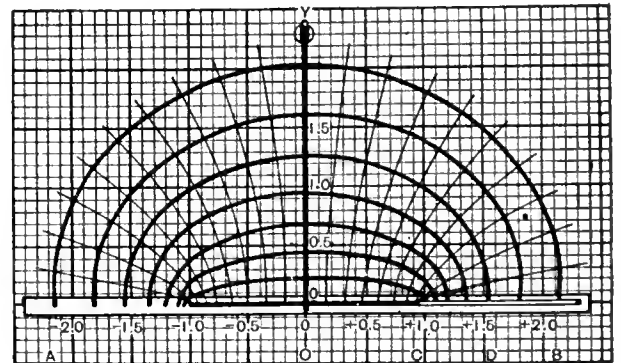


FIG. 47—PLAN VIEW OF MODEL
From a point on the $O Z$ axis, 15 units from O .

If the real part of the hyperbolic angle is negative, according to the expression $\cosh (-\theta_1 \pm j\theta_2)$; then since $\cosh (-\theta_1 \mp j\theta_2) = \cosh (\theta_1 \mp j\theta_2)$, we proceed as in the case of a positive real component, but with a change in the sign of the imaginary component.

The operation of tracing $\cosh (\pm \theta_1 \pm j\theta_2)$ on the $X Y$ plane, thus calls for two successive orthogonal projections onto that plane; namely (1) the projection corresponding to $\cosh (\pm \theta_1)$ as though $j\theta_2$ did not exist, and then (2), the projection corresponding to $\cosh j\theta_2 = \cos \theta_2$ independently of θ_1 , except that the radius of the circle, and its plane, are both conditioned by the magnitude of θ_1 .

If we trace the locus of $\cosh (\theta_1 \pm j\theta_2)$, where θ_1 is held constant, it is evident from Fig. 47 that we shall remain on one circle, which projects into the same corresponding ellipse on the $X Y$ plane. That is, the locus of $\cosh (\theta_1 \pm j\theta_2)$ with θ_1 held constant, is an ellipse, whose semi major and minor diameters are $\cosh \theta_1$ and $\sinh \theta_1$, respectively. If, on the other hand, we trace $\cosh (\pm \theta_1 + j\theta_2)$ with θ_2 held constant, we shall run over a certain tie wire bridging all the circles in the model, which tie wire is $\sin \theta_2$ dm. above the board, and its projection on the board, in the plane $X Y$ of projection, is part of a hyperbola.

PROCEDURE FOR $\sinh (\theta_1 + j\theta_2)$

It would be readily possible to produce a modification of this model here described, which would enable the sine of a complex angle to be projected on the $X Y$ plane following constructions already referred to. The transition to a new model for sines is, however, unnecessary. It suffices to use the cosine

model here described in a slightly different way. One has only to recall that

$$\sinh \theta = -j \cosh \left(\theta + j \frac{\pi}{2} \right)$$

or

$$\sinh (\theta_1 + j\theta_2) = -j \cosh \left[\theta_1 + j \left(\theta_2 + \frac{\pi}{2} \right) \right]$$

Consequently, in order to find the sine of a complex hyperbolic angle, we proceed on the model as though we sought the cosine of the same angle, increased by $\frac{\pi}{2}$ radians or one quadrant, in the imaginary or circular component. We then operate with $-j$ on the plane vector so obtained; i. e., we rotate it through one quadrant in the $X Y$ plane and in the clockwise direction. An equivalent step is, however, to rotate the X and Y axes of reference in that plane through one quadrant in the reverse or

positive direction. That is, we may omit the $-j$ operation, if, in dealing with sine projections, we treat $O Y$ as an $O X$ axis, and $-O X$ as an $O Y$ axis, or read off the projections on the $X Y$ plane to the $-Y O Y$ axis as initial line.

The only difference, therefore, between projecting the cosine and the sine of a complex hyperbolic angle in the model, is that in the latter case the circular component is increased by one quadrant and the projected plane vector is read off to the $O Y$ reference axis as initial line. The model thus gives the projection of either $\cosh (\pm \theta_1 \pm j\theta_2)$ or $\sinh (\pm \theta_1 \pm j\theta_2)$ within the limits of ± 1.4 and -1.4 for θ_1 , and for θ_2 between the limits $+\infty$ and $-\infty$. For accurate numerical work, reference would, of course, be made to the charts and tables of such functions already published, and which enable such functions to be obtained either directly or by interpolation, for all ordinary values of θ_1 and θ_2 .

CHAPTER XI

PERFORMANCE OF LONG TRANSMISSION LINES

(RIGOROUS SOLUTION BY HYPERBOLIC FUNCTIONS)

AS STATED in the discussion of the convergent series solution, the performance of an electric circuit is completely determined by its physical characteristics;—resistance, reactance, conductance and capacitance and the impressed frequency. These five quantities are accurately and fully accounted for in the two complex quantities.

$$\begin{aligned} \text{Impedance } Z &= R + jX \\ \text{Admittance } Y &= G + jB \end{aligned}$$

Having determined the numerical values for these two complex quantities, no further consideration need be given to the physical quantities of the circuit or to the frequency.

In the hyperbolic theory the circuit is said to subtend a certain complex angle, $\theta = \sqrt{ZY}$. This quantity represents in a sense the electrical length of the circuit. The numerical value of this angle θ is expressed in hyperbolic radians. If the circuit is very long electrically the numerical value of the angle will be comparatively large. Conversely, if the circuit is electrically short, it will be comparatively small. The numerical value of the angle θ is, therefore, a measure of the electrical length of the circuit and an indication of how much distortion in the distribution of voltage and current is to be expected as an effect of the capacitance and leakage of the circuit.

In order to give an idea of the extent of the variation in the complex θ and its functions $\cosh \theta$ and $\sinh \theta$ for power transmission circuits of various lengths corresponding to 25 and 60 cycle frequencies approximate values have been calculated, as shown in Table O.

This tabulation indicates that for circuits of from 100 to 500 miles in length, operated at frequencies of 25 and 60 cycles, the complex hyperbolic angle of the circuit (which is a plane-vector quantity) has a maximum modulus, or size of 0.41 for 25 cycles and of 1.05 for 60 cycles. It has an argument, or slope, lying between 70 and 78 degrees for 25 cycles and between 80 and 85 degrees for 60 cycles.

In the convergent series solution, the three so-called auxiliary constants A , B and C determine the performance of the circuit. These three auxiliary constants are simply expressions for certain hyperbolic functions of the complex hyperbolic angle θ of the circuit.

Thus

$$A = \cosh \theta$$

$$B = \sinh \theta \sqrt{\frac{Z}{Y}} = Z \frac{\sinh \theta}{\theta} = Z'$$

$$C = \sinh \theta \frac{1}{\sqrt{\frac{Z}{Y}}} = Y' \frac{\sinh \theta}{\theta}$$

ADDITIONAL SYMBOLS

In addition to the symbols previously listed, the following will be employed in the hyperbolic treatment.

- α = Linear hyperbolic angle expressed in hyps per mile. It is a complex quantity consisting of a real component α_1 and an imaginary component α_2 . It is also known as the attenuation constant or the propagation constant of the circuit.
- α_1 = The real component of the linear hyperbolic angle α , expressed in hyps. It is a measure of the shrinkage or loss in amplitude of the traveling wave, per unit length of line traversed.
- α_2 = The imaginary component of the linear hyperbolic angle α , expressed in circular radians. It is a measure of the loss in phase angle of the traveling wave, per unit length of line traversed.
- θ = The complex hyperbolic angle subtended by the entire circuit, expressed in hyps. It differs from α in that it embraces the entire circuit, whereas α embraces unit length of circuit (in this case one mile), $\theta = \alpha \times L$, where L is the length of the circuit expressed in miles.
- θ_1 = The real component of the complex hyperbolic angle of the circuit expressed in hyps, and defines the shrinkage or loss in amplitude or size of a traveling wave, in traversing the whole length of the line.
- θ_2 = The imaginary component of the complex hyperbolic angle of the circuit expressed in circular radians, expressing the loss in phase angle or slope of the traveling wave, in traversing the whole length of line.
- e = 2.7182818 which is the base of the Napierian system of logarithms. $\text{Log}_{10} e = 0.4342945$.
- θ_s = Position angle at sending end.
- θ_r = Position angle at receiving end.
- θ_p = Position angle at point P on a circuit.
- δ = Impedance load to ground or zero potential at receiving end line, in ohms at an angle.

- $z_0 = \sqrt{\frac{Z}{Y}}$ = Surge impedance of a conductor in ohms at an angle.
- $y_0 = \frac{Y}{z_0}$ = Surge admittance of a conductor in mhos at an angle.

TABLE O—GENERAL EFFECT OF DISTANCE AND FREQUENCY UPON THE COMPLEX HYPERBOLIC ANGLE AND ITS FUNCTIONS

LENGTH OF CIRCUIT (MILES)	Z	Y	ZY	$\theta = \sqrt{ZY}$	$\cosh \theta$	$\sinh \theta$
25 CYCLES						
100	43.3150°	0.000230190°	0.00996140°	0.10170°	0.99102°	0.10170°
200	80.6160°	0.000430190°	0.03466150°	0.19173°	0.98103°	0.19173°
300	109.165°	0.000670190°	0.07303153°	0.27172°	0.96103°	0.27172°
400	143.165°	0.000900190°	0.12870153°	0.36172°	0.94103°	0.35172°
500	156.162°	0.001100190°	0.17160157°	0.41172°	0.91103°	0.39172°
60 CYCLES						
100	84.8171°	0.000560190°	0.04749161°	0.22182°	0.98103°	0.22182°
200	165.124°	0.001030190°	0.17925166°	0.42183°	0.91103°	0.41183°
300	244.124°	0.001600190°	0.39040169°	0.63182°	0.81103°	0.60182°
400	324.124°	0.002130190°	0.70090170°	0.84182°	0.67103°	0.74182°
500	407.182°	0.002700190°	1.09890170°	1.05183°	0.51103°	0.87182°

These values are but roughly approximate to illustrate the general effect for certain circuits.

DETERMINATION OF THE AUXILIARY CONSTANTS

It was shown in Chart XI how values for the auxiliary constants A , B and C may be determined mathematically by convergent series form of solution, using problem X as an example. Chart XVI gives information as to how these same auxiliary constants may be determined by the use of real hyperbolic functions.

The solution for the auxiliary constants by real hyperbolic functions is given completely for problem X in Chart XVI. Vector diagrams are given to assist in following the solution. In the solution for the auxiliary constants by convergent series, the operations were carried out by aid of rectangular co-ordinates of the complex, or vector quantities. In Chart XVI, the operations are to a large extent carried out by the aid of polar co-ordinates. In the case of convergent series, most of the operations consist of adding the various terms of the series together. As addition and subtraction

of complex quantities can be most readily carried out when expressed in rectangular co-ordinates, this form of expression is used for the convergent-series solution. On the other hand, powers and roots of complex quantities are most readily obtained by polar co-ordinate expression. In the solution by real hyperbolic functions Chart XVI, operations for powers and roots predominate, and for this reason polar expressions have been quite generally employed. The solution by real hyperbolic functions is briefly this:—

The impedance Z and the admittance Y are first set down in complex form and their product obtained.

square root of this product gives the complex angle $\theta = \sqrt{ZY}$ of the circuit. This angle is then expressed in rectangular co-ordinates as $\theta_1 + j\theta_2$ for the purpose of determining the numerical value of its real part θ_1 (expressed in hyps) and its imaginary or circular part θ_2 expressible in circular radians. This circular part

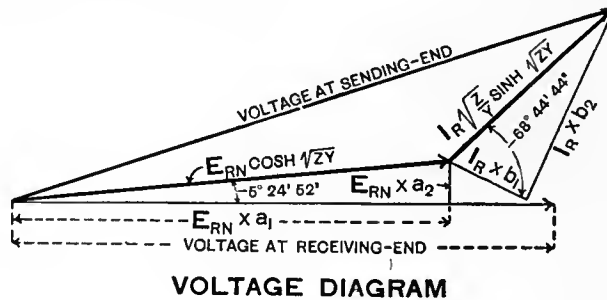
CHART XVI—RIGOROUS SOLUTION FOR AUXILIARY CONSTANTS OF PROBLEM X BY REAL HYPERBOLIC FUNCTIONS

CHARACTERISTICS OF CIRCUIT

LENGTH 300 MILES. CYCLES 80.
 CONDUCTORS—3 # 000 STRANDED COPPER.
 SPACING OF CONDUCTORS 10 X 10 X 20 FEET.
 EQUIVALENT DELTA SPACING=12.8 FT.

LINEAR CONSTANTS OF CIRCUIT
 TOTAL PER CONDUCTOR

$R = 0.350 \times 300 = 105$ OHMS TOTAL RESISTANCE AT 25° C.
 $X = 0.830 \times 300 = 249$ OHMS TOTAL REACTANCE.
 $B = 5.21 \times 300 \times 10^{-6} = .001563$ MHO TOTAL SUSCEPTANCE.
 $G = 0 \times 300 = 0$ MHO TOTAL CONDUCTANCE.
 $g =$ (IN THIS CASE TAKEN AS ZERO).



SOLUTION FOR $\theta = \sqrt{ZY}$

$\theta_2 = \tan^{-1} \frac{X}{R} = \tan^{-1} \frac{249}{105} = 67^\circ 8' 8''$
 $Z = R + jX = 105 + j249$
 $= \sqrt{105^2 + 249^2} = 270.233$
 $= 270.233 / 67^\circ 8' 8''$

$\theta_Y = \tan^{-1} \frac{B}{G} = \tan^{-1} \frac{0.001563}{0} = 90^\circ$
 $Y = G + jB = 0 + j0.001563$
 $= \sqrt{0^2 + 0.001563^2} = 0.001563$
 $= 0.001563 \angle 90^\circ$

$\theta_{ZY} = \theta_Z + \theta_Y = 67^\circ 8' 8'' + 90^\circ = 157^\circ 8' 8''$
 $ZY = 270.233 \times 0.001563 = 0.4223745$
 $= 0.4223745 \angle 157^\circ 8' 8''$

$\theta = \sqrt{ZY} = 0.8499035 / 78^\circ 34' 4''$ HYP.
 $= 0.128817 + j0.637009$ HYP.
 $= L(\alpha_1 + j\alpha_2)$

$\sqrt{\frac{Z}{Y}} = \sqrt{\frac{270.233 / 67^\circ 8' 8''}{0.001563 \angle 90^\circ}}$
 $= \sqrt{172893 \angle 22^\circ 51' 52''}$
 $= 415.805 \angle 11^\circ 25' 56''$

WAVE LENGTH
 $\alpha_2 = +j0.637009$ HYP.
 $\alpha_2 = \frac{0.637009}{300} = 0.00212338$

WAVE LENGTH = $\frac{2X}{\alpha_2} = \frac{6.2831853072}{0.00212338} = 2959$ MILES

SOLUTION FOR (A)

(A) = $\cosh \sqrt{ZY} = (\cosh \theta_1 \cos \theta_2 + j \sinh \theta_1 \sin \theta_2)$
 $\theta_1 = 0.128817$ HYP $\theta_2 = \frac{380}{2\pi} \times 0.837009 = 38^\circ 29' 52''$
 $\text{LOG COSH } \theta_1 = 0.003594$ $\text{LOG SINH } \theta_1 = \bar{1}.111172$
 $\text{LOG COS } \theta_2 = \bar{1}.905194$ $\text{LOG SIN } \theta_2 = \bar{1}.774359$
 $\text{LOG } a_1 = \bar{1}.908788$ $\text{LOG } a_2 = \bar{1}.885631$
 $a_1 = 0.81058$ $a_2 = 0.07583$
 $0.81058 + j0.07583$
(A) = 0.842 / 5° 24' 52''

SOLUTION FOR (B)

(B) = $\sqrt{\frac{Z}{Y}} \sinh \sqrt{ZY} = \sqrt{\frac{Z}{Y}} (\sinh \theta_1 \cos \theta_2 + j \cosh \theta_1 \sin \theta_2)$
 $\text{LOG SINH } \theta_1 = \bar{1}.111172$ $\text{LOG COSH } \theta_1 = 0.003594$
 $\text{LOG COS } \theta_2 = \bar{1}.905194$ $\text{LOG SIN } \theta_2 = \bar{1}.774359$
 $\frac{1}{0.108368}$ $\frac{1}{\bar{1}.777953}$
 $\text{SINH } \theta_1 \cos \theta_2 = 0.10383$ $\text{COSH } \theta_1 \sin \theta_2 = 0.59973$
 $\text{TAN}^{-1} \frac{0.59973}{0.10383} = 80^\circ 10' 40''$
 $\text{SINH } \theta_1 \cos \theta_2 + j \cosh \theta_1 \sin \theta_2 =$
 $= 0.10383 + j0.59973$
 $= 0.80885 / 80^\circ 10' 40''$
 $\sqrt{\frac{Z}{Y}} = 415.805 / 11^\circ 25' 56''$
(B) = 263.08 / 11° 25' 56'' x 0.80885 / 80° 10' 40''
 $= 263.08 / 68^\circ 44' 44''$

SOLUTION FOR (C)

(C) = $\frac{1}{\sqrt{Y}} \sinh \sqrt{ZY}$
 $= \frac{1}{415.805 / 11^\circ 25' 56''} \times 0.80885 / 80^\circ 10' 40''$
 $= 0.002405 / 11^\circ 25' 56'' \times 0.80885 / 80^\circ 10' 40''$
(C) = 0.001464 / 91° 38' 36''

As a check against possible serious errors in the calculations, the calculated values may be compared with values read from the Wilkinson Charts. The above results check exactly with those obtained by convergent series. (See Chart XI).

is converted to degrees by multiplying by $57^\circ .29578$. The hyperbolic cosine and sine of this complex angle are next obtained by the aid of logarithms of the functions of the component parts of the hyperbolic complex angle θ . The equation for $\cosh \theta$ and $\sinh \theta$ is given just above the solution. With a view of eliminating the necessity of calculation for each complex angle, $\cosh \theta$ and $\sinh \theta$, Dr. Kennelly has prepared tables and charts from which these two functions (and others) may be obtained directly, thus very materially shortening the solution by hyperbolic functions. Since complex angles have two variable components ($\theta_1 + j \theta_2$) tables of functions of such angles would have to be quite extensive in order that the steps for which values for the functions are given be not excessive. Although tables of functions of complex angles are not as complete as is desired they are a great help in the solution of ordinary power circuits. Functions corresponding to angles lying between the values for angles in these tables may readily be approximated by simple proportion, giving values sufficiently accurate for ordinary power transmission circuits. They have been calculated in Chart XVI for the purpose of illustrating such procedure and also as a high degree of accuracy was here desired for the purpose of illustrating the agreement of the results as obtained by different rigorous methods. Ordinarily these values would be taken from tables.

SOLUTION BY NOMINAL π METHOD

By this method, in place of considering the admittance of the circuit as being distributed (as it is in the actual circuit) it is based upon the assumption that the total conductor admittance may be lumped at two points, one half being placed at each end of the circuit. Such an artificial circuit is known as a "nominal π " circuit since the nominal values of impedance and admittance are ascribed to this circuit. On the above assumption, the current per conductor is the vector sum of the receiving end load and the receiving end condenser currents. The sending end current is the vector sum of the conductor and the sending end condenser currents. The performance of such a circuit may be determined either graphically or mathematically.

If the circuit is not of great electrical length, (say not over 100 miles at 60 cycles or 200 miles at 25 cycles) the performance of the corresponding nominal π circuit will not be materially different from that of the actual circuit having distributed constants which it imitates. If, however, the circuit is of great electrical length the performance of the nominal π circuit no longer closely imitates the performance of the actual circuit which it represents, owing to an error due to the lumpiness of the artificial circuit. Dr. Kennelly has shown that by making certain modifications in the linear or fundamental constants for the impedance and admittance of the nominal π circuit, the lumpiness error will vanish, so that the artificial circuit will then truly represent at the terminals the behavior under steady state

operation, taking distributed admittance into account. Such a corrected artificial circuit is known as the "equivalent" π circuit, because it then becomes externally equivalent to the actual circuit, having distributed constants, in every respect.

The complex numbers which must be applied to the impedance, Z and the admittances, $\frac{Y}{2}$ and $\frac{Y'}{2}$ of the nominal π circuit in order to correct these nominal values into the equivalent circuit are called the correcting factors of the nominal π circuit. The nominal values of the impedance Z and the admittances $\frac{Y}{2}$ of the circuit must be multiplied by these vector correcting factors in order to convert them into the "equivalent" values; thus:—

$$Z' = Z \frac{\sinh \theta}{\theta}$$

$$\frac{Y'}{2} = \frac{Y}{2} \frac{\tanh \theta/2}{\theta/2}$$

Where $\theta = \sqrt{ZY}$ is the hyperbolic complex angle subtended by the circuit.

Complete tables of hyperbolic functions are not always available; then again, many engineers have a natural aversion to the use of such functions. In order to avoid these objections as well as to simplify calculations, Dr. Kennelly has charted these "correcting factors" for hyperbolic complex angles up to $\theta = 1.0$ radian in steps of 0.01 in size and 1 degree in slope. The writer is particularly indebted to Dr. Kennelly for these charts, which are reproduced herewith for the first time, as Charts XVIII, XIX, XX and XXI. It is believed that the use of these charts will greatly simplify the calculation of the performance of electric power transmission circuits by hyperbolic functions. They enable the vector values of these ratios to be read to at least three decimal places in sizes and to two decimal places in slope, and their availability makes the use of tables of hyperbolic functions unnecessary. The corrected conductor impedance Z' is the same as the familiar auxiliary constant B .

EQUIVALENT π SOLUTION FOR PROBLEM X

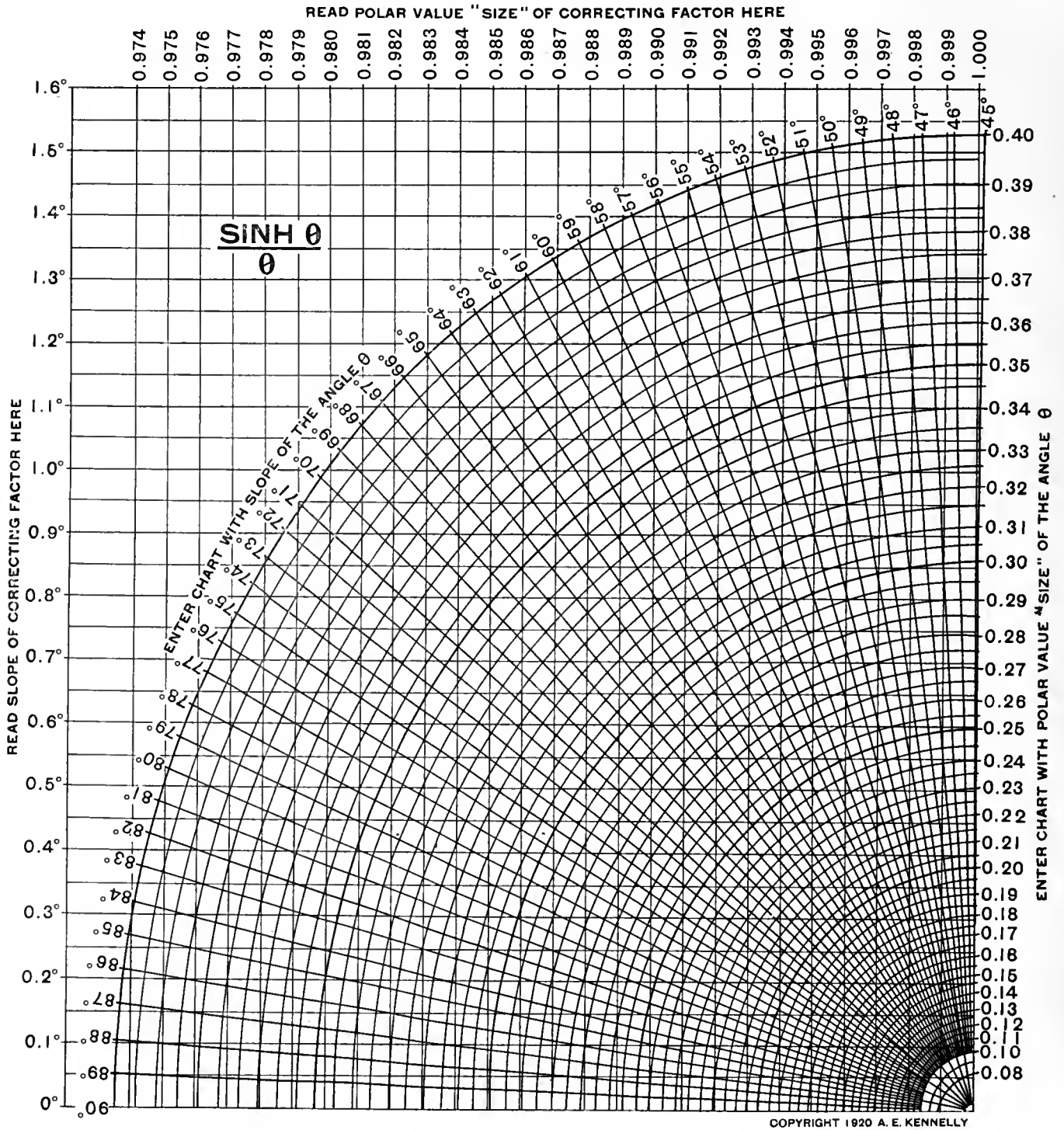
The solution for problem X by the equivalent π method is given in Chart XVII. At the top of the sheet are two diagrams, one a diagram for one conductor of the circuit of problem X and the other a corresponding vector diagram of the currents and the voltages at both ends. The numerical values of the angles and the quantities pertaining to problem X are placed upon the two diagrams for the purpose of assisting in following the mathematical solution.

The physical properties of the circuit are first set down, its linear constants obtained from the tables of constants and multiplied by the length of the circuit to obtain the total values per conductor. The next procedure is to calculate the hyperbolic angle θ of the circuit. To do this the impedance and the admittance of the circuit are set down as complex quantities in the form of polar co-ordinates and multiplied together by multiplying their slopes and adding their angles. The square root of the resulting vector is obtained by tak-

CHART XVIII

KENNELLY CHART FOR IMPEDANCE CORRECTING FACTOR

(FOR ANGLES HAVING POLAR VALUES BETWEEN 0 AND 0.40)



To find the vector "correcting factor" corresponding to any complex line angle θ , of a circuit, the angle θ is expressed in polar form with the slope in fractional degrees. The correcting factor as read from the chart will be in polar form with its slope in fractional degrees. Consult Table P for rapid conversion to minutes and seconds. For example:—

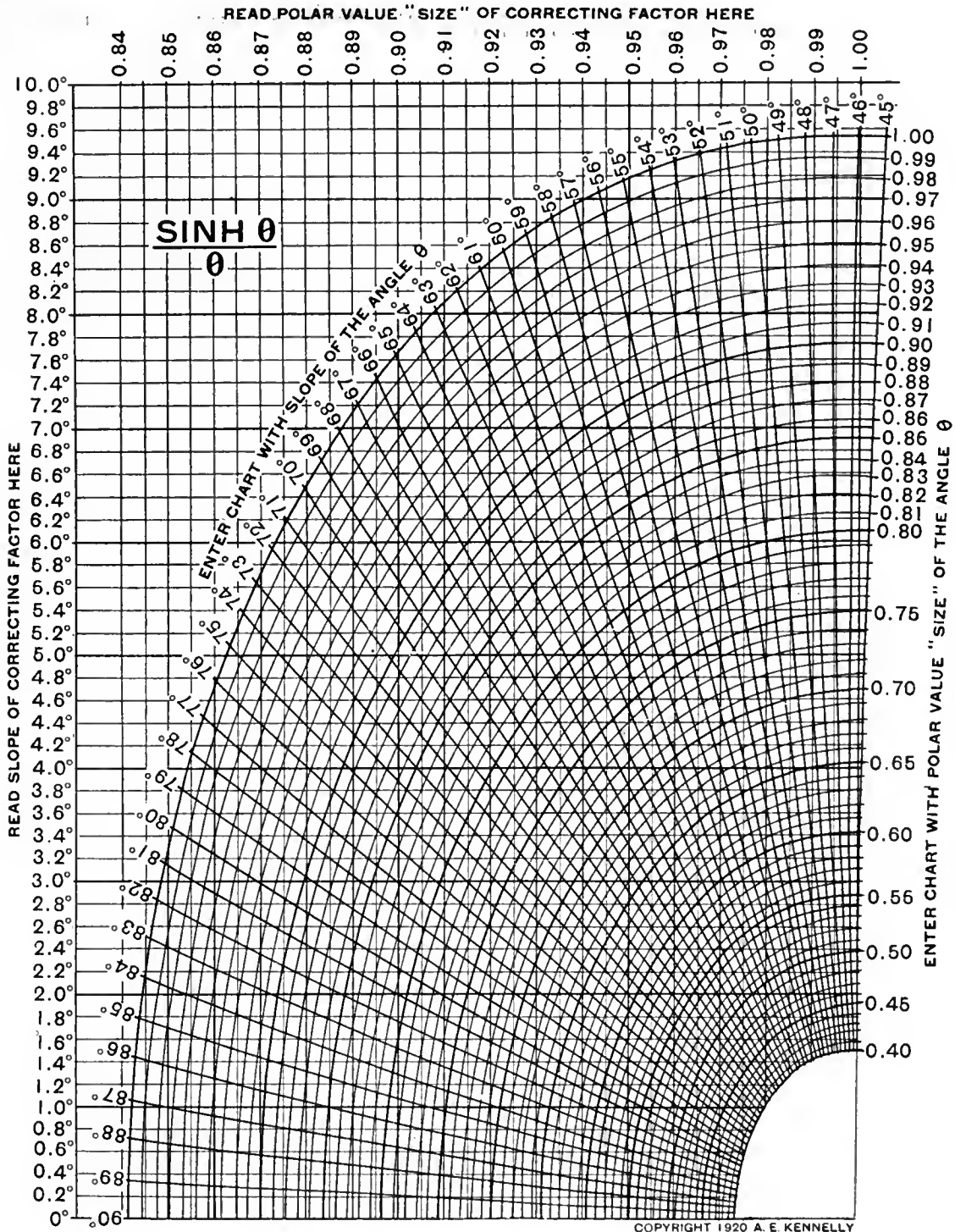
$$\theta = 0.3 \angle 68^\circ, \text{ correcting factor} = 0.9893 \angle 0^\circ .60 = 0.9893 \angle 0^\circ 36' 00''$$

$$\theta = 0.215 \angle 80^\circ .5, \text{ correcting factor} = 0.9927 \angle 0^\circ .149 = 0.9927 \angle 0^\circ 08' 56''$$

CHART XIX

KENNELLY CHART FOR IMPEDANCE CORRECTING FACTOR

(FOR ANGLES HAVING POLAR VALUES BETWEEN 0.40 AND 1.0)



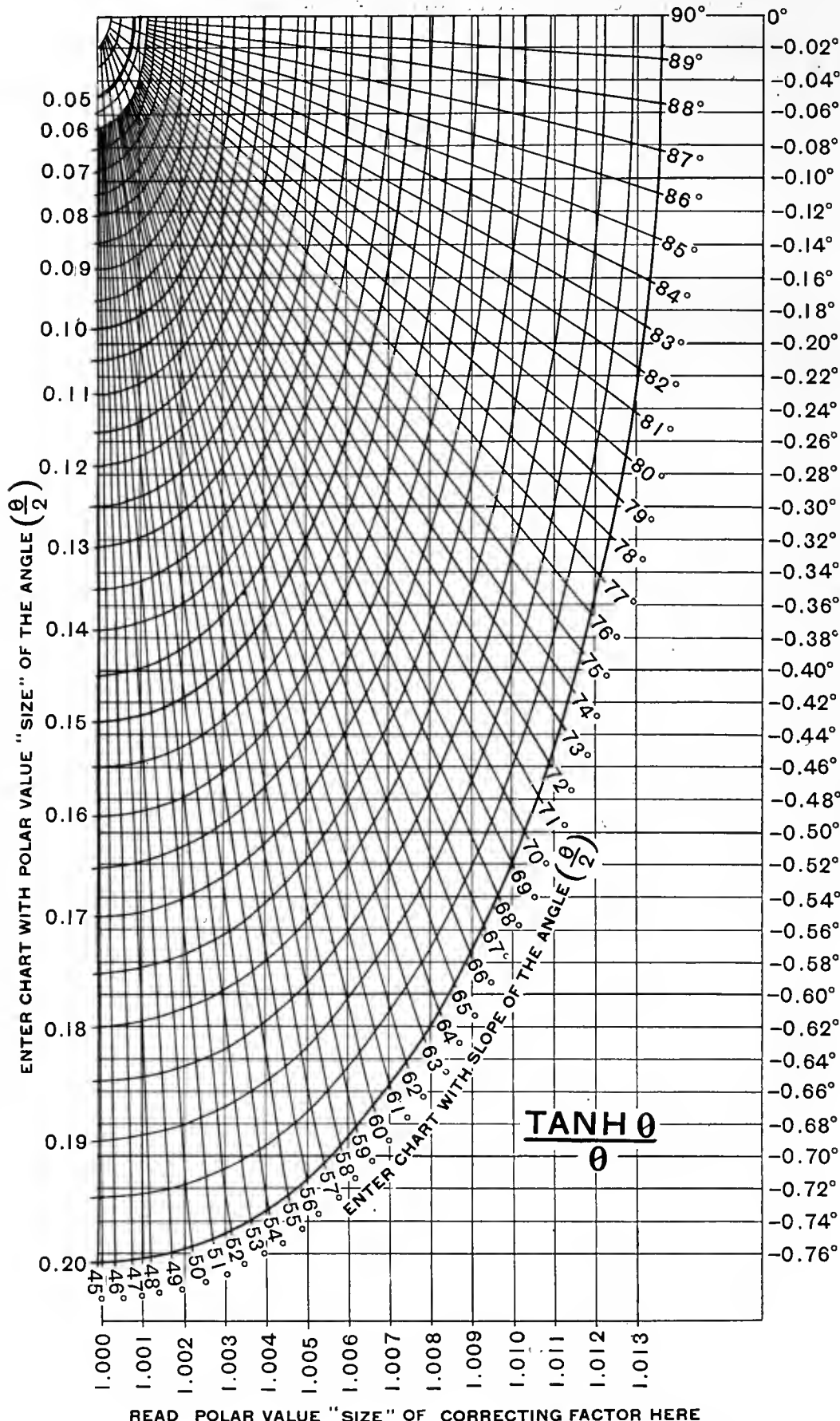
To find the vector "correcting factor" corresponding to any complex line angle θ , of a circuit, the angle θ is expressed in polar form with the slope in fractional degrees. The correcting factor as read from the chart will be in polar form with its slope in fractional degrees. Consult Table P for rapid conversion to minutes and seconds. For example:—

$$\theta = 0.8 \angle 62^\circ, \text{ correcting factor} = 0.943 \angle 5^\circ.19 = 0.943 \angle 5^\circ 11' 24''$$

$$\theta = 0.6499 \angle 78^\circ.57, \text{ correcting factor} = 0.9365 \angle 1^\circ.61 = 0.9365 \angle 1^\circ 36' 36''$$

CHART XX KENNELLY CHART FOR ADMITTANCE CORRECTING FACTOR

(FOR ANGLES HAVING POLAR VALUES BETWEEN 0 AND 0.20)



READ SLOPE OF CORRECTING FACTOR HERE

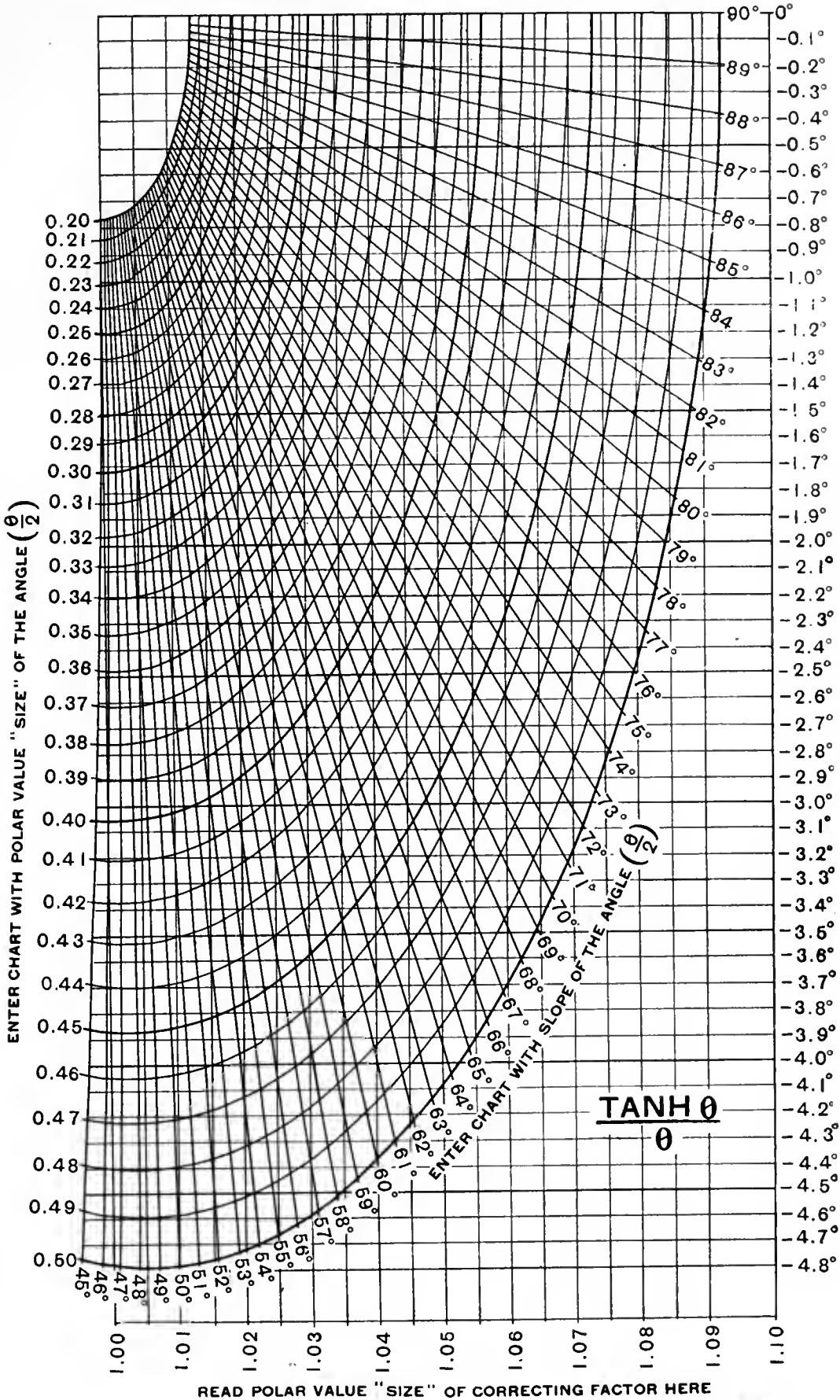
Consult Table P for rapid conversion to minutes and seconds. For example:—
 $\theta = 0.4 \angle 61^\circ, \left(\frac{\theta}{2}\right) = 0.2 \angle 61^\circ$, correcting factor = $1.007 \sqrt{0^\circ.055}$
 $= 1.007 \sqrt{0^\circ 39' 18''}$
 $\theta = 0.326 \angle 75^\circ.5, \left(\frac{\theta}{2}\right) = 0.163 \angle 75^\circ.5$, correcting factor = $1.0078 \sqrt{0^\circ.25} = 1.0078 \sqrt{0^\circ 15' 00''}$

To find the vector correcting factor corresponding to any complex line angle θ of a circuit, the angle θ is expressed in polar form with the slope in fractional degrees. The angle of the line, θ , is then divided by $2 \left(\frac{\theta}{2}\right)$ as it is necessary to enter the admittance charts with half the angle θ . The correcting factor as read from the chart will be in polar form with its slope in fractional degrees.

$$\frac{\text{TANH } \theta}{\theta}$$

CHART XXI KENNELLY CHART FOR ADMITTANCE CORRECTING FACTOR

(FOR ANGLES HAVING POLAR VALUES BETWEEN 0.20 AND 0.50)



To find the vector correcting factor corresponding to any complex line angle θ , of a circuit, the angle θ is expressed in polar form with the slope in fractional degrees. The angle of the line θ is then divided by 2 ($\frac{\theta}{2}$) as it is necessary to enter the admittance charts with half the angle θ . The correcting factor as read from the chart will be in polar form with its slope in fractional degrees. Consult Table P for rapid conversion to minutes and seconds. For example:—

ing the square root of the slope and halving the angle. The result is the hyperbolic angle θ of the circuit expressed in hyps.

The ratio charts XIX and XXI are next consulted and the correcting values $\frac{\sinh \theta}{\theta}$ and $\frac{\tanh \theta/2}{\theta/2}$ corresponding to the hyperbolic angle of the circuit read off. Having thus obtained the correcting factors corresponding to this circuit, the linear impedance Z and linear admittance Y per conductor are multiplied respectively by the \sinh and the \tanh correcting factors.

If the circuit under consideration is electrically short the effect of these correcting factors upon the linear constants will be small and possibly negligible but, as the circuit becomes longer, their effect becomes increasingly greater. The effect of the correcting factors for problem X is to change the linear impedance Z from $270.233 / 67^\circ 08' 08''$ to $Z' = 253.083 / 68^\circ 44' 41''$ and to change the linear admittance Y from $0.001563 / 90^\circ$ to $Y' = 0.001615512 / 89^\circ 10' 45''$. In other words this circuit will behave in the steady state at 60 cycles as though its conductor resistance were reduced from 105 to 91.7486 ohms and its inductive reactance reduced from 249 to 235.866 ohms. Similarly it will behave as though a non-inductive leak of 11.571 micromhos, has been applied to each condenser in shunt.

In order to illustrate the exact agreement in the results as obtained by the equivalent π method with those obtained by either the convergent series or pure hyperbolic solution, the ratio values used for this problem were calculated and not obtained graphically. The accuracy in the performance resulting from the use of ratio values taken from the charts is well within the requirements of practical power circuits. The mathematical solution for these factors is given in Fig. 48.

Having determined the corrected values for the impedance Z' and the admittance Y' which will produce exact results, the remainder of the solution may be carried out graphically as indicated by the vector diagram in the upper right hand part of Chart XVII or mathematically as indicated under this vector diagram.

EQUIVALENT T SOLUTION

Dr. Kennelly has shown that the correcting factors which convert the nominal π into the equivalent π of the conjugate smooth line, are the same as those which convert the nominal T into the equivalent T , but in inverse order;—that is the correcting factors for the nominal T line are

$$Z' = Z \frac{\tanh \theta/2}{\theta/2}$$

$$Y' = Y \frac{\sinh \theta}{\theta}$$

Either the equivalent π or the equivalent T solution may be used by applying the two correcting factors properly. Usually less arithmetical work will be required for the equivalent π solution.

ELECTRICAL CONDITIONS AT INTERMEDIATE POINTS

In the foregoing, the behavior of circuits at their terminals has been considered. In some cases it may

be desirable to predetermine the voltage and the current at points along the circuit between the terminals. This may be particularly desirable in case of circuits of great electrical length and consequently having a pronounced bend or hump in the voltage graphs representing the voltage at points along the circuit. In Fig. 21 voltage and current graphs were shown for the circuit of problem X corresponding to zero load; also load conditions. Accompanying this stated was the step-by-step method by which the current and voltage at these intermediate points had been determined. In a corresponding manner the intermediate electrical conditions may be determined by the employment of hyperbolic functions. It is usual, however, when employing hyperbolic functions for determining the voltage or the current at points along a smooth circuit, in the steady state, to take advantage of the following facts relative to the variation in current and potential from point to point in such a circuit.

The potentials of any and all points of such a circuit are as the sines and the currents as the cosines of the corresponding position angles. This means that if the position angles corresponding to two points of a smooth circuit in the steady state are known, and the voltage or the current at one of these points is also known, then the voltage or current at any other point will be directly proportional to the sine or the cosine respectively of the corresponding position angles. In a similar manner, the impedance follows the tangents, the admittance the cotangents and the volt-amperes the sines of twice the angles. Herein lies the beauty of the application of hyperbolic functions of complex angles for determining the electrical performance of electric circuits. The relationship expressed above (taken from Dr. Kennelly's "Artificial Electric Lines") are given in equation form below for ready reference:—

$$\frac{E_p}{E_c} = \frac{\sinh \theta_p}{\sinh \theta_c} \text{ numeric } \angle$$

$$\frac{I_p}{I_c} = \frac{\cosh \theta_p}{\cosh \theta_c} \text{ numeric } \angle$$

$$\frac{Z_p}{Z_c} = \frac{\tanh \theta_p}{\tanh \theta_c} \text{ numeric } \angle$$

$$\frac{Y_p}{Y_c} = \frac{\coth \theta_p}{\coth \theta_c} \text{ numeric } \angle$$

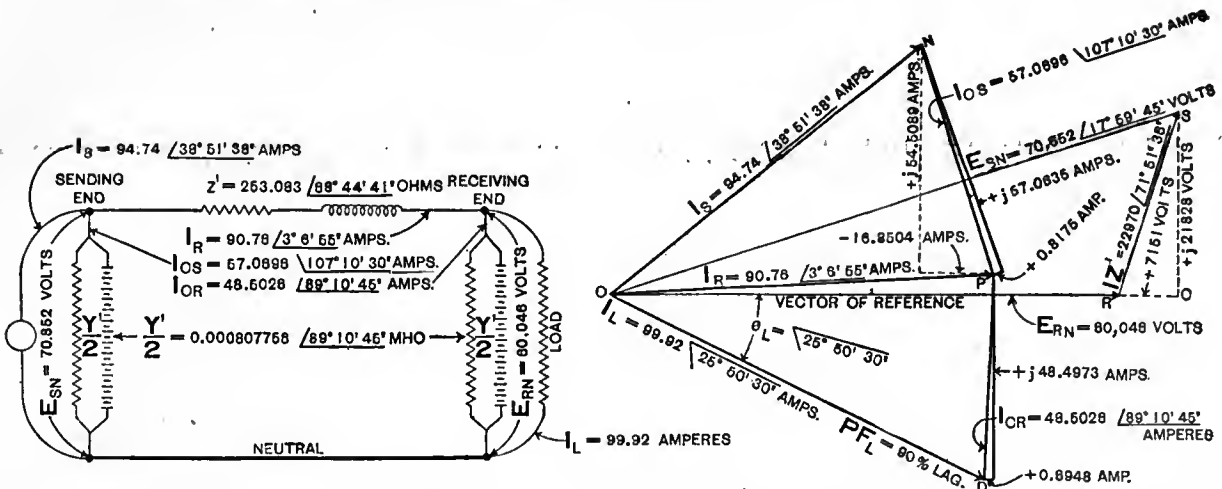
$$\left| \frac{Kv-a_p}{Kv-a_c} \right| = \left| \frac{\sinh 2 \theta_p}{\sinh 2 \theta_c} \right| \text{ numeric } \angle$$

Where p and c are points along the circuit, c being some point where the electrical conditions are known, and p the point for which they are to be computed. The vertical lines enclosing the two parts of the last equation are for the purpose of indicating that the "size" of these complex quantities are referred to in this equation.

POSITION ANGLES

Reference has been made to the line as subtending a certain complex hyperbolic angle θ . Since the circuit through the load also encounters resistance and reactance, the load may be said to subtend also a certain complex hyperbolic angle, so that the receiving end of the circuit occupies an angular position θ_r . The total

CHART XVII—RIGOROUS EQUIVALENT π SOLUTION OF PROBLEM X



CHARACTERISTICS OF CIRCUIT

LENGTH, 300 MILES. CYCLES, 60.
 CONDUCTORS—3 # 000 STRANDED COPPER.
 SPACING OF CONDUCTORS 10 X 10 X 20 FEET.
 EQUIVALENT DELTA SPACING=12.8 FT.

LINEAR CONSTANTS OF CIRCUIT

FROM TABLES PER MILE

TABLE NO. 2. $r = 0.350$ OHM AT 25° O.
 TABLE NO. 6. $x = 0.830$ OHM (BY INTERPOLATION).
 TABLE NO. 10. $b = 6.21 \times 10^{-8}$ MHO (BY INTERPOLATION)
 $g =$ (IN THIS CASE TAKEN AS ZERO).

TOTAL PER CONDUCTOR

$R = 0.350 \times 300 = 105$ OHMS TOTAL RESISTANCE.
 $X = 0.830 \times 300 = 249$ OHMS TOTAL REACTANCE.
 $B = 6.21 \times 300 \times 10^{-8} = .001583$ MHO TOTAL SUSCEPTANCE.
 $G = 0 \times 300 = 0$ MHO TOTAL CONDUCTANCE.

SOLUTION FOR HYPERBOLIC ANGLE $\theta = \sqrt{ZY}$

$$Z = 105 + j249 \quad Y = 0 + j0.001583$$

$$= 270.233 / 67^\circ 8' 8'' \quad = 0.001583 / 90^\circ$$

$$\theta = \sqrt{270.233 / 67^\circ 8' 8'' \times 0.001583 / 90^\circ}$$

$$= \sqrt{0.4223745 / 167^\circ 8' 8''}$$

$$= 0.8499035 / 78^\circ 34' 4'' \text{ HYP.}$$

$$= 0.8499035 / 78^\circ.5878 \text{ HYP.}$$

$$= 0.1288168 + j0.8370092 \text{ HYP.}$$

FROM DR. KENNELLY'S CHARTS

CHART XIX $\frac{\sinh \theta}{\theta} = 0.9365396 / 1^\circ.8094 = 0.9365396 / 1^\circ 36' 33''$
 CHART XXI $\frac{\tanh \theta/2}{\theta/2} = 1.033598 / 0^\circ.8206 = 1.033598 / 0^\circ 49' 15''$

★ THESE VALUES WERE CALCULATED IN ORDER TO OBTAIN A HIGH DEGREE OF ACCURACY FOR THE PURPOSE OF DEMONSTRATING THE FUNDAMENTAL ACCURACY OF THIS METHOD.

CORRECTION OF LINEAR CONSTANTS

$$Z' = 270.233 / 67^\circ 8' 8'' \times 0.9365396 / 1^\circ 36' 33''$$

$$= 253.083 / 68^\circ 44' 41'' \text{ (WHICH IS AUXILIARY CONSTANT (B))}$$

$$= 91.7486 + j235.868 \text{ OHMS}$$

$$Y' = 0.001583 / 90^\circ \times 1.033598 / 0^\circ 49' 15''$$

$$= 0.001615512 / 89^\circ 10' 45'' \text{ MHO}$$

$$\frac{Y'}{2} = 0.000807756 / 89^\circ 10' 45''$$

$$= 0.000011571 + j0.00080767$$

$$= 1238 / 88^\circ 10' 45'' \text{ OHMS REACTANCE.}$$

CALCULATION OF PERFORMANCE ★

PER PHASE TO NEUTRAL

$$KV_{AR_N} = \frac{18,000}{3} = 6,000. \quad KW_{RN} = \frac{18,200}{3} = 5,400.$$

$$E_{RN} = \frac{104,000}{1.732} = 80,048. \quad I_R = \frac{6,000 \times 1,000}{60,048} = 99.92.$$

$PF_R = 90\% \text{ LAGGING.}$

RECEIVING-END CONDITIONS

$$I_{CR} = 80,048 \times 0.000807756 / 89^\circ 10' 45'' = 48.5028 / 89^\circ 10' 45''$$

$$= 0.8948 + j48.4973 \text{ AMP.}$$

$$I_R = 99.92 (0.90 - j0.438) + 0.8948 + j48.4973$$

$$= 90.623 + j4.8322 \text{ AMPS.}$$

$$= 90.76 / 3^\circ 6' 55'' \text{ AMPS.}$$

$PF_R = \cos 3^\circ 6' 55'' = 99.95\% \text{ LEADING.}$

$$KW_{CR} = 80,048 (0.8948 + j48.4973)$$

$$= 41.72 + j2912.089$$

$$KW_{RN} = 8000 (0.90 - j0.438) + 41.72 + j2912.07$$

$$= 5441.72 + j2988.07$$

$$I_R Z' = 90.76 / 3^\circ 6' 55'' \times 253.083 / 68^\circ 44' 41''$$

$$= 22970 / 71^\circ 51' 38'' \text{ VOLTS}$$

$$= 7151 + j21,828 \text{ VOLTS}$$

SENDING-END CONDITIONS

$$E_{SN} = 80,048 + 7151 + j21,828$$

$$= 87,197 + j21,828 \text{ VOLTS}$$

$$= 70,852 / 17^\circ 59' 45'' \text{ VOLTS}$$

$$I_{CS} = 70,852 \times 0.000807756 / 89^\circ 10' 45''$$

$$= 0.8175 + j57.0635 \text{ (TO SUPPLY END VOLTAGE)}$$

$$= 57.0698 / 107^\circ 10' 30'' \text{ TO VECTOR OF REFERENCE}$$

$$= -18.8504 + j54.5089$$

$$I_S = (90.623 + j4.8322) + (-18.8504 + j54.5089)$$

$$= 73.77 + j59.44 \text{ AMPS.}$$

$$= 94.74 / 38^\circ 51' 38'' \text{ AMPS.}$$

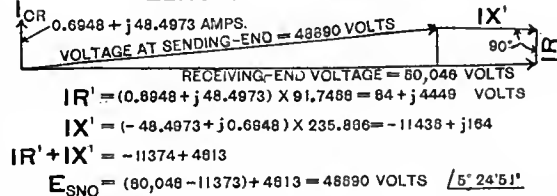
$$PF_S = \cos 38^\circ 51' 38'' - 17^\circ 59' 45'' = 93.44\% \text{ LEADING}$$

$$KW_{SN} = 70,852 \times 94.74 \times 0.9344 = 6255 \text{ KW PER PHASE}$$

LOSS = 6255 - 5400 = 855 KW PER PHASE

$$EFF = \frac{5400 \times 100}{6255} = 86.33\%$$

ZERO LOAD CONDITIONS



*The above results check with those obtained by convergent series. (See Chart XIII).

angle of the circuit (line and load) will be $\theta_r + \theta = \theta_s$. By similar reasoning all points lying between the receiving and sending ends of a line will occupy or assume an angular position θ_p . If that part of the linear angle θ of the line between the receiving end and the point p be designated as θ_{pr} , then the angular position of the point p will be $\theta_p = \theta_r + \theta_{pr}$. Thus, at a point in the middle of the line, the position angle will be $\theta_p = \theta_r + \theta_{pr} = \theta_r + \theta/2$.

If the line is grounded or short-circuited at the receiving end, there will be no load containing resistance and reactance, and consequently no load angle. In such case $\theta_r = 0$ and the distribution of position angles along the line will be purely a linear function of the total line angle θ . In such a case $\theta_s = \theta$.

Load Conditions—In Fig. 49 the procedure is shown which may be followed for determining by complex functions of position angles the current and the voltage vectors at points 25 miles apart along problem X circuit, under load conditions.

The procedure is first to determine the complex angle θ_r , at the receiving end resulting from the load. The mathematical determination of this load angle is tedious. Such determination is given for problem X circuit under stated load in Fig. 49. This complex angle θ_r of the load (that is the position angle at the receiving end) is such that its complex tangent equals the impedance load δ to ground, or zero potential, at the receiving end of line (ohms \angle) divided by the surge impedance Z_0 of a conductor (ohms \angle). That is,—

$$\tanh \theta_r = \frac{\delta}{Z_0}$$

Since we are here interested only in the ratio between the load impedance and the surge impedance, the values may be taken either per unit length or total per conductor. Although $\tanh \theta_r$ is readily calculated, as may be seen by consulting Fig. 49, the subsequent calculation for the corresponding angle θ_r is tedious. After having calculated the $\tanh \theta_r$, the corresponding angle θ_r may be obtained with sufficient accuracy from a table of tangents of complex angles or, more readily still, from a chart of such functions.* After having determined the angle θ_r by consulting a chart of tangents of complex angles, or by mathematical calculation, as in Fig. 49, the position angles at points along the circuit may easily and readily be determined as follows:

The change in the position angle from point to point along the circuit, due to the line impedance and the line admittance is purely a linear function of the line angle θ . This is the case whether the line is grounded, loaded or free at the receiving end.

Referring to Fig. 49, the angular position of the receiving end, due to the load conditions assumed, was calculated to be $0.48047 + j 1.06354$. It is therefore necessary to add this angle to each of the linear line angles of the various points along the line in order to obtain the position angles of the points in question.

Thus the linear line angle of the middle point of the circuit is $0.0644084 + j 0.3185046$ and adding to this the load angle $0.48047 + j 1.06354$ gives $0.544874 + j 1.3820446$, which corresponds with the entry in the tabulation of Fig. 51 for the position angle at the middle of the circuit. In a similar manner position angles for the load assumed are readily determined for points 25 miles apart. Having determined the position angles for the various points along the circuit, the sines and the cosines corresponding to these position angles may be approximated closely from tables or charts of such complex functions, or may be calculated accurately by following the equations at the lower left hand corner of Fig. 51. Since the receiving end voltage and current are known to be 60 046 volts and 99.92 amperes respectively, the voltage and currents at all other points of this circuit will be as the sines and cosines of the corresponding position angles. From the vector quantities that have been assigned to the voltage and current at the points along the circuit, the power-factors at these points are readily determined.

The current and voltage graphs at the bottom of Fig. 51 were plotted from values as determined by the use of functions of position angles. These check exactly with similar graphs as determined by the Wilkinson charts and step-by-step process (See Fig. 21).

Zero Load Condition—The procedure which may be followed for determining the position angles under zero load, their functions and the corresponding current and voltage distribution is the same as given above for load conditions and is shown in Fig. 50. In this case, however, there is no load and consequently no real part to the load angle. On the other hand the impedance of the load is infinite, that is $\delta = \infty$ so that $\theta_r = \tanh^{-1} \frac{\infty}{Z_0} = j \frac{\pi}{2}$. The effect of this supersurge impedance load at the receiving end at zero load is to cause a phase rotation of 90 degrees or one quadrant, $j \frac{\pi}{2} = 1.57080$ circular radians. Thus, at zero load, $\theta_{r0} = (0 + j \frac{\pi}{2}) = 0 + j 1.57080$ and this angle must be added to each of the linear position angles of the points along the line. With the position angles corresponding to zero load thus obtained, and assigned to the points along the circuit, the voltage will be found to follow the sines, and the current the cosines, etc. of these position angles.

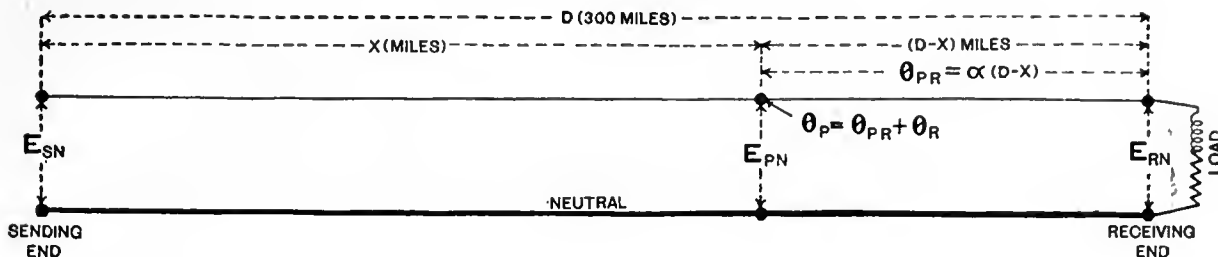
POLAR DIAGRAM OF CURRENT VOLTAGE

In Fig. 52 are shown the polar graphs of the voltage and the current for problem X, corresponding to load, and also to zero load conditions. These polar graphs were plotted from the vector values for current and voltage as tabulated in Figs. 49 and 50 for each 25 miles of circuit.

*Such as that worked out by Dr. Kennelly and published by the Harvard University Press. The chart atlas referred to contains graphs of complex tangents of complex angles, and by following the chart in the reverse from the usual direction the complex angle corresponding to any complex tangent may be read off directly.

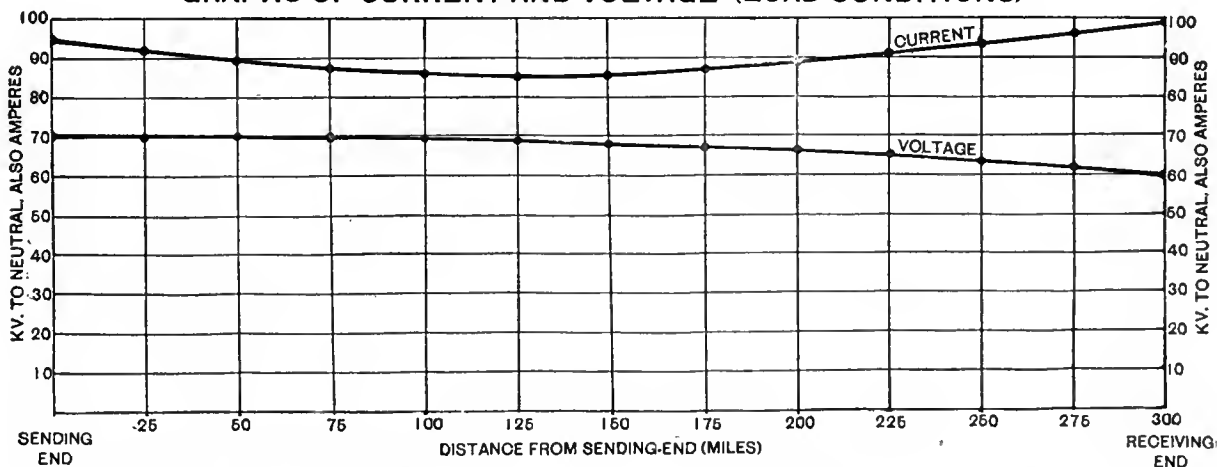
CURRENT & VOLTAGE DISTRIBUTION

(LOAD CONDITIONS)



(D-X) MILES	X MILES	POSITION ANGLE $\theta_P = \theta_{PR} + \theta_R$	SINH θ_P (THE VOLTAGE FOLLOWS THIS COMPLEX FUNCTION)	E_{PN} VOLTS \angle	COSH θ_P (THE CURRENT FOLLOWS THIS COMPLEX FUNCTION)	I_P AMPERES \angle	PF P %
0	300	$0.48047 + j1.06354$ $\theta_2 = 60^\circ 56' 11''$	$0.24249 + j0.977693$ $= 1.00657 \angle 76^\circ 03' 35''$	60.046 $\angle 0^\circ 0' 0''$	$0.54294 + j0.43632$ $= 0.69654 \angle 38^\circ 47' 10''$	99.92 $\angle 25^\circ 50' 31''$	-90.00
25	275	$0.49120 + j1.11662$ $\theta_2 = 63^\circ 58' 40''$	$0.22426 + j1.0092$ $= 1.0338 \angle 77^\circ 28' 16''$	61.670 $\angle 1^\circ 24' 41''$	$0.49272 + j0.45937$ $= 0.67364 \angle 42^\circ 59' 38''$	96.64 $\angle 21^\circ 38' 03''$	-93.83
50	250	$0.50194 + j1.16971$ $\theta_2 = 67^\circ 01' 10''$	$0.20430 + j1.0391$ $= 1.0590 \angle 78^\circ 52' 36''$	63.173 $\angle 2^\circ 49' 01''$	$0.44064 + j0.48176$ $= 0.65288 \angle 47^\circ 33' 08''$	93.66 $\angle 17^\circ 04' 33''$	-94.04
75	225	$0.51267 + j1.22279$ $\theta_2 = 70^\circ 03' 39''$	$0.18259 + j1.0663$ $= 1.0819 \angle 80^\circ 16' 59''$	64.540 $\angle 4^\circ 13' 24''$	$0.38682 + j0.50333$ $= 0.63480 \angle 52^\circ 27' 25''$	91.06 $\angle 12^\circ 10' 20''$	-95.94
100	200	$0.52341 + j1.27587$ $\theta_2 = 73^\circ 06' 09''$	$0.15917 + j1.0909$ $= 1.1025 \angle 81^\circ 41' 55''$	65.770 $\angle 5^\circ 38' 20''$	$0.33139 + j0.52399$ $= 0.61999 \angle 57^\circ 41' 22''$	88.94 $\angle 6^\circ 56' 19''$	-97.60
125	175	$0.53414 + j1.32895$ $\theta_2 = 76^\circ 08' 38''$	$0.13409 + j1.1127$ $= 1.1207 \angle 83^\circ 07' 43''$	66.854 $\angle 7^\circ 04' 08''$	$0.27447 + j0.54361$ $= 0.60815 \angle 63^\circ 12' 39''$	87.24 $\angle 1^\circ 25' 02''$	-98.90
150	150	$0.54488 + j1.38204$ $\theta_2 = 79^\circ 11' 07''$	$0.10735 + j1.1317$ $= 1.1368 \angle 84^\circ 34' 52''$	67.815 $\angle 8^\circ 31' 17''$	$0.21618 + j0.56197$ $= 0.60211 \angle 68^\circ 57' 32''$	86.37 $\angle 4^\circ 19' 51''$	-99.73
175	125	$0.55561 + j1.43512$ $\theta_2 = 82^\circ 13' 36''$	$0.07908 + j1.1477$ $= 1.1504 \angle 86^\circ 03' 30''$	68.626 $\angle 9^\circ 59' 55''$	$0.15667 + j0.57927$ $= 0.60080 \angle 74^\circ 51' 57''$	86.34 $\angle 10^\circ 14' 16''$	+99.99
200	100	$0.56635 + j1.48821$ $\theta_2 = 85^\circ 16' 05''$	$0.04926 + j1.1607$ $= 1.1618 \angle 87^\circ 34' 11''$	69.306 $\angle 11^\circ 30' 36''$	$0.09608 + j0.59508$ $= 0.60279 \angle 80^\circ 49' 42''$	86.47 $\angle 16^\circ 12' 01''$	+99.66
225	75	$0.57708 + j1.54129$ $\theta_2 = 88^\circ 18' 35''$	$0.01798 + j1.1707$ $= 1.1708 \angle 89^\circ 07' 13''$	69.843 $\angle 13^\circ 03' 38''$	$0.03455 + j0.60939$ $= 0.60962 \angle 86^\circ 45' 18''$	87.45 $\angle 22^\circ 07' 39''$	+98.75
250	50	$0.58782 + j1.59438$ $\theta_2 = 91^\circ 21' 04''$	$-0.01471 + j1.1775$ $= 1.1775 \angle 90^\circ 42' 57''$	70.243 $\angle 14^\circ 39' 22''$	$-0.02784 + j0.62207$ $= 0.62270 \angle 92^\circ 33' 44''$	89.33 $\angle 27^\circ 56' 03''$	+97.32
275	25	$0.59855 + j1.64746$ $\theta_2 = 94^\circ 23' 34''$	$-0.04863 + j1.1811$ $= 1.1821 \angle 92^\circ 21' 28''$	70.517 $\angle 16^\circ 17' 53''$	$-0.09073 + j0.63306$ $= 0.63953 \angle 98^\circ 09' 22''$	91.74 $\angle 34^\circ 11' 41''$	+95.17
300	0	$0.60929 + j1.70055$ $\theta_2 = 97^\circ 26' 03''$	$-0.08381 + j1.1814$ $= 1.1844 \angle 94^\circ 03' 28''$	70.652 $\angle 17^\circ 59' 53''$	$-0.15416 + j0.64226$ $= 0.66050 \angle 103^\circ 29' 45''$	94.75 $\angle 38^\circ 52' 04''$	+93.43

GRAPHS OF CURRENT AND VOLTAGE (LOAD CONDITIONS)



$\text{SINH}(\theta_1 + j\theta_2) = (\text{SINH} \theta_1 \text{COS} \theta_2 + j \text{COSH} \theta_1 \text{SIN} \theta_2)$

$\text{COSH}(\theta_1 + j\theta_2) = (\text{COSH} \theta_1 \text{COS} \theta_2 + j \text{SINH} \theta_1 \text{SIN} \theta_2)$

ANGLE AT RECEIVING END $\theta_R = 0.48047 + 1.06354$

ANGLE OF LINE $\theta = 0.12882 + 0.63701$

$\theta_S = \theta + \theta_R = 0.60929 + 1.70055$

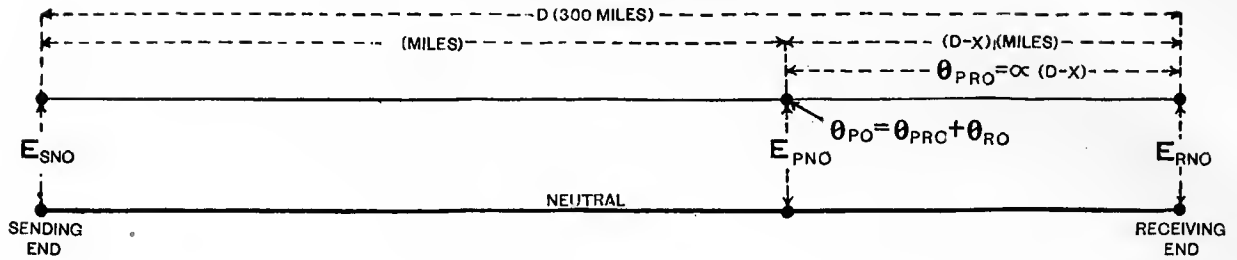
ONE QUADRANT = 1.57079632 CIRCULAR RADIAN.

ONE CIRCULAR RADIAN = 206264.8062" = 67° 17' 44.8"

$\alpha = 0.00042839 + j0.00212336$

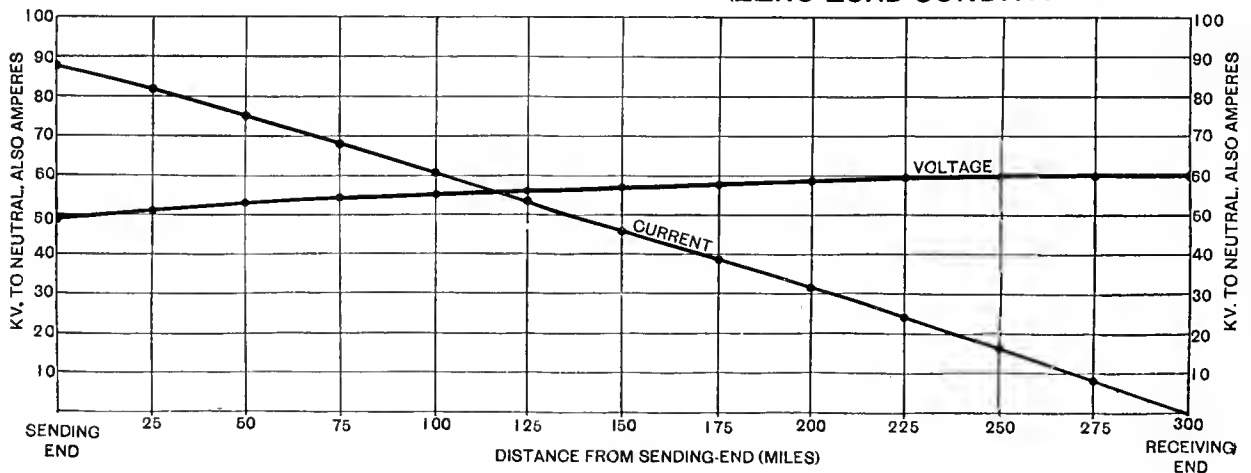
FIG. 51—CURRENT AND VOLTAGE DISTRIBUTION For problem X by position angles (load conditions).

CURRENT AND VOLTAGE DISTRIBUTION
(ZERO LOAD CONDITION)



(D-X) MILES	X MILES	POSITION ANGLE $\theta_{PO} = \theta_{PRO} + \theta_{RO}$	SINH θ_{PO} (THE VOLTAGE FOLLOWS THIS COMPLEX FUNCTION)	E_{PNO} VOLTS \angle	COSH θ_{PO} (THE CURRENT FOLLOWS THIS COMPLEX FUNCTION)	I_{PO} AMPERES \angle	PF $_{PO}$ %
0	300	$0 + j1.57080$ $\theta_2 = 90^\circ 00' 00''$	$0 + j1.00000$ $= 1.00000 \angle 90^\circ$	60046 $\angle 0^\circ$	0 $0 \angle 90^\circ$	0 $\angle 11^\circ 25' 56''$	
25	275	$0.01073 + j1.62388$ $\theta_2 = 93^\circ 02' 29''$	$0.00057 + j0.99865$ $= 0.99865 \angle 89^\circ 58' 03''$	59965 $\angle 0^\circ 01' 57''$	$0.05307 + j0.01070$ $= 0.05414 \angle 11^\circ 23' 58''$	7.82 $\angle 90^\circ 01' 58''$	0
50	250	$0.02146 + j1.67696$ $\theta_2 = 96^\circ 04' 58''$	$0.00227 + j0.99460$ $= 0.99460 \angle 89^\circ 52' 09''$	59803 $\angle 0^\circ 07' 51''$	$0.10598 + j0.02133$ $= 0.10816 \angle 11^\circ 22' 48''$	15.62 $\angle 90^\circ 03' 08''$	+00.12
75	225	$0.03220 + j1.73004$ $\theta_2 = 99^\circ 07' 28''$	$0.00511 + j0.98785$ $= 0.98786 \angle 89^\circ 42' 13''$	59317 $\angle 0^\circ 17' 47''$	$0.15866 + j0.03179$ $= 0.16181 \angle 11^\circ 19' 48''$	23.37 $\angle 90^\circ 06' 08''$	+00.32
100	200	$0.04294 + j1.78313$ $\theta_2 = 102^\circ 09' 57''$	$0.00905 + j0.97844$ $= 0.97847 \angle 89^\circ 28' 12''$	58753 $\angle 0^\circ 31' 48''$	$0.21090 + j0.04199$ $= 0.21503 \angle 11^\circ 15' 35''$	31.05 $\angle 90^\circ 10' 21''$	+00.61
125	175	$0.05367 + j1.83621$ $\theta_2 = 105^\circ 12' 26''$	$0.01409 + j0.96638$ $= 0.96648 \angle 89^\circ 09' 50''$	58033 $\angle 0^\circ 50' 10''$	$0.26269 + j0.051842$ $= 0.26776 \angle 11^\circ 09' 50''$	38.66 $\angle 90^\circ 16' 06''$	+00.99
150	150	$0.06441 + j1.88930$ $\theta_2 = 108^\circ 14' 56''$	$0.02018 + j0.95168$ $= 0.95188 \angle 88^\circ 47' 07''$	57156 $\angle 1^\circ 12' 53''$	$0.31380 + j0.06120$ $= 0.31970 \angle 11^\circ 02' 10''$	46.17 $\angle 90^\circ 23' 46''$	+1.42
175	125	$0.07514 + j1.94238$ $\theta_2 = 111^\circ 17' 25''$	$0.02731 + j0.93436$ $= 0.93476 \angle 88^\circ 19' 33''$	56129 $\angle 1^\circ 40' 27''$	$0.36417 + j0.07006$ $= 0.37085 \angle 11^\circ 53' 22''$	53.55 $\angle 90^\circ 32' 33''$	+1.98
200	100	$0.08588 + j1.99546$ $\theta_2 = 114^\circ 19' 54''$	$0.03543 + j0.91452$ $= 0.91522 \angle 87^\circ 46' 53''$	54955 $\angle 2^\circ 13' 07''$	$0.41354 + j0.07835$ $= 0.42090 \angle 10^\circ 43' 41''$	60.77 $\angle 90^\circ 42' 15''$	+2.65
225	75	$0.09661 + j2.04854$ $\theta_2 = 117^\circ 22' 24''$	$0.04449 + j0.89218$ $= 0.89328 \angle 87^\circ 08' 43''$	53638 $\angle 2^\circ 51' 17''$	$0.46194 + j0.08593$ $= 0.46986 \angle 10^\circ 32' 16''$	67.85 $\angle 90^\circ 53' 40''$	+3.40
250	50	$0.10735 + j2.10164$ $\theta_2 = 120^\circ 24' 53''$	$0.05445 + j0.86735$ $= 0.86905 \angle 86^\circ 24' 28''$	52183 $\angle 3^\circ 35' 32''$	$0.50917 + j0.09275$ $= 0.51755 \angle 10^\circ 19' 26''$	74.73 $\angle 91^\circ 06' 30''$	+4.33
275	25	$0.11808 + j2.15473$ $\theta_2 = 123^\circ 27' 22''$	$0.06525 + j0.84014$ $= 0.84267 \angle 85^\circ 33' 33''$	50599 $\angle 4^\circ 26' 27''$	$0.55514 + j0.09874$ $= 0.56385 \angle 10^\circ 05' 07''$	81.42 $\angle 91^\circ 20' 49''$	+5.41
300	0	$0.12882 + j2.20781$ $\theta_2 = 126^\circ 29' 52''$	$0.07683 + j0.81056$ $= 0.81420 \angle 84^\circ 35' 08''$	48889 $\angle 5^\circ 24' 52''$	$0.59973 + j0.10384$ $= 0.60865 \angle 9^\circ 49' 22''$	87.89 $\angle 91^\circ 36' 34''$	+6.64

GRAPHS OF CURRENT AND VOLTAGE (ZERO LOAD CONDITIONS)



$\text{SINH}(\theta_1 + j\theta_2) = (\text{SINH} \theta_1 \text{COS} \theta_2 + j \text{COSH} \theta_1 \text{SIN} \theta_2)$
 $\text{COSH}(\theta_1 + j\theta_2) = (\text{COSH} \theta_1 \text{COS} \theta_2 + j \text{SINH} \theta_1 \text{SIN} \theta_2)$

ANGLE AT RECEIVING END $\theta_{RO} = 0 + j1.57080$
 ANGLE OF LINE $\theta = 0.12882 + j0.83701$
 $\theta_{SO} = \theta + \theta_{RO} = 0.12882 + j2.20781$

ONE QUADRANT = 1.57079632 CIRCULAR RADIAN
 ONE CIRCULAR RADIAN = 206264.8062" = 57° 17' 44.8"

$\alpha = 0.00042939 + j0.00212338$

FIG. 50—CURRENT AND VOLTAGE DISTRIBUTION
 For problem X by position angles (zero load conditions).

<p style="text-align: center;">PROBLEM "X"</p> <p>R=105 OHMS, X=249 OHMS. G=0 MHO, Y=0.001563 MHO. Z=270.233 / 67°08'08" Y=0.001563 / 90° $\theta = \sqrt{ZY} = \sqrt{0.4223745 / 157°08'08"}$ = 0.6499035 / 78°34'04" = 0.1288168 + j0.6370092</p>	<p style="text-align: center;">CALCULATION FOR $\frac{\theta}{2}$</p> <p>$\frac{\theta}{2} = 0.3249518 / 78°34'04"$ = 0.0644084 + j0.3185046 $\sinh \frac{\theta}{2} = 0.06121122 + j0.3137963$ = 0.3197107 / 78°57'43" $\cosh \frac{\theta}{2} = 0.9516754 + j0.0201832$ = 0.9518894 / 1°12'54"</p> <p>$\text{TANH } \frac{\theta}{2} = \frac{\sinh(\theta/2)}{\cosh(\theta/2)} = \frac{0.3197107 / 78°57'43"}{0.9518894 / 1°12'54"}$ = 0.3358696 / 77°44'49" $\text{TANH } (\theta/2) = \frac{0.3358696 / 77°44'49"}{0.3249518 / 78°34'04"}$ = 1.033598 / 0°49'15" = ADMITTANCE CORRECTING FACTOR.</p>
<p style="text-align: center;">CALCULATION FOR $\frac{\theta}{2}$</p> <p>$\sinh \theta = 0.1038393 + j0.599735$ = 0.6086583 / 80°10'38" *</p> <p>$\frac{\sinh \theta}{\theta} = \frac{0.6086583 / 80°10'38"}{0.6499035 / 78°34'04"}$ = 0.9365365 / 1°36'33" = IMPEDANCE CORRECTING FACTOR.</p>	

CHECK, $\sinh \theta = 2 \sinh \frac{\theta}{2} \cosh \frac{\theta}{2} = 2 \times 0.3197107 / 78°57'43" \times 0.9518894 / 1°12'54"$
= 0.6086583 / 80°10'38" (WHICH CHECKS WITH *).

FIG. 48—MATHEMATICAL DETERMINATION OF CORRECTING FACTORS FOR EQUIVALENT π SOLUTION

<p style="text-align: center;">PROBLEM "X"</p> <p>Z=105+j249=270.233 / 67°08'08" OHMS. Y=0+j0.001563=0.001563 / 90° MHO. $\theta = \sqrt{ZY} = 0.1288168 + j0.6370092$ HYP. KV-$A_{RN} = 6000000 \sqrt{25°50'31"}$ WATTS. = 5400000 - j2615340. $E_{RN} = 60044.4$ VOLTS TO NEUTRAL. $I_R = 99.92605 \sqrt{25°50'31"}$</p>	<p style="text-align: center;">SOLUTION FOR $\text{TANH } \theta_R$</p> <p>$\delta = \frac{E_{RN}}{I_R} = 600.888 / 25°50'31"$ OHMS. $Z_0 = \sqrt{\frac{Z}{Y}} = 415.805 \sqrt{11°25'56"}$ OHMS. $\text{TANH } \theta_R = \frac{600.888 / 25°50'31"}{415.805 \sqrt{11°25'56"}}$ = 1.44512 / 37°16'27" = 1.14995 + j0.875209 * = ($\theta_1 + j\theta_2$)</p>
<p style="text-align: center;">SOLUTION FOR ANGLE θ_R</p> $\text{TANH}^{-1}(\theta_1 \pm j\theta_2) = \frac{1}{2} \text{LOGH} \sqrt{\frac{(1+\theta_1)^2 + \theta_2^2}{(1-\theta_1)^2 + \theta_2^2}} + j \left[\frac{\pi - \text{TAN}^{-1}\left(\frac{\theta_1+1}{\pm\theta_2}\right) + \text{TAN}^{-1}\left(\frac{\theta_1-1}{\pm\theta_2}\right)}{2} \right]$ $\theta_R = \frac{1}{2} \text{LOGH} \sqrt{\frac{(1+1.14995)^2 + 0.875209^2}{(1-1.14995)^2 + 0.875209^2}} + j \left[\frac{180^\circ - \text{TAN}^{-1} \frac{2.14995}{0.875209} + \text{TAN}^{-1} \frac{0.14995}{0.875209}}{2} \right]$ $= \frac{1}{2} \text{LOGH} \sqrt{\frac{4.62229 + 0.76599}{0.022485 + 0.76599}} + j \frac{(180^\circ - 67^\circ 50' 58" + 9^\circ 43' 20")}{2}$ $= \frac{1}{2} (\text{LOGH } 2.61415) + j60^\circ 56' 11"$ $= \frac{1}{2} (0.960939) + j1.0635397$ <p>$\theta_R = 0.4804695 + j1.0635397$ $\theta = 0.1288168 + j0.6370092$ $\theta_s = 0.6092863 + j1.7005489$</p> <p style="text-align: center;">CHECK</p> <p>$\sinh \theta_R = 1.006572 / 76^\circ 03' 36"$ $\cosh \theta_R = 0.69653 / 38^\circ 47' 09"$ $\text{TANH } \theta_R = \frac{\sinh \theta_R}{\cosh \theta_R} = \frac{1.006572 / 76^\circ 03' 36"}{0.69653 / 38^\circ 47' 09"}$ = 1.14995 + j0.875209 (WHICH CHECKS WITH *).</p>	

FIG. 49—POSITION ANGLE θ_R AT RECEIVING-END
Mathematical determination at load conditions.

CHOICE OF VARIOUS METHODS

Two graphical and two mathematical forms of solution for circuits of long electrical length have been described thus far. These four methods have been given for the purpose of providing a choice of procedure for the beginner. Graphical solutions are more simple and more readily performed than mathematical solutions and, if used correctly and made to a large scale, will yield results well within the limits of permissible error for power transmission circuits. There is always a possibility of error with any method, even though the solution is carefully checked. For this reason it is desirable that errors be guarded against by the use of two different forms of solution. For instance

wave, then it will be necessary to take their effect into account, if high accuracy is essential. In such a case there is an independent solution required of potential and current for each single frequency in turn, as though the others did not exist, and then the r.m.s. value at any point on the line is the perpendicular sum of the separate frequency values.

A detail discussion of the manner of including the effect of harmonic components in the current and voltage waves is quoted below from Dr. Kennelly's "Artificial Electric Lines."

"The ordinary complex harmonic impressed e.m.f. contains a fundamental frequency associated with multiple frequency harmonics. The *n*th multiple of the frequency is called the *n*th harmonic. The fundamental may thus be included as the first harmonic.

"In order to deal with the plural-frequency case quantitatively, it is necessary to analyze the impressed potential wave into its harmonic components. As is well known, the complete Fourier analysis of a complex wave may be written

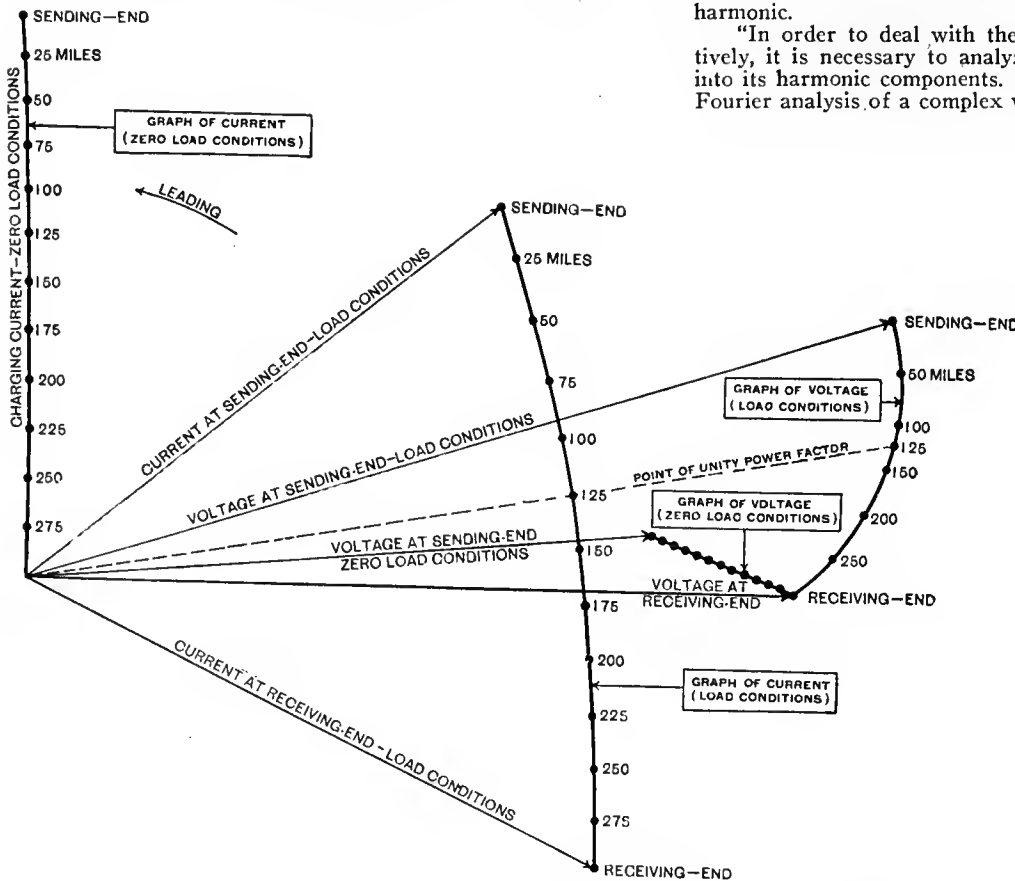


FIG. 52—POLAR DIAGRAM OF CURRENT AND VOLTAGE DISTRIBUTION FOR PROBLEM X

the first solution could be made by making use of the Wilkinson charts followed by its accompanying graphical solution. The second solution could then be made by means of Dr. Kennelly's ratio charts XVIII to XXI, followed by its accompanying graphical solution. These two methods would then yield results obtained by two entirely different routes and methods of procedure. The use of two such methods would constitute check against errors being made in either solution.

EFFECT OF HARMONIC CURRENTS AND VOLTAGES

The foregoing discussion is based upon the assumption that the fundamental wave is of sine shape and consequently free from harmonics. If harmonics of considerable magnitude are present in the fundamental

$$V_0 + V'_1 \sin \omega t + V'_2 \sin 2\omega t + V'_3 \sin 3\omega t + V'_4 \sin 4\omega t + \dots + V''_1 \cos \omega t + V''_2 \cos 2\omega t + V''_3 \cos 3\omega t + V''_4 \cos 4\omega t + \dots \text{ volts (1)}$$

where V_0 is a continuous potential, such as might be developed by a storage battery, ordinarily absent in an a. c. generator wave, V'_1, V''_1, V'_2, V''_2 , etc., maximum cyclic amplitudes of the various sine and cosine components. The even harmonics are ordinarily negligible in an a. c. generator wave; so that V'_2, V''_2, V'_4, V''_4 , etc., are ordinarily all zeros. If we count time from some moment when the fundamental component passes through zero in the positive direction, $V''_1 = 0$ and the series becomes

$$V'_1 \sin \omega t + V'_3 \sin 3\omega t + V'_5 \sin 5\omega t + \dots + V''_3 \cos 3\omega t + V''_5 \cos 5\omega t + \dots \text{ volts (2)}$$

Compounding sine and cosine harmonic components into resultant harmonics of displaced phase, this may be expressed as $V_{r1} \sin \omega t + V_{r3} \sin (3\omega t + \beta_3) + V_{r5} \sin (5\omega t + \beta_5) + \dots \text{ volts (3)}$

where $V_{rn} = \sqrt{V'^n_{} + V''^n_{}}$ volts (4)

and $\tan \beta_n = \frac{V''_n}{V'_n}$ numeric (5)

Formulas (1) and (2) give the wave analysis in sine and cosine harmonics, while (3) gives it in resultant sine harmonics.

"When considering a plural-frequency alternating-current line, we require to know the harmonic analysis of the impressed potential, either in sine and cosine harmonics, or in resultant harmonics, the latter analysis is preferable, as being shorter and containing fewer terms. A decision must be made as to the number of frequencies or upper harmonics which must be taken into account.

"Ordinarily, the sizes of the harmonics diminish as their order increases; but there are numerous exceptions to this rule, as when some particular tooth frequency in the alternating-current generator establishes a prominent size for that harmonic. Care must therefore be exercised not to exclude any important harmonics. On the other hand, the fewer the harmonics to be dealt with, the better, because the labor involved in correctly solving the problem increases in nearly the same ratio as the number of harmonics retained.

"The rule is to work out the position angle, r.m.s. potential, and r.m.s. current distributions, over the artificial or conjugate smooth line, for each harmonic component in turn, as though it existed alone, and then to combine them, at each position, in the well-known way for root mean squares.

"Combination of Components of Different Frequencies into a R.m.s. Resultant.—Let the r.m.s. value of each alternating-current harmonic component be obtained by dividing its amplitude with $\sqrt{2}$ in the usual way, and let

$$V_n = \frac{V_{rn}}{\sqrt{2}} = \sqrt{\frac{V'_n{}^2 + V''_n{}^2}{2}} \quad \text{r.m.s. volts (6)}$$

be the r.m.s. value of the n th harmonic. Then the r.m.s. value of all the harmonics together, over any considerable number of cycles, will be

$$V = \sqrt{V_1^2 + V_2^2 + V_3^2 + \dots} \quad \text{r.m.s. volts (7)}$$

or, as is well known, the joint r.m.s. value of a plurality of r.m.s. values of different frequency, is the square root of the sum of their squares. If a continuous potential V_0 be present, this may be regarded as a r.m.s. harmonic of zero frequency, and be included thus:

$$V = \sqrt{V_0^2 + V_1^2 + V_2^2 + V_3^2 + \dots} \quad \text{r.m.s. volts (8)}$$

Moreover, from (4), it is evident that the squares of the r.m.s. values of the sine and cosine terms of any harmonic may be

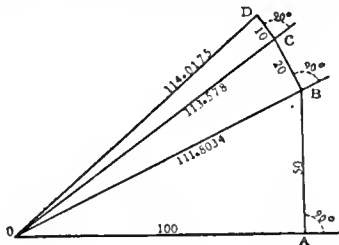


FIG. 53—GEOMETRICAL REPRESENTATION OF A JOINT R.M.S. VALUE OF PLURAL-FREQUENCY COMPONENTS BY PERPENDICULAR SUMMATION OR "CRAB ADDITION"

substituted for the square of their resultant; or that, in this respect, the sine and cosine terms may be treated as though they were components of different frequencies.

"The same procedure applies to plural-frequency currents. Find the r.m.s. resultant harmonics. The r.m.s. value of all together will be the square root of the sum of their squares. A continuous current, if present, may be included, as the r.m.s. value of an alternating current of zero frequency.

"Graphical Representation of R.m.s. Plural-frequency Combination.—The process represented algebraically in (7) or (8) may be represented graphically by the process of successive perpendicular summation, or "crab addition." An example will suffice to make this clear. A fundamental alternating current of 100 amp. r.m.s., is associated with a continuous current of 50 amp., and with two other alternating currents of other frequencies of 20 and 10 amp. r.m.s., respectively. What will be the joint r.m.s. current? Here by (8),

$$I = \sqrt{100^2 + 50^2 + 20^2 + 10^2} = \sqrt{10000 + 2500 + 400 + 100} = \sqrt{13000} = 114.0175 \text{ amp. r.m.s.}$$

"In Fig. 53, OA represents the fundamental r.m.s. current. AB , added perpendicularly to OA represents the continuous current, or current of 50 r.m.s. amp. at zero frequency. The perpendicular sum of OA and AB is $OB = 111.8034$ amp. Adding similarly the other frequency components BC and CD ,

the total perpendicular sum is $OD = 114.0175$ amp. The order in which the components are added manifestly does not affect the final result, and it is a matter of insignificance whether the various frequencies coacting are "harmonic," i. e., are integral multiples of a fundamental, or not, so long as they are different.

"The complete solution of an alternating-current line with complex harmonic potentials and currents thus requires an independent solution of potential and current for each single frequency in turn, as though the others were non-existent, and then the r.m.s. value at any point on the line is the perpendicular sum of the separate frequency values. The powers and energies of the different frequencies are independent of each other, and the total transmitted energy is the sum of the energies transmitted at the separate component frequencies."

BIBLIOGRAPHY

In order to give due prominence to some of the valuable contributions on the subject of performance of electrical circuits and as an acknowledgment to their authors of the assistance received from a study of them, the following publications are suggested as representing a very helpful and valuable addition to the library of the transmission engineer. They are given in the approximate order of their publication:—

Calculation of the High Tension Line and Output and Regulation in Long Distance Lines by Percy H. Thomas. (Published in *A. I. of E. E. Trans.* Vol. XXVIII, Part, 1, 1909). The former paper introduces a so-called "wave formula" for determining the performance of long lines having considerable capacity which embodies the use of algebra only. The second paper suggests the use of split conductors in order to adjust the ratio of the capacity and inductance of the line so that the leading and lagging components more nearly neutralize each other.

Formulae, Constants and Hyperbolic Constants by W. E. Miller. (Published in *G. E. Review*, supplement dated May 1910). This is a treatise upon the subject wherein hyperbolic functions of complex angles are tabulated for sinh and cosh ($x + jy$) up to $x = 1, y = 1$ in steps of 0.02.

Transmission Line Formulas by H. B. Dwight. (Published by John Wiley & Sons, Inc.). This book introduces what are known as "Dwight's 'K' formulas," which permit the solution of transmission problems without the use of mathematics higher than arithmetic. It also contains working formulas based upon convergent series and the solution of many problems both by the K formulas and by convergent series.

Tables of Complex Hyperbolic and Circular Functions by Dr. A. E. Kennelly. (Published by the Harvard University Press). This book gives functions of complex angles for polar values up to 3.0 by steps of 0.1 and for angles from 45° to 90° by steps of one degree; also functions in terms of rectangular coordinates $x + jy$ to $x = 10$ by steps of 0.05 and of y virtually to infinity by steps of 0.05.

Chart Atlas of Complex Hyperbolic and Circular Functions by Dr. A. E. Kennelly. (Published by Harvard University Press in large charts, 48 by 48 cm.) Presenting curves for all the tables published in above referred to "Tables of Complex Hyperbolic and Circular Functions" for rapid graphical interpolation.

Constant Voltage Transmission by H. B. Dwight. (Published by John Wiley & Son, Inc.). Embraces a very complete study of the use of over-excited synchronous motors for controlling the voltage of transmission.

The Application of Hyperbolic Functions to Electrical Engineering Problems by Dr. A. E. Kennelly. (Published by the McGraw-Hill Book Company). Every student should have a copy of this book because of its simplicity and completeness in explaining the application of hyperbolic functions to transmission circuit problems. It also contains a very complete bibliography of publications upon this general subject.

Artificial Electric Lines by Dr. A. E. Kennelly. (Published by McGraw-Hill Book Co.). This is a valuable treatise in which the subject is treated in accordance with the hyperbolic theory.

Electrical Phenomena in Parallel Conductors by Dr. F. E. Pernot. (Published by John Wiley & Son, Inc.). Being a very recent treatise, this book contains much practical and many readily understandable explanations for both the beginner and those further advanced in the study of this subject. It contains a six-place table of logarithms of real hyperbolic functions for values of x from 0.000 to 2.000 for intervals of 0.001 in the argument. This is the most complete table of real hyperbolic functions which the author has seen.

TABLE P—SUBDIVISIONS OF A DEGREE

SECONDS TO DEGREES		MINUTES TO DEGREES		DEGREES TO MINUTES AND SECONDS					
// =	o	/ =	o	o =	/	//	o =	/	//
01	0.0003	01	0.0167	0.001	00	03.6	0.006	00	21.6
02	0.0006	02	0.0333	0.002	00	07.2	0.007	00	25.2
03	0.0008	03	0.0500	0.003	00	10.8	0.008	00	28.8
04	0.0011	04	0.0667	0.004	00	14.4	0.009	00	32.4
05	0.0014	05	0.0833	0.005	00	18.0	0.010	00	36.0
06	0.0017	06	0.1000						
07	0.0019	07	0.1167						
08	0.0022	08	0.1333						
09	0.0025	09	0.1500						
10	0.0028	10	0.1667						
11	0.0031	11	0.1833	0.01	00	36	0.51	30	36
12	0.0033	12	0.2000	0.02	01	12	0.52	31	12
13	0.0036	13	0.2167	0.03	01	48	0.53	31	48
14	0.0039	14	0.2333	0.04	02	24	0.54	32	24
15	0.0042	15	0.2500	0.05	03	00	0.55	33	00
16	0.0044	16	0.2667	0.06	03	36	0.56	33	36
17	0.0047	17	0.2833	0.07	04	12	0.57	34	12
18	0.0050	18	0.3000	0.08	04	48	0.58	34	48
19	0.0053	19	0.3167	0.09	05	24	0.59	35	24
20	0.0055	20	0.3333	0.10	06	00	0.60	36	00
21	0.0058	21	0.3500	0.11	06	36	0.61	36	36
22	0.0061	22	0.3667	0.12	07	12	0.62	37	12
23	0.0064	23	0.3833	0.13	07	48	0.63	37	48
24	0.0067	24	0.4000	0.14	08	24	0.64	38	24
25	0.0069	25	0.4167	0.15	09	00	0.65	39	00
26	0.0072	26	0.4333	0.16	09	36	0.66	39	36
27	0.0075	27	0.4500	0.17	10	12	0.67	40	12
28	0.0078	28	0.4667	0.18	10	48	0.68	40	48
29	0.0081	29	0.4833	0.19	11	24	0.69	41	24
30	0.0083	30	0.5000	0.20	12	00	0.70	42	00
31	0.0086	31	0.5167	0.21	12	36	0.71	42	36
32	0.0089	32	0.5333	0.22	13	12	0.72	43	12
33	0.0092	33	0.5500	0.23	13	48	0.73	43	48
34	0.0094	34	0.5667	0.24	14	24	0.74	44	24
35	0.0097	35	0.5833	0.25	15	00	0.75	45	00
36	0.0100	36	0.6000	0.26	15	36	0.76	45	36
37	0.0103	37	0.6167	0.27	16	12	0.77	46	12
38	0.0106	38	0.6333	0.28	16	48	0.78	46	48
39	0.0108	39	0.6500	0.29	17	24	0.79	47	24
40	0.0111	40	0.6667	0.30	18	00	0.80	48	00
41	0.0114	41	0.6833	0.31	18	36	0.81	48	36
42	0.0117	42	0.7000	0.32	19	12	0.82	49	12
43	0.0119	43	0.7167	0.33	19	48	0.83	49	48
44	0.0122	44	0.7333	0.34	20	24	0.84	50	24
45	0.0125	45	0.7500	0.35	21	00	0.85	51	00
46	0.0128	46	0.7667	0.36	21	36	0.86	51	36
47	0.0130	47	0.7833	0.37	22	12	0.87	52	12
48	0.0133	48	0.8000	0.38	22	48	0.88	52	48
49	0.0136	49	0.8167	0.39	23	24	0.89	53	24
50	0.0139	50	0.8333	0.40	24	00	0.90	54	00
51	0.0141	51	0.8500	0.41	24	36	0.91	54	36
52	0.0144	52	0.8667	0.42	25	12	0.92	55	12
53	0.0147	53	0.8833	0.43	25	48	0.93	55	48
54	0.0150	54	0.9000	0.44	26	24	0.94	56	24
55	0.0153	55	0.9167	0.45	27	00	0.95	57	00
56	0.0156	56	0.9333	0.46	27	36	0.96	57	36
57	0.0159	57	0.9500	0.47	28	12	0.97	58	12
58	0.0162	58	0.9667	0.48	28	48	0.98	58	48
59	0.0164	59	0.9833	0.49	29	24	0.99	59	24
60	0.0167	60	1.0000	0.50	30	00	1.00	60	00

EXAMPLES

$0^{\circ}.41 = 0^{\circ}.24'36''$ $0^{\circ}.41'00'' = 0^{\circ}.6833$
 $0^{\circ}.005 = 0^{\circ}.00'18''$ $0^{\circ}.00'48'' = 0^{\circ}.0128$

CHAPTER XII

COMPARISON OF VARIOUS METHODS

The "localized capacitance" or "localized admittance" methods are discussed below for the two following reasons. A discussion of them is of academic interest and a tabulation of the magnitude of the errors in the results as obtained by these approximate methods when applied to circuits of different lengths and frequencies should be helpful. These methods may be carried out either graphically or mathematically, but since they are only approximate the simpler graphical solution should suffice. Their principle virtue is the fact that they simplify the determination of performance, but this is obtained at the expense of accuracy. The more accurate of these methods is somewhat tedious to carry out. The graphical solution previously described in connection with the Wilkinson charts will be generally more accurate and shorter than these localized capacitance methods.

THE LOCALIZED CAPACITANCE methods are:—the single end condenser method; the middle condenser or *T* method; the split condenser or nominal π method and Dr. Steinmetz three condenser method. These four lumped capacitance methods assume the total capacitance of the circuit as being divided up and "lumped" in the form of condensers shunted across the circuit at one or more points.

methods, usually an approximation to the true value may be obtained.

The middle condenser or *T* method assumes that the total capacitance may be shunted across the circuit at the middle point. On this assumption the total charging current will flow over one half the length of the circuit. This method is therefore more nearly accurate than the single-condenser method.

The split condenser or π method assumes one half

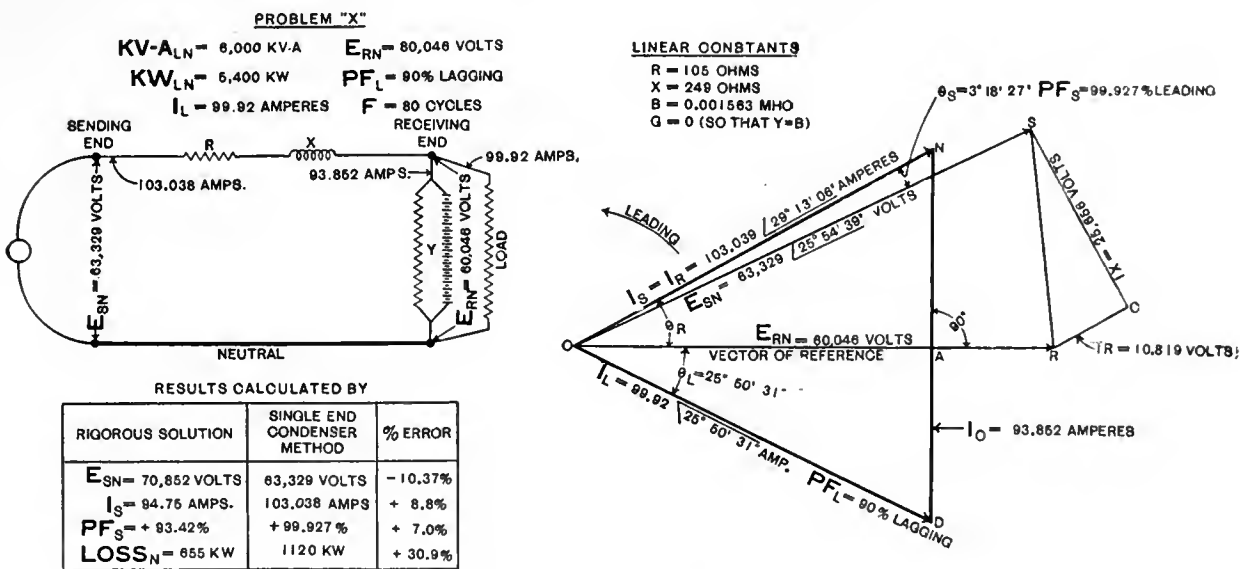


FIG. 54—SINGLE END CONDENSER METHOD
Problem X.

The single condenser method assumes the total capacitance as being lumped or shunted across the circuit at the receiving-end. On this assumption the total charging current for the circuit would flow over the entire circuit. Actually the charging current is distributed along the circuit so that the entire charging current does not flow over the entire circuit. Obviously the assumption of the total capacitance being lumped at the receiving-end will therefore give over compensation for the effect of the charging current upon the voltage regulation of the circuit. This method of solution yields a voltage too low at the sending end by nearly the same amount that the straight impedance method gives it too high. By averaging the values, as

the capacitance being shunted across the circuit at each end. In this case one-half of the charging current flows over the entire circuit. This assumed distribution of the charging current also more nearly represents the actual distribution than the single-condenser method.

Dr. Steinmetz has proposed a method assuming three condensers shunted across the circuit. One in the middle, of two-thirds, and one at each end, each of one sixth the total capacitance of the circuit. This method is equivalent to assuming that the electrical quantities are distributed along the circuit in a way representing an arc of a parabola. This method assumes one-sixth the charging current flowing over one half the entire circuit and five sixth the charging current flowing over the other half of the circuit. This method gives quite

accurate results unless the circuit is very long and the frequency high.

Figs. 54-57 show leaky condensers placed at different points of the circuits, that is they indicate that there is a leak G , as well as a susceptance B . For simplicity pure condensers have been assumed in the accompanying calculations; that is we have assumed $G=0$. This is the usual assumption in such cases, for the reason that G is usually very small, and localized capacitance methods are approximations at best. In the equivalent π solution previously given, we have indicated the treatment when the condensers have a leak. In such case, however, the equivalent π method produces exact results, and the nature of such solution may demand a condenser having a material leak.

AUXILIARY CONSTANTS

Mr. T. A. Wilkinson and Dr. Kennelly have worked out the algebraic expressions for the auxiliary

receiving-end. In such case the entire charging current would flow over the total length of the circuit.

Solution by Impedance Method—The diagrams of connections and corresponding graphical vector solution for problem X by the single-end condenser method is indicated by Fig. 54. The current DN consumed by the condenser (zero leakage assumed) leads the receiving-end voltage OR by 90 degrees and is,—

$$I_c = 0.001563 \times 60.046 = 93.852 \text{ amperes.}$$

The load current of 99.92 amperes, lagging $25^\circ 50' 30''$ (90% power-factor) has a component OA of $99.92 \times 0.90 = 89.928$ amperes in phase with the receiving-end voltage and a component AD of $99.92 \times 0.4359 = 43.555$ amperes in lagging quadrature with the receiving-end voltage. This lagging component is therefore in opposite direction to the charging current, the effect of which is to neutralize an equivalent amount of charging current. The remaining current AN in leading quadrature with the receiving-end voltage is

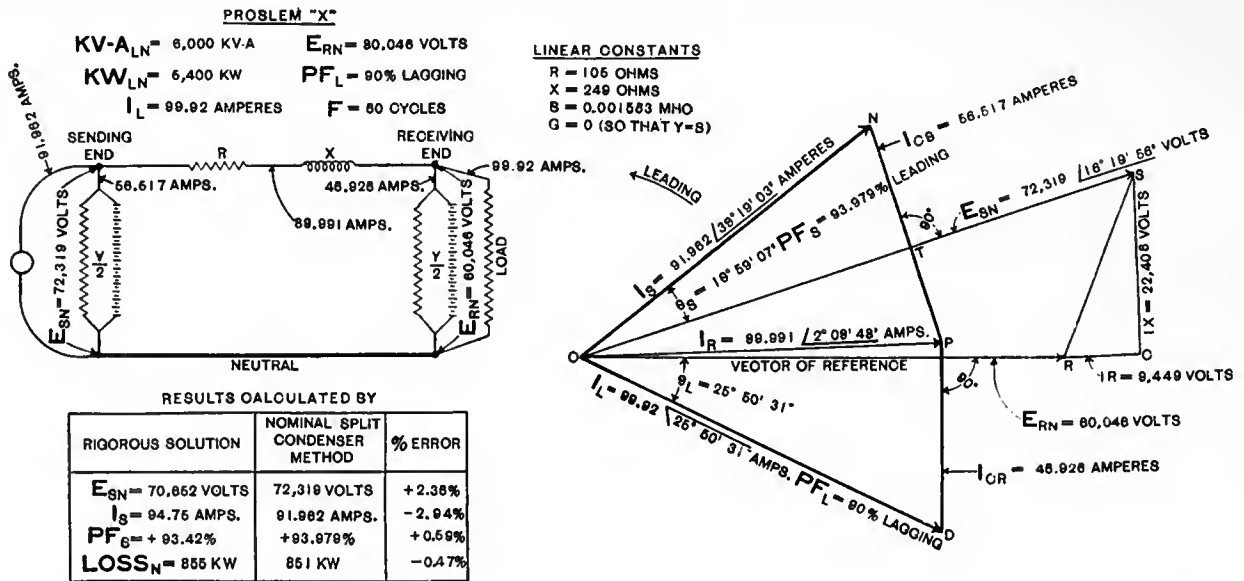


FIG. 55—NOMINAL π OR SPLIT CONDENSER METHOD Problem X.

constants corresponding to these four circuits of localized capacitance. These are given in Table Q. It may be interesting to observe to what extent each of the four localized capacitance methods takes account of the three linear line constants R , X and B . The rigorous or exact expression for the auxiliary constants is given under Table Q for comparison with the values corresponding to the localized condenser methods. The numerals under the algebraic expressions correspond to problem X; that is, to a certain 60 cycle circuit, 300 miles long. They are given to illustrate for a long circuit, the account taken of the fundamental constants for each of the five methods listed. These numerals may be compared with the rigorous or exact values as given under the rigorous expressions at the bottom of the table.

SINGLE END CONDENSER METHOD

This method assumes that the total capacitance of the circuit may be concentrated across the circuit at the

$93.852 - 43.555 = 50.297$ amperes. The current ON in the conductor is therefore:—

$$I_s = \sqrt{(89.928^2 + (50.297)^2} = 103.038 \text{ amperes.}$$

The current at the sending-end leads the voltage at the receiving-end by the angle θ_R whose tangent is,—

$$\frac{50.297}{89.928} = 29^\circ 13' 06''$$

The voltage consumed by the resistance, and the reactance of each conductor is,—

$$IR = 103.038 \times 105 = 10819 \text{ Volts (resistance drop)}$$

$$IX = 103.038 \times 249 = 25656 \text{ Volts (reactance drop)}$$

The receiving-end conditions are thus,—

$$I_R = 103.038 \text{ amperes}$$

$$\theta_R = 29^\circ 13' 06''$$

$$\cos \theta_R = 0.8772$$

$$\sin \theta_R = 0.4881$$

and from (40)

$$E_{in} = \sqrt{(60.046 \times 0.8727 + 10.819)^2 + (60.046 \times 0.4881 - 25.656)^2}$$

$$= 63.329 \sqrt{3^\circ 18' 27''} \text{ volts to vector } ON$$

$$= 63.329 \sqrt{25^\circ 54' 39''} \text{ volts to vector of reference.}$$

$$PF_s = \cos 3^\circ 18' 27'' = 99.927 \text{ percent leading.}$$

$$KV \cdot A_{in} = 103.038 \times 63.329 = 6525 \text{ kv-a.}$$

$$KW_{in} = 6525 \times 0.99927 = 6520 \text{ kw.}$$

$$Loss_s = 6520 - 5400 = 1120 \text{ kw.}$$

Solution by Complex Quantities—From Table Q the auxiliary constants corresponding to the single end condenser method are found as follows:—

$$a_1 = 1 - XB = 0.610813$$

$$a_2 = RB = 0.164115$$

$$b_1 = R = 105 \text{ ohms.}$$

$$b_2 = X = 249 \text{ ohms.}$$

$$c_1 = 0$$

$$c_2 = B = 0.001563 \text{ mho.}$$

The voltage at the sending end is determined as follows:—

$$I_L (\cos \theta_L - j \sin \theta_L) = 89.928 - j 43.555$$

$$\times (b_1 + j b_2) = 20.286 + j 17.819$$

$$+ E_{in} (a_1 + j a_2) = 36.677 + j 9.854$$

$$E_{in} = \frac{56.963 + j 27.673}{63.329 \sqrt{25^\circ 54' 39''}} \text{ volts.}$$

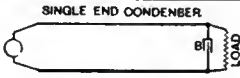
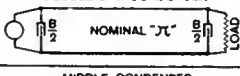
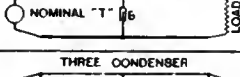
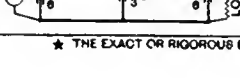
end is completely determined by the load current at the receiving-end and the vector addition thereto of the current supplied at that end to the condenser under receiving-end voltage. For determining the sending-end voltage $A'_v = I + YZ$ and $B'_v = Z$; but for determining the sending-end current $A'_I = I$ and $C'_I = Y$. If the condenser were applied symmetrically A'_v and A'_I would be identical.

SPLIT CONDENSER OR NOMINAL π SOLUTION

This method assumes that the total capacitance of the circuit may be concentrated at the two ends, one-half being placed across the circuit at either end. In this case one-half the charging current flows over the entire circuit. The total resistance and the total reactance of one conductor is placed between the two terminal condensers.

With this assumption the current consumed by the condenser across the receiving-end of the circuit is added vectorially to the load current and the power-factor of the combined currents calculated. With these new load conditions determined the conditions at the

TABLE Q—AUXILIARY CONSTANTS
CORRESPONDING TO CIRCUITS OF LOCALIZED CAPACITANCE

METHOD	a_1	a_2	b_1	b_2	c_1	c_2	EQUIVALENT CONVERGENT SERIES FORM OF EXPRESSION *
IMPEDANCE	1	0	R =105	X =+j249	0	0	$A' = 1 \quad B' = Z \quad C' = 0$
SINGLE END CONDENSER 	$1 - XB$ =0.610813	RB =+j0.164115	R =105	X =+j249	0	B =+j0.001563	$A' = 1 + YZ \quad B' = Z \quad C' = Y$
DOUBLE END CONDENSER 	$1 - \frac{XB}{2}$ =0.8064086	$\frac{RB}{2}$ =+j0.082058	R =105	X =+j249	$-\frac{B^2R}{4}$ =-0.0000641	$8 - \frac{B^2X}{4}$ =+j0.001411	$A' = (1 + \frac{YZ}{2}) \quad B' = Z \quad C' = Y(1 + \frac{YZ}{4})$
MIDDLE CONDENSER 	$1 - \frac{XB}{2}$ =0.8064086	$\frac{RB}{2}$ =+j0.082058	$R - \frac{RXB}{2}$ =84.6877	$X - \frac{B}{4}(X^2 - R^2)$ =+j229.081	0	B =+j0.001563	$A' = (1 + \frac{YZ}{2}) \quad B' = Z(1 + \frac{YZ}{4})$ $C' = Y$
THREE CONDENSER 	$1 - \frac{XB}{2} + \frac{B^2}{36}(X^2 - R^2)$ =0.806888	$\frac{RB}{2} - \frac{RXB^2}{18}$ =+j0.0786091	$R - \frac{RXB}{3}$ =91.3786	$X - \frac{B}{6}(X^2 - R^2)$ =+j236.721	$-\frac{6RB^2}{36} + \frac{RXB^3}{108}$ =-0.0000347	$B - \frac{B^2X}{36} + \frac{B^3}{216}(X^2 - R^2)$ =+j0.0014794	$A' = (1 + \frac{YZ}{2} + \frac{Y^2Z^2}{36}) \quad B' = Z(1 + \frac{YZ}{6})$ $C' = Y(1 + \frac{8YZ}{36} + \frac{Y^2Z^2}{216})$

* THE EXACT OR RIGOROUS EXPRESSIONS FOR THE AUXILIARY CONSTANTS ARE GIVEN BELOW THE NUMERICAL FIGURES CORRESPOND TO PROBLEM "X"

$$A = (1 + \frac{YZ}{2} + \frac{Y^2Z^2}{24} + \frac{Y^3Z^3}{720} + \frac{Y^4Z^4}{40320} + \dots)$$

$$= \cosh \theta = 0.61088 + j0.07883$$

$$B = Z(1 + \frac{YZ}{2} + \frac{Y^2Z^2}{120} + \frac{Y^3Z^3}{6,040} + \frac{Y^4Z^4}{382,880} + \dots)$$

$$= Z, \sinh \theta = 91.7486 + j236.666$$

$$C = Y(1 + \frac{YZ}{6} + \frac{Y^2Z^2}{120} + \frac{Y^3Z^3}{6,040} + \frac{Y^4Z^4}{382,880} + \dots)$$

$$= Y, \sinh \theta = -0.0000411 + j0.0014634$$

which checks exactly with the results as obtained previously by the impedance method.

The current at the sending end may be determined as follows:—

$$I_s (\cos \theta_s - j \sin \theta_s) = 89.928 - j 43.555$$

$$+ E_{in} (c_1 + j c_2) = 0 + j 93.852$$

$$I_s = \frac{89.928 + j 50.297}{103.038 \sqrt{29^\circ 13' 06''}} \text{ amperes.}$$

which also checks exactly with the result as previously determined by the impedance method.

It should be noted here that in determining the sending-end current, the auxiliary constant $(a + j a_2)$ did not enter into the calculation as it does in the rigorous solution; this is owing to the inherent dissymmetry of the single-end condenser. This is the only case in which the capacitance is applied dissymmetrically, consequently the current entering the line at the sending-

end are calculated by the impedance method. This is the only calculation required when employing the nominal π method for determining the sending-end voltage. The voltage at the sending-end is therefore more readily calculated by this method than by the T method which requires the calculation of the two separate halves of the circuit. If, however, the current, power-factor and kw input are required, a second calculation must be made to determine them. In such cases the current consumed by the condenser at the sending-end must be added vectorially to that of the line conductors.

Solution by Impedance Method—The diagrams of connections and corresponding graphical vector solutions for problem X by the nominal π method is indicated in Fig. 55. The charging current consumed by the condenser (zero leakage assumed) at the receiving-

end of the circuit leads the receiving-end voltage by 90 degrees and is,—

$$I_{cr} = \frac{0.001563}{2} \times 60.046 = 46.926 \text{ amperes.}$$

The current I_r in each conductor is the vector sum of the load and condenser currents and may be determined as follows:—

$$I_r = \sqrt{(99.92 \times 0.90)^2 + (I_{cr} + 99.92 \times -0.4359)^2}$$

$$= 89.991 \angle 2^\circ 08' 48'' \text{ amperes.}$$

$$PF_r = \cos 2^\circ 08' 48'' = 99.33 \text{ percent leading.}$$

The voltage consumed by the resistance, and the reactance of each conductor is,—

$$IR = 89.991 \times 105 = 9449 \text{ volts (resistance drop)}$$

$$IX = 89.991 \times 249 = 22408 \text{ volts (reactance drop)}$$

and from (40),—

$$E_{sn} = \sqrt{(60.046 \times 0.9933 + 9449)^2 + (60.046 \times 0.037458 - 22408)^2}$$

$$= 72319 \angle 16^\circ 11' 08'' \text{ volts to current vector } OP.$$

$$= 72319 \angle 18^\circ 19' 56'' \text{ volts to vector of reference } OR.$$

The charging current consumed by the condenser at the sending-end (zero leakage assumed) leads the voltage at the sending-end by 90° and is,—

$$I_{cs} = \frac{0.001563}{2} \times 72319 = 56.517 \text{ amperes.}$$

The current at the sending-end is the vector sum of the current in the conductor and the current consumed by the condenser at the sending-end. It may be calculated as follows:—

$$OT = 89.991 (\cos 16^\circ 11' 08'') = 86.424 \text{ amperes.}$$

$$TP = 89.991 (\sin 16^\circ 11' 08'') = 25.085 \text{ amperes.}$$

$$TN = 56.517 - 25.085 = 31.432 \text{ amperes.}$$

therefore,—

$$I_s = \sqrt{86.424^2 + 31.432^2}$$

$$= 91.962 \angle 19^\circ 59' 07'' \text{ amperes to vector } OS.$$

$$= 91.962 \angle 38^\circ 19' 03'' \text{ to vector of reference } OR.$$

$$PF_s = \cos 19^\circ 59' 07'' = 93.979 \text{ percent leading.}$$

$$KV \cdot A_{sn} = 91.962 \times 72.319 = 6651 \text{ kv-a.}$$

$$KW_{sn} = 6651 \times 0.93979 = 6251 \text{ kw.}$$

$$Loss_{sn} = 6251 - 5400 = 851 \text{ kw.}$$

$$Eff = \frac{5400 \times 100}{6251} = 86.37 \text{ percent.}$$

Solution by Complex Quantities—From Table Q the auxiliary constants corresponding to the nominal π method of solution are found as follows:—

$$a_1 = 1 - \frac{XB}{2} = 0.8054065.$$

$$a_2 = \frac{RB}{2} = 0.0820575.$$

$$b_1 = R = 105 \text{ ohms.}$$

$$b_2 = X = +j 249 \text{ ohms.}$$

$$c_1 = -\frac{B^2R}{4} = -0.0000641 \text{ mho.}$$

$$c_2 = B - \frac{B^2X}{4} = 0.001411 \text{ mho.}$$

The voltage at the sending-end is determined as follows:—

$$I_L (\cos \theta_L - j \sin \theta_L) = 89.928 - j43.555.$$

$$\times (b_1 + jb_2) = 20286 + j17819 \text{ volts.}$$

$$+ E_{rs} (a_1 + ja_2) = 48361 + j4927 \text{ volts.}$$

$$E_{sn} = 68647 + j22746.$$

$$= 72319 \angle 18^\circ 19' 56'' \text{ volts.}$$

The current at the sending-end may be determined as follows:—

$$I_L (\cos \theta_L - j \sin \theta_L) = 89.928 - j43.555.$$

$$\times (a_1 + ja_2) = +76.003 - j27.700 \text{ amperes.}$$

$$+ E_{rs} (C_1 + jC_2) = -3.849 + j84.718 \text{ amperes.}$$

$$I_s = 72.154 + j57.018.$$

$$= 91.962 \angle 38^\circ 19' 03'' \text{ amperes.}$$

The above results check exactly with those previously obtained by impedance calculations. This agreement indicates that the nominal π solution may, if desired, be used with complex quantities, assuming values for the auxiliary constants as indicated in Table Q.

Convergent Series Expression—Table Q indicates that the nominal π solution is equivalent to using the following values for the auxiliary constants in the convergent series form of solution,—

$$A' = \left(1 + \frac{YZ}{2}\right), \quad B' = Z, \quad C' = Y \left(1 + \frac{YZ}{4}\right)$$

We will now show that the above expressions yield the same values for the auxiliary constants as given in Table Q. From chart XI the following values corresponding to problem X are taken.

$$ZY = -0.389187 + j0.164115$$

therefore,

$$A' = 1.0000000$$

$$-0.1945935 + j0.0820575$$

$$A' = 0.8054065 + j0.0820575$$

$$B' = 105 + j249$$

$$C' = 1.0000000$$

$$-0.0972967 + j0.0410287$$

$$= Y (0.9027033 + j0.0410287)$$

$$C' = -0.0000641 + j0.001411$$

Thus the values for the auxiliary constants as determined by the above incomplete convergent series expression check with those as determined above from the equations in Table Q.

MIDDLE CONDENSER OR NOMINAL T METHOD

THIS METHOD assumes that the total capacitance of the circuit may be concentrated at its middle point. In such a case the entire charging current would flow over half of the circuit. The resistance and the reactance on each side of the capacitance or condenser is equal respectively to half the total conductor resistance and conductor reactance.

From an inspection of the diagram of such a circuit, Fig. 56, it is evident that two calculations will be required. Starting with the known receiving-end conditions, the conditions at the middle of the circuit are first calculated by the simple impedance method. To these calculated results the current consumed by the condenser shunted across the middle of the circuit must be vectorially added. This will give the load condition at the middle of the circuit from which the sending-end conditions may be calculated.

Solution by Impedance Method—The diagram of connections and the corresponding graphical vector solution for problem X by the nominal T method is indicated by Fig. 56. The electrical conditions at the middle of the circuit may be determined as follows:—

$$I_R \frac{R}{2} = 99.92 \times 52.5 = 5246 \text{ volts (resistance drop)}$$

$$I_R \frac{X}{2} = 99.92 \times 124.5 = 12440 \text{ volts (reactance drop)}$$

$$E_{mn} = \sqrt{(60.046 \times 0.9 + 5246)^2 + (60.046 \times 0.4359 + 12440)^2}$$

$$= 70753 \text{ } \angle 33^\circ 04' 36'' \text{ to current vector } OD$$

$$= 70753 \text{ } \angle 7^\circ 14' 05'' \text{ to vector of reference } OR$$

The current consumed by the condenser (zero leakage assumed) leads the voltage *OM* at the middle of the circuit by 90 degrees and is:—

$$I_c = 0.001563 \times 70753 = 110.587 \text{ amperes}$$

The voltage consumed by the condenser current flowing back to the sending-end is:—

$$I_c \frac{R}{2} = 110.587 \times 52.5 = 5806 \text{ volts (resistance drop)}$$

$$= FC$$

$$I_c \frac{X}{2} = 110.587 \times 124.5 = 13768 \text{ volts (reactance drop)}$$

$$= FM$$

The voltage vector *OC* upon which the impedance triangle corresponding to the receiving-end load current $I_R = I_L$ flowing over the sending-end half of the circuit is constructed, may be found as follows:—

$$OC = \sqrt{(70753 - 13768)^2 + 5806^2}$$

$$= 57280 \text{ } \angle 5^\circ 49' 03'' \text{ volts to vector } OM$$

$$= 57280 \text{ } \angle 13^\circ 03' 08'' \text{ volts to vector of reference } OR$$

The voltage *OC* leads the receiving-end current *OD* by the angle $33^\circ 04' 36'' + 5^\circ 49' 03'' = 38^\circ 53' 39''$ which angle corresponds to a power-factor of 77.831

percent. The voltage at the sending-end will therefore be:—

$$E_{sn} = \sqrt{(57280 \times 0.77831 + 5246)^2 + (57280 \times 0.62788 + 12440)^2}$$

$$= 69467 \text{ } \angle 44^\circ 10' 14'' \text{ volts to vector } OD$$

$$= 69467 \text{ } \angle 18^\circ 19' 43'' \text{ volts to vector of reference } OR$$

If desired, the receiving-end current and the condenser current may be combined and the corresponding impedance triangle for the sending-end half of the circuit constructed on the end of vector *OM* as indicated by the dotted lines.

The current at the sending-end may be determined as follows:—

$$OB = 99.92 \cos 33^\circ 04' 36'' = 83.727 \text{ amperes.}$$

$$BD = 99.92 \sin 33^\circ 04' 36'' = 54.532 \text{ amperes.}$$

$$BN = 110.587 - 54.532 = 56.055 \text{ amperes.}$$

$$I_s = ON = \sqrt{(83.727)^2 + (56.055)^2}$$

$$= 100.76 \text{ } \angle 33^\circ 48' 06'' \text{ amperes to vector } OB.$$

$$= 100.76 \text{ } \angle 41^\circ 02' 11'' \text{ amperes to vector of reference } OR.$$

The current at the sending-end leads the voltage at the sending-end by the angle $41^\circ 02' 11'' - 18^\circ 19' 43'' = 22^\circ 42' 28''$, which corresponds to a power-factor at the sending-end of 92.25 percent leading.

The power at the sending-end is:—

$$Kv\text{-}a_{sn} = 100.76 \times 69467 = 7000 \text{ kv}\text{-}a.$$

$$Kw_{sn} = 7000 \times 0.9225 = 6457 \text{ kw.}$$

$$Loss_{sn} = 6457 - 5400 = 1057 \text{ kw.}$$

Solution by Complex Quantities—From table Q the auxiliary constants corresponding to the nominal T method of solution are found as follows:

$$a_1 = 1 - \frac{XB}{2} = 0.8054065$$

$$a_2 = \frac{RB}{2} = 0.0820575$$

$$b_1 = R - \frac{RXB}{2} = 84.5677$$

$$b_2 = X - \frac{B}{4}(X^2 - R^2) = 229.081$$

$$c_1 = 0$$

$$c_2 = B = 0.001563$$

The voltage at the sending-end is obtained as follows:—

$$I_R (\cos \theta_R - j \sin \theta_R) = 89.928 - j 43.554$$

$$\times (b_1 + j b_2) = 17582 + j 16918$$

$$+ E_{mn} (a_1 + j a_2) = 48361 + j 4927$$

$$E_{sn} = 65943 + j 21845$$

$$= 69467 \text{ } \angle 18^\circ 19' 43''$$

The current at the sending-end may be calculated as follows:—

$$I_R (\cos \theta_R - j \sin \theta_R) = 89.928 - j 43.554$$

$$\times (a_1 + j a_2) = 76.0026 - j 27.6994$$

$$+ E_{mn} (c_1 + j c_2) = 0 + j 93.8519$$

$$I_s = 76.0026 + j 66.1525$$

$$= 100.76 \text{ } \angle 41^\circ 02' 11'' \text{ amperes}$$

The above results check with those previously obtained by impedance calculations. This agreement indicates that the nominal T solution may, if desired, be made by complex quantities, assuming values for the auxiliary constants as indicated in Table Q.

Convergent Series Expression—Table Q indicates that the nominal T solution is equivalent to using the following values for the auxiliary constants in the convergent series form of solution:—

$$A' = \left(1 + \frac{ZY}{2}\right)$$

$$B' = Z \left(1 + \frac{ZY}{4}\right)$$

$$C' = Y$$

Comparing the above expressions for the auxiliary constants with the complete expression yielding rigorous values the following difference may be noted.

For auxiliary constant A' the first two terms in the complete series for the hyperbolic cosine are used and

expressions check exactly with those as determined above from the equations in Table Q.

THREE CONDENSER METHOD

This method (proposed by Dr. Chas. P. Steinmetz) assumes that the admittance of the circuit may be lumped or concentrated across the circuit at three points, one-sixth being localized at each end and two-thirds at the middle of the circuit. This is equivalent to assuming that the electrical quantities are distributed along the circuit in a manner represented by the arc of a parabola. It is evident that this method more nearly approaches the actual distribution of the impedance and the admittance of the circuit than any of the three previously described localized admittance methods, and therefore yields more accurate results.

From an inspection of the diagram of such a circuit, Fig. 57, it will be evident that it is necessary to calculate the performance of the two halves of the cir-

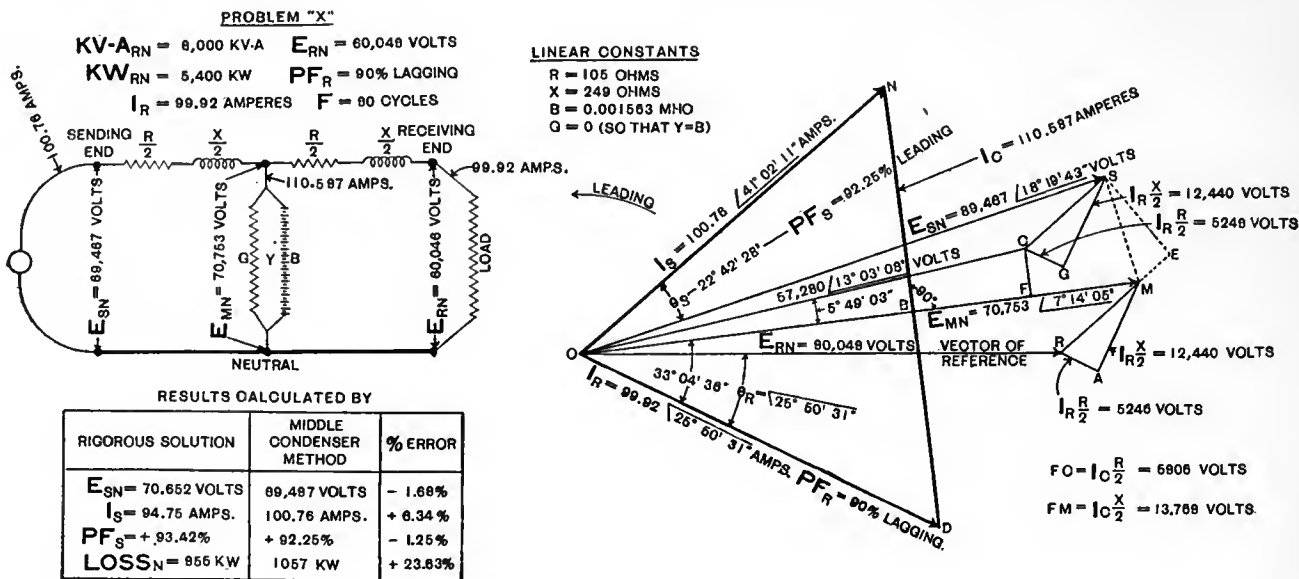


FIG. 56—NOMINAL T OR MIDDLE CONDENSER METHOD

all terms beyond omitted. For auxiliary constant B' the first two terms of the complete series are also used except that the coefficient of the second term is given as $1/4$, whereas in the complete series it is $1/6$. Auxiliary constant C' is equivalent to the first term only of the complete expression.

We will now show that the above expressions yield the same values for the auxiliary constants as given in Table Q. From Chart XI the following values corresponding to problem X are taken:—

$$Z = 105 + j249$$

$$ZY = -0.389187 + j0.164115$$

Therefore $A' = \frac{1.000000}{-0.1945935 + j0.0820575}$

$$A' = +0.8054065 + j0.0820575$$

$$B' = \frac{1.000000}{-0.09729675 + j0.04102875}$$

$$= Z (0.90270325 + j0.04102875)$$

$$B' = 84.5677 + j229.081$$

$$C' = 0 + j0.001563$$

Thus the values for the auxiliary constants as determined by the above incomplete convergent series

circuit in order to arrive at the sending-end voltage and an additional calculation will be required to determine the sending-end current, power and power-factor.

Solution by Impedance Method—The diagram of connections and corresponding graphical vector solution for problem X by the three condenser method is indicated by Fig. 57. The charging current consumed by the condenser (zero leakage assumed) at the receiving-end leads the receiving-end voltage by 90 degrees and is:—

$$I_{cr} = \frac{0.001563}{6} \times 60046 = 15.642 \text{ amperes.}$$

The current per conductor for the receiving-end half of the circuit is:—

$$I_r = \sqrt{(99.92 \times 0.9)^2 + (99.92 \times 0.4359 - 15.642)^2}$$

$$= 94.16 \angle 17^\circ 14' 38'' \text{ amperes}$$

$$PF_r = \cos 17^\circ 14' 38'' = 95.505 \text{ lagging}$$

The voltage consumed by the resistance and the reactance per conductor between the receiving-end and the middle of the circuit is:—

by the impedance method of procedure, it will be seen that they are in exact agreement.

Convergent Series Expression—Dr. F. E. Pernot in "Electrical Phenomena in Parallel Conductors," Vol. I, shows that the above described three condenser solution is equivalent to using the following values for the auxiliary constants in the convergent series form of solution:—

$$A' = \left(1 + \frac{ZY}{2} + \frac{Z^2 Y^2}{36} \right)$$

$$B' = Z \left(1 + \frac{ZY}{6} \right)$$

$$C' = Y \left(1 + \frac{5ZY}{36} + \frac{Z^2 Y^2}{216} \right)$$

Comparing the above expressions for the auxiliary constants with the complete expressions yielding rigorous values, the following differences may be noted. For constant A' the first two terms are the same as in the complete series, but the third term is less than in the complete series, and all terms beyond the third are omitted. For constant B' the first two terms are the same as in the complete series, but all terms beyond the second are omitted. For constant C' both the ZY and the $Z^2 Y^2$ terms are smaller than in the complete series and all terms beyond the third are omitted.

The above expressions yield the same values for the auxiliary constants as given in Table Q. Thus from chart XI, the following values corresponding to problem X are taken:—

$$ZY = -0.389187 + j 0.164115$$

$$Z^2 Y^2 = +0.124532 - j 0.127742$$

Therefore

$$A' = \begin{array}{l} 1.000000 \\ -0.194593 + j 0.0820575 \\ \quad \quad \quad 0.003459 - j 0.0035484 \end{array}$$

$$A' = 0.808866 + j 0.0785091$$

$$B' = \begin{array}{l} 1.000000 \\ -0.0648645 + j 0.0273525 \end{array}$$

$$B' = \begin{array}{l} Z (0.9351355 + j 0.0273525) \\ 91.3785 + j 235.7208 \end{array}$$

$$C' = \begin{array}{l} 1.000000 \\ -0.0540538 + j 0.0227938 \\ + 0.0005765 - j 0.0005914 \end{array}$$

$$C' = \begin{array}{l} Y (0.9465227 + j 0.0222024) \\ -0.0000347 + j 0.0014794 \end{array}$$

It will be seen that the above convergent series expression for the auxiliary constants check exactly with those as determined by the equations in Table Q.

COMPARATIVE ACCURACY OF VARIOUS METHODS

In order to determine the inherent error in various methods of solution, when applied to circuits of increasing length; also for frequencies of both 25 and 60 cycles, 64 problems were solved. These problems embrace thirty-two 25 cycle circuits, varying in length from 20 to 500 miles and in voltage from 10 000 to 200 000 volts. Fixed receiving-end load conditions were assumed for unity, and also for 80 percent power-factor lagging. These same problems were also solved for a frequency of 60 cycles.

These 64 problems with corresponding linear constants and assumed load conditions are stated on Chart

XXII. This is followed by columns in which have been tabulated the error in voltage at the sending-end of these circuits as determined by nine different methods. The errors are expressed in percent of receiving-end voltage. Obviously the inherent error corresponding to various methods will vary widely for conductors of various resistances and to some extent for different receiving-end loads. The tabulated values should therefore be looked upon as comparative rather than absolute for all conditions.

Rigorous Solution—The column headed "Rigorous Solution" contains values for the sending-end voltage which are believed to be exact. These values were obtained by calculating values for the auxiliary constants by means of convergent series and then calculating the performance mathematically. The calculations were carried out to include the sixth place and terms in convergent series were used out to the point where they did not influence the results.

The first values calculated were checked by a second set of values calculated independently at another time and where differences were found the correct values were determined and substituted. This corrected list of values was again checked by a third independent calculation. It is therefore believed that the values contained in this column are exact, representing 100 percent.

Semi-Graphical Solution—The next column contains the error in the results as derived by the combination of an exact mathematical solution for the auxiliary constants and a graphical solution from there on. This combination gave results in which the maximum error does not exceed eight one hundredths of one percent of receiving-end voltage for either frequency. In other words, since the values for the auxiliary constants used in this method were exact, the maximum error of eight one hundredths of one percent occurs in the construction and reading of the graphical constructions.

Complete Graphical Solution—This solution employs Wilkinson's charts for obtaining graphically the auxiliary constants, the remainder of the solution being also made graphically as previously described. It will be seen that the maximum error as obtained by this complete graphical solution is seven hundredths of one percent for the 25 cycle and twenty-five hundredths of one percent for the 60 cycle circuits. These errors represent the combined result of various errors. First there is a slight fundamental error in the basis upon which the Wilkinson Charts are constructed when used for circuits employing conductors of various sizes and spacings, the introduction of this error making possible the simplification attained. Then there is the inherent limitation of precision obtainable in the construction and reading of the charts and vector diagrams.

These results show that the inherent accuracy of this simplified, all graphical solution is sufficiently accurate for all practical power circuits up to 300 miles long.

Dwight's "K" Formulas—The high degree of accuracy resulting by the use of H. B. Dwight's "K" formulas should be noted. This error is a maximum of eleven hundredths of one percent for these 32 twenty-five cycle problems. The statement is therefore justified that these "K" formulas are sufficiently accurate for all 25 cycle power circuits.

For the 60 cycle problems the maximum error by the "K" formulas for problems up to and including 200 miles is one-fourth of one percent of receiving-end voltage. For 300 mile circuits this error is one-half of one percent and increases rapidly as the circuit exceeds 300 miles in length. The accuracy of the "K" formulas for 60 cycle circuits is therefore well within that of the assumed values of the linear constants for circuits up to approximately 300 miles in length.

The "K" formulas are based upon the hyperbolic formula expressed in the form of convergent series. In the development of these formulas, use was made of the fact that the capacitance multiplied by the reactance of non-magnetic transmission conductors is a constant quantity to a fairly close approximation. This assumption has enabled the "K" formulas to be expressed in comparatively simple algebraic form without the use of complex numbers. To those not familiar or not in position to make themselves familiar with the operation of complex numbers, such as is used in the convergent

series or hyperbolic treatments, the availability of the Dwight "K" formulas will be apparent.*

Localized Capacitance Methods—The next four columns contain values indicating the error in results as determined by the four different localized capacitance methods previously described in detail. It is interesting to note the high degree of accuracy inherent in Dr. Steinmetz's three condenser method. It is also interesting to note that three of these methods over compensate (that is, give receiving-end voltages too low) and one (the split condenser method) gives under compensation.

Impedance Method—The values of the sending-end voltage as obtained by the impedance method (which takes no account of capacitance) are always too high when applied to circuits containing capacitance. The results by this method are included here simply to serve as an indication of how great is the error for this method when applied to circuits of various lengths and frequencies of 25 and 60 cycles. Some engineers prefer to use this method for circuits of fair length and allow for the error. These tabulations will give an approximation of the necessary allowance to be made.

*These have been included with much other valuable material in "Transmission Line Formulas" by H. B. Dwight, published by D. Van Nostrand Co. of New York City.

CHAPTER XIII

CABLE CHARACTERISTICS

Heating Limits for Cables

THE MAXIMUM safe-limiting temperatures in degrees C at the surface of conductors in cables is given in the Standardization Rules of the A. I. E. E. (1918) as follows:—

- For impregnated paper insulation (85—E)
- For varnished cambric (75—E)
- For rubber insulation (60—0.25 E)

Where *E* represents the effective operating e.m.f. in kilovolts between conductors and the numerals represent temperature in degrees C. Thus, at a working pressure of 5 kv, the maximum safe limiting temperature at the surface of the conductors in a cable would be:—

- For impregnated paper insulation (80 degrees C)
- For varnished cambric insulation (70 degrees C)
- For rubber compound insulation (58.75 degrees C)

The actual maximum safe continuous current load for any given cable is determined primarily by the temperature of the surrounding medium and the rate of radiation. This current value is greater with direct than with alternating current and decreases with increasing frequency, being less for a 60 cycles than for 25 cycles. The carrying capacity of cables will therefore be less in hot climates than in cooler climates and will be considerably increased during the winter.

Cables immersed in water, carry at least 50 per cent more than when installed in a four-duct line, and when buried in the earth 15 to 30 per cent more than in a duct line, depending upon the character of soil moisture, etc. Circulating air or water through conduits containing lead covered cables will increase their capacity. From the above it is evident that no general rule relative to carrying capacity can be formulated to apply in all cases, and it is necessary, therefore, to consider carefully the surroundings when determining the size of cables to be used.

The practicability of tables which specify carrying capacity for cables installed in ducts will generally be questioned, for the reason that operating conditions are frequently more severe than those upon which table values are based. A duct line may operate at a safe temperature throughout its entire length, except at one isolated point adjacent to a steam pipe or excessive local temperatures due to some other cause. If larger cables are not employed at this point, burnouts may occur here when the remainder of the cable line is operating well within the limits of safe operating temperature. The danger in using table values for carrying capacity without carefully considering the condition of earth temperatures throughout the entire duct length is thus evident.

HEATING OF CABLES—TABLE XXIV

The basis upon which the data in Table XXIV has been calculated is covered by foot notes below the table. The kv-a values are determined from the current in amperes and are based upon 30 degree C rise and a maximum of 3000 volts.* Expressing the carrying capacity of cables in terms of kv-a (corrected for the varying thickness of insulation required for various voltages) may be found more convenient than the usual manner of expressing it in amperes. It will be noted that the kv-a values of the table are on the basis of a four-duct line and that for more than four ducts in the line the table kv-a values will be reduced to the following:—

- For a 4 duct line—100 percent.
- For a 6 duct line—88 percent.
- For an 8 duct line—79 percent.
- For a 10 duct line—71 percent.
- For a 12 duct line—63 percent.
- For a 16 duct line—60 percent.

When applied to all sizes of cables, the above values are only approximate. The reduction of carrying capacity caused by the presence of many cables is more for large cables than for small ones. Also, where load factors are small, the reduction due to the presence of many cables is less than the value assigned, although the carrying capacity of a small number of cables is only slightly affected.

REACTANCE OF THREE-CONDUCTOR CABLES

Tables XXV and XXVI contain values for the inductance, reactance and impedance of round three-conductor cables of various sizes and for the thicknesses of insulation indicated. All values in the tables are on the basis of one conductor of the cable one mile long.

The table values were calculated from the fundamental equation (4),

$$L = 0.08047 + 0.741 \log_{10} \frac{D}{R}$$

where *L* = the inductance in millihenries per mile of each conductor, *R* the actual radius of the conductor and *D* the distance between conductor centers expressed in the same units as *R*. As indicated in Section I, under Inductance,** this formula has been derived on the basis of solid conductors. In the case of cables, the effective radius is actually slightly less than that of the stranded conductor. The values for

*These current values are taken from General Electric Bulletin No. 49302 dated March 1917. They are in general slightly higher than those published by the Standard Underground Cable Company in their Hand Book dated 1906.

**Chapter I.

TABLE XXIV—CARRYING CAPACITY OF INSULATED COPPER CONDUCTORS

The following values for carrying capacity must not be assumed unless it is positively known that the conditions upon which they are based will not be exceeded in service.

THREE CONDUCTOR CABLES

B & S NO.	XX CARRYING CAPACITY IN AMPERES DIRECT CURRENT BASED UPON 30° C RISE AND A MAXIMUM OF 3000 VOLTS. PAPER INSULATION	K.V.A. WHICH MAY BE TRANSMITTED AT THREE PHASE AND THE FOLLOWING VOLTAGES OVER PAPER INSULATED LEAD COVERED CABLES INSTALLED IN A FOUR DUCT LINE WITH 30° C RISE IN TEMPERATURE BASED UPON THE ASSUMPTION THAT ALL DUCTS CARRY LOADED CABLES AND UPON A NORMAL EARTH TEMPERATURE OF 20° C FOR A 8 DUCT LINE THESE K.V.A. VALUES WOULD BE REDUCED TO APPROXIMATELY 88 PER CENT FOR AN 8 DUCT LINE TO 79 PER CENT FOR A 10 DUCT LINE TO 71 PER CENT FOR A 12 DUCT LINE TO 63 PER CENT AND FOR A 16 DUCT LINE (4 WIDE AND 4 HIGH) TO 60 PER CENT OF THE TABLE VALUES X X X X.																
		220 VOLTS	440 VOLTS	650 VOLTS	1100 VOLTS	2200 VOLTS	3300 VOLTS	4000 VOLTS	6000 VOLTS	6600 VOLTS	10000 VOLTS	11000 VOLTS	12000 VOLTS	13200 VOLTS	15000 VOLTS	20000 VOLTS	22000 VOLTS	25000 VOLTS
1/4	18	7	14	17	34	68	103	124	181	202	300	328	356	390	438	570	620	693
1/2	22	9	17	21	42	84	125	152	225	247	367	400	435	477	536	704	757	847
3/8	27	11	21	26	51	102	151	182	270	293	440	478	513	561	629	824	881	987
1/2	30	12	23	28	57	114	171	206	307	336	500	547	585	636	710	924	985	1115
5/8	35	14	26	32	64	128	192	230	345	378	550	598	639	693	777	1016	1081	1231
3/4	40	15	28	35	70	140	210	252	375	410	600	650	693	750	837	1096	1165	1327
7/8	45	16	30	37	74	148	222	268	400	438	630	680	723	780	870	1136	1209	1381
1	50	17	32	39	78	156	234	282	420	460	660	710	753	810	900	1176	1251	1431
1 1/8	55	18	34	41	82	164	246	296	440	480	690	740	783	840	930	1216	1291	1471
1 1/4	60	19	36	43	86	172	258	312	460	500	720	770	813	870	960	1248	1323	1503
1 3/8	65	20	38	45	90	180	270	324	480	520	750	800	843	900	990	1280	1355	1535
1 1/2	70	21	40	47	94	188	282	336	500	540	780	830	873	930	1020	1312	1387	1567
1 5/8	75	22	42	49	98	196	294	348	520	560	810	860	903	960	1050	1344	1419	1599
1 3/4	80	23	44	51	102	204	306	360	540	580	840	890	933	990	1080	1376	1451	1639
1 7/8	85	24	46	53	106	212	318	372	560	600	870	920	963	1020	1110	1408	1483	1679
2	90	25	48	55	110	220	330	384	580	620	900	950	993	1050	1140	1440	1515	1719
2 1/8	95	26	50	57	114	228	342	396	600	640	930	980	1023	1080	1170	1472	1547	1759
2 1/4	100	27	52	59	118	236	354	408	620	660	960	1010	1053	1110	1200	1504	1579	1799
2 3/8	105	28	54	61	122	244	366	420	640	680	990	1040	1083	1140	1230	1536	1611	1839
2 1/2	110	29	56	63	126	252	378	432	660	700	1020	1070	1113	1170	1260	1568	1643	1881
2 5/8	115	30	58	65	130	260	390	444	680	720	1050	1100	1143	1200	1290	1600	1675	1919
2 3/4	120	31	60	67	134	268	402	456	700	740	1080	1130	1173	1230	1320	1632	1707	1959
2 7/8	125	32	62	69	138	276	414	468	720	760	1110	1160	1203	1260	1350	1664	1739	1999
3	130	33	64	71	142	284	426	480	740	780	1140	1190	1233	1290	1380	1696	1771	2039
3 1/8	135	34	66	73	146	292	438	492	760	800	1170	1220	1263	1320	1410	1728	1803	2079
3 1/4	140	35	68	75	150	300	450	504	780	820	1200	1250	1293	1350	1440	1760	1835	2119
3 3/8	145	36	70	77	154	308	462	516	800	840	1230	1280	1323	1380	1470	1792	1867	2159
3 1/2	150	37	72	79	158	316	474	528	820	860	1260	1310	1353	1410	1500	1824	1899	2199
3 5/8	155	38	74	81	162	324	486	540	840	880	1290	1340	1383	1440	1530	1856	1931	2239
3 3/4	160	39	76	83	166	332	498	552	860	900	1320	1370	1413	1470	1560	1888	1963	2279
3 7/8	165	40	78	85	170	340	510	564	880	920	1350	1400	1443	1500	1590	1920	1995	2319
4	170	41	80	87	174	348	522	576	900	940	1380	1430	1473	1530	1620	1952	2027	2359
4 1/8	175	42	82	89	178	356	534	588	920	960	1410	1460	1503	1560	1650	1984	2059	2399
4 1/4	180	43	84	91	182	364	546	600	940	980	1440	1490	1533	1590	1680	2016	2085	2439
4 3/8	185	44	86	93	186	372	558	612	960	1000	1470	1520	1563	1620	1710	2048	2119	2479
4 1/2	190	45	88	95	190	380	570	624	980	1020	1500	1550	1593	1650	1740	2080	2155	2519
4 5/8	195	46	90	97	194	388	582	636	1000	1040	1530	1580	1623	1680	1770	2112	2183	2559
4 3/4	200	47	92	99	198	396	594	648	1020	1060	1560	1610	1653	1710	1800	2144	2215	2599
4 7/8	205	48	94	101	202	404	606	660	1040	1080	1590	1640	1683	1740	1830	2176	2247	2639
5	210	49	96	103	206	412	618	672	1060	1100	1620	1670	1713	1770	1860	2208	2279	2679
5 1/8	215	50	98	105	210	420	630	684	1080	1120	1650	1700	1743	1800	1890	2240	2315	2719
5 1/4	220	51	100	107	214	428	642	696	1100	1140	1680	1730	1773	1830	1920	2272	2343	2759
5 3/8	225	52	102	109	218	436	654	708	1120	1160	1710	1760	1803	1860	1950	2304	2375	2799
5 1/2	230	53	104	111	222	444	666	720	1140	1180	1740	1790	1833	1890	1980	2336	2407	2839
5 5/8	235	54	106	113	226	452	678	732	1160	1200	1770	1820	1863	1920	2010	2368	2439	2879
5 3/4	240	55	108	115	230	460	690	744	1180	1220	1800	1850	1893	1950	2040	2400	2475	2919
5 7/8	245	56	110	117	234	468	702	756	1200	1240	1830	1880	1923	1980	2070	2432	2503	2959
6	250	57	112	119	238	476	714	768	1220	1260	1860	1910	1953	2010	2100	2464	2535	2999
6 1/8	255	58	114	121	242	484	726	780	1240	1280	1890	1940	1983	2040	2130	2496	2567	3039
6 1/4	260	59	116	123	246	492	738	792	1260	1300	1920	1970	2013	2070	2160	2528	2599	3079
6 3/8	265	60	118	125	250	500	750	804	1280	1320	1950	2000	2043	2100	2190	2560	2635	3119
6 1/2	270	61	120	127	254	508	762	816	1300	1340	1980	2030	2073	2130	2220	2592	2667	3159
6 5/8	275	62	122	129	258	516	774	828	1320	1360	2010	2060	2103	2160	2250	2624	2695	3199
6 3/4	280	63	124	131	262	524	786	840	1340	1380	2040	2090	2133	2190	2280	2656	2727	3239
6 7/8	285	64	126	133	266	532	798	852	1360	1400	2070	2120	2163	2220	2310	2688	2759	3279
7	290	65	128	135	270	540	810	864	1380	1420	2100	2150	2193	2250	2340	2720	2795	3319
7 1/8	295	66	130	137	274	548	822	876	1400	1440	2130	2180	2223	2280	2370	2752	2823	3359
7 1/4	300	67	132	139	278	556	834	888	1420	1460	2160	2210	2253	2310	2400	2784	2855	3399
7 3/8	305	68	134	141	282	564	846	900	1440	1480	2190	2240	2293	2350	2440	2816	2887	3439
7 1/2	310	69	136	143	286	572	858	912	1460	1500	2220	2270	2333	2390	2480	2848	2919	3479
7 5/8	315	70	138	145	290	580	870	924	1480	1520	2250	2300	2373	2430	2520	2880	2955	3519
7 3/4	320	71	140	147	294	588	882	936	1500	1540	2280	2330	2413	2470	2560	2912	2983	3559
7 7/8	325	72	142	149	298	596	894	948	1520	1560	2310	2360	2453	2510	2600	2944	3015	3599
8																		

TABLE XXV—INDUCTANCE, REACTANCE AND IMPEDANCE, AT 25 CYCLES, PER MILE OF SINGLE CONDUCTOR FOR THREE CONDUCTOR CABLES

AREA IN CIRCULAR MILS B & S NO.	DIAMETER IN INCHES	RESISTANCE PER MILE IN OHMS ★	INSULATION THICKNESS IN 64THS OF AN INCH ★★											
			3/64 BY 3/64			4/64 BY 4/64			5/64 BY 5/64			6/64 BY 6/64		
			IND. M.H.	REAC. OHMS	IMP. OHMS	IND. M.H.	REAC. OHMS	IMP. OHMS	IND. M.H.	REAC. OHMS	IMP. OHMS	IND. M.H.	REAC. OHMS	IMP. OHMS
500 000	.814	.116	.338	.0530	.128	.349	.0547	.129	.340	.0565	.130	.370	.0580	.130
450 000	.772	.129	.340	.0533	.140	.351	.0552	.140	.342	.0568	.141	.373	.0585	.142
400 000	.728	.145	.343	.0537	.155	.354	.0559	.155	.346	.0572	.155	.377	.0592	.157
350 000	.681	.166	.346	.0542	.175	.357	.0560	.174	.370	.0581	.176	.380	.0596	.177
300 000	.630	.194	.349	.0547	.204	.361	.0567	.204	.374	.0587	.205	.386	.0605	.207
250 000	.575	.233	.353	.0554	.240	.366	.0575	.240	.381	.0597	.240	.394	.0619	.242
0000	.528	.275	.357	.0564	.281	.372	.0585	.281	.387	.0607	.282	.403	.0633	.282
000	.470	.346	.362	.0567	.352	.379	.0595	.352	.397	.0623	.352	.411	.0645	.353
00	.418	.437	.369	.0579	.441	.388	.0607	.442	.406	.0637	.442	.423	.0665	.442
0	.373	.550	.377	.0592	.552	.398	.0625	.554	.417	.0653	.554	.432	.0677	.554
1	.332	.695	.384	.0603	.697	.405	.0635	.698	.427	.0673	.698	.447	.0700	.699
2	.292	.879	.393	.0617	.882	.417	.0655	.882	.441	.0697	.882	.463	.0727	.882
3	.260	1.11	.403	.0633	1.11	.431	.0675	1.11	.454	.0712	1.11	.476	.0746	1.11
4	.232	1.40	.413	.0648	1.40	.442	.0695	1.40	.469	.0736	1.40	.494	.0775	1.40
6	.184	2.21	.437	.0685	2.21	.470	.0737	2.21	.501	.0785	2.21	.529	.0830	2.21
7/64 BY 7/64 8/64 BY 8/64 9/64 BY 9/64 10/64 BY 10/64														
500 000	.814	.116	.379	.0595	.130	.389	.0610	.131	.398	.0625	.132	.407	.0640	.133
450 000	.772	.129	.384	.0602	.143	.393	.0617	.143	.403	.0634	.144	.411	.0645	.145
400 000	.728	.145	.389	.0610	.158	.396	.0622	.158	.409	.0642	.159	.417	.0655	.160
350 000	.681	.166	.395	.0620	.177	.402	.0630	.178	.415	.0652	.178	.423	.0664	.179
300 000	.630	.194	.399	.0626	.207	.409	.0642	.205	.421	.0660	.205	.431	.0675	.206
250 000	.575	.233	.409	.0642	.242	.419	.0658	.242	.430	.0675	.242	.442	.0693	.243
0000	.528	.275	.415	.0652	.283	.427	.0673	.283	.441	.0690	.284	.452	.0708	.285
000	.470	.346	.429	.0673	.353	.440	.0690	.353	.455	.0714	.354	.466	.0730	.355
00	.418	.437	.439	.0690	.443	.455	.0714	.443	.469	.0735	.443	.483	.0758	.444
0	.373	.550	.453	.0712	.554	.466	.0731	.554	.485	.0760	.554	.503	.0780	.556
1	.332	.695	.466	.0732	.698	.483	.0757	.697	.501	.0785	.699	.516	.0810	.700
2	.292	.879	.483	.0758	.882	.502	.0787	.882	.521	.0816	.883	.537	.0843	.885
3	.260	1.11	.499	.0782	1.11	.519	.0814	1.11	.538	.0845	1.11	.558	.0875	1.11
4	.232	1.40	.518	.0813	1.40	.538	.0845	1.40	.558	.0875	1.40	.577	.0905	1.40
6	.184	2.21	.557	.0873	2.21	.580	.0910	2.21	.601	.0943	2.21	.622	.0975	2.21
11/64 BY 11/64 12/64 BY 12/64 13/64 BY 13/64 14/64 BY 14/64														
500 000	.814	.116	.417	.0655	.133	.427	.0670	.133	.434	.0687	.134	.441	.0697	.135
450 000	.772	.129	.423	.0665	.145	.431	.0675	.145	.439	.0690	.146	.448	.0705	.147
400 000	.728	.145	.429	.0673	.160	.436	.0683	.160	.446	.0700	.161	.457	.0717	.162
350 000	.681	.166	.436	.0685	.180	.444	.0700	.180	.453	.0710	.180	.464	.0729	.181
300 000	.630	.194	.444	.0697	.206	.456	.0715	.206	.461	.0722	.207	.473	.0742	.208
250 000	.575	.233	.454	.0712	.244	.465	.0730	.244	.475	.0745	.245	.486	.0762	.245
0000	.528	.275	.465	.0730	.285	.476	.0745	.285	.486	.0760	.286	.498	.0782	.287
000	.470	.346	.481	.0755	.355	.493	.0775	.355	.503	.0790	.355	.516	.0810	.356
00	.418	.437	.498	.0780	.445	.510	.0800	.445	.521	.0816	.445	.535	.0840	.446
0	.373	.550	.514	.0805	.556	.528	.0828	.556	.539	.0845	.556	.554	.0870	.557
1	.332	.695	.531	.0830	.700	.546	.0855	.700	.559	.0877	.700	.573	.0900	.700
2	.292	.879	.554	.0870	.882	.570	.0895	.882	.583	.0915	.883	.598	.0938	.884
3	.260	1.11	.574	.0900	1.11	.591	.0927	1.11	.606	.0950	1.11	.618	.0970	1.11
4	.232	1.40	.596	.0935	1.40	.613	.0962	1.40	.627	.0983	1.40	.643	.1010	1.40
6	.184	2.21	.643	.1010	2.21	.661	.1037	2.21	.678	.1063	2.21	.696	.1090	2.21
16/64 BY 16/64 18/64 BY 18/64 20/64 BY 20/64 22/64 BY 22/64														
500 000	.814	.116	.457	.0717	.136	.474	.0744	.138	.487	.0764	.140	.501	.0785	.141
450 000	.772	.129	.462	.0725	.148	.481	.0754	.150	.496	.0778	.151	.509	.0800	.152
400 000	.728	.145	.471	.0738	.163	.487	.0764	.164	.505	.0792	.165	.519	.0815	.166
350 000	.681	.166	.480	.0753	.182	.496	.0778	.183	.513	.0805	.185	.529	.0830	.186
300 000	.630	.194	.491	.0770	.208	.511	.0802	.210	.526	.0825	.211	.541	.0848	.212
250 000	.575	.233	.505	.0792	.246	.524	.0822	.247	.541	.0848	.248	.557	.0875	.249
0000	.528	.275	.517	.0810	.287	.536	.0840	.288	.556	.0870	.289	.573	.0900	.290
000	.470	.346	.536	.0840	.357	.556	.0870	.357	.575	.0905	.358	.592	.0930	.360
00	.418	.437	.552	.0865	.446	.578	.0907	.446	.599	.0940	.447	.618	.0970	.448
0	.373	.550	.575	.0902	.558	.601	.0942	.558	.621	.0972	.558	.641	.1005	.559
1	.332	.695	.598	.0938	.700	.623	.0980	.700	.645	.1010	.701	.666	.1045	.702
2	.292	.879	.623	.0978	.884	.649	.1017	.884	.674	.1060	.885	.693	.1085	.886
3	.260	1.11	.649	.1018	1.11	.674	.1060	1.11	.698	.1095	1.11	.721	.1130	1.12
4	.232	1.40	.673	.1055	1.41	.701	.1100	1.41	.725	.1138	1.41	.746	.1170	1.41
6	.184	2.21	.725	.1155	2.22	.754	.1180	2.22	.780	.1225	2.22	.809	.1270	2.22

*Resistance based upon 100 percent conductivity at 25 degrees C (77 degrees F), including two percent allowance for spiral of strands and two percent allowance for spiral of conductors. For a temperature of 65 degrees C (149 degrees F) these resistance values would be increased 15 percent.

**The inductance is in millihenries; the reactance and the impedance are in ohms.

The table values were derived from the equation $L = 0.08047 + 0.741 \log_{10} \frac{D}{R}$ where R is the radius of conductor, D the distance between centers of conductors expressed in the same terms as R , and L the inductance in millihenries per mile of each conductor. All values in the table are single-phase and based upon a single conductor one mile long.

TABLE XXVI—INDUCTANCE, REACTANCE AND IMPEDANCE, AT 60 CYCLES, PER MILE OF SINGLE CONDUCTOR FOR THREE CONDUCTOR CABLES

AREA IN CIRCULAR MILS B & S NO.	DIAMETER IN INCHES	RESISTANCE PER MILE IN OHMS ★	INSULATION THICKNESS IN 64THS OF AN INCH ★★											
			3/64 BY 3/64			4/64 BY 4/64			6/64 BY 5/64			6/64 BY 6/64		
			IND. M.H.	REAC. OHMS	IMP. OHMS	IND. M.H.	REAC. OHMS	IMP. OHMS	IND. M.H.	REAC. OHMS	IMP. OHMS	IND. M.H.	REAC. OHMS	IMP. OHMS
500 000	.814	.116	.338	.127	.172	.349	.131	.175	.360	.136	.178	.370	.140	.182
450 000	.772	.129	.340	.128	.181	.351	.132	.184	.362	.137	.189	.373	.141	.191
400 000	.728	.145	.343	.129	.195	.354	.134	.197	.364	.138	.201	.377	.142	.194
350 000	.681	.166	.346	.130	.211	.357	.135	.214	.370	.140	.217	.380	.143	.220
300 000	.630	.194	.349	.132	.235	.361	.136	.237	.374	.141	.240	.386	.145	.244
250 000	.575	.233	.353	.133	.268	.366	.138	.271	.381	.144	.274	.394	.149	.277
0000	.528	.275	.357	.135	.308	.372	.140	.309	.387	.146	.313	.403	.152	.316
000	.470	.346	.362	.136	.373	.379	.143	.375	.397	.150	.378	.411	.155	.381
00	.418	.437	.369	.139	.460	.388	.146	.461	.406	.153	.464	.423	.160	.466
0	.373	.550	.377	.142	.569	.398	.150	.571	.417	.157	.572	.432	.163	.573
1	.332	.695	.384	.145	.711	.405	.152	.713	.429	.162	.715	.447	.168	.716
2	.292	.879	.393	.148	.893	.417	.157	.894	.441	.166	.896	.463	.174	.896
3	.260	1.10	.403	.152	1.14	.431	.162	1.12	.454	.177	1.12	.476	.180	1.12
4	.232	1.40	.413	.156	1.41	.441	.167	1.41	.469	.187	1.42	.498	.187	1.42
6	.184	2.21	.437	.165	2.22	.470	.177	2.22	.501	.199	2.22	.529	.200	2.22
7/64 BY 7/64 8/64 BY 8/64 9/64 BY 9/64 10/64 BY 10/64														
500 000	.814	.116	.379	.143	.184	.389	.146	.186	.399	.150	.190	.407	.153	.192
450 000	.772	.129	.384	.145	.194	.393	.148	.195	.403	.152	.199	.411	.155	.202
400 000	.728	.145	.389	.147	.206	.396	.149	.208	.409	.154	.212	.417	.157	.230
350 000	.681	.166	.395	.149	.222	.402	.151	.224	.415	.157	.229	.423	.160	.231
300 000	.630	.194	.399	.150	.245	.409	.154	.246	.421	.158	.251	.431	.162	.254
250 000	.575	.233	.409	.154	.279	.419	.158	.282	.430	.162	.285	.442	.166	.286
0000	.528	.275	.415	.157	.318	.427	.161	.320	.441	.166	.323	.452	.170	.323
000	.470	.346	.429	.162	.383	.440	.166	.385	.455	.172	.388	.466	.176	.389
00	.418	.437	.439	.166	.467	.455	.171	.471	.469	.177	.473	.483	.182	.474
0	.373	.550	.453	.171	.578	.466	.176	.578	.485	.183	.580	.498	.188	.582
1	.332	.695	.466	.176	.718	.483	.182	.697	.501	.189	.720	.516	.195	.721
2	.292	.879	.483	.182	.900	.502	.189	.900	.521	.196	.902	.537	.202	.902
3	.260	1.10	.499	.188	1.13	.519	.195	1.13	.538	.203	1.13	.558	.211	1.13
4	.232	1.40	.518	.195	1.41	.538	.203	1.41	.558	.211	1.42	.577	.218	1.42
6	.184	2.21	.557	.210	2.22	.580	.219	2.22	.601	.226	2.22	.622	.234	2.22
11/64 BY 11/64 12/64 BY 12/64 13/64 BY 13/64 14/64 BY 14/64														
500 000	.814	.116	.417	.157	.195	.427	.161	.198	.434	.164	.202	.441	.166	.202
450 000	.772	.129	.423	.160	.209	.431	.164	.208	.439	.165	.211	.449	.170	.224
400 000	.728	.145	.429	.161	.218	.436	.164	.210	.446	.168	.218	.457	.172	.224
350 000	.681	.166	.436	.164	.235	.444	.168	.237	.453	.171	.240	.464	.178	.240
300 000	.630	.194	.444	.167	.256	.456	.172	.256	.461	.174	.262	.473	.182	.264
250 000	.575	.233	.454	.171	.289	.465	.175	.292	.475	.179	.295	.486	.183	.296
0000	.528	.275	.465	.175	.328	.476	.180	.330	.484	.183	.332	.498	.188	.334
000	.470	.346	.481	.181	.392	.493	.184	.395	.503	.190	.396	.516	.194	.398
00	.418	.437	.498	.188	.476	.510	.192	.479	.521	.196	.480	.535	.202	.482
0	.373	.550	.514	.194	.584	.528	.199	.586	.539	.203	.589	.554	.209	.590
1	.332	.695	.531	.200	.724	.546	.206	.725	.559	.211	.726	.573	.216	.728
2	.292	.879	.554	.209	.905	.570	.215	.906	.583	.220	.908	.598	.225	.910
3	.260	1.10	.574	.216	1.13	.591	.222	1.13	.606	.228	1.13	.638	.233	1.14
4	.232	1.40	.596	.224	1.42	.613	.231	1.42	.627	.236	1.42	.663	.242	1.42
6	.184	2.21	.643	.242	2.22	.661	.249	2.22	.678	.256	2.22	.696	.262	2.22
16/64 BY 16/64 18/64 BY 18/64 20/64 BY 20/64 22/64 BY 22/64														
500 000	.814	.116	.457	.172	.208	.474	.179	.212	.487	.183	.217	.501	.189	.222
450 000	.772	.129	.462	.174	.218	.481	.181	.224	.496	.187	.228	.509	.192	.232
400 000	.728	.145	.471	.178	.230	.487	.183	.235	.505	.190	.240	.519	.196	.244
350 000	.681	.166	.480	.181	.246	.496	.187	.252	.513	.193	.254	.529	.200	.260
300 000	.630	.194	.491	.185	.270	.511	.192	.274	.526	.199	.279	.541	.204	.264
250 000	.575	.233	.505	.190	.302	.524	.197	.306	.541	.204	.311	.557	.210	.314
0000	.528	.275	.517	.195	.338	.534	.202	.342	.556	.210	.348	.573	.216	.352
000	.470	.346	.536	.202	.403	.556	.209	.409	.575	.217	.410	.592	.223	.415
00	.418	.437	.552	.208	.486	.578	.218	.490	.599	.226	.494	.618	.233	.496
0	.373	.550	.575	.217	.592	.601	.226	.596	.621	.234	.599	.641	.242	.602
1	.332	.695	.598	.225	.732	.623	.235	.734	.645	.243	.737	.666	.251	.740
2	.292	.879	.623	.235	.912	.649	.245	.914	.674	.254	.917	.693	.261	.920
3	.260	1.10	.649	.245	1.14	.674	.254	1.14	.698	.262	1.14	.721	.272	1.14
4	.232	1.40	.673	.254	1.42	.701	.264	1.43	.725	.273	1.43	.746	.281	1.43
6	.184	2.21	.725	.273	2.22	.754	.284	2.23	.780	.294	2.23	.809	.305	2.23

*Resistance based upon 100 percent conductivity at 25 degrees C (77 degrees F), including two percent allowance for spiral of strands and two percent allowance for spiral of conductors. For a temperature of 65 degrees C (149 degrees F) these resistance values would be increased 15 percent.

**The inductance is in millihenries; the reactance and the impedance are in ohms.

The table values were derived from the equation $L = 0.08047 + 0.741 \text{Log}_{10} \frac{D}{R}$ where R is the radius of conductor, D the distance between centers of conductors expressed in the same terms as R , and L the inductance in millihenries per mile of each conductor. All values in the table are single-phase and based upon a single conductor one mile long.

inductance, as determined by the fundamental formula, would thus tend to give values several percent less than the actual when applied to three-conductor cable calculations. On the other hand spiraling the conductors of three conductor cables tends to increase their reactance by several percent. It may, therefore, be

assumed that the use of the fundamental formula in the case of three-conductor cables give results approximately correct. Skin effect on the larger cables will, however, tend to decrease the reactance slightly, particularly at 60 cycles.

CAPACITANCE OF 3 CONDUCTOR CABLES

Formulas for determining the approximate capacitance of three-conductor cables are cumbersome. They give reasonably accurate results only in the case of a homogeneous dielectric and in cases where the conductors are small compared to the radius of the sheath. They give inaccurate results in cases of large conductors closely spaced. Fig. 58* illustrates the various

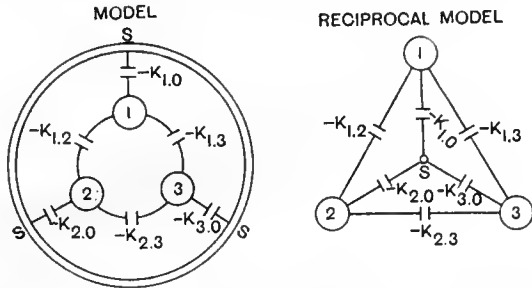


FIG. 58—REPRESENTATION OF CAPACITANCES OF A SYMMETRICAL THREE-PHASE CABLE

capacitances of a three-conductor cable. Formulas taken from Russel's "Alternating Currents" have been combined and converted to common logarithms and are given below. They were derived by the method of images and on the assumption that the conductors are round and symmetrically spaced with respect to the axis of the sheath.

$$C_1 = \frac{I}{13.82 \log_{10} \frac{R^2 - d^2}{3 R^2 d^2 r}} + \frac{I}{6.91 \log_{10} \left(\frac{1.73d}{r} \times \frac{R^2 - d^2}{(R^4 + R^2 d^2 + d^4)^{1/2}} \right)} \times 0.179 \times K \quad (70)$$

$$C_{12} = \frac{I}{13.82 \log_{10} \frac{R^2 - d^2}{3 R^2 d^2 r}} - \frac{I}{13.82 \log_{10} \left(\frac{1.73d}{r} \times \frac{R^2 - d^2}{(R^4 + R^2 d^2 + d^4)^{1/2}} \right)} \times 0.179 \times K \quad (71)$$

Where,—

- R = inside radius of sheath in centimeters (Fig. 59).
- r = radius of conductor in centimeters.
- d = distance between axis of conductor and axis of sheath in centimeters.
- K = the dielectric constant. For impregnated paper insulation it varies between 3 and 4; for varnished cambric insulation it varies between 4 and 6; for rubber insulation it varies between 4 and 9.
- C₁ = capacitance in microfarads per mile between one conductor and the other two conductors plus the sheath.
- C₁₋₂ = mutual capacitance in microfarads per mile between any two conductors. The capacitance to neutral is twice this value.
- C₁₂ is used in determining the capacitance for various combinations or arrangements as explained below.

CAPACITANCE AND SUSCEPTANCE—TABLE XXVII

Table XXVII contains values for capacitance and susceptance of three conductor paper insulated cable for the various sizes of conductors and thicknesses of insulation indicated. All values are based upon a value for K of 3.5 and, as indicated, a thickness of insulation for the jacket the same as that surrounding each con-

ductor. The values were calculated by equations (70) and (71).

The susceptance values given for 25 and 60 cycles are to neutral. In calculating the voltage regulation of circuits, it is general practice to calculate the regulation on the basis of one conductor to neutral. The susceptance between two of the conductors would be half the table values to neutral. The values for susceptance were calculated from the equation,—

$$\text{Susceptance to neutral in micromhos} = 2 \pi f C$$

Thus No. 0 three-conductor cable with 7/64 and 7/64 insulation has a capacitance between conductors of 0.195 microfarads (0.39 microfarads to neutral). The susceptance to neutral at 60 cycles therefore is,—
 $2 \pi 60 \times 0.39 = 147$ microfarads, as indicated by the table.

INTER-RELATION OF CAPACITANCE OF THREE-CONDUCTOR CABLES

The following equations for determining the effective capacitance for various arrangements of the three conductors and the sheath are given in Russell's "Alternating Currents."

$$\text{Capacitance between 1 and 2} = \frac{1}{2} (C_1 - C_{12}) \dots \dots (72)$$

$$\text{Capacitance between 1 and 2, 3} = \frac{2}{3} (C_1 - C_{12}) \dots \dots (73)$$

$$\text{Capacitance between 1 and S (2 and 3 insulated)} = \frac{(C_1 - C_{12})(C_1 + 2C_{12})}{C_1 + C_{12}} \dots \dots (74)$$

$$\text{Capacitance between 1 and S, 2 (3 insulated)} = \frac{(C_1 - C_{12})(C_1 + C_{12})}{C_1} \dots \dots (75)$$

$$\text{Capacitance between 1 and S, 2, 3} = C_1 \dots \dots (76)$$

$$\text{Capacitance between S and 1, 2, (3 insulated)} = \frac{2(C_1 - C_{12})(C_1 + 2C_{12})}{C_1} \dots \dots (77)$$

$$\text{Capacitance between 1, S and 2, 3} = 2(C_1 + C_{12}) \dots \dots (78)$$

$$\text{Capacitance between S and 1, 2, 3} = 3(C_1 + 2C_{12}) \dots \dots (79)$$

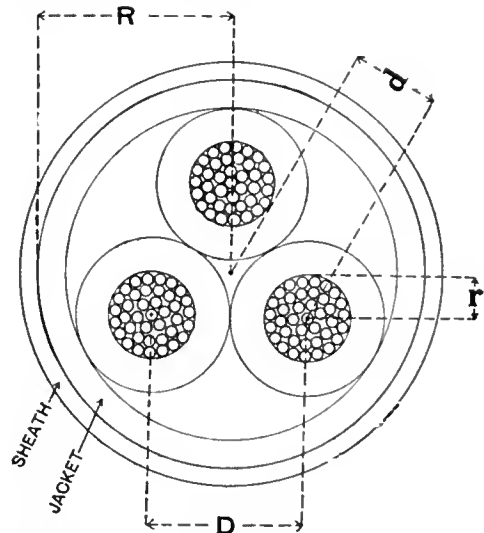


FIG. 59—DIMENSIONS OF A SYMMETRICAL THREE-PHASE CABLE

C₁ (76) may be measured in the ordinary way, by reading the throw of a mirror galvanometer and comparing with the throw given by a standard condenser. A further measurement of (78) or (79) will give a simple equation to find C₁₂. For instance, if measurements were taken of (78) and (79) and were found to be:—

*Reproduced from Alexander Russel's "Alternating Currents."

TABLE XXVII—CAPACITANCE AND SUSCEPTANCE PER MILE OF THREE CONDUCTOR PAPER INSULATED CABLES

AREA IN CIRCULAR MILS B & S NO.	INSULATION THICKNESS IN 64THS OF AN INCH																			
	3/64 BY 3/64					4/64 BY 4/64					5/64 BY 5/64					6/64 BY 6/64				
	CAPACITANCE			SUSCEPTANCE TO NEUTRAL		CAPACITANCE			SUSCEPTANCE TO NEUTRAL		CAPACITANCE			SUSCEPTANCE TO NEUTRAL		CAPACITANCE			SUSCEPTANCE TO NEUTRAL	
	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES
500 000	.680	-.217	.448	141	337	.613	-.175	.394	124	297	.555	-.154	.354	111	267	.505	-.137	.321	101	242
450 000	.667	-.197	.432	136	325	.590	-.169	.379	119	286	.538	-.149	.343	108	259	.488	-.130	.309	97	233
400 000	.657	-.194	.425	133	320	.570	-.159	.364	114	274	.517	-.142	.329	103	248	.475	-.125	.300	94	226
350 000	.640	-.189	.414	130	313	.560	-.158	.359	113	270	.506	-.138	.322	101	242	.460	-.119	.289	91	218
300 000	.606	-.176	.391	123	294	.545	-.153	.349	110	263	.490	-.131	.310	97	234	.446	-.116	.281	88	212
250 000	.590	-.171	.385	120	286	.518	-.142	.330	104	249	.468	-.125	.296	93	223	.427	-.109	.268	84	202
0000	.570	-.160	.365	111	265	.500	-.134	.317	100	239	.445	-.115	.280	88	211	.407	-.103	.255	80	192
000	.535	-.147	.341	107	257	.475	-.125	.300	94	228	.420	-.107	.262	82	198	.384	-.095	.239	75	180
00	.513	-.140	.327	103	246	.447	-.116	.281	88	212	.398	-.101	.249	78	187	.364	-.088	.226	71	170
0	.494	-.123	.308	97	232	.422	-.107	.264	83	199	.374	-.090	.232	73	175	.342	-.081	.211	66	159
1	.462	-.119	.290	91	219	.398	-.099	.248	78	187	.356	-.086	.221	69	167	.323	-.074	.198	62	149
2	.420	-.107	.263	83	198	.373	-.091	.232	73	175	.332	-.077	.203	64	153	.305	-.070	.187	59	141
3	.402	-.101	.251	79	189	.352	-.084	.218	69	165	.314	-.072	.193	61	145	.284	-.062	.173	54	131
4	.378	-.100	.239	75	180	.330	-.077	.203	64	153	.295	-.066	.180	57	136	.270	-.059	.164	52	124
6	.342	-.081	.211	66	159	.301	-.063	.182	57	137	.264	-.056	.160	50	121	.239	-.050	.144	45	108
7/64 BY 7/64 8/64 BY 8/64 9/64 BY 9/64 10/64 BY 10/64																				
CAPACITANCE					SUSCEPTANCE TO NEUTRAL		CAPACITANCE					SUSCEPTANCE TO NEUTRAL		CAPACITANCE					SUSCEPTANCE TO NEUTRAL	
C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	
500 000	.468	-.124	.296	93	224	.435	-.115	.275	86	207	.410	-.104	.257	81	193	.392	-.097	.244	77	184
450 000	.454	-.119	.286	90	216	.427	-.107	.267	84	201	.405	-.103	.254	79	191	.380	-.093	.236	74	178
400 000	.442	-.116	.279	88	210	.415	-.105	.260	82	196	.392	-.099	.245	77	184	.368	-.090	.229	72	173
350 000	.426	-.108	.267	84	201	.398	-.099	.248	78	187	.380	-.093	.236	74	178	.358	-.087	.222	70	167
300 000	.415	-.105	.260	82	196	.390	-.096	.243	76	183	.365	-.089	.227	71	171	.348	-.081	.215	68	162
250 000	.400	-.101	.250	79	188	.370	-.089	.229	72	173	.352	-.087	.219	69	165	.332	-.078	.205	65	155
0000	.380	-.094	.237	75	178	.354	-.085	.220	69	166	.334	-.076	.205	64	155	.316	-.073	.194	61	146
000	.358	-.086	.222	70	168	.332	-.079	.205	64	155	.315	-.073	.194	61	146	.296	-.066	.181	57	136
00	.336	-.080	.208	65	157	.313	-.071	.192	60	145	.295	-.067	.181	57	136	.278	-.061	.169	53	127
0	.317	-.073	.195	61	147	.293	-.065	.179	56	135	.279	-.061	.170	54	128	.263	-.056	.159	50	120
1	.299	-.068	.180	54	138	.284	-.061	.160	50	121	.257	-.054	.158	50	119	.247	-.050	.145	47	114
2	.279	-.061	.170	54	128	.264	-.056	.160	50	121	.247	-.052	.150	47	113	.233	-.048	.140	44	106
3	.264	-.056	.160	50	121	.248	-.052	.150	47	113	.232	-.048	.140	44	106	.222	-.044	.133	42	100
4	.250	-.053	.151	47	114	.233	-.048	.140	44	106	.221	-.045	.133	42	100	.210	-.041	.125	39	94
6	.221	-.045	.133	42	100	.209	-.041	.125	39	94	.198	-.037	.117	37	88	.188	-.036	.112	35	85
11/64 BY 11/64 12/64 BY 12/64 13/64 BY 13/64 14/64 BY 14/64																				
CAPACITANCE					SUSCEPTANCE TO NEUTRAL		CAPACITANCE					SUSCEPTANCE TO NEUTRAL		CAPACITANCE					SUSCEPTANCE TO NEUTRAL	
C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	
500 000	.371	-.084	.239	72	173	.355	-.085	.220	69	164	.343	-.082	.212	67	160	.329	-.078	.203	64	153
450 000	.364	-.087	.225	71	170	.352	-.085	.218	69	164	.332	-.082	.205	64	155	.321	-.075	.198	62	149
400 000	.356	-.085	.220	69	166	.338	-.080	.209	66	157	.326	-.076	.201	63	152	.310	-.071	.190	60	143
350 000	.340	-.080	.210	66	158	.328	-.077	.202	63	152	.313	-.073	.195	61	147	.300	-.068	.184	58	139
300 000	.329	-.078	.203	64	153	.313	-.071	.192	60	145	.307	-.069	.186	59	140	.290	-.065	.177	56	133
250 000	.316	-.072	.199	63	150	.308	-.068	.183	58	138	.288	-.064	.176	55	133	.276	-.061	.168	53	127
0000	.302	-.069	.185	58	140	.285	-.067	.176	55	133	.278	-.061	.169	53	127	.264	-.056	.160	50	121
000	.282	-.061	.171	54	129	.271	-.060	.165	52	124	.261	-.056	.158	50	119	.251	-.053	.152	48	115
00	.267	-.059	.162	51	122	.255	-.059	.154	48	116	.247	-.052	.150	47	113	.237	-.048	.142	45	107
0	.250	-.053	.151	48	114	.241	-.050	.145	46	109	.233	-.048	.140	44	106	.222	-.044	.133	42	100
1	.237	-.050	.143	45	108	.228	-.047	.137	43	103	.220	-.044	.132	42	100	.212	-.042	.127	40	96
2	.222	-.045	.133	42	100	.216	-.044	.130	41	98	.208	-.041	.124	39	94	.199	-.039	.118	37	89
3	.212	-.042	.127	40	96	.204	-.039	.121	38	91	.195	-.037	.116	36	88	.190	-.036	.113	36	85
4	.201	-.039	.120	38	91	.192	-.037	.114	36	86	.186	-.034	.110	35	83	.180	-.033	.106	33	80
6	.181	-.033	.107	34	81	.174	-.031	.102	32	77	.168	-.030	.099	31	75	.163	-.029	.096	30	73
16/64 BY 16/64 18/64 BY 18/64 20/64 BY 20/64 22/64 BY 22/64																				
CAPACITANCE					SUSCEPTANCE TO NEUTRAL		CAPACITANCE					SUSCEPTANCE TO NEUTRAL		CAPACITANCE					SUSCEPTANCE TO NEUTRAL	
C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	C ₁	C ₁₂	C _{1&2}	25 CYCLES	60 CYCLES	
500 000	.308	-.071	.189	59	143	.288	-.065	.176	55	133	.276	-.061	.168	53	127	.263	-.057	.160	50	121
450 000	.302	-.069	.185	58	140	.282	-.062	.172	54	130	.269	-.058	.163	51	123	.255	-.054	.154	48	116
400 000	.292	-.066	.179	56	135	.276	-.060	.168	53	127	.260	-.055	.157	50	118	.249	-.053	.151	47	113
350 000	.282	-.061	.171	54	129	.266	-.057	.161	51	121	.252	-.053	.152	48	115	.240	-.049	.144	45	108
300 000	.271	-.059	.165	52	125	.256	-.054	.155	49	117	.242	-.050	.146	46	110	.232	-.047	.139	44	105
250 000	.260	-.055	.157	49	118	.244	-.051	.147	46	111	.234	-.048	.139	44	105	.222	-.044	.133	42	100
0000	.248	-.052	.150	47	113	.233	-.049	.141	44	106	.222	-.044	.133	42	100	.211	-.041	.126	40	96
000	.234	-.048	.141	44	106	.221	-.046	.132	42	100	.211	-.041	.126							

TABLE XXVIII—THREE-PHASE CHARGING KV-A PER MILE OF THREE-PHASE CIRCUIT OF THREE CONDUCTOR PAPER INSULATED CABLES

25 CYCLES

AREA IN CIRCULAR MILS B & S NO.	CHARGING KV-A PER MILE (EXPRESSED IN KV-A 3 PHASE) FOR PAPER INSULATED THREE CONDUCTOR CABLES BASED UPON A VALUE FOR 'K' OF 3.5 AND UPON A THICKNESS OF INSULATION SURROUNDING THE CONDUCTORS AND OF THE JACKET INDICATED.										
	220 VOLTS	440 VOLTS	550 VOLTS	1100 VOLTS	2200 VOLTS	4400 VOLTS	6000 VOLTS		6600 VOLTS		6900 VOLTS
	4 64	4 64	4 64	5 64	6 64	8 64	10 64	14 64	10 64	14 64	10 64
500 000	.03600	.0240	.0376	.134	.488	1.66	2.76	2.26	3.35	2.79	3.66
450 000	.03375	.0230	.0360	.131	.469	1.62	2.66	2.22	3.22	2.70	3.52
400 000	.03150	.0220	.0346	.125	.455	1.58	2.58	2.15	3.13	2.61	3.42
350 000	.02925	.0218	.0332	.122	.440	1.51	2.51	2.08	3.04	2.52	3.33
300 000	.02700	.0213	.0323	.117	.425	1.47	2.44	2.01	2.96	2.44	3.23
250 000	.02500	.0201	.0315	.113	.406	1.39	2.33	1.90	2.83	2.31	3.07
000 000	.00483	.0193	.0303	.106	.387	1.33	2.19	1.79	2.65	2.18	2.90
000 000	.00454	.0182	.0285	.109	.363	1.24	2.04	1.72	2.48	2.09	2.71
000 000	.00424	.0170	.0266	.104	.343	1.16	1.90	1.61	2.31	1.96	2.52
0	.00400	.0160	.0250	.108	.319	1.08	1.79	1.51	2.18	1.83	2.37
1	.00376	.0151	.0236	.103	.300	1.04	1.68	1.43	2.05	1.74	2.23
2	.00352	.0141	.0221	.102	.275	.965	1.58	1.33	1.92	1.61	2.09
3	.00333	.0133	.0209	.107	.240	.908	1.51	1.29	1.83	1.57	2.00
4	.00309	.0124	.0194	.106	.218	.855	1.40	1.28	1.70	1.44	1.85
6	.00275	.0110	.0173	.106	.218	.855	1.26	1.08	1.52	1.31	1.66

25 CYCLES										
10,000 VOLTS	11,000 VOLTS		13,200 VOLTS		16,500 VOLTS	20,000 VOLTS		22,000 VOLTS	25,000 VOLTS	
12 64	12 64	14 64	12 64	16 64	14 64	18 64	18 64	18 64	18 64	20 64
500 000	6.93	8.35	7.77	12.00	10.25	17.35	23.46	22.00	26.66	34.5
450 000	6.83	8.23	7.50	11.80	10.10	16.80	23.3	21.6	26.1	33.8
400 000	6.62	7.98	7.26	11.50	9.75	16.25	22.4	21.2	25.6	33.2
350 000	6.33	7.62	7.02	10.95	9.40	15.75	21.6	20.4	24.6	32.0
300 000	6.02	7.27	6.78	10.45	9.05	15.20	20.8	19.6	23.6	30.7
250 000	5.82	7.02	6.42	10.10	8.52	14.60	19.6	18.4	22.2	28.8
000 000	5.53	6.66	6.05	9.56	8.17	13.55	18.8	17.6	21.3	27.6
000 000	5.23	6.30	5.82	9.05	7.65	13.05	17.6	16.8	20.3	26.5
000 000	4.82	5.80	5.45	8.36	7.30	12.25	16.8	15.6	18.8	24.4
0	4.62	5.57	5.09	8.00	6.78	11.40	15.6	14.8	17.8	23.2
1	4.32	5.21	4.84	7.48	6.43	10.85	14.8	14.0	16.9	21.9
2	4.12	4.97	4.48	7.13	6.26	10.05	14.4	13.2	15.9	20.7
3	3.82	4.60	4.36	6.60	5.72	9.78	13.6	12.8	15.5	20.1
4	3.65	4.36	4.06	6.27	5.57	8.95	12.8	12.0	14.5	18.8
6	3.22	3.87	3.63	5.37	4.87	8.15	11.2	10.8	13.1	16.3

60 CYCLES										
220 VOLTS	440 VOLTS	550 VOLTS	1100 VOLTS	2200 VOLTS	4400 VOLTS	6000 VOLTS		6600 VOLTS		6900 VOLTS
4 64	4 64	4 64	5 64	6 64	8 64	10 64	14 64	10 64	14 64	10 64
500 000	.0143	.0574	.0900	.323	1.17	4.00	6.58	5.49	8.00	6.65
450 000	.0138	.0554	.0858	.313	1.13	3.88	6.40	5.35	7.75	6.48
400 000	.0132	.0530	.0830	.301	1.09	3.79	6.20	5.13	7.53	6.22
350 000	.0131	.0523	.0818	.292	1.05	3.61	5.98	4.98	7.25	6.07
300 000	.0127	.0510	.0798	.283	1.02	3.54	5.80	4.78	7.05	5.80
250 000	.0120	.0483	.0755	.270	.975	3.35	5.55	4.56	6.75	5.52
000 000	.0115	.0463	.0725	.255	.925	3.22	5.24	4.33	6.35	5.26
000 000	.0105	.0433	.0685	.240	.870	3.05	4.88	4.14	5.99	5.00
000 000	.0102	.0410	.0643	.226	.820	2.80	4.55	3.84	5.52	4.65
0	.0096	.0385	.0602	.212	.768	2.61	4.30	3.59	5.22	4.35
1	.0090	.0362	.0566	.202	.720	2.48	4.08	3.44	4.95	4.17
2	.0084	.0339	.0530	.185	.680	2.34	3.80	3.19	4.60	3.87
3	.0080	.0320	.0500	.176	.632	2.18	3.59	3.05	4.35	3.70
4	.0074	.0296	.0465	.165	.600	2.05	3.37	2.87	4.08	3.48
6	.0066	.0265	.0415	.147	.522	1.82	3.05	2.62	3.70	3.17

60 CYCLES										
10,000 VOLTS	11,000 VOLTS		13,200 VOLTS		16,500 VOLTS	20,000 VOLTS		22,000 VOLTS	25,000 VOLTS	
12 64	12 64	14 64	12 64	16 64	14 64	18 64	18 64	18 64	18 64	20 64
500 000	16.7	20.1	18.5	28.9	24.9	41.5	57.3	53.3	64.2	83.3
450 000	16.5	19.8	18.1	28.5	24.9	40.3	56.0	52.0	62.7	81.5
400 000	15.8	19.0	17.3	27.2	23.5	38.8	54.0	50.9	61.3	79.5
350 000	15.3	18.4	16.8	26.4	22.4	37.6	51.6	48.5	58.3	75.8
300 000	14.6	17.6	16.1	25.2	21.8	36.0	50.0	46.8	56.5	73.3
250 000	13.9	16.7	15.4	24.0	20.5	34.5	47.3	44.3	53.6	69.5
000 000	13.4	16.1	14.7	23.2	19.6	32.8	45.3	42.5	51.2	66.5
000 000	12.5	15.0	13.9	21.6	18.4	31.2	42.5	40.0	48.3	62.7
000 000	11.7	14.1	13.0	20.2	17.4	29.0	40.0	37.6	45.4	59.0
0	11.0	13.2	12.1	18.9	16.3	27.2	37.6	35.7	43.0	55.8
1	10.4	12.5	11.6	17.9	15.5	26.0	35.6	33.6	40.5	52.5
2	9.85	11.9	10.8	17.0	14.9	24.1	33.5	31.7	38.2	49.5
3	9.15	11.0	10.3	15.8	14.1	23.0	32.4	30.4	36.7	47.6
4	8.65	10.4	9.7	15.0	13.2	21.7	30.4	28.4	34.3	44.5
6	7.75	9.3	8.8	13.4	11.7	19.8	26.9	25.6	30.9	38.9

The values in Table XXVIII are based upon a value for the dielectric constant K of 3.5. For all other values of K the table values will change in direct proportion. Values for K will usually be found between the following limits; for impregnated paper 3.0 to 4.0; for varnished cambric 4.0 to 6.0 and for rubber 4.0 to 9.0.

$2 C_1 + 2 C_{12} = 0.410 \text{ mf. per mile} \dots\dots\dots (78)$

And $3 C_1 + 6 C_{12} = 0.450 \text{ mf. per mile} \dots\dots\dots (79)$

Therefore $C_1 = 0.26 \text{ mf. per mile}$

$C_{12} = -0.055 \text{ mf per mile}$

Numerical Examples—From Table XXVII for a 250 000 circ. mil., three-conductor cable having a band of insulation surrounding each conductor of 16/64 of an inch and an insulation jacket surrounding all three conductors of the same thickness, the following values are obtained:—

$C_1 = 0.260 \text{ mf. per mile.}$

$C_{12} = -0.055 \text{ mf. per mile.}$

Then, in the order in which the capacitance increases,—

Capacitance between 1 and 2 = 0.157 mf. per mile.... (72)

Capacitance between 1 and 2, 3 = 0.210 mf. per mile.. (73)

Capacitance between 1 and S (2 and 3 insulated) = 0.230 mf. per mile..... (74)

Capacitance between 1 and S, 2 (3 insulated) = 0.248 mf. per mile..... (75)

Capacitance between 1 and S, 2, 3 = 0.260 mf. per mile..... (76)

Capacitance between S and 1, 2 (3 insulated) = 0.363 mf. per mile..... (77)

$$\begin{aligned} \text{Capacitance between } 1, S \text{ and } 2, 3 &= 0.410 \text{ mf. per mile} \dots\dots\dots (78) \\ \text{Capacitance between } S \text{ and } 1, 2, 3 &= 0.450 \text{ mf. per mile} \dots\dots\dots (79) \end{aligned}$$

COMPARISON OF CALCULATED CAPACITANCE WITH TEST RESULTS

The difference between measured results of capacitance and the results calculated by the above formulas are given in Fig. 60. It will be seen that in all cases these calculated results are less than the corresponding test results, the discrepancy being greater as the conductor becomes larger and the separation less. The differences vary from zero to as much as eleven percent for the largest cable, at the minimum spacing shown. The discrepancy is greatest with the minimum thickness of insulation. Since such cables would be used only for low-voltage service, the charging current would be small and consequently this error would probably be of little importance. For 6600 volt cables the results by the formula would seem to be approximately five percent too low.

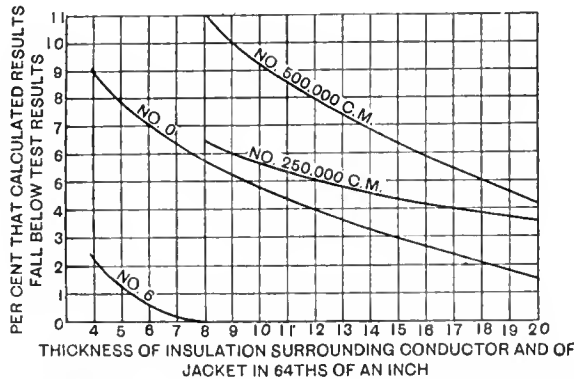


FIG. 60—COMPARISON OF CALCULATED AND MEASURED CAPACITANCES Tests made on three conductor paper insulated cables, K = 3.5.

The cause of the discrepancy between the formula and test results is as follows:—In order to obtain a mathematical solution, Russell found it necessary to make certain approximations to the true physical conditions. Thus the resulting mathematical formula cannot give exact results. The approximation made by Russell is very close to the actual physical fact where the conductors are small compared with the insulation thickness, but it is not very close where the conductors are large compared with the insulation.

CHARGING KV-A—TABLE XXVIII

Table XXVIII contains values for charging current (expressed in kv-a, three-phase) for three-conductor paper insulated cables, both 25 and 60 cycles, based upon a value for K of 3.5. For other values of K, the table values would vary in proportion. For other thicknesses of insulation, the kv-a values would vary as the susceptance values corresponding to the thickness of insulation (See Table XXVII). In some cases, such for instance, as grounded neutral systems, the thickness

of insulation of the jacket may be less than that surrounding the conductors. In such cases it might be desirable to calculate the susceptance and charging current, if accurate results were desired. The values for charging current corresponding to two thicknesses of insulation are included for some of the commonly employed transmission voltages.

These kv-a values were calculated by using the values for susceptance in Table XXVII which, in turn, were derived from the capacitance in the same table obtained by formulae (70) and (71). Thus a 350 000 circ. mil cable with 10/64 and 10/64 paper insulation has a 60 cycle susceptance to neutral of 167 micromhos per mile. Since the charging current in amperes to neutral equals the susceptance to neutral \times volts to neutral $\times 10^{-6}$ and assuming 6600 volts, three-phase between conductors, we have:—

$$167 \times \frac{6600}{1.73} \times 10^{-6} = 0.637 \text{ amperes to neutral.}$$

$$\text{Charging kv-a} = 0.637 \times 3815 \times 3 = 7.25 \text{ kv-a,}$$

as indicated in Table XXVIII.

VALUES FOR K

The capacitance of any cable depends upon the dielectric constant of the insulating material and a dimension term or form factor. The dielectric constant should be determined from actual cables and not from samples of material. The usual range in value for K is given below.

	Value of K
Impregnated Paper	3.0 to 4.0
Varnish Cambric	4.0 to 6.0
Rubber	4.0 to 9.0

All values in Tables XXVII and XXVIII are based upon a value of K of 3.5. For all other values of K all table values will vary in the same proportion as their K values. The actual value of permittivity of most paper insulation runs about ten percent less than the value 3.5 which has been used in calculating the accompanying table values. The true alternating-current capacitance is always considerably lower than the capacitance measured with ballistic galvanometer.

REFERENCES

“Electric Power Conductors,” by W. S. Del Mar.
 “Electric Cables,” by Coyle and Howe, London, England.
 “The Heating of Cables with Current,” by Melsom & Booth, *Journal I. E. E.*, Vol. 47—1911.
 “Current and Rating of Electric Cables,” by Atkinson & Fisher, *Trans. A. I. E. E.*, 1913, p. 325.
 “The Heating of Cables Carrying Current,” by Dushman, *Trans. A. I. E. E.*, 1913, p. 333.
 “Effect of Moisture in the Earth on Temperature of Underground Cables,” by Imlay, *Trans. A. I. E. E.*, 1915, Part I, p. 223.
 “Temperature Rise of Insulated Lead Covered Cables,” by Richard A. Powell, *Trans. A. I. E. E.*, 1916, Part II, p. 1017.
 “The Restoration of Service After a Necessary Interruption,” by Ricketts, *Trans. A. I. E. E.*, 1916, Part II, p. 635.
 “The Dielectric Field in an Electric Power Cable,” by Atkinson, *Trans. A. I. E. E.*, June 1919.
 “The Current Carrying Capacity of Lead Covered Cables,” by Atkinson, *Journal A. I. E. E.*, Sept. 1920.

CHAPTER XIV

SYNCHRONOUS MOTORS AND CONDENSERS FOR POWER-FACTOR IMPROVEMENT

BEFORE discussing the employment of synchronous machinery for improving the power-factor of circuits, it may be desirable to review how a change in power-factor affects the generators supplying the current.

Fig. 61 shows the effect of in-phase, lagging and leading components of armature current upon the field strength of generators*. A single-coil armature is illustrated as revolving between the north and south poles of a bipolar alternator. The coil is shown in four positions 90 degrees apart, corresponding to one complete revolution of the armature coil. The direction of the field flux is assumed to be constant as indicated by the arrows on the field poles of each illustration. In addition to this field flux, when current flows through the armature coil another magnetic flux is set up, magnetizing the iron in the armature in a direction at right angles to the plane of the armature coil. This will be referred to as armature flux.

This armature flux varies with the armature current, being zero in a single-phase generator when no armature current flows, and reaching a maximum when full armature current flows. It changes in direction relative to the field flux as the phase angle of the armature current changes.

The revolving armature coil generates an alternating voltage the graph of which follows closely a sine wave, as shown in Fig. 61. When it occupies a vertical plane marked *start* no voltage is generated, for the reason that the instantaneous travel of the coil, is parallel with the field flux.** As the coil moves forward in a clockwise direction, the field enclosed by the armature coil decreases; at first slowly but then more rapidly until the rate of change of flux through the coil becomes a maximum when the coil has turned 90 degrees, at which instant the voltage generated becomes a maximum. As the horizontal position is passed the voltage decreases until it again reaches zero when the coil has traveled 180 degrees or occupies again a vertical plane. As the travel continues the voltage again starts to increase but since the motion of the coil

relative to the fixed magnetic field is reversed the voltage in the coil builds up in the reverse direction during the second half of the revolution. When the coil has reached the two 270 degree position the voltage has again become maximum but in the opposite direction to that when the coil occupied the position of 90 degrees. When the coil returns to its original position at the start the voltage has again dropped to zero, thus completing one cycle.

If the current flowing through this armature coil is in phase with the voltage, it will produce cross magnetization in the armature core, in a vertical direction, as indicated by the arrows at the 90 and 270 degree positions. The cross magnetization neither opposes nor adds to the field flux at low loads and therefore has comparatively little influence on the field flux. At heavy loads, however, this cross magnetization has considerable demagnetizing effect, due to the shift in rotor position resulting from the shifting of the field flux at heavy loads.

If the armature is carrying lagging current, this current will tend to magnetize the armature core in such a direction as to oppose the field flux. This action is shown by the middle row of illustrations of Fig. 61. Under these illustrations is shown a current wave lagging 90 degrees representing the component of current required to magnetize transformers, induction motors, etc. When the lagging component of current reaches its maximum value the armature coil will occupy a vertical position (position marked *start*, 180 degrees and 360 degrees) and in this position the armature flux will directly oppose the field flux, as indicated by the arrows. The result is to reduce the flux threading the armature coil and thus cause a lowering of the voltage. This lagging current encounters resistance and a relatively much greater reactance, each of which consumes a component of the induced voltage, as shown in Fig. 62. When the armature current is lagging, the voltage induced by armature inductance is in such a direction as to subtract from the induced voltage, and thus the voltage is still further lowered, as a result of the armature self induction. In order to bring the voltage back to its normal value it will be necessary to increase the field flux by increasing the field current. Generators are now usually designed of sufficient field capacity to compensate for lagging loads of 80 per cent power-factor.

If the armature is carrying a leading current this leading component will tend to magnetize the armature core in such a direction as to add to the field flux.

*For a more detailed discussion of this subject the reader is referred to excellent articles by F. D. Newbury in the *ELECTRIC JOURNAL* of April 1918, "Armature Reaction of Poly-phase Alternators"; and of July 1918, "Variation of Alternator Excitation with Load".

**For the sake of simplicity this and the following statements are based upon the assumption that armature reaction does not shift the position of the field flux. Actually, under load, the armature reaction causes the position of the field flux to be shifted toward one of the pole tips, so that the position of the armature coil is not quite vertical at the instant of zero voltage in the coil.

This action is shown by the bottom row of illustrations of Fig. 61. Under these illustrations is shown a current wave leading the voltage wave by 90 degrees. When the leading component of current reaches its maximum values, the armature coil will again occupy vertical positions, but the armature flux will add to that of the field flux, as indicated by the arrow. The resulting flux threading the armature coil is thus increased causing a rise in voltage. This leading current flowing through the generator armature encounters resistance and a relatively much greater reactance, each of which consumes a component of the induced voltage, as shown in Fig. 62. When the armature current is lead-

ture current than in an alternator having armature ampere turns large compared with its field ampere turns.

Modern alternators are of such design that when carrying rated lagging current at zero power-factor they require approximately 200 to 250 percent of their no-load field-current and when carrying rated leading current at zero power-factor they require approximately -15 to +15 percent of their no-load field current. Thus with lagging armature current the iron will be worked at a considerable higher point on the saturation curve and the heating of the field coils will increase because of the greater field current required.

The voltage diagrams of Fig. 62 are intended to show only the effect of armature resistance and armature reactance upon voltage variation. Voltage regu-

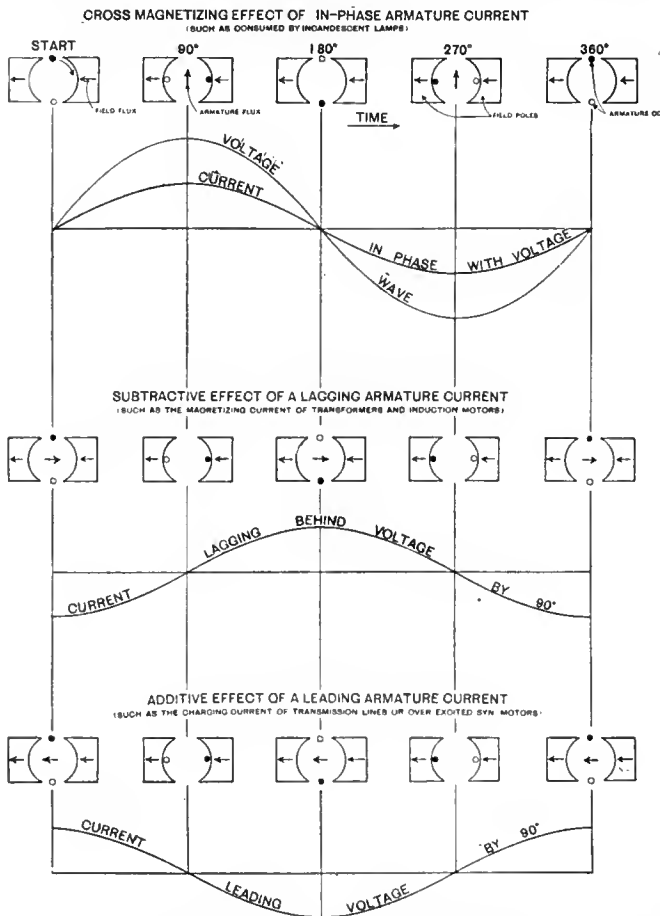


FIG. 61—EFFECT OF ARMATURE CURRENT UPON FIELD EXCITATION OF ALTERNATING-CURRENT GENERATORS

ing, the voltage induced by armature inductance is in such a direction as to add to the induced voltage and thus the voltage at the alternator terminals is still further increased as the result of armature self-induction. In order to reduce the voltage to its normal value it is necessary to decrease the field flux by decreasing the field current.

With alternators of high reaction the magnetizing or de-magnetizing effect of leading or lagging current will be greater than in cases where the armature reaction is low. For instance if the alternator is so designed that the ampere turns of the armature at full armature current are small compared to its field ampere turns, the voltage of such a machine would be less disturbed with a change in power-factor of the arma-

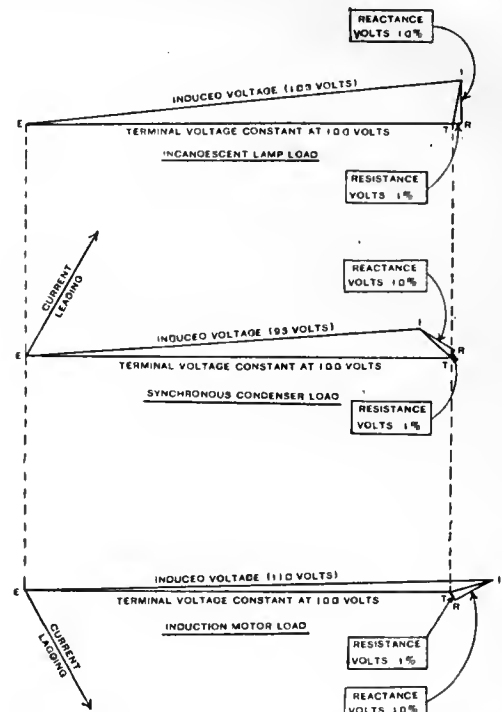


FIG. 62—VECTORS ILLUSTRATING THE EFFECT OF ARMATURE REACTANCE AND RESISTANCE UPON THE TERMINAL VOLTAGE FOR IN-PHASE, LEADING AND LAGGING CURRENTS

lation is the combined effect of armature impedance and armature reaction. Turbogenerators have, for instance, very low armature reactance but their armature reaction is higher, so that the resulting voltage regulation may not be materially different from that of a machine with double the armature reactance. Under normal operation armature reaction is a more potent factor in determining the characteristics of a generator than armature reactance. In the case of a generator with a short circuit ratio of unity, this total reactive effect may be due, 15 percent to armature reactance and 85 percent to armature reaction.

For the case illustrated by Fig. 62 the field flux corresponds to the induced voltage indicated, but the field current does not. The field current corresponds to a value obtained by substituting the full synchronous impedance drop for that indicated.

SYNCHRONOUS CONDENSERS AND PHASE MODIFIERS

The term "synchronous condenser" applies to a synchronous machine for raising the power-factor of circuits. It is simply floated on the circuit with its fields over excited so as to introduce into the circuit a leading current. Such machines are usually not intended to carry a mechanical load. When this double duty is required they are referred to as synchronous motors for operation at leading power-factor. On long transmission circuits, where synchronous condensers are used in parallel with the load for varying the power-factor, thereby controlling the transmission voltage, it is sometimes necessary to operate them with under excited fields at periods of lightloads. They are then no longer synchronous condensers but strictly speaking become synchronous reactors.

Whether synchronous motors for operation at leading power-factor, synchronous condensers or synchronous reactors be used they virtually do the same thing, that is; their function is to change the power-factor of the load by changing the phase angle between the armature current and the terminal voltage. They

TABLE R—SYNCHRONOUS CONDENSER LOSSES

Kv-a	Loss (Kw)	Kv-a	Loss (Kw)
100	12	3500	180
200	18	5000	220
300	22	7500	320
500	32	10000	420
750	47	15000	620
1000	55	20000	820
1500	70	25000	1000
2000	120	35000	1400
2500	130	50000	2000

are, therefore, sometimes referred to as "phase modifiers." This latter name seems more appropriate when the machine is to be operated both leading and lagging, as when used for voltage control of long transmission lines.

Rating — Synchronous condensers as regularly built may be operated at from 30 to 40 percent of their rating lagging, depending upon the individual design. Larger lagging loads result in unstable operation on account of the weakened field. Phase modifiers can be designed to operate at full rating, both leading and lagging, but they are larger, require larger exciters, have a greater loss and cost 15 to 20 percent more than standard condensers.

Starting—Condensers are furnished with squirrel-cage damper windings, to prevent hunting, which also provides a starting torque of approximately 30 percent of normal running torque. They have a pull-in torque of around 15 percent of running torque. The line current at starting varies from 50 to 100 percent of normal. The larger units are sometimes equipped for forced oil lubrication, which raises the rotor sufficiently to permit of oil entering the bearing, thus reducing the starting current.

Mechanical Load—Synchronous condensers are generally built for high speeds and equipped with shafts of small diameter. If they are to be used to transmit some mechanical power it may be necessary to equip them with larger shafts and bearings, particularly if belted rather than direct connected. If a phase modifier is to furnish mechanical energy and at the same time to operate lagging at times of light load for the purpose of holding down the voltage on an unloaded transmission line there may be danger of the machine falling out of step, if a heavy mechanical load occurs when the machine is operating with a weak field.

Losses—At rated full load leading power-factor the total losses, including those of the exciter, will vary from approximately 12 percent for the smallest capacity to approximately four percent for the larger capacity 60 cycle synchronous condensers. The approximate

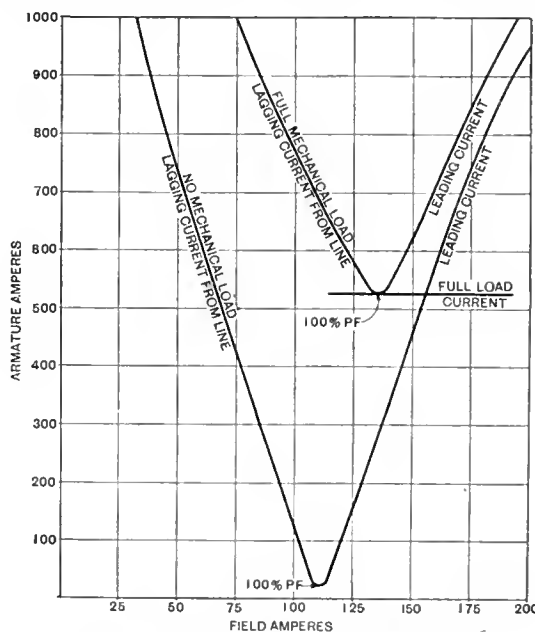


FIG. 63—V-CURVES OF A PHASE MODIFIER

values given in Table R may be of service for preliminary purposes.

"V" Curves—The familiar V curves shown in Fig. 63 serve to give some idea of the variation in field current for a certain phase modifier when operating between full load lagging and full load leading kv-a.* For this particular machine the excitation must be increased from 112 amperes at no load minimum input or unity power-factor to 155 amperes at full kv-a output leading or a range of 1.4 to 1 in. field excitation. For operation between full lagging and full leading, with no mechanical work done, the range of excitation is from 67 to 155 or 2.3 to 1.

Generators as Condensers—Ordinary alternators may be employed as synchronous condensers or synchronous motors by making proper changes in their field poles and windings to render them self-starting

*These curves have been reproduced from H. B. Dwight's book "Constant Voltage Transmission".

and safely insulated against voltages induced in the field when starting.

Where transmission lines feed into a city net work and a steam turbine generator station is available these generating units can serve as synchronous condensers by supplying just enough steam to supply their losses and keep the turbine cool. When operated in this way they make a reliable standby to take the important load quickly in case of trouble on a transmission line.

Location for Condensers—The nearer the center of load that the improvement in power-factor is made the better, as thereby the greatest gain in regulation, greatest saving in conductors and apparatus are made since distribution lines, transformers, transmission lines and generators will all be benefited.

How High to Raise the Power-Factor—Theoretically for most efficient results the system power factor should approach unity. The cost of synchronous apparatus having sufficient leading current capacity to raise the power-factor to unity increases so rapidly as unity is approached, as to make it uneconomical to carry the power-factor correction too high. Not only the cost but also the power loss chargeable to power-factor improvement mounts rapidly as higher power-factors are reached. This is for the reason that the reactive kv-a in the load corresponding to each percent change in power-factor is a maximum for power-factors near unity. It usually works out that it doesn't pay to raise the power factor above 90 to 95 percent, except in cases where the condenser is used for voltage control, rather than power-factor improvement.

DETERMINING THE CAPACITY OF SYNCHRONOUS MOTORS AND CONDENSERS FOR POWER-FACTOR IMPROVEMENT

A very simple and practical method for determining the capacity of synchronous condensers to improve the power-factor is by aid of cross section paper. A very desirable paper is ruled in inch squares, sub-ruled into 10 equal divisions. With such paper, no other equipment is required.

With a vector diagram it is astonishing how easy it is to demonstrate on cross section paper, the effect of any change in the circuit. A few typical cases are indicated in Fig. 64. These diagrams are all based upon an original circuit of 3000 kv-a at 70 percent power-factor lagging, shown by (1). It is laid off on the cross section paper as follows. The power of the circuit is 70 percent of 3000 or 2100 kw, which is laid off on line *AB*, by counting 21 sub-divisions, making each sub-division represent 100 kw or 100 kv-a. Now lay a strip of blank paper over the cross section paper and make two marks on one edge spaced 30 sub-divisions apart. This will then be the length of the line *AC*. This blank sheet is now laid over the cross section paper with one of the marks at the edge held at the point *A*. The other end of the paper is moved downward until the second mark falls directly below the point *B* thus locating point *C*. The length of the

line *BC* represents the lagging reactive kv-a in the circuit, in this case 2140 kv-a.

Diagram (2) shows the effect of adding a 1500 kv-a synchronous condenser to the original circuit. The full load loss of this condenser is assumed as 70 kw. The resulting kv-a and power-factor are determined as follows: Starting from the point *C* trace to the right a line 0.7 of a division long. This is parallel to the line *AB* for the reason that it is true power, so that there is now 2170 kw true energy. The black triangle represents the condenser, the line *CD*, 15 divisions long, representing the rating of the condenser. In this case, however, the vertical line is traced upward in place of downward, because the condenser kv-a is leading. This condenser results in decreasing the load from 3000 kv-a at 70 percent power-factor to 2275 kv-a at 95.4 percent power-factor. The line *AD* represents in magnitude and direction, the resulting kv-a in this circuit. The power-factor of the resulting circuit is the ratio of the true energy in kw to the kv-a or 95.4 percent, in this case. Since the line *AD* lays below the line *AB*, that is in the lagging direction, the power-factor is lagging.

Diagram (3) is the same as (2) except that the condenser is larger, being just large enough to neutralize all of the lagging component of the load, resulting in a final load of 2215 kw at 100 percent power-factor. Diagram (4) is similar to (3) except that a still larger condenser is shown. This condenser not only neutralizes all of the lagging kv-a of the load but in addition introduces sufficient leading kv-a into the circuit to give a leading resultant power-factor of 94 percent with an increase in kv-a of the resulting circuit from 2215 of (3) to 2400 kv-a of (4).

Diagram (5) illustrates the addition to the original circuit of a 100 percent power-factor synchronous motor of 600 hp. rating. As this motor has no leading or lagging component, there is no vertical projection. The power-factor of the circuit is raised from 70 to 77 percent as the result of the addition of 500 kw true power (load plus loss in motor) to the circuit. A resistance load would have this same effect.

Diagram (6) shows a 450 kw (600 hp.) synchronous motor of 625 kv-a input at 80 percent leading power-factor added to the original circuit. The input to this motor (including losses) is assumed to be 500 kw. The resulting load for the circuit is 3150 kv-a at 82.5 percent lagging power-factor.

The Diagram (7) shows an 850 kw, (1140 hp.) synchronous motor generator of 1666 kv-a input at 60 percent power-factor leading added to the original circuit. This gives a resulting load of 3200 kv-a at 96.9 percent lagging power-factor.

Diagram (8) shows the addition to the original circuit of the following loads, including losses.

- A 550 kw synchronous converter at 100 percent power-factor.
- A 650 kw induction motor at 70 percent lagging power-factor.
- A 500 kw synchronous motor.

The resultant load of this circuit is 3800 kw, and if a power-factor of 95 percent lagging is desired the total kv-a will be 4000. The line AD may be located by a piece of marked paper and the capacity of the necessary synchronous motor scaled off. This is found to be 1650 kv-a at 30.3 percent power-factor.

The Circle Diagram—The circle diagram in Fig. 65 shows the fundamental relations between true kw, reactive kv-a and apparent kv-a corresponding to different power-factors, the values upon the chart being read to any desired scale to suit the numerical values of the problem under consideration. This diagram is suffi-

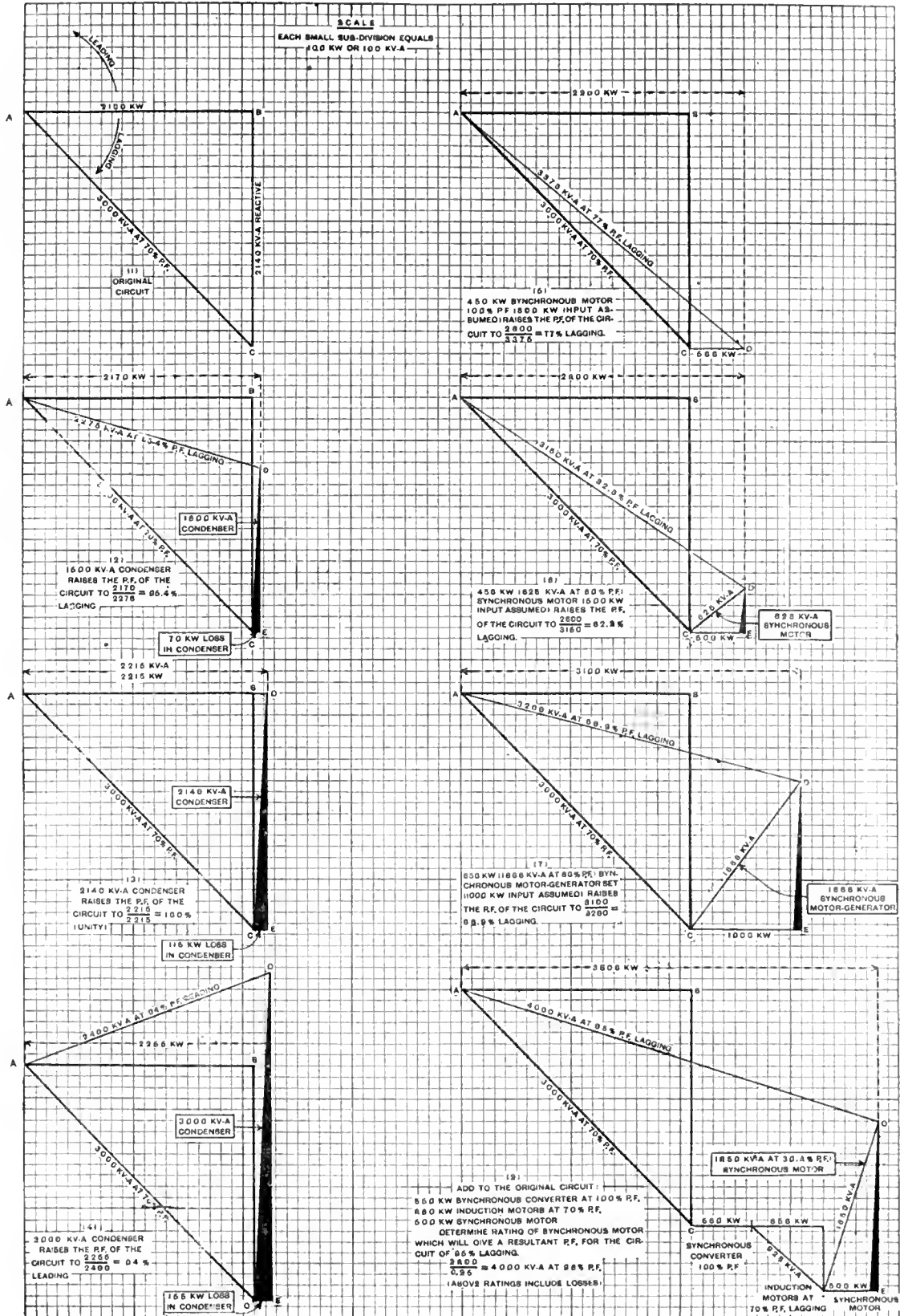


FIG. 64—EXAMPLES IN POWER-FACTOR IMPROVEMENT

ciently accurate for ordinary power-factor problems. In place of drawing out the vector diagrams as just explained they are traced out with a pin point on the circle diagram.

Assume again a load of 2100 kw at 70 percent power-factor lagging, and that the power-factor is to be raised to 95.4 percent as in (2) of Fig. 64, and that the loss in the condenser necessary to accomplish this is again taken as 70 kw. The capacity of the synchronous condenser may be traced on the circle diagram as follows: From the true power load of 2100 kw (top horizontal line) follow vertically downward

of the condenser would be the hypotenuse rather than the vertical projection. The error in assuming the vertical projection as the rating of the condenser is negligible unless the condenser furnishes mechanical power, in which case the hypotenuse should be marked on a separate strip of paper and its length determined from the kv-a scale.

ADVANTAGE OF HIGH POWER-FACTOR

Less Capacity Installed—Low power-factors demand larger generators, exciters, transformers, switching equipment and conductors. Loads of 70 percent

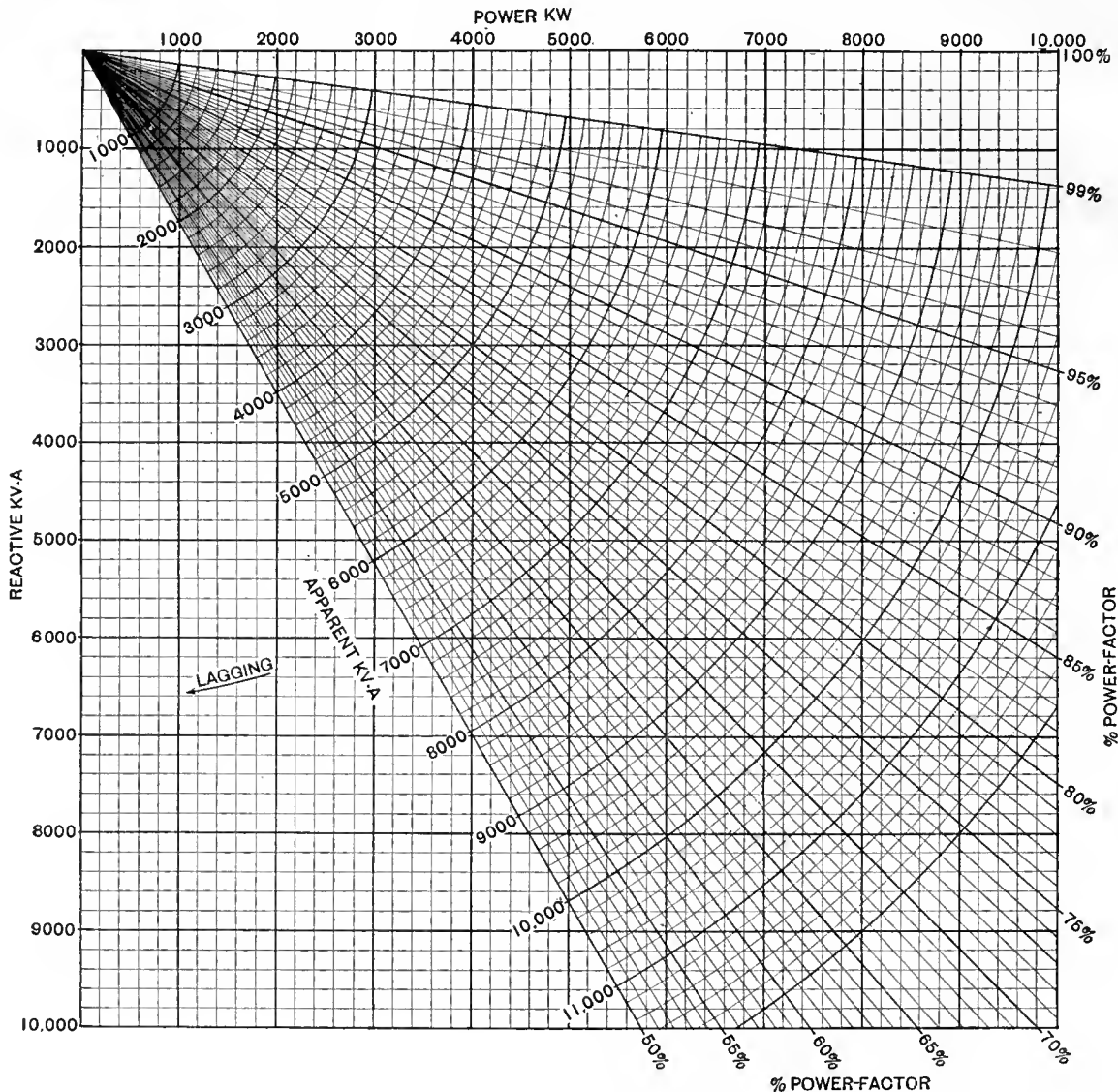


FIG. 65—RELATION BETWEEN ENERGY LOAD, APPARENT LOAD AND REACTIVE KV-A FOR DIFFERENT POWER FACTORS

until the diagonal line representing 70 percent power-factor is reached. This is opposite 2140 kv-a reactive component. From the point thus obtained, go horizontally to the right a distance representing 70 kw power. From this point go vertically upward until the diagonal line representing 95.4 percent power-factor is reached. Then read the amount of reactive kv-a (640) corresponding to this last point. The original lagging component of 2140—640=1500 kv-a which is approximately the capacity of the condenser necessary to accomplish the above results. Actually the rating

power-factor demand equipment of 28 percent greater capacity than would be required if the power-factor were 90 percent. The cost of apparatus for operation at 70 percent power-factor would be approximately 15 percent greater than the cost of similar apparatus for 90 percent power-factor operation, since the capacity of apparatus to supply a certain amount of energy is inversely proportional to the power-factor.

Higher Efficiency—Assume that the power-factor of a 1000 kv-a (700 kw at 70 percent power-factor) transmission circuit is raised to 90 percent. As the cop-

per loss varies as the square of the current, raising the power-factor reduces the copper loss approximately 40 percent. If we assume an efficiency for the generator of 93 percent (one percent copper loss); for combined raising and lowering transformers 94 percent (three percent copper loss) and for the transmission line 92 percent, the saving in copper loss corresponding to 90 percent power-factor operation would be as follows:

Generators	0.4 percent
Transformers	1.2 percent
Transmission line	3.2 percent
Total	4.8 percent or approximately 33 kw.

To raise the power-factor to 90 percent would require a synchronous condenser of 375 kv-a capacity. This size condenser would have a total loss of about 30 kw, resulting in a net gain in loss reduction of three kw. Against this gain would be chargeable, the interest and depreciation of the condenser cost with its accessories, also any cost of attendance which there might be in connection with its operation. It is evident that in this case it would not pay to install a condenser if increased efficiency were the only motive.

TABLE S—COST OF POWER-FACTOR CORRECTION WITH SYNCHRONOUS MOTORS

Syn. Motor Kv-a	Motor Will Furnish		Chargeable to Power-Factor Correction	
	Mech. Kw	Leading Kv-a	Loss Kw	Difference in Price
140	100	100	1.6	\$500.00
280	200	200	2.5	500.00
420	300	300	5.0	500.00
700	500	500	8.0	800.00
1050	750	750	9.0	1000.00
1400	1000	1000	14.0	1200.00

The improvement in power-factor can be more cheaply and efficiently obtained by the installation of one or more synchronous motors designed for operation at leading power-factor. Sufficient capacity of these will give, in addition to mechanical load, sufficient leading current to raise the power-factor to 90 percent. The extra expense and increased loss of synchronous motors enough larger to furnish the necessary leading component for power-factor correction is very small. Table S gives in a very approximate way, some idea of the amount of loss and proportional cost of synchronous motors chargeable to power-factor improvement when delivering both mechanical power and leading current.

Thus if a synchronous condenser is used on the above circuit there is a loss of 30 kw, chargeable to power-factor improvement, whereas if a synchronous motor of sufficient capacity (530 kv-a) to give 375 kw mechanical work and at the same time the necessary 375 kv-a leading current for power-factor improvement, the extra loss chargeable to power-factor improvement would be something like six kw. The increased cost of a synchronous motor to furnish 375 kv-a leading current in addition to 375 kw power would be about \$600 whereas the cost of a 375 kv-a

condenser would be in the neighborhood of \$4000. Varying costs and designs make cost and loss values unreliable. They are given here only to illustrate the points which should be considered when considering synchronous motors vs synchronous condensers.

Improved Voltage Regulation—The voltage drop under load for generators, transformers and transmission lines rapidly increases as the power-factor goes down. Table T gives an idea of the variation in voltage drop corresponding to various power-factors at 60 cycles.

Automatic voltage regulation may be used to hold the voltage constant at the generators or at some other point, but it cannot prevent voltage changes at all points of the system.

Increased Plant Capacity—The earlier alternators were designed for operation at 100 percent power-factor with prime movers, boilers, etc. installed on the same basis. Increasing induction motor loads have resulted in power-factors of 70 and 80 percent. As a result, some of the older generating stations are being operated with prime movers, boilers etc. underloaded because the 100 percent power-factor generators which

TABLE T—EFFECT OF POWER-FACTOR ON VOLTAGE DROP

Percent Power-Factor	100	90	80	70
Generators *(older design)	8.0	-	25.0	-
Transformers	1.2	4.1	4.9	5.5
Transmission line	7.9	13.0	14.2	15.2

they drive limit the amount of power that can be generated without endangering the generator windings. This condition some times makes it necessary to operate three units, where two might be sufficient to carry the load at unity power-factor. The shutting down of a unit would result in a considerable saving in steam consumption. A recent case came up of a transmission line 30 miles long, fed at each end by a small generating station. On account of heavy line drop it was necessary to operate both stations to furnish the comparatively light night load. Investigation developed that by installing a synchronous condenser at one of these terminal stations for reducing the voltage drop in the line, one generating station could be shut down during the night, thereby resulting in a very large annual saving in coal and labor bills.

A station may have some generating units designed for 100 percent power-factor and other units designed for 80 percent power-factor; or again, where two generating stations feed into the same transmission system, one may have 100 percent power-factor generating units and the other 80 percent power-factor

*The present-day design of maximum rated generators with a short-circuit ratio of about unity will barely circulate full-load current with normal no-load excitation. Under such conditions the terminal voltage would be practically zero regardless of the power-factor.

generating units. In such cases, the field strength of the generators may be so adjusted as to cause the 80 percent power-factor units to take all the lagging current, thus permitting the 100 percent power-factor units to be loaded to their full kw rating.

BEHAVIOR OF A. C. GENERATORS WHEN CHARGING A TRANSMISSION LINE*

It has been shown above how leading armature current, by increasing the field strength, causes an increase in the voltage induced in the armature of an alternator and consequently an increase in its terminal voltage. It was also shown that the terminal voltage is further increased as result of the voltage due to self induction adding vectorially to the voltage induced in the armature.

If an alternator with its fields open is switched onto a dead transmission line having certain electrical characteristics, it will become self exciting, provided there is sufficient residual magnetism present to start the phenomenon. In such case, the residual magnetism in the fields of the generator will cause a low voltage to be generated which will cause a leading line charging current to flow through the armature. This leading current will increase the field flux which in turn will increase the voltage, causing still more charging current to flow, which in turn will still further increase the line voltage. This building up will continue until stopped by saturation of the generator fields. This is the point of stable operation. Whether or not a particular generator becomes self exciting when placed upon a dead transmission line depends upon the relative slope of the generator and line characteristics.

In Fig. 66 are shown two curves for a single 45 000 kv-a, 11 000 volt generator, the charging current of the transmission line being plotted against generator terminal voltage. One curve corresponds to zero excitation, the other curve to 26.6 percent of normal excitation. A similar pair of curves correspond to two duplicate generators in parallel**. The straight line representing the volt-ampere characteristics of the transmission line fed by these generators corresponds to a 220 kv, 60 cycle, three-phase transmission circuit, 225 miles long, requiring 69 000 kv-a to charge it with the line open at the receiving end.

The volt-ampere charging characteristic of a transmission line is a straight line, that is, the charging current is directly proportional to the line voltage. On the other hand the exciting volt-ampere characteristic for the armature has the general slope of an ordinary saturation curve.

*For a more detailed discussion of this subject see the following articles:—"Characteristics of Alternators when Excited by Armature Currents" by F. T. Hague, in the JOURNAL for Aug. 1915; "The Behavior of Alternators with Zero Power-Factor Leading Current" by F. D. Newbury, in the JOURNAL for Sept. 1918; "The Behavior of A. C. Generators when Charging a Transmission Line" by W. O. Morris, in the *General Electric Review* for Feb. 1920.

**It is assumed that with the assumed field current such generators can be synchronized and held together during the process of charging the line.

If the alternator characteristic lie above the line characteristic at a point corresponding to a certain charging current the leading charging current will cause a higher armature terminal voltage than is required to produce that current on the line. As a result the current and voltage will continue to rise until, on account of saturation, the alternator characteristic falls until it crosses the line characteristic. At this point the voltage of the generator and that of the line are the same for the corresponding current. If on the other hand the alternator characteristic falls below the line characteristic the alternator will not build up without permanent excitation.

As stated previously, whether or not a generator becomes self-exciting when connected to a dead transmission line depends upon the relative slopes of generator and transmission line characteristics. The relative slopes of these curves depend upon:—

- a—The magnitude of the line charging current.
- b—The rating of the generators compared to the full voltage charging kv-a of the line.
- c—The armature reaction. High armature reaction, (that is low short-circuit ratio) favors self-excitation of the generators.
- d—The armature reactance. High armature reactance also favors self-excitation of the generators.

Methods of Exciting Transmission Lines—If the relative characteristics of an alternator and line are such as to cause the alternator to be self-exciting, this condition may be overcome by employing two or more

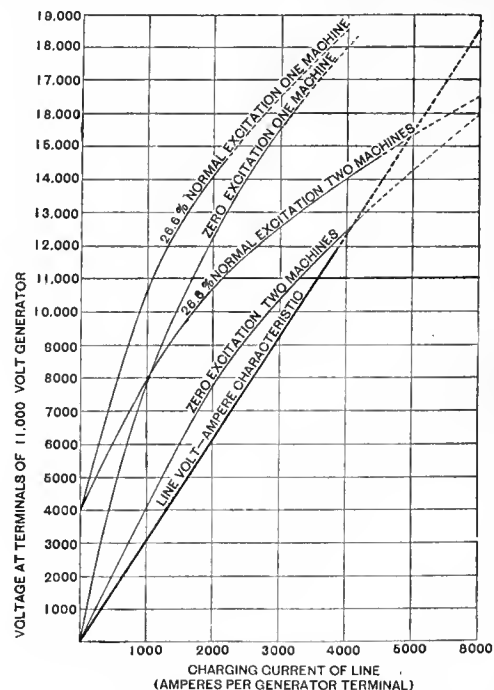


FIG. 66—VOLT AMPERE CHARACTERISTICS OF ONE 45 000 KV-A, 11 000 VOLT GENERATOR; TWO DUPLICATE 45 000 KV-A GENERATORS; AND A THREE-PHASE, SINGLE-CIRCUIT, 220 KV TRANSMISSION LINE

alternators (provided they are available for this purpose) to charge the transmission line. The combined characteristics of two or more alternators may be such as to fall under the line characteristic, in which case the alternator will not be self-exciting. In such case, the alternators could be brought up to normal speed, and given sufficient field charge to enable them to be

synchronized and held in step, after which they could be connected to the dead transmission line and their voltage raised to normal.

Generators as normally designed will carry approximately 40 percent of their rated current at zero leading power-factor. If more than this current is demanded of them they are likely to become unstable in operation. By modifying the design of normal alternators so as to give low armature reaction, they may be made to carry a greater percentage of leading current. If the special design is such that with zero

Fig. 66, and there were sufficient residual magnetism to start the phenomenon, the generator voltage would rise to approximately double normal value before the point of staple operation is reached. If, however, two generators having 26.6 percent of normal excitation were paralleled and connected to this circuit, a point of staple operation would be reached at a terminal voltage of approximately 15 500 volts. Actually stable operation would be reached at a somewhat less terminal voltage for the reason that the line would probably not be open at the receiving end, but

would probably have the lowering transformers connected to it. In such case the magnetizing current required for lowering transformers would lower the receiving end voltage, resulting in less line charging current.

In either case the curves of Fig. 66 show that either more than two generators will be required to charge the line when unloaded, or some other method of charging must be resorted to. Reactance coils could be used at the receiving end to furnish lagging current for neutralizing some of the line charging current,

but there might be difficulty in removing these from the circuit when the line is fully charged. At the present time it is expected that the problem of charging long transmission lines may usually be solved by starting one or more generators with sufficient field strength to permit them to be synchronized and held in step. One or more phase modifiers with under-excited fields may then be connected to the line at the receiving end and brought up to normal speed with the generators. Such a method of solving this problem has been employed by the Southern California Edison Company.

TABLE U—INSTALLATIONS OF LARGE PHASE MODIFIERS (1921)
By American Manufacturers

Kv-a	R.P.M.	Volts	Cycles	No. of Units	Date of Order	NAME AND LOCATION
30 000	600	6600	50	1	1919	So. Cal. Ed. Co., Los Angeles, Cal.
20 000	600	11 000	60	2	1921	Pacific Gas & Elec.
15 000	375	6600	50		1912	Southern Cal. Ed. Co., Los Ang., Cal.
15 000	375	6600	50	1	1912	Pacific Lt. & Pr. Co.
12 500	500	22 000	50	2	1918	Andhra Valley, India
7500	400	6600	60	2	1913	Utah Pr. & Lt. Co., Salt Lake, Utah
7500	400	6600	60	2	1916	Canton El. Co., Canton, Ohio
7500	600	13 800	60	1	1917	Blackstone Valley Gas & Elec. Co., Pawtucket, R. I.
7500	600	13 800	60	1	1917	New England Pr. Co., Worcester, Mass.
7500	720	13 800	60	1	1918	New England Pr. Co., Fitchburg, Mass.
7500	800	11 500	40	1	1918	Adirondack El. Pwr. Corp., Watervliet, New York
7500	750	11 000	50	1	1919	Energia Electrica de Cataluna, Barcelona, Spain
7500	600	11 000	60	1	1920	Duquesne Light Co.
7500	600	1200	60	2	1918	J. G. White, Engineers
7500	600	11 000	60	1	1918	Duquesne Light Co.
7500	600	11 000	60	1	1916	Duquesne Light Co.
7500	600	11 000	60	2	1917	Duquesne Light Co.
6500	750	2200	50	1	1917	Shanghai Municipal Council, Shanghai, China
6000	600	16 500	50	1	1914	So. Cal. Ed. Co., Los Angeles, Cal.
5000	600	7200	60	1	1916	Pac. Pwr. & Lt., Kennewick, Wash.
5000	500	6600	50	2	1915	Tata Hydro El. Pr. & S. Co., India
5000	750	6600	50	3	1917	Ebro Irrigation & Pr. Co., Barcelona, Spain
5000	750	11 500	50	1	1919	Societa Lombarda Distribuziona Energia Elettrics, Italy
5000	600	2300	60	1	1918	Turnbull Steel Co., Warren, Ohio
5000	720	2300/ 4000	60		1921	Public Service of N. Ill.
5000	720	11 000	60	1	1921	Tskata & Co., Japan.
5000	600	13 200	60	1	1919	Conn. Lt. & Pr. Co.

voltage field excitation when carrying half the line charging kv-a, the armature voltage will not exceed 70 percent of normal, this reduced voltage will result in a line charging kv-a of half of normal value. Specially designed alternators usually result in larger and more costly machines and the gain resulting in the special design is usually not sufficient to warrant the extra cost.

If a single generator with its field circuit open were connected to a dead transmission circuit such as the one whose volt-ampere characteristics are shown in

CHAPTER XV

PHASE MODIFIERS FOR VOLTAGE CONTROL

WITH alternating-current transmission there is a voltage drop resulting from the resistance of the conductors, which is in phase with the current. In addition there is a reactance voltage drop; that is a voltage of self-induction generated within the conductors which varies with and is proportional to the current, and may add to or decrease the line voltage. If the line is long, the frequency high or the amount of power transmitted large, this induced voltage will be large, influencing greatly the line drop. By employment of phase modifiers the phase or direction of this induced voltage may be controlled so that it will be exerted in a direction that will result in the desired sending end voltage.

A certain amount of self-induction in a transmission circuit is an advantage, allowing the voltage at the receiving end to be held constant under changes in load by means of phase modifiers. It may even be made to reduce the line voltage drop to zero, so that the voltage at the two ends of the line is the same for all loads. Self-induction also reduces the amount of current which can flow in case of short-circuits, thus tending to reduce mechanical strains on the generator and transformer windings, and making it easier for circuit breaking devices to function successfully. On the other hand, high self-induction reduces the amount of power which may be transmitted over a line and may, in case of lines of extreme length, make it necessary to adopt a lower frequency. It also increases the capacity of phase modifiers necessary for voltage control. High reactance also increases the surge over-voltage that a given disturbance will set up in the system.

On the long lines, the effect of the distributed leading charging current flowing back through the line inductance is to cause, at light loads, a rise in voltage from generating to receiving end. At heavy loads, the lagging component in the load is usually sufficient to reverse the low-load condition; so that a drop in voltage occurs from generating to receiving end. The charging current of the line is, to a considerable extent, an advantage; for it partially neutralizes the lagging component in the load, thus raising the power-factor of the system and reducing the capacity of synchronous condensers necessary for voltage control.

The voltage at the receiving end of the line should be held constant under all loads. To partially meet this condition, the voltage of the generators could be varied to a small extent. On the longer lines, however, the voltage range required of the generators would be too great to permit regulation in this

manner. In such cases, phase modifiers operating in parallel with the load are employed. The function of phase modifiers is to rotate the phase of the current at the receiving end of the line so that the self-induced voltage of the line (always displaced 90 degrees from the current) swings around in the direction which will result in the desired line drop. In some cases a phase modifier is employed which has sufficient capacity not only to neutralize the lagging component at full load, but, in addition, to draw sufficient leading current from the circuit to compensate entirely for the ohmic and reactance voltage drops of the circuit. In this case, the voltage at the two ends of the line may be held the same for all loads. This is usually accomplished by employing an automatic voltage regulator which operates on the exciter fields of the phase modifier. The voltage regulator may, if desired, be arranged to compound the substation bus voltage with increasing load.

CHECKING THE WORK

A most desirable method of determining line performance is by means of a drawing board and an engineer's scale. A vector diagram of the circuit under investigation, with all quantities drawn to scale, greatly simplifies the problem. Each quantity is thus represented in its true relative proportion, so that the result of a change in magnitude of any of the quantities may readily be visualized. Graphical solutions are more readily performed, and with less likelihood of serious error than are mathematical solutions. The accuracy attainable when vector diagrams are drawn 20 to 25 inches long and accurate triangles, T squares, straight edges and protractors are employed is well within practical requirements. Even the so-termed "complete solution" may be performed, graphically with ease and accuracy. A very desirable virtue of the graphical solution which follows is that it exactly parallels the fundamental, mathematical solution. For this reason this graphical solution is most helpful even when the fundamental mathematical solution is used, for it furnishes a simple check against serious errors. The result may be checked graphically after each individual mathematical operation by drawing a vector in the diagram paralleling the mathematical operation. Thus, any serious error in the mathematical solution may be detected as soon as made.*

*A method of checking arithmetical operations which requires little time and is an almost sure preventative of errors is that known as "casting out the nines." This method is given in most older arithmetics but has been dropped from many of the modern ones. A complete discussion is given in Robinson's "New Practical Arithmetic" published by The American Book Company.

When converting a complex quantity mathematically from polar to rectangular co-ordinates, or vice versa, the results may readily be checked by tracing the complex quantity on cross-section paper and measuring the ordinates and polar angle, or for approximate work the conversion may be made graphically to a large scale. For instance, in using hyperbolic functions, polar values will be required for obtaining powers and roots of the complex quantity. For approximate work much time will be saved by obtaining the polar values graphically.

In the graphical solution of line performance it will usually be desirable to check the line loss by a mathematical solution in cases which require exact loss values. Since the line loss may be five percent or less of the energy transmitted, a small error in the overall results might correspond to a large error in the value of the line loss.

EFFECT OF TRANSFORMERS IN THE CIRCUIT

Usually long transmission circuits have transformers installed at both ends of the circuit and one or more phase modifiers in parallel with the load. Such a transmission circuit must transmit the power loss of the phase modifiers and of the receiver transformers. In addition to this power loss, a lagging reactive current is required to magnetize the transformer iron. A complete solution of such a composite circuit (generator to load) requires that the losses of the phase modifiers and transformers be added vectorially to the load at the point where they occur so that their complete effect may be included in the calculation of the performance of the circuit. A complete solution also requires that three separate solutions be made for such a circuit.* First with the known or assumed conditions at the load side of the lowering transformers the corresponding electrical conditions at the high voltage side of the transformers is determined by the usual short line impedance methods. With the electrical conditions at the receiving end of the high-tension line thus determined, the electrical conditions at the sending end of the line are determined by one of the various methods which take into account the distributed quantities of the circuit. With the electrical condition at the sending end thus determined the electrical conditions at the generating side of the raising transformers are determined. The above complete method of procedure, is tedious if carried out mathematically, but if carried out graphically is comparatively simple.

It is the general practice to neglect the effect of condenser and lowering transformer loss in traveling over the line, but to add this loss to the loss in the high-tension line after the performance has been calculated. If the loss in condensers and lowering transformers is five percent of the power transmitted the

error in the calculated results would probably be less than 0.5 percent, a rather small amount.

In order to simplify calculations, it is the general practice to consider the lumped transformer impedance as though it were distributed line impedance by adding it to the linear constants of the line and then proceeding with the calculations as though there were no transformers in the circuit. This simplifies the solution but at the expense of accuracy, particularly if the line is very long, the frequency high or the ratio transformer to line impedance high. This simplified solution introduces maximum errors of less than two percent in the results for a 225 mile, 60-cycle line.

It has been quite general practice to disregard the effect of the magnetizing current consumed by transformers. The magnetizing current required to excite transformers containing the older transformer iron was about two percent and therefore its effect could generally be ignored. Later designs of transformers employ silicon steel, and their exciting current varies from about 20 percent for the smaller of distribution type transformers, to about 12 percent on transformers of 100 kv-a capacity and about five percent for the very largest capacity transformers. The average magnetizing current for power transformers is between six and eight percent. This magnetizing current is important for the reason that it is practically in opposition to the current of over-excited phase modifiers used to vary the power-factor. If in a line having 100 000 kv-a transformer capacity at the receiving end, the magnetizing current is five percent, there will be a 5000 kv-a lagging component. If the capacity of phase modifiers required to maintain the proper voltage drop under this load is 50 000 kv-a the lagging magnetizing component of 5000 kv-a will subtract this amount from the effective rating of the phase modifiers, with a resulting error of ten percent in the capacity of the phase modifiers required.

In the diagrams and calculations which follow, the transformer leakage, consisting of an in-phase component of current (iron loss) and a reactive lagging component of current (magnetizing current), is considered as taking place at the low-tension side of the transformers. A more nearly correct location would be to consider the leak as at the middle of the transformer, that is, to place half the transformer impedance on each side of the leak. To solve such a solution it would be necessary to solve two complete impedance diagrams for the transformers at each end of the circuit. The gain in accuracy of results would not, for power transmission lines, warrant the increased arithmetical work and complication necessary.

In the case of lowering transformers, it would seem that the magnetizing current would be supplied principally from synchronous machines connected to the load. If phase modifiers are located near the lowering transformers, the transformers would probably draw most of their magnetizing current from

*A method for calculating a transmission line with transformers at each end in one solution is given in the articles by Messrs. Evans and Sels in the JOURNAL for July, August, September, *et seq.* 1921.

them rather than from the generators at the distant end of the line. Partly for this reason, but more particularly for simplicity, the leak of the lowering transformers will be considered as taking place at the load side of the transformers. On this basis we first

current also from the low side; that is from the generators. Both the complete and the approximate methods of solving long line problems which follow, include the effect of not only the magnetizing current consumed by the transformers, but also the losses in

TABLE V—COMPARISON OF RESULTS AS OBTAINED BY FIVE DIFFERENT METHODS OF CALCULATIONS

75,000 KW (88,236 KVA AT 85% PF) 3 PHASE, 50 CYCLES RECEIVER VOLTAGE HELD CONSTANT AT 220 KV. 50,000 KVA CONDENSER AT RECEIVING END
LENGTH OF TRANSMISSION 226 MILES ALL TABULATED VALUES REFERRED TO NEUTRAL

AREA IN CIRCULAR MILS	METHOD	RECEIVING END TO NEUTRAL						SENDING END TO NEUTRAL						LOSSES IN KW TO NEUTRAL												
		LOW TENSION SIDE OF TRANSFORMERS			HIGH TENSION SIDE OF TRANSFORMERS			HIGH TENSION SIDE OF TRANSFORMERS			LOW TENSION SIDE OF TRANSFORMERS			LOWERING TRANSFORMERS		CONDENSER		RAISING TRANSFORMERS		TOTAL LOSS						
		VOLTS	AMPS	PF	VOLTS	AMPS	PF	VOLTAGE		CURRENT	PF _s	VOLTAGE		CURRENT	PF _{GEN}	IRON	COPPER	CONDENSER	KW _N	LOSS IN % OF KW _L	IRON	COPPER	KW _N	LOSS IN % OF KW _L		
		E _{LN}	I _L	LEAD	E _{RN}	I _R	LAG	E _{SN}	%	I _S	%	LEAD	E _{GEN-N}	%	I _{GEN}	%	LEAD	"	"	"	"	"	"	"	"	
605000	A	127020	2023	99.90	127556	2049	99.63	129090	100	2278	100	93.77	126920	100	2261	100	97.49	23.5	130	666	1542	6.16	23.5	165	3973	11.97
"	B	"	"	"	"	"	"	124247	963	2365	103.9	93.35	124657	98.2	2323	102.8	95.14	"	"	"	1634	6.53	"	178	3078	12.31
"	C	"	"	"	"	"	"	126783	98.4	2287	100.4	94.32	127537	100.5	2246	99.3	95.87	"	"	"	1383	6.33	"	172	3021	12.08
"	D	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1553	6.21	"	166	2985	11.94
"	E	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1510	6.04	"	160	2936	11.74
715500	A	"	"	"	127556	2049	99.63	127911	100	2287	100	93.50	125668	100	2264	100	97.36	"	"	"	1320	5.98	"	166	2753	11.01
"	B	"	"	"	"	"	"	123041	962	2375	103.9	93.09	123409	98.2	2331	102.8	94.93	"	"	"	1408	5.63	"	180	2854	11.42
"	C	"	"	"	"	"	"	125576	98.2	2277	100.4	94.09	126292	100.5	2254	99.5	95.67	"	"	"	1338	5.35	"	173	2777	11.11
"	D	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1347	5.40	"	168	2738	11.13
"	E	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1273	5.09	"	162	2701	10.90
795000	A	"	"	"	127556	2049	99.63	127196	100	2293	100	93.33	124909	100	2274	100	97.24	"	"	"	1192	4.77	"	167	2625	10.50
"	B	"	"	"	"	"	"	122313	962	2384	103.8	92.94	122648	98.2	233.5	102.7	94.80	"	"	"	1260	5.04	"	181	2707	10.83
"	C	"	"	"	"	"	"	124846	98.1	2307	100.4	93.94	125532	100.5	226.8	99.3	95.55	"	"	"	1177	4.71	"	174	2617	10.47
"	D	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1198	4.79	"	169	2639	10.53
"	E	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1126	4.50	"	162	2589	10.39
954000	A	"	"	"	127556	2049	99.63	126132	100	2304	100	92.93	123740	100	228.4	100	96.99	"	"	"	976	3.90	"	169	2411	9.67
"	B	"	"	"	"	"	"	121212	961	2394	103.9	92.55	121488	98.2	235.1	102.9	94.51	"	"	"	1059	4.28	"	183	2608	10.03
"	C	"	"	"	"	"	"	123737	98.1	231.5	100.5	93.58	124369	100.5	227.3	99.5	95.31	"	"	"	1020	4.08	"	177	2463	9.85
"	D	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	1014	4.05	"	170	2459	9.80
"	E	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	"	984	3.93	"	165	2415	9.66

*A—Transformer impedances treated as lumped at the ends of the line. This is the most nearly accurate of the five methods. It is referred to in the text as the complete solution.

B—This assumes the impedance of the lowering transformers as line impedance. It takes no account of the leakage of the lowering transformers.

C—This assumes the impedance of both lowering and raising transformers as line impedance—It takes no account of the leakage of the lowering and raising transformers.

D—This is the same as B except that the leakage of the lowering transformers has been added to the load—It is referred to in the text as the approximate solution.

E—This is the same as C except that the leakage of the lowering transformers has been added to the load.

have a load current expressed in rectangular coordinates with the load voltage as a temporary vector of reference. To this we add algebraically a phase modifier current (loss + j or leading) and to this we add the transformer leakage (loss - j or lagging). In other words, these three components of current at the receiving end of the line add up algebraically upon a

transformers and phase modifiers flowing over the line.

For the purpose of determining the magnitude of errors in the calculated results corresponding to simplified methods of calculation where transformers are required at both ends of the line, the calculations shown in Table V were made. Five methods of calculations were made for each of four sizes of cable. A con-

TABLE W—PERCENTAGE ERRORS IN RESULTS, AS DETERMINED BY VARIOUS METHODS OF CALCULATION.

These methods do not take complete account of the effects of the transformers in the circuit

Method	At Generator Percent Error			At Sending End Percent Error			Line Loss Percent Error	Transformer Account
	E _{gen}	I _{gen}	PF _{gen}	E _s	I _s	PF _s		
A	0	0	0	0	0	0	0	Complete method—Assumed for comparison as resulting in 100 percent values.
B	-3.7	+3.9	-0.42	+0.37	Leak of lowering transformers ignored. Impedance of lowering transformers assumed as line impedance.
C	-1.8	+2.8	-2.35	+0.17	Leaks of raising and lowering transformers ignored. Impedance of raising and lowering transformer assumed as line impedance.
D	-1.6	+0.4	+0.55	+0.05	Same as B except that the transformer leak has been added to the load.
E	+0.5	-0.7	-1.62	-0.12	Same as C except that the transformer leak has been added to the load.

common vector of reference, thus making it very easy to obtain the resulting load at the receiving end of the line.

The transformers at the sending end of the line have been considered as receiving their magnetizing

current also from the low side; that is from the generators. Both the complete and the approximate methods of solving long line problems which follow, include the effect of not only the magnetizing current consumed by the transformers, but also the losses in

pedance as line impedance, gives the sending end voltage too low by 3.7 percent and the current too high by 3.9 percent.

Table X contains approximate data upon transformers of various capacities 25 and 60 cycles. Since such data will vary greatly for different voltages it must be considered as very approximate but may be found useful in the absence of specific data for the problem at hand.

Fig. 67 shows complete current and voltage diagrams for both short and long lines. The diagram illustrating short lines is based upon the current having the same value and direction at all points of the circuit. On this basis the IR drops of the line and of the raising and lowering transformers will be in the same direction. Likewise their individual IX drops will also be in the same direction. It is evident, therefore, that, for short lines where the capacitance

voltage circuit in order to combine properly with the linear constants of the line. Although all calculations are made in terms of the high-voltage circuit the results may, if desired, be converted to terms of the low voltage circuit, by applying the ratio of transformation.

The transformer impedance to neutral is one-third the equivalent single-phase value. The reason for this is that the I^2R and I^2X for one phase is identical whether to neutral or between phases. Since the current between phases is equal to the current to neutral divided by $\sqrt{3}$, the square of the phase current would be one-third the square of the current to neutral; therefore, R and X to neutral will be one-third the phase values. Another way of looking at this is that the resistance and reactance ohms vary with the square of the voltage, and since the phase voltage is $\sqrt{3}$ times the voltage to neutral, the phase resistance and phase reactance would be three times that to neutral. In

TABLE X—APPROXIMATION OF RESISTANCE AND REACTANCE VOLTS, OF IRON AND COPPER LOSSES AND OF MAGNETIZING CURRENT FOR TRANSFORMERS OF VARIOUS CAPACITIES

Capacity of Transformer KV-A	60 CYCLES PER SECOND					25 CYCLES PER SECOND				
	Percent *Resistance	Percent *Reactance	Percent Loss		Percent Magnetizing Current	Percent Resistance	Percent Reactance	Percent Loss		Percent Magnetizing Current
			Iron	Copper				Iron	Copper	
200	1.5	5.5	1.4	1.5	10	2.6	4.0	1.1	2.6	10
300	1.3	5.6	1.3	1.3	9	2.15	4.0	1.0	2.15	10
500	1.2	6.0	1.2	1.2	8	1.85	4.1	1.0	1.85	9
750	1.1	6.3	1.0	1.1	8	1.65	4.2	0.9	1.65	9
1000	1.1	6.5	0.9	1.1	7	1.55	6.0	0.8	1.55	8
1500	0.9	7.0	0.8	0.9	6	1.4	6.2	0.8	1.4	8
2000	0.8	7.0	0.7	0.8	6	1.3	6.4	0.7	1.3	8
3000	0.75	7.0	0.7	0.75	6	1.2	6.8	0.6	1.2	7
5000	0.65	7.0	0.6	0.65	6	1.1	7.2	0.5	1.1	7
7500	0.6	8.0	0.6	0.6	5	1.0	7.8	0.5	1.0	7
10000	0.6	8.9	0.5	0.6	5	1.0	8.0	0.5	1.0	6
15000	0.55	8.5	0.5	0.55	5	0.95	8.0	0.6	0.95	6
25000	0.5	9.0	0.6	0.5	5	0.9	8.0	0.6	0.9	6
35000	0.5	9.5	0.6	0.5	5	0.9	9.0	0.6	0.9	6
50000	0.5	10.0	0.6	0.5	5	0.9	9.0	0.6	0.9	6

*The actual ohms resistance and ohms reactance will vary as the square of the voltage. The values in above table must be considered as only roughly approximate. They will vary materially with transformers wound for different voltages

is negligible, the transformer impedance may be added directly to the line impedance, provided the electrical characteristics on the high-tension side of the transformers are not required.

As the line becomes longer, the current changes in both amount and direction from point to point, as a result of the superimposed distributed charging current of the line. The result of this is that the impedance triangles of the line and of lowering and raising transformers change in both size and relative position; so that their individual impedances can no longer be added together and considered as all line impedance, without accepting an error in the results thus obtained. The complete diagram for long lines shown by Fig. 67 will be considered later.

TRANSFORMER IMPEDANCE TO NEUTRAL*

Transformer constants are referred to the high

calculating the impedance to neutral, the results will be the same whether star or delta connection is used.

Even if the transformers at both ends of the transmission line are duplicates their impedance will not be the same if operated on different taps of the windings to accommodate different voltages. In such cases, their impedances will vary as the square of the voltages. For instance, if they are operated at 220 and 230 kv at the receiving and sending end respectively, then their impedances will have the relation of $\frac{220^2}{230^2} = 0.915$. In other words, if the resistance and reactance of the receiving end transformers is 3.185 and 39.82 ohms respectively, the sending end transformers will have resistances and reactances of 3.481 and 43.52 ohms respectively; provided transformer taps corresponding to this higher voltage are used.

The impedance in ohms of an 18 000 kv-a, three-phase, or of three 6000 kv-a single-phase transformers, connected in a bank, may be determined as fol-

*The writer desires to express his appreciation of helpful assistance and useful data on transformer characteristics received from Mr. J. F. Peters.

flows. Assume that they are operated at 104 000 volts between conductors (60 046 to neutral) and that the resistance voltage is 1.04 percent and reactance voltage is 4.80 percent.

The single-phase values are:—

$$\frac{6000000}{104000} = 57.7 \text{ amperes}$$

$$R_t = \frac{104000 \times 0.0104}{57.7} = 18.75 \text{ ohms resistance}$$

$$X_t = \frac{104000 \times 0.048}{57.7} = 86.52 \text{ ohms reactance}$$

The values to neutral are, as stated above, one-third of the above; but, for the sake of uniformity in determining values to neutral, should preferably be determined as follows:—

$$\frac{6000000}{60046} = 99.92 \text{ amperes to neutral}$$

$$R_{tn} = \frac{60046 \times 0.0104}{99.92} = 6.25 \text{ ohms resistance to neutral}$$

$$X_{tn} = \frac{60046 \times 0.048}{99.92} = 28.84 \text{ ohms reactance to neutral}$$

If two or more banks operate in parallel, the resultant impedance Z_r can be obtained by taking the re-

sultant impedance Z_r can be obtained by taking the re-

$$Z_r = \frac{4.91 \times 9.82}{4.91 + 9.82} = 3.27 \text{ percent at 6000 kva.}$$

$$= 0.69 \text{ percent resistance volts at 6000 kva.}$$

$$= 3.19 \text{ percent reactance volts at 6000 kva.}$$

If the impedance triangles of the two banks to be paralleled are considerably different (that is their ratio of resistance to reactance) it will be necessary to express the impedances in complex form. We have assumed above that the triangles are proportional, otherwise they would not divide the load evenly at all power-factors. Solving the preceding problem for the resultant impedance by complex notation, we get:

$$Z_r = \frac{(1.04 + j4.8) \times (2.08 + j9.6)}{(1.04 + j4.8) + (2.08 + j9.6)}$$

$$= \frac{-43.917 + j19.968}{3.12 + j14.4}$$

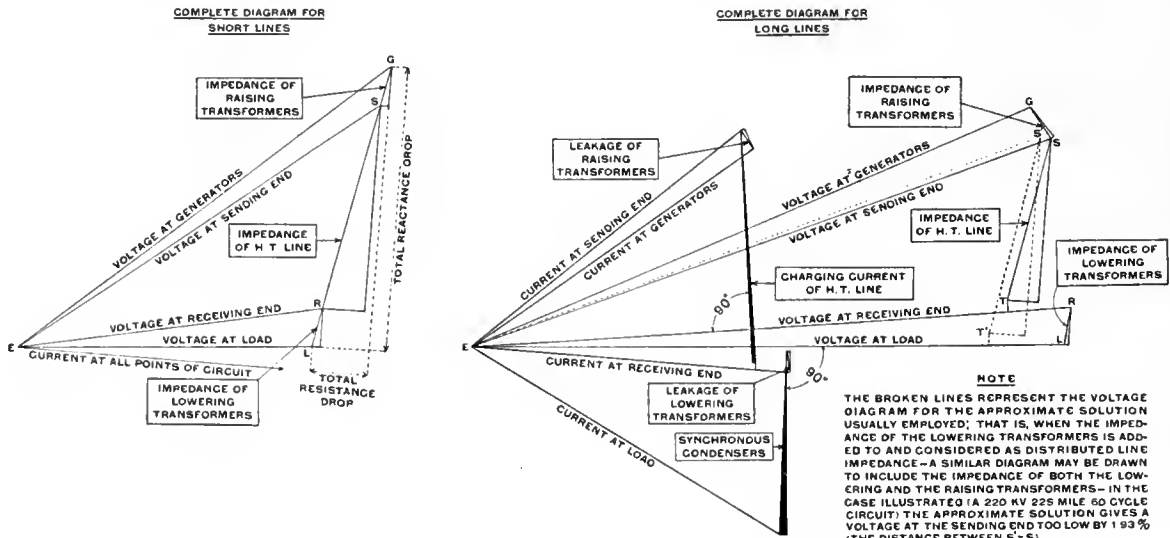


FIG. 67—VECTOR DIAGRAMS FOR SHORT AND LONG LINES

iprocal of the sum of the reciprocals of the individual impedance. Thus:—

$$Z_r = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

In the above example $Z_t = \sqrt{1.04^2 + 4.8^2} = 4.91$ percent.

To parallel two banks containing transformers duplicates of the above, we get, by the above rule, the following resultant impedance:—

$$Z_r = \frac{4.91 \times 4.91}{4.91 + 4.91} = 2.45 \text{ percent}$$

Which is just half the impedance of a single bank, as is evident without applying the rule.

Where two or more banks are to be operated in parallel consisting of transformers not duplicates, then the above rule must be applied to determine the resultant impedance. If the impedances are expressed in percent, as is usual, then they must be both referred

to the same kv-a base. For instance, if a 6000 kv-a and a 3000 kv-a transformer each have a resistance of 1.04 percent and a reactance of 4.8 percent, their impedance is 4.91 percent. Before combining the impedances, that of the 3000 kv-a unit should be put in terms of the 6000 kv-a, and the resultant would be:—

$$= \frac{48.25 \sqrt{155^{\circ} 32' 58''}}{14.734 \sqrt{177^{\circ} 46' 29''}}$$

$$= 3.27 \sqrt{177^{\circ} 46' 29''} \text{ ohms}$$

$$= 0.69 + j 3.19 \text{ ohms}$$

THE AUXILIARY CONSTANTS

The graphical construction for short lines represented typically by the Mershon Chart is so generally known and understood that a similar construction modified to take into accurate account the distribution effect of long lines will readily be followed. Both the short and the long line diagrams are reproduced in Fig. 68. From these diagrams it will be seen how the three auxiliary constants correct or modify the short line diagram adapting it to long line problems. The two mathematical and three graphical methods of obtaining the auxiliary constants are indicated at the

bottom of this figure. Since the auxiliary constants are functions of the physical properties of the circuit and of the frequency only, they are entirely independent of the voltage or the current. Having determined

Constants a_1 and a_2 —If the line is short electrically the charging current, and consequently its effect upon the voltage regulation is small. In such a case constant a_1 would be unity and constant a_2 would be zero, and the line impedance triangle would be attached to the end of the vector ER representing the receiving end voltage, since this vector also represents the sending end voltage at zero load.

If, however, the circuit contains appreciable capacitance, the e.m.f. of self-induction resulting from the charging current will result in a lower voltage at zero load at the sending end than at the receiving end of the line. Obviously, the load impedance triangle must be attached to the end of the vector representing the voltage at the sending end of the circuit at zero load. This is the vector ER' of the long line diagrams of Fig. 68. In such a circuit the effect of the charging current is sufficiently great to cause the shifting of the point R for a short line to the position R' for the long line. The constants a_1 and a_2 therefore, determine the length and position of the vector representing the sending end voltage at zero load. Actually the constant a_2 represents the volts resistance drop due to the charging current for each volt at the receiving end of the circuit. That is, the line FR' equals approximately one-half the charging current times the resistance R , taking into account, of course, the distributed nature of the circuit. For a short line, it would be sufficiently accurate to assume that the total charging current flows through one-half the resistance of the circuit. To make this clear, it will be shown later that, for a 220 kv problem, the resistance per conductor is $R = 34.65$ ohms and the auxiliary constant $C_2 = 0.001211$ mho. Thus, this line will take 0.001211 amperes charging current, at zero load, for each volt maintained at the receiving end, and since $FR' =$ approxi-

mately $I_{cc} \times \frac{R}{2}$ we have FR' or $a_2 = 0.001211 \times \frac{34.65}{2} = 0.020980$. The exact value of a_2 as calculated

by hyperbolic functions, taking into account the distributed nature of the circuit is 0.020234. Since the charging current is in leading quadrature with the voltage ER , the resistance drop FR' due to the charging current is also at right angles to ER .

The length of the line FR or $(one-a_1)$, represents the voltage consmed by the charging current flowing through the inductance of the circuit. This may also be expressed with small error if the circuit is not of

great electrical length as $I_{cc} \times \frac{X}{2}$. The reactance per conductor for the 220 kv problem is 178.2 ohms. Therefore, $FR = 0.001211 \times \frac{178.2}{2} = 0.107900$ and

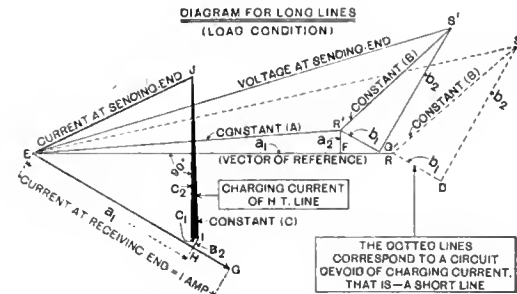
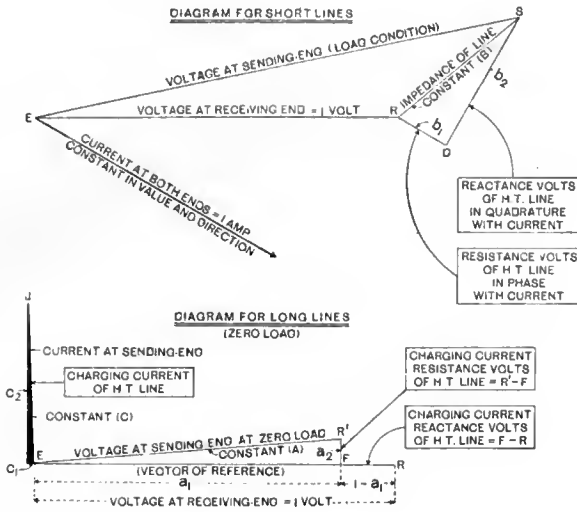
$a_1 = 1 - 0.107900 = 0.892100$. The exact value of a_1 as calculated rigorously, is 0.893955.

Constants b_1 and b_2 —These constants represent respectively the resistance and the reactance in ohms,

VECTORS BASED UPON ONE VOLT AND ONE AMPERE AT 65% POWER FACTOR BEING DELIVERED AT THE RECEIVING END—THE DIAGRAMS CORRESPOND TO A LONG LINE

$$E_s = E_R (a_1 + j a_2) + I_R (b_1 + j b_2)$$

$$I_s = I_R (c_1 + j c_2) + E_R (c_1 + j c_2)$$



HOW THE AUXILIARY CONSTANTS MAY BE OBTAINED

(A) $(a_1 + j a_2) = \left[1 + \frac{YZ}{2} + \frac{Y^2 Z^2}{24} + \frac{Y^3 Z^3}{720} + \frac{Y^4 Z^4}{40,320} + \text{ETC.} \right]$ (BY CONVERGENT SERIES—SEE CHART XI)
 = $\cosh \theta$ (BY REAL HYPERBOLIC FUNCTIONS—SEE CHART XVI)
 = $\frac{\sinh \theta / \theta}{\tanh \theta / \theta}$ (GRAPHICAL—SEE KENNELLY'S CORRECTING FACTOR CHARTS XVIII-XIX-XX-XXI)
 = $\cosh \theta$ (GRAPHICAL—SEE KENNELLY'S CHART ATLAS, HARVARD PRESS)
 = $\cosh \theta$ (ALL GRAPHICAL FROM WILKINSON'S CHART "A"—SEE CHART V)

(B) $(b_1 + j b_2) = Z \left[1 + \frac{YZ}{2} + \frac{Y^2 Z^2}{24} + \frac{Y^3 Z^3}{720} + \frac{Y^4 Z^4}{40,320} + \text{ETC.} \right]$ (BY CONVERGENT SERIES—SEE CHART XI)
 = $\frac{Z}{\sqrt{2}} \sinh \theta$ (BY REAL HYPERBOLIC FUNCTIONS—SEE CHART XVI)
 = $Z \frac{\sinh \theta / \theta}{\tanh \theta / \theta}$ (GRAPHICAL—SEE KENNELLY'S CORRECTING FACTOR CHARTS XVIII-XIX)
 = $\frac{Z}{\sqrt{2}} \sinh \theta$ (GRAPHICAL—SEE KENNELLY'S CHART ATLAS, HARVARD PRESS)
 = $\frac{Z}{\sqrt{2}} \sinh \theta$ (ALL GRAPHICAL FROM WILKINSON'S CHART "B"—SEE CHART VI)

(C) $(c_1 + j c_2) = Y \left[1 + \frac{YZ}{2} + \frac{Y^2 Z^2}{24} + \frac{Y^3 Z^3}{720} + \frac{Y^4 Z^4}{40,320} + \text{ETC.} \right]$ (BY CONVERGENT SERIES—SEE CHART XI)
 = $\frac{1}{\sqrt{2}} \sinh \theta$ (BY REAL HYPERBOLIC FUNCTIONS—SEE CHART XVI)
 = $Y \frac{\sinh \theta / \theta}{\tanh \theta / \theta}$ (GRAPHICAL—SEE KENNELLY'S CORRECTING FACTOR CHARTS XVIII-XIX)
 = $\frac{1}{\sqrt{2}} \sinh \theta$ (GRAPHICAL—SEE KENNELLY'S CHART ATLAS, HARVARD PRESS)
 = $\frac{1}{\sqrt{2}} \sinh \theta$ (ALL GRAPHICAL FROM WILKINSON'S CHART "C"—SEE CHART VII)

WHERE $\theta = \sqrt{ZY}$

FIG. 68—HOW THE AUXILIARY CONSTANTS MODIFY SHORT LINE DIAGRAMS ADAPTING THEM TO LONG LINE PROBLEMS

by any of the five methods referred to, the value for the auxiliary constants corresponding to a given circuit, the remainder of the solution for any receiving end current or voltage is readily performed graphically.

as modified by the distributed nature of the circuit. The values for these constants, multiplied by the current in amperes at the receiving end of the circuit, give the IR and IX volts drop consumed respectively by the resistance and the reactance of the circuit. To illustrate this, the values of R and X for the 220 kv problem are 34.65 ohms and 178.2 ohms per conductor. The distributed effect of the circuit modifies these linear values of R and X so that their effective values are $b_1 = 32.198$ and $b_2 = 172.094$ ohms. The line impedance triangle, as modified to take into exact account the distributed nature of the circuit, is therefore smaller than it would be if the circuit were without capacitance.

Constants c_1 and c_2 —These constants represent respectively the conductance and susceptance in mhos as modified by the distributed nature of the circuit. The values for these constants, multiplied by the volts at the receiving end of the circuit, give the current consumed respectively by the conductance and the susceptance of the circuit. To illustrate, the linear value of c_2 for the 220 kv problem is 0.001211 mho. The distribution effect of the circuit modifies this linear value so that its effective value $c_2 = 0.001168$. The value of c_1 is so small that its effect is negligible for all except for long circuits. An exception to this statement would be that if the line loss is very small compared to the amount of power transmitted the percent error in the value of line loss may be considerably increased if the effect of c_1 is not included in the solution. If c_1 is ignored, c_2 will represent the charging current at zero load per volt at the receiving end. Thus c_2 multiplied by the receiving end voltage, gives the charging current at zero load for the circuit. For the 220 kv problem $c_2 = 0.001168$ and this multiplied by 127 020, the re-

ceiving end voltage to neutral, gives 148.36 amperes charging current per conductor.

Referring to the formulas at the top of Fig. 68, $E_r (a_1 + j a_2)$ is that part of E_s which would have to be impressed at the sending end if $I_r = 0$, or the line was freed at the receiving end with E_r steadily maintained there. It may be called "free" component of E_s *. Again $I_r (b_1 + j b_2)$ is that other part of E_s which would have to be impressed at the sending end, if $E_r = 0$, or the line was short-circuited at the receiving end, with I_r steadily maintained there. It may be called the "short" component of E_s .

Similarly, the term $I_r (a_1 + j a_2)$ is the component of I_s necessary to maintain I_r at the receiving end without any voltage there ($E_r = 0$); while $E_r (c_1 + j c_2)$ is the component of I_s necessary to maintain E_r at the receiving end without any current there ($I_r = 0$). The reason that c_1 is likely to be negative in ordinary power lines is because the complex hyperbolic angle of any good power transmission line has a large slope, being usually near 88 degrees. The sinh of such an angle, within the range of line lengths and sizes of b ordinarily present, is also near 90 degrees in slope.

The surge impedance $Z_0 = \sqrt{\frac{Z}{Y}}$ of such a line is not far from being reactanceless; but it usually develops a small negative or condensive slope. This means that the surge admittance $Y_0 = \frac{I}{Z_0}$ usually develops a small positive slope. Consequently, C or the product $E_r (c_1 + j c_2)$ usually slightly exceeds 90 degrees in slope; or c_1 becomes a small negative rectilinear component.

*See paper by Houston and Kennelly on "Resonance in A. C. Lines" in Trans. A. I. E. E. April, 1895

CHAPTER XVI

A TYPICAL 220 KV PROBLEM

TO illustrate the method of determining the performance of long lines requiring phase modifiers for voltage control, the following 220 kv problem will be considered, which is typical of many likely to be considered in the near future. A line necessitating such large expenditure would warrant a thorough investigation before determining the final design. The conclusions are given only for the purpose of illustrating the procedure.

The Problem—It is assumed that 300 000 kw at 85 percent lagging power-factor is to be delivered a distance of 225 miles, at 220 kv, three-phase, 60 cycles. Two lines will be required, so that in case one is under repair, the other will transmit the entire 300 000 kw load. Since the self-induced voltage would be excessive if the 300 000 kw were transmitted in emergency over a single-circuit tower line, we will assume that each tower line will support two three-phase circuits. The cost of two three-phase circuits per tower line will not be greatly in excess of a single circuit tower line employing conductors of double the cross-section. On this basis each of the four three-phase circuits will normally transmit 75 000 kw and, under emergency condition, each of the two circuits on one tower line will transmit 150 000 kw. Such a transmission is illustrated by Fig. 69*

Economic Size of Conductors—For a fixed transmission voltage and material of conductors, the most economic size of conductor will be found by applying Kelvin's law extended to include, in addition to the cost of conductors, that part of the cost of towers, insulators, line construction, phase modifiers, etc. which increases directly with the cost of conductors. Kelvin's law is as follows:—

"The most economical section of a conductor is that which makes the annual cost of the I^2R losses equal to the annual interest on the capital cost of the conducting material plus the necessary annual allowance for depreciation". Stated another way, "The annual cost of the energy wasted, added to the annual allowance for depreciation and interest on first cost shall be a minimum".

*The calculations and the illustrations in this article were made in such a way as to be equally suited for the series of articles on "Electrical Characteristics of Transmission Circuits" and the Superpower Survey. Figs. 69, 70, 72 and 75 and Charts XXIII, XXV and XXVII appear also in the report of the latter, which is printed as *Professional Paper 123* by the United States Geological Survey. Similarly, Charts XXIV, XXVI and XXVIII appear in the Paper by L. E. Imlay in the *Journal of the A. I. E. E.* for June, 1921. (Ed.)

In Table Y is shown a comparison of values of capitalized losses vs. first costs of conductors for four sizes of aluminum-steel cables considered in connection with this 220 kw problem**. The cost of power losses is based upon rates of 0.3, 0.4 and 0.5 cents per kw hour, an average load corresponding to 80 percent of the full load loss and a capitalization of these losses at 15 percent. The cost of the cables is based upon 29 cents per pound for the complete cable (aluminum plus the steel). All tabulated data is based upon four three-phase circuits. The losses include those in the high voltage line only. If the capacity of transformers or phase modifiers varies materially for different conductors, the difference in their losses should be included.

If the base load power generated in such a large amount by water power costs 0.3 of a cent per kw-hr.,

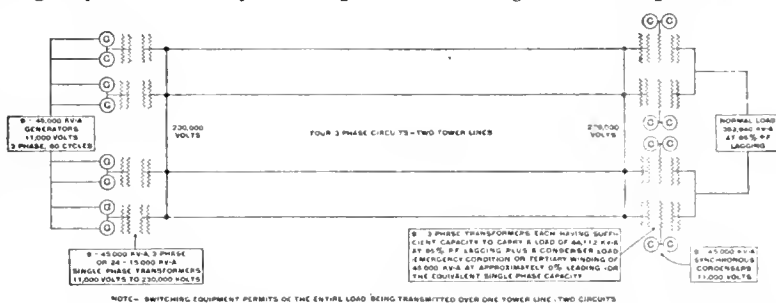


FIG. 69—THE TRANSFORMER AND CONDENSER ARRANGEMENT UPON WHICH THE CALCULATIONS FOR THE 220KV PROBLEM HAVE BEEN BASED.

It is not intended that this arrangement would, upon a complete study of the problem, be found to be the most desirable. If single-phase transformers were selected, possibly three banks for each double circuit would be found more desirable than four banks, as indicated above.

The values in Table Y show that the smallest size cable, 605 000 circ. mil. will be the cheapest to install. At 0.3 cents per kw-hr. the power loss for this cable, capitalized at 15 percent, represents the equivalent of an investment of \$2 593 000 for the four three-phase circuits, whereas the cost of the conductors is \$3 224 000. If the cost of power loss is taken as 0.4 cents per kw-hr., the next larger cable will be the most economical size to use, provided that there is no increased cost of towers, insulators, etc. If the losses in transformers or condensers vary for the different sizes of cables compared such losses should be included with the conductor losses.

There is always a question as to what price should be charged in Kelvin's equation in estimating the cost of power loss. If all power saved could be promptly sold, the cost to allow might be considered the cost at the consumers meter. If, on the contrary, none of the power saved can be sold under any circumstances,

**An interesting graphic presentation of Kelvin's Law is given in the article by Mr. L. J. Moore in the *Electrical World* for Sept. 24, 1921, p. 612.

then the cost to allow is the cost at the generating switchboard. Intermediate cases may occur.

The conductor losses of Table Y were taken from the calculated values by the complete method A listed in Table V*. It is usually sufficient to calculate the

TABLE Y—APPLICATION OF KELVIN'S LAW

Conductors Circ. Mill.	Total Loss in 12 Conductors Kw	Cost of Power lost in 12 Conductors, Capitalized at 15%			Cost of 12 Conductors at 29c per Lb.
		At 0.3c per Kw-hr.	At 0.4c per Kw-hr.	At 0.5c per Kw-hr.	
*605 000	18 504	\$2 593 000	\$3 458 000	\$4 322 000	\$3 224 000
715 500	15 840	\$2 220 000	\$2 960 000	\$3 700 000	\$3 837 000
795 000	14 304	\$2 040 000	\$2 673 000	\$3 341 000	\$4 244 000
954 000	11 712	\$1 641 000	\$2 188 000	\$2 736 000	\$5 011 000

*This is the smallest conductor which is, in this case, permissible on account of corona limitations. These tabulations are total for four three-phase circuits. It will usually be sufficiently accurate to calculate the conductor I²R loss for one size of conductor and assume that the loss for other sizes will be proportional to their resistances. This assumes that the distribution of current throughout the length of circuit will be approximately the same for the different sizes of conductors compared. The above data is based upon 75 000 kw at 85 percent power-factor, three-phase, 60 cycles, delivered over each of the four circuits a distance of 225 miles at 220 kv with a 50 000 kv-a condenser in parallel with the load on each of the four circuits and an average load equivalent to 80 per cent of full load. It should be noted that the third, fourth and fifth columns do not give the actual cost of the power lost, but give instead the values at which these losses are capitalized.

loss in the conductors for one size of cable and to estimate it for other sizes of cable, assuming that this loss varies as the resistance of the conductors, that is, for a given line, frequency, load, delivery voltage and condenser capacity the current distribution in the line is approximately the same for various sizes of conductors likely to be considered. Since the conductor loss varies as the square of the current and directly as the resistance, it will be sufficient to estimate the loss for other conductors as being inversely proportional to their resistance.

The various constants corresponding to the four sizes of conductors considered are listed in Table Z. It may be interesting to note the variation in these constants corresponding to the different sizes of cable for the high-tension line alone, and also when the transformer impedances are included with the line impedance.

SOLUTION OF THE 220 KV PROBLEM

Assuming that 605 000 circ. mil. aluminum-steel cables work out as the most economical size, the next step is the determination of the auxiliary constants A, B, and C for this size of conductor, spacing and 60 cycles. (These constants would have previously been determined when determining the most economical size). Mathematically these constants may be calculated by real hyperbolic functions (Chart XVI) or by convergent series (Chart XI). Graphically, they may be obtained from Wilkinson's charts (Charts V, VI and VII) or through the medium of Dr. Kennelly's charts

(Charts XVIII, XIX, XX and XXI). When using charts it is desirable to read the results from them at two different times as a check against errors in reading, or the constants may be read from both the Wilkinson and Kennelly charts and the results compared. From Table V we find $r = 0.154$ ohms, so that $R = 0.154 \times 225 = 34.65$ ohms and $x = 0.792$ so that $X = 0.792 \times 225 = 178.2$ ohms. From Table X we obtain $b = 5.38 \times 10^{-6}$ so that $B = 5.38 \times 225 \times 10^{-6} = 0.001211$ mho. G is assumed here as zero.

From Wilkinson Charts—

$$a_1 = 0.892$$

$$\text{and since } rb = 0.828$$

$$a_2 = 0.020$$

$$b_1 = 32.2 \text{ ohms}$$

$$b_2 = 173.5 \text{ ohms}$$

$$\text{and since } rb^2 = 4.457$$

$$c_1 = (\text{too small to read})$$

$$c_2 = 0.001175$$

From Dr. Kennelly's Charts—We must first obtain the hyperbolic complex angle of the circuit as follows:—

$$Z = 34.65 + j 178.2$$

$$= 181.54 \angle 78^\circ 59' 46''$$

$$Y = 0 + j 0.001211$$

$$= 0.001211 \angle 90^\circ$$

$$ZY = 0.21984 \angle 168^\circ 59' 46''$$

$$\theta = \sqrt{ZY} = 0.4689 \angle 84^\circ 29' 53''$$

$$\frac{\text{Sinh } \theta}{\theta} = 0.964 \angle 0.4^\circ$$

$$= 0.964 \angle 0^\circ 24' 00''$$

$$\frac{\text{Tanh } \theta}{\theta} = 1.0785 \angle 0.88^\circ$$

$$= 1.0785 \angle 0^\circ 52' 48''$$

TABLE Z—CABLE AND CIRCUIT CONSTANTS CORRESPONDING TO A THREE-PHASE, 60 CYCLE CIRCUIT, 225 MILES LONG CONSISTING OF FOUR SIZES OF ALUMINUM CABLES OF AN ARRANGEMENT EQUIVALENT TO 21 FEET DELTA

ALUM	STEEL	TOTAL	COPPER EQUIV	DIA. OF ALUM COND.	STRANDB		LINEAR CONSTANTS OF LINE TO NEUTRAL								IMPEDANCE TO NEUTRAL OF 60 000 KV-A BANK OF TRANSFORMERS *			
					AL	ST	r	x	g	b	R	X	G	B	R _{TN}	X _{TN}	G _{TN}	B _{TN}
605,000	78,000	683,500	380,400	0.952	54	7	0.154	0.792	0	5.38	34.65	178.2	0	12.11	6.37	78.64	0	0
715,500	92,900	808,900	430,000	1.036	54	7	0.131	0.782	0	5.45	29.98	175.9	0	12.26	6.37	78.64	0	0
795,000	103,100	898,100	500,000	1.092	54	7	0.111	0.775	0	5.49	24.33	174.4	0	12.35	6.37	78.64	0	0
954,000	123,700	1,077,700	600,000	1.196	54	7	0.0778	0.764	0	5.58	22.00	171.9	0	12.56	6.37	78.64	0	0

*Since two 50 000 kv-a banks of transformers will be required at each end the corresponding values for impedance will be half these amounts.

$$A = \frac{\text{Sinh } \theta}{\text{Tanh } \theta} = \frac{0.964 \angle 0^\circ 24' 00''}{1.0785 \angle 0^\circ 52' 48''}$$

$$= 0.8939 \angle 1^\circ 16' 48''$$

$$a_1 = 0.8937$$

$$a_2 = 0.01996$$

*In the JOURNAL for Dec. 1921, p. 544.

$$B = Z \frac{\text{Sinh } \theta}{\theta} = 181.54 \frac{78^\circ 59' 46''}{\theta} \times 0.964 \frac{6^\circ 24' 00''}{\theta}$$

$$= 175.0 \frac{79^\circ 23' 46''}{\theta} \text{ ohms}$$

$$b_1 = 32.2 \text{ ohms}$$

$$b_2 = 172 \text{ ohms}$$

$$C = Y \frac{\text{Sinh } \theta}{\theta} = 0.001211 \frac{90^\circ}{\theta} \times 0.964 \frac{6^\circ 24' 00''}{\theta}$$

$$= 0.001167 \frac{90^\circ 24' 00''}{\theta} \text{ mho}$$

$$c_1 = -0.000008 \text{ mho}$$

$$c_2 = 0.001167 \text{ mho}$$

The auxiliary constants as obtained graphically and by exact mathematical solution, are given in Table ZZ. It is thus seen that the Kennelly charts, although primarily intended for correcting the linear impedance and the linear admittance of circuits for the equivalent π solution, are highly adaptable to determining the values of the auxiliary constants to a very close degree of accuracy. The use of these charts for obtaining auxiliary constants requires more arithmetical work than the use of the Wilkinson charts. For instance the hybolic angle, $\theta = \sqrt{ZY}$ of the circuit must first be calculated before the charts can be employed. The results, read from charts, must then be multiplied by the impedance and the admittance of the circuit for obtaining auxiliary constants B and C . Auxiliary constant A cannot be taken directly from a single Kennelly chart. To obtain this auxiliary constant from these charts it is necessary to divide the values read from two of these

charts since $A = \frac{\sinh \theta / \theta}{\tanh \theta / \theta}$. Chart $\tanh \theta / \theta$ is constructed for angles up to and including 0.50 polar values. This makes it adapted to angles up to 1.0 polar value when used for determining correcting factors for the equivalent π solution. This is for the reason that for obtaining such correcting factors we enter this chart with $\theta/2$. However for obtaining auxiliary constant A by means of values read from these charts we must enter this chart with θ in place of $\theta/2$. This limits the use of the Kennelly charts for obtaining auxiliary constant A to circuit angles not exceeding 0.5 polar values. In case the circuit angle has a polar value greater than 0.5, Wilkinson chart A may be used provided the line is not over 300 miles long. If the circuit is over 300 miles long the auxiliary constants should be determined by mathematical calculation.

In the following discussion the calculated values for the auxiliary constants will be used, since exact results are required for the purpose of comparing the results with those obtained by the approximate method, a description of which follows the complete solution.

NORMAL LOAD—COMPLETE SOLUTION

The complete solution for normal load is given by Chart XXIII. At the top is illustrated the circuit diagrammatically. Underneath this is stated the load conditions, linear and the auxiliary constants for this circuit. The transformer data and method of determining the amperes iron loss, magnetizing current and impedance to the neutral of the lowering transformer is

also shown. Actually the impedance of raising and lowering transformers, even when duplicates, is slightly different when the connections are not made to similar taps. This difference is so slight (and so far as the raising transformer is concerned so unimportant) that for simplicity, we are assuming that both raising and lowering transformers have the same impedance. This comprises all the data required for a complete mathematical or graphical solution of this circuit.

Following the data is a complete graphical vector solution of this circuit with symbols placed on all vectors indicating the manner of obtaining their values. At the lower left hand corner is placed a complete mathematical solution of the problem, which parallels the graphical solution (one method of solution checking the other). In the calculations of the high-voltage circuit the current, in order to include the power-factor, must always be expressed in complex form referred to the vector of reference, as indicated by a dot under the symbol I .

At the lower right hand corner a method is indicated of determining the transmission loss from the calculated quantities. The loss in the high-tension line

TABLE ZZ—AUXILIARY CONSTANTS FOR 220 KV PROBLEM APPROXIMATE SOLUTION

	Calculated	From Wilkinson Chart	From Kennelly Chart
a_1	0.893955 = 100 %	0.892 = 99.78 %	0.8937 = 99.97 %
a_2	0.020234 = 100 %	0.020 = 98.85 %	0.01996 = 98.65 %
b_1	32.198 = 100 %	32.2 = 100 %	32.2 = 100 %
b_2	172.094 = 100 %	173.5 = 100.82 %	172 = 99.95 %
c_1	-0.000008 = 100 %	can't read	-0.000008 = 100 %
c_2	0.001168 = 100 %	0.001175 = 100.6 %	0.001167 = 99.91 %

can be determined graphically by scaling off the voltage and the current at each end of the high-tension line and measuring the angle between the vectors representing the current and the voltage. The current times the voltage times the cosine of this angle will give the power at the point considered and the difference between the power as so determined at the two ends of the high-tension line is the line loss. The losses in transformers and condensers are known and stated at the top of the chart.

The complete vector diagram is constructed as follows: First draw the horizontal line representing E_{LN} , the voltage at the load to neutral. This should be drawn to as large a scale as possible. All other voltage vectors will of course be drawn to the same scale. The vector I_L representing the load current is now drawn to as large a scale as can be used without mixing the current vectors with the voltage vectors. This is drawn at an angle of $31^\circ 47'$ from E_{LN} in the lagging direction, corresponding to a lagging load of 85 percent power-factor. It usually works out that for normal load the power-factor at the receiving end should be slightly lagging and at the sending end slightly leading so that the average power-factor of the line will be close to unity. This will necessitate a phase modifier in parallel with the load, having approximately the capacity of the lagging kv-a in the load.

The lagging kv-a in the load is equal to the kv-a of the load times the sine of the angle of the load. In this case it is $88\,235 \times \sin 31^\circ 47' = 46\,500$ kv-a. The vector diagram is constructed on the basis of a 45 000 kv-a condenser in parallel with the load. This condenser has a power loss of 4.72 amperes to neutral and since this is in phase with the load voltage, we trace from the end of the load current vector horizontally to the right a distance representing 4.72 amperes by the current scale. The current per terminal for the condenser is 118.09 amperes so that the leading component of the current input of the condenser is 118.00 amperes. Since this is leading it is drawn vertically upward from the last point determined. Actually we will not need to determine the 118 amperes leading component, but will complete the solid black condenser triangle, since the length of the input line is 118.09 amperes. To the vector sum of load and condenser currents thus determined we now add the leakage current of the lowering transformers, the lagging component of which materially effects the capacity of the phase modifiers required because of its nearly direct opposition to it under load. We have assumed that the leakage current required by the lowering transformers will be supplied by the phase modifier on account of its close electrical proximity to the lowering transformers. On this assumption the triangle representing this transformer leakage will be located as indicated. There is a loss current of 1.85 amperes in phase with the load voltage and a magnetizing current of 13.9 amperes in lagging quadrature with the load voltage. We thus find that the current I_R at the receiving end of the line is 204.17 amperes, lagging $5^\circ 1' 16''$ behind the load voltage. In this case the magnetizing current of the lowering transformer reduces the effective capacity of the phase modifier by an amount of 13.9 amperes; that is by 5.3 percent of the total capacity of the lowering transformers.

We next determine the voltage at the high-voltage side of the lowering transformers; that is the voltage E_{RN} at the receiving end of the transmission line. Knowing the resistance and reactance of the lowering transformer banks to neutral and the current I_R , the transformer resistance voltage drop is plotted in phase with the current I_R and the reactance voltage drop in quadrature with the resistance drop as indicated. The voltage at the sending end E_{SN} of the transmission line is next determined by applying auxiliary constants A and B to the voltage and current respectively of the receiving end.

The base of the impedance triangle for the high-tension line $I_R \times b_1$ represents the resistance drop of the high-tension line in phase with the receiving end current. In quadrature to this is the reactance volts drop of the line $I_R \times b_2$. The voltage at the sending end is thus determined to be 131 858 volts which corresponds to slightly less than 230 000 volts between conductors. An arc of a circle corresponding to the voltage to be maintained at the sending end will serve as

a guide in determining the proper capacity condenser necessary to maintain this sending end voltage. An increase in condenser capacity rotates the vector I_R in a counter-clockwise direction, swinging the line impedance triangle also in a counter-clockwise direction thus decreasing the voltage E_{SN} and reducing the line drop. A decrease in condenser capacity rotates the vector I_R in a clockwise direction, swinging the line impedance triangle also in a clockwise direction, thus increasing the voltage E_{SN} and increasing the line drop. Thus the effect upon line voltage drop may be readily determined for condensers of various capacities.

The next step is to determine the current at the sending end. This is done by applying auxiliary constants A and C to the current and voltage respectively of the receiving end. It will be noted that the charging current is drawn as leading by 90 degrees the high-tension voltage at the receiving end, which voltage is taken as the vector of reference as in previous discussions. The current at the sending end is thus determined to be 220.34 amperes leading the vector of reference by $35^\circ 12'$. The impedance triangle for the raising transformers may now be drawn in, the resistance drop of same being drawn parallel with I_S . This then gives the voltage at the generators. The current at the generators is determined by adding vectorially to I_S the leakage of the raising transformers. It is assumed that the raising transformers will receive their excitation from the generators, in which case the leakage triangle will occupy the position shown, resulting in a current at the generators of 218.88 amperes.

NORMAL LOAD—APPROXIMATE SOLUTION

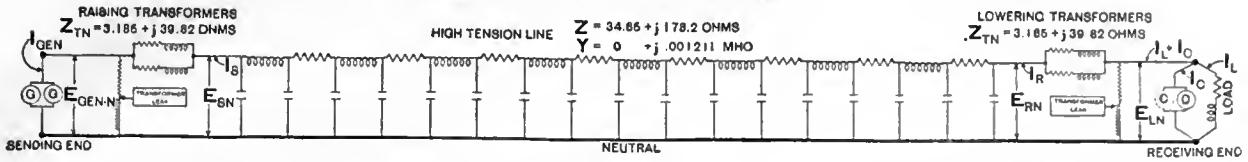
The approximate solution for normal load is given in Chart XXIV. It differs from the complete solution in that the impedance of the lowering transformers is added to and considered as a part of the line impedance so that there are no transformer impedance triangles to construct. It differs also in that, in the case illustrated, the conditions at the sending end only are obtained, whereas in the complete diagram the conditions at both sending end and generators were determined. If the condition at the generators in place of at the sending end is required, the impedance of the raising transformers would also be added to that of the line, the general construction of the diagram remaining the same as for the complete solution.

If it is not necessary to know conditions at both sides of the raising and lowering transformer banks, then it will be seen from a comparison of the two diagrams that the approximate solution will be simpler, although the results will be somewhat incorrect. For instance, for the 220 kv problem illustrated, the errors in the results will, according to tabulations in the lower right hand corner, vary from 0.88 to 2.38 percent. If the losses in condensers and transformers were not added to the load (as they are in both these complete and approximate methods) and the transformer mag-

CHART XXIII—220 KV PROBLEM—NORMAL LOAD

(COMPLETE SOLUTION)

(LOW TENSION VALUES REFERRED TO THE HIGH TENSION CIRCUIT)



NORMAL LOAD

PER 3 PHASE CIRCUIT	PER PHASE TO NEUTRAL
KV-A _L = 88.236	KV-A _{LN} = 20.412
KW _L = 76.000	KW _{LN} = 26.000
PF _L = 86% LAG.	PF _{LN} = 86% LAG.
E _L = 220,000	E _{LN} = 127,020
I _L = 231.65	I _{LN} = 231.65
60 CYCLES	

CONDENSER

(ONE REQUIRED)

3 PHASE	TO NEUTRAL
KV-A _C = 46.000	KV-A _{CN} = 15.000
E _C = 220,000	E _{CN} = 127,020
I _C = 118.09	I _{CN} = 118.09
KW _{C-LOSS} = 1.800	KW _{C-LOSS-N} = 800
I _{C-LOSS} = 4.72	I _{C-LOSS-N} = 4.72

NOTE - THE CONDENSER INDICATED BY BROKEN LINE CIRCLE SERVES AS A SPARE DURING NORMAL OPERATION BUT IS REQUIRED FOR THE EMERGENCY CONDITION.

LINEAR CONSTANTS

Z = 34.85 + j178.2 OHMS
Y = 0 + j.001211 MHO

AUXILIARY CONSTANTS

(A) = (a₁ + j a₂) = cosh θ = 893.955 + j.020,234 = 884.10 / 1° 17' 47"
(B) = (b₁ + j b₂) = Z sinh θ = 32.189 + j 172.094 OHMS
(C) = (c₁ + j c₂) = Y sinh θ = -0.000,008 + j.001,188 MHO
WHERE θ = √ZY

TRANSFORMERS

(TWO BANKS IN PARALLEL AT EACH END OF THE LINE)
ON BASIS OF 100,000 KV-A RATING FOR TWO BANKS

RESISTANCE VOLTS	= 0.658%
REACTANCE VOLTS	= 6.226%
MAGNETIZING CURRENT	= 5.300%
IRON LOSS	= 0.705%

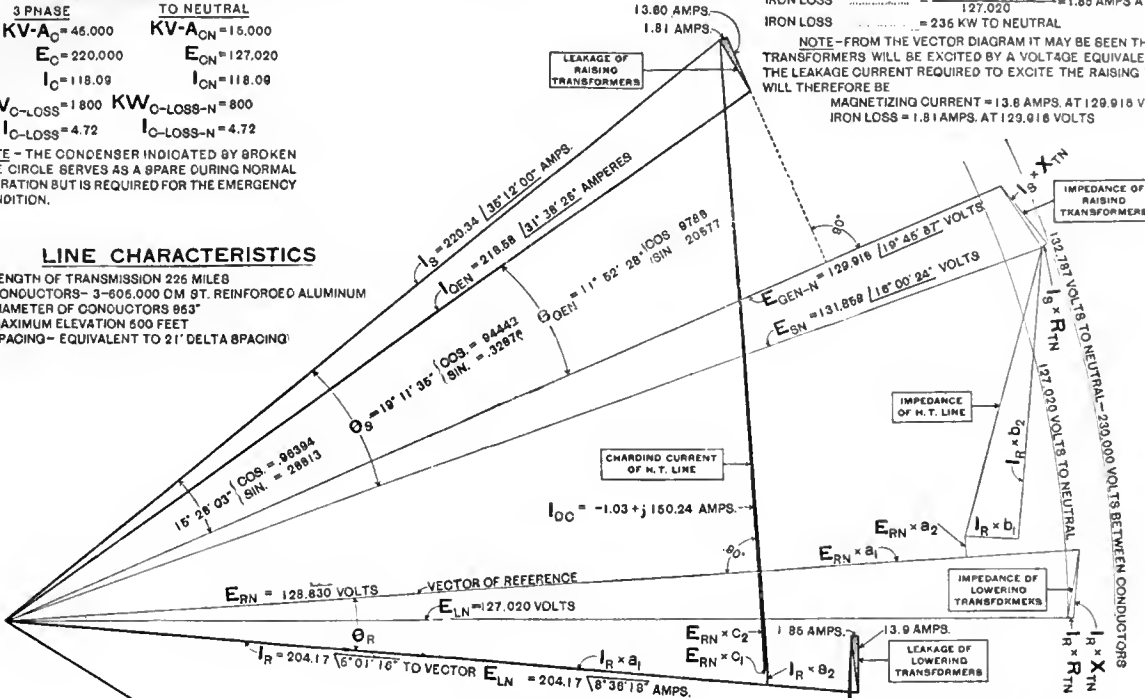
VALUES TO NEUTRAL

KV-A _{TN} = 33,333	E _{TN} = 127,020	I _{TN} = 282.4
R _{TN} = 0.0858 × 127,020 / 282.4 = 3.185 OHMS RESISTANCE		
X _{TN} = 0.8225 × 127,020 / 282.4 = 39.82 OHMS REACTANCE		
MAGNETIZING CURRENT = 0.630 × 33,333.333 / 127,020 = 13.9 AMPS AT 127,020 VOLTS		
IRON LOSS = 0.0705 × 33,333.333 / 127,020 = 1.85 AMPS AT 127,020 VOLTS		
IRON LOSS = 236 KW TO NEUTRAL		

NOTE - FROM THE VECTOR DIAGRAM IT MAY BE SEEN THAT THE RAISING TRANSFORMERS WILL BE EXCITED BY A VOLTAGE EQUIVALENT TO 129,918 - THE LEAKAGE CURRENT REQUIRED TO EXCITE THE RAISING TRANSFORMERS WILL THEREFORE BE
MAGNETIZING CURRENT = 13.8 AMPS. AT 129,918 VOLTS
IRON LOSS = 1.81 AMPS. AT 129,918 VOLTS

LINE CHARACTERISTICS

LENGTH OF TRANSMISSION 225 MILES
CONDUCTORS - 3-805,000 CM ST. REINFORCED ALUMINUM
DIAMETER OF CONDUCTORS 953"
MAXIMUM ELEVATION 600 FEET
SPACING - EQUIVALENT TO 21' DELTA SPACING



CALCULATION FOR RECEIVING-END CURRENT AND VOLTAGE

I_L = 198.82 - j121.97 AMPS. TO VECTOR E_{LN}
I_C = 4.72 + j118.09 OOS 5° 01' 18" = .99818
I_T = 1.86 - j 13.90 SIN 6° 01' 18" = .087623
I_R = 203.30 - j 17.87 OOS 8° 38' 19" = .98874
I_R = 203.30 - j 17.87 SIN 8° 38' 19" = 14.983
= 204.17 √5° 01' 18" AMPS. TO VECTOR E_{LN}
= 204.17 √8° 38' 19" AMPS. TO VECTOR OF REFERENCE
= 201.87 - j 30.56

CALCULATION FOR HIGH TENSION CIRCUIT

E_{RN}(A) = 114.890 + j 2.803 I_R(A) = 181.08 - j 23.23
I_R(B) = 11.757 + j33.757 E_{RN}(C) = -1.03 + j160.24
E_{SN} = 128.746 + j38.380 I₈ = 180.06 + j127.01
= 131.858 / 16° 09' 24" VOLTS = 220.34 / 35° 12' 00" AMPS.

CALCULATION FOR GENERATOR VOLTAGE AND CURRENT

E_{GEN-N} = √[(131.858 + 84442 + 220.34 × 3.185)² + (131.858 × 32876 - 220.34 × 39.82)²] = 129,918 / 15° 25' 03" VOLTS TO VECTOR I₈
= 129,918 / 19° 45' 57" VOLTS
KV-A₈ = 131.858 × 220.34 × 3 = 87,163 PER 3 PHASE CIRCUIT
KV-A_{GEN} = 129,918 × 218.88 × 3 = 85,308 PER 3 PHASE CIRCUIT
KV-A_{CD} = 131.858 × 160.24 × 3 = 69,431 PER 3 PHASE CIRCUIT

DETERMINATION OF LOSSES

KW _{LN} = 26.000
KW _{RN} = 128.830 × 204.17 (COS 8° 38' 19") = 26.988
KW _{BN} = 131.858 × 220.34 (COS 16° 11' 36") = 27.439
KW _{GEN-N} = 129.918 × 218.88 (COS 11° 52' 28") = 27.827

LOSSES TO NEUTRAL

LOWERING TRANSFORMERS AND CONDENSER	25.988 - 25.000 = 0.988
HIGH TENSION LINE	27.439 - 25.988 = 1.471
RAISING TRANSFORMERS	27.827 - 27.439 = 0.388
TOTAL LOSS TO NEUTRAL	28.27

AS A PARTIAL CHECK

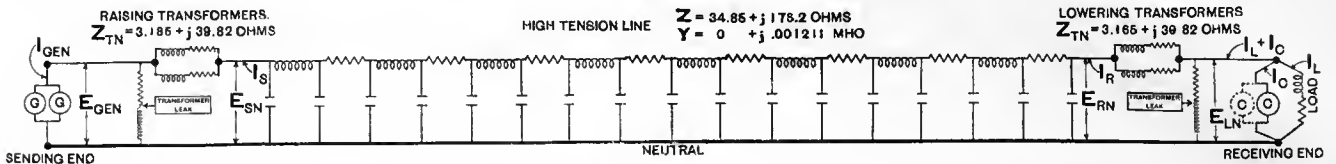
LOWERING TRANSFORMERS	(IRON LOSS 1.85 × 127.02) = 236
(COPPER LOSS 204.17² × 3.185) = 132	
SYNCHRONOUS CONDENSER	= 800
RAISING TRANSFORMERS	(IRON LOSS 1.81 × 129.918) = 236
(COPPER LOSS 220.34² × 3.185) = 154	
HIGH TENSION LINE	(VALUE ABOVE ASSUMED AS CORRECT) = 1471
TOTAL	2827

EFFICIENCY

EFFICIENCY, (HIGH TENSION LINE)	= 25.988 / 27.439 = 94.64%
EFFICIENCY, (GENERATORS TO LOAD)	= 25.000 / 27.827 = 89.84%

CHART XXIV—220 KV PROBLEM—NORMAL LOAD (APPROXIMATE SOLUTION)

THIS APPROXIMATE SOLUTION ASSUMES THAT THE IMPEDANCE OF THE LOWERING TRANSFORMERS MAY BE ADDED TO THE LINE IMPEDANCE AND TREATED AS THOUGH IT WERE DISTRIBUTED LINE IMPEDANCE—THIS ASSUMPTION SIMPLIFIES THE SOLUTION AT THE EXPENSE OF ACCURACY (SEE LOWER RIGHT HAND CORNER OF PAGE). ALSO TEXT—THE SOLUTION BELOW IS BASED UPON THE VOLTAGE BEING HELD CONSTANT AT THE LOAD SIDE OF THE LOWERING TRANSFORMERS AND AT THE HIGH TENSION SIDE OF THE RAISING TRANSFORMERS—IF THE VOLTAGE IS TO BE HELD CONSTANT AT THE GENERATOR BUS, THE IMPEDANCE OF THE RAISING TRANSFORMERS MUST ALSO BE ADDED TO THAT OF THE LINE—ALL LOW TENSION VALUES ARE REFERRED TO THE HIGH TENSION CIRCUIT.



NORMAL LOAD

PER 3 PHASE CIRCUIT	PER PHASE TO NEUTRAL
$KV-A_L = 88,236$	$KV-A_{LN} = 29,412$
$KW_L = 76,000$	$KW_{LN} = 26,000$
$PF_L = 86\% \text{ LAG.}$	$PF_{LN} = 86\% \text{ LAG.}$
$E_L = 220,000$	$E_{LN} = 127,020$
$I_L = 231.65$	$I_{LN} = 231.65$
80 CYCLES	

LINEAR CONSTANTS

$Z = 37.636 + j218.02 \text{ OHMS } \star$
 $Y = 0 + j.001211 \text{ MHO}$

\star THIS INCLUDES IMPEDANCE OF LOWERING TRANSFORMERS

AUXILIARY CONSTANTS

(A) = $(a_1 + j a_2) = \cosh \theta = .870763 + j.021911$
 (B) = $(b_1 + j b_2) = Z \frac{\sinh \theta}{\theta} = \sqrt{\frac{Z}{Y}} \sinh \theta = 34.5853 + j208.83 \text{ OHMS}$
 (C) = $(c_1 + j c_2) = Y \frac{\sinh \theta}{\theta} = \frac{1}{\sqrt{\frac{Z}{Y}}} \sinh \theta = -.000,009 + j.001168 \text{ MHO.}$
 WHERE $\theta = \sqrt{ZY}$

TRANSFORMERS

(TWO BANKS IN PARALLEL AT EACH END OF THE LINE)

ON BASIS OF 100,000 KV-A RATING FOR TWO BANKS	RESISTANCE VOLTS	REACTANCE VOLTS	MAGNETIZING CURRENT	IRON LOSS
	= 0.858 %	= 8.225 %	= 5.300 %	= 0.705 %

VALUES TO NEUTRAL

$KV-A_{TN} = 33,333.$	$E_{TN} = 127,020.$	$I_{TN} = 262.4$
$R_{TN} = \frac{.00658 \times 127,020}{262.4} = 3.186 \text{ OHMS RESISTANCE}$		
$X_{TN} = \frac{.08225 \times 127,020}{262.4} = 36.82 \text{ OHMS REACTANCE}$		
MAGNETIZING CURRENT = $\frac{.0530 \times 33,333.333}{127,020} = 13.9 \text{ AMPS AT } 127,020 \text{ VOLTS}$		
IRON LOSS = $\frac{.00705 \times 33,333.333}{127,020} = 1.86 \text{ AMPS AT } 127,020 \text{ VOLTS}$		
IRON LOSS = 235 KW TO NEUTRAL		

CONDENSER

(ONE REQUIRED)

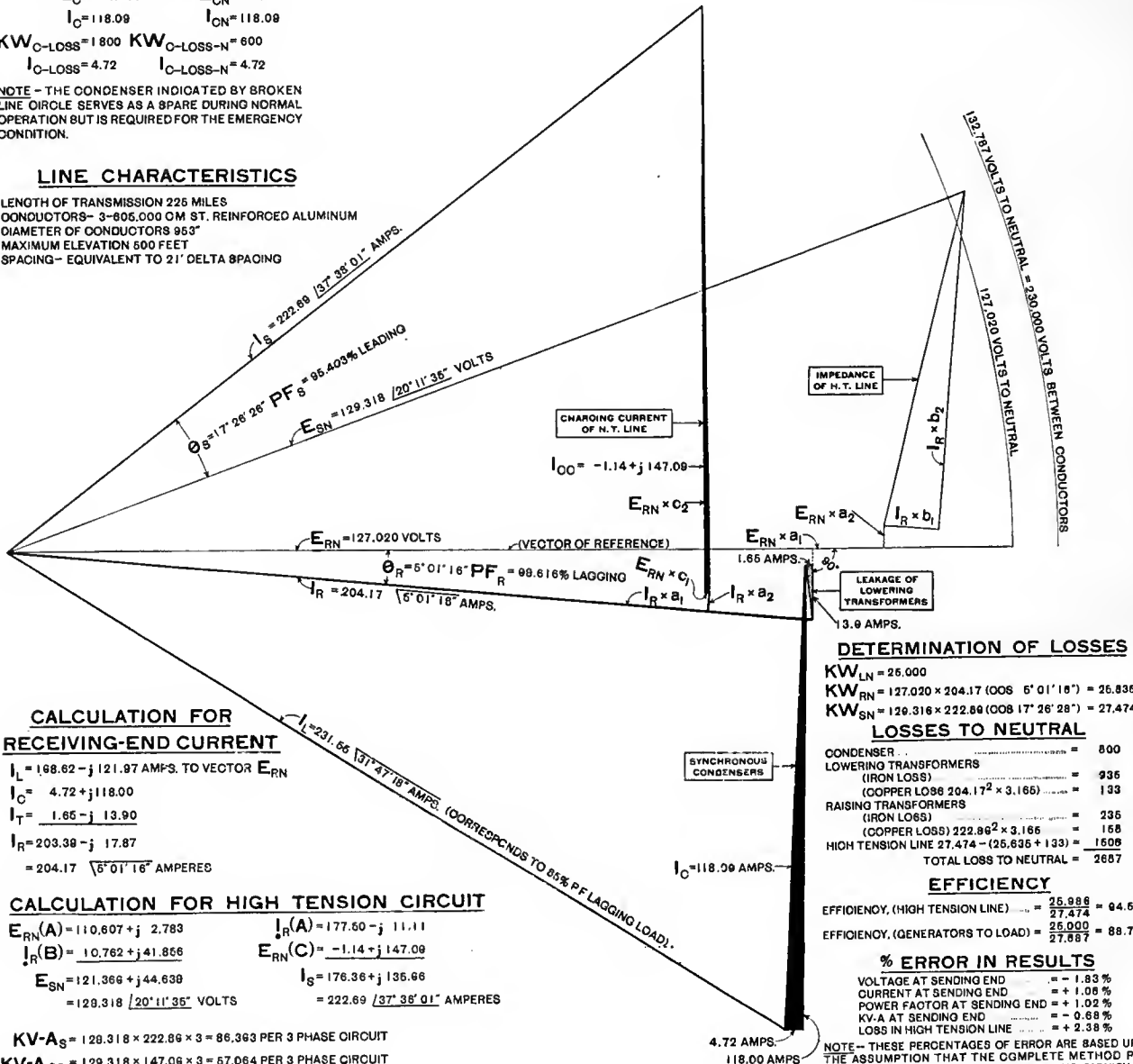
3 PHASE	TO NEUTRAL
$KV-A_C = 45,000$	$KV-A_{CN} = 15,000$
$E_C = 220,000$	$E_{CN} = 127,020$
$I_C = 118.09$	$I_{CN} = 118.09$

$KW_{C-LOSS} = 1800$	$KW_{C-LOSS-N} = 600$
$I_{C-LOSS} = 4.72$	$I_{C-LOSS-N} = 4.72$

NOTE - THE CONDENSER INDICATED BY BROKEN LINE CIRCLE SERVES AS A SPARE DURING NORMAL OPERATION BUT IS REQUIRED FOR THE EMERGENCY CONDITION.

LINE CHARACTERISTICS

LENGTH OF TRANSMISSION 226 MILES
 CONDUCTORS— 3-806,000 CM ST. REINFORCED ALUMINUM
 DIAMETER OF CONDUCTORS 96"
 MAXIMUM ELEVATION 500 FEET
 SPACING— EQUIVALENT TO 21' DELTA SPACING



CALCULATION FOR RECEIVING-END CURRENT

$I_L = 168.62 - j 121.97 \text{ AMPS. TO VECTOR } E_{RN}$
 $I_C = 4.72 + j 118.00$
 $I_T = 1.85 - j 13.90$
 $I_R = 203.39 - j 17.87$
 $= 204.17 \angle 6^{\circ} 01' 16'' \text{ AMPERES}$

CALCULATION FOR HIGH TENSION CIRCUIT

$E_{RN}(A) = 110,607 + j 2,783$ $I_R(A) = 177.60 - j 11.11$
 $I_R(B) = 10,762 + j 41.866$ $E_{RN}(C) = -1.14 + j 147.09$
 $E_{SN} = 121,369 + j 44,639$ $I_S = 176.36 + j 135.66$
 $= 222.69 \angle 37^{\circ} 38' 01'' \text{ AMPERES}$
 $= 129,318 \angle 20^{\circ} 11' 35'' \text{ VOLTS}$

$KV-A_S = 129,318 \times 222.69 \times 3 = 86,363 \text{ PER 3 PHASE CIRCUIT}$
 $KV-A_{CO} = 129,318 \times 147.09 \times 3 = 67,064 \text{ PER 3 PHASE CIRCUIT}$

DETERMINATION OF LOSSES

$KW_{LN} = 26,000$
$KW_{RN} = 127,020 \times 204.17 \text{ (COS } 5^{\circ} 01' 16'') = 26,836$
$KW_{SN} = 129,318 \times 222.69 \text{ (COS } 17^{\circ} 26' 28'') = 27,474$

LOSSES TO NEUTRAL

CONDENSER	800
LOWERING TRANSFORMERS (IRON LOSS)	936
(COPPER LOSS $204.17^2 \times 3.186$)	133
RAISING TRANSFORMERS (IRON LOSS)	235
(COPPER LOSS $222.69^2 \times 3.186$)	158
HIGH TENSION LINE $27,474 - (26,836 + 133)$	1608
TOTAL LOSS TO NEUTRAL =	2687

EFFICIENCY

EFFICIENCY, (HIGH TENSION LINE)	$\frac{26,988}{27,474} = 94.52$
EFFICIENCY, (GENERATORS TO LOAD)	$\frac{26,000}{27,887} = 88.71$

% ERROR IN RESULTS

VOLTAGE AT SENDING END	= -1.83 %
CURRENT AT SENDING END	= +1.06 %
POWER FACTOR AT SENDING END	= +1.02 %
KV-A AT SENDING END	= -0.98 %
LOSS IN HIGH TENSION LINE	= +2.38 %

NOTE— THESE PERCENTAGES OF ERROR ARE BASED UPON THE ASSUMPTION THAT THE COMPLETE METHOD PRODUCES 100 % VALUES.—THE MINUS SIGNS SIGNIFY RESULTS TOO LOW; THE PLUS SIGNS RESULTS TOO HIGH.

netizing current were not taken into account, (as it also is in both these methods) the error resulting from the use of the approximate method would be considerably greater than the above values.

The simplified graphical approximate solution illustrated by Chart XXIV will yield results sufficiently accurate for preliminary work, although for final results it should be supplemented by a mathematical solution and, in cases of very long lines, a complete mathematical solution might be desirable. A complete solution as given by Chart XXIII may be followed as a guide in such cases.

The method of obtaining the auxiliary constants corresponding to the approximate solution is given below. The linear constants of the circuit including transformer impedance are determined as follows:—

	Resistance (Ohms)	Reactance (Ohms)
Line.....	34.650	178.20
Transformers.....	3.185	39.82
Total.....	37.835	218.02

Dividing these total values by 225 we obtain the following as the impedance per mile of the combined circuit.

$$r = 0.1681 \text{ ohms}$$

$$x = 0.969 \text{ ohms}$$

TABLE ZZZ—AUXILIARY CONSTANTS FOR 220 KV PROBLEM, APPROXIMATE SOLUTION

Calculated	From Wilkinson Chart	From Kennelly Chart
$a_1 = 0.870783 = 100\%$	0.892 = 102.44% 0.868 = 99.68% (corrected)	0.8713 = 100.05%
$a_2 = 0.021911 = 100\%$	0.0221 = 100.86%	0.02206 = 100.68%
$b_1 = 34.5653 = 100\%$	34.3 = 99.23%	34.561 = 99.99%
$b_2 = 208.83 = 100\%$	211.2 = 101.14%	208.92 = 100.04%
$c_1 = -0.000009 = 100\%$	-0.00001 = 111.11%	-0.000009 = 100%
$c_2 = 0.001158 = 100\%$	0.001163 = 100.43%	0.001159 = 100.09%

The admittance per mile is assumed the same as before namely:—

$$b = 5.38 \times 10^{-6} \text{ mho}$$

$$g = 0$$

From Wilkinson's Charts

$$a_1 = 0.892$$

and since $rb = 0.904$

$$a_2 = 0.221$$

$$b_1 = 34.3 \text{ ohms}$$

$$b_2 = 211.2 \text{ ohms}$$

and since $rb^2 = 4.865$

$$c_1 = -0.000010$$

$$c_2 = 0.001163$$

From Dr. Kennelly's Charts

$$Z = 37.835 + j 218.02$$

$$= 221.28 \angle 80^\circ 09' 23''$$

$$Y = 0 + j 0.001211$$

$$= 0.001211 \angle 90^\circ$$

$$ZY = 0.26797 \angle 170^\circ 09' 23''$$

$$\theta = \sqrt{ZY} = 0.5177 \angle 85^\circ 04' 41''$$

from Chart XIX $\frac{\text{Sinh } \theta}{\theta} = 0.957 \angle 6^\circ 45'$

$$= 0.957 \angle 6^\circ 27' 00''$$

from Chart XXI $\frac{\text{Tanh } \theta}{\theta} = 1.098 \angle 1^\circ 00' 00''^*$

$$A = \frac{\text{Sinh } \theta / \theta}{\text{Tanh } \theta / \theta} = \frac{0.957 \angle 6^\circ 27' 00''}{1.098 \angle 1^\circ 00' 00''}$$

$$= 0.8716 \angle 1^\circ 27' 00''$$

$$a_1 = 0.8713$$

$$a_2 = 0.02206$$

$$B = Z \frac{\text{Sinh } \theta}{\theta} = 221.28 \angle 80^\circ 09' 23'' \times 0.957 \angle 6^\circ 27' 00''$$

$$= 211.76 \angle 80^\circ 36' 23'' \text{ ohms}$$

$$b_1 = 34.561 \text{ ohms}$$

$$b_2 = 208.92 \text{ ohms}$$

$$C = Y \frac{\text{Sinh } \theta}{\theta} = 0.001211 \angle 90^\circ \times 0.957 \angle 6^\circ 27' 00''$$

$$= 0.0011589 \angle 90^\circ 27' 00''$$

$$c_1 = -0.000009$$

$$c_2 = 0.001159$$

The auxiliary constants as obtained graphically and by exact mathematical results are given in Table ZZZ.

The same remarks in regard to use of the Kennelly charts for obtaining the auxiliary constants as given under the complete solution also apply when the approximate solution is used. Wilkinson chart A, if used when transformer impedance is added to the line impedance, as in the approximate method, requires a correction to constant a_1 . Constant a_2 as read from this chart will be correct but constant a_1 as read from the chart will be too high for the following reason. Constant c_1 accounts for the rise in voltage along the line at zero load due to the charging current flowing through the line inductance adding directly to the sending end voltage. The section of Wilkinson chart A applying to constant a_1 is based upon distance and frequency only, so that values read from this section would be the same for a given

distance and frequency regardless of whether or not transformer impedance is included with the line constants. This section of chart A therefore takes account only of the voltage lowering effect of the charging current flowing through the line inductance. In addition to this, it flows also through the transformer inductance, which further lowers the value of a_1 . The value of a_1 read from the chart must therefore be reduced. From the chart, $a_1 = 0.892$ volt corresponding to a voltage rise of 0.108 volt which results from a linear conductance reactance of 178.02 ohms. Actually the reactance of the circuit including lowering transformers is 218.02 ohms or 22.5 percent greater. Increasing 0.108 volt by 22.5 percent we get 0.132 volt rise, so that a_1 becomes $1.000 - 0.132 = 0.868$, which is 99.68 percent of the calculated results.

In the following solutions calculated values for the auxiliary constants are used since exact results are required for the purpose of comparing the results with those previously obtained by the complete solution.

*This was interpolated since this angle lies beyond the range of this chart.

EMERGENCY LOAD—COMPLETE SOLUTION

The complete solution for emergency load conditions shown by Chart XXV follows the same construction as covered by Chart XXIII for normal load. The difference being that the load is doubled and the condenser capacity for a circuit increased nearly four times. Thus to force double the amount of power through the line and transformer impedance, with the same voltage drop, it is necessary in this case, nearly to quadruple the condenser capacity per circuit. Thus to meet the emergency condition nearly double the total condenser capacity will be required. This large increase in condenser capacity necessitated drawing the current vectors to one half the scale used for current vectors in the normal load diagram.

EMERGENCY LOAD—APPROXIMATE SOLUTION

The approximate solution for emergency load shown by Chart XXVI follows the same construction as in Chart XXIV for normal load with the exception of increased load and condenser capacity.

ZERO LOAD—COMPLETE SOLUTION

The complete solution for zero load is shown by Chart XXVII. In this case the load is made up of a lagging phase modifier load and the leakage of the lowering transformers. The same constructions are used as for the other complete solutions.

ZERO LOAD—APPROXIMATE SOLUTION

The approximate solution for zero load is shown by Chart XXVIII. It may be seen from the tabulated errors that this approximate method produces at zero load larger errors than the corresponding errors for loaded conditions. This is usually of little importance, however, as the light load conditions are generally not important.

PHASE MODIFIER CURVES

Frequently the normal and maximum amount of power to be transmitted is known; that is the transmission line, condensers and transformers are designed for a certain maximum load and it is of little importance what condenser capacity would be required for other loads or for various sending end voltages. At other times, especially in preliminary surveys, such data may be very necessary.

In Fig. 70 are plotted curves* showing the phase modifier capacity required to produce certain voltages at the sending end corresponding to various receiving-end loads at 85 percent power-factor and 220 kv. At 85 percent power-factor and 220 kv 200 000 kw is approximately the maximum amount of power which may be transmitted through the lowering transformers and over this line of three 605 000 circ. mil. cables if the sending end voltage is not permitted to exceed 230 kv. This is indicated by the fact that the curve corresponding to this load becomes flat when it reaches the 230 kv horizontal line. To deliver this maximum load at 220 kv through the impedance of this line will require a total condenser capacity of about 300 000 kv-a. The economic capacity of the line is reached at loads very much below the maximum theoretical limit of 200 000 kw.

The sending end voltages corresponding to various

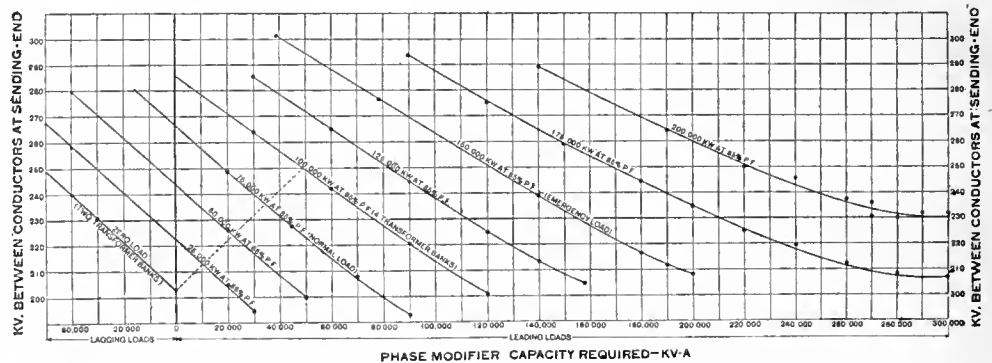


FIG. 70—PHASE MODIFIER CAPACITY REQUIRED TO MAINTAIN CONSTANT RECEIVER VOLTAGE.

These curves indicate for a constant load power-factor of 85 percent lagging and constant load voltage of 220 kv, the amount of energy which may be delivered to the load over one 225 mile, 60 cycle, three-phase circuit consisting of three 605 000 circ. mil aluminum-steel conductors corresponding to various voltages between conductors at the high-tension side of the raising transformers. The values by which these curves were drawn were determined graphically. For 230 kv at the sending end the maximum amount of power which can be transmitted is approximately 200 000 kw and to force this amount of power through the line impedance will require approximately 300 000 kv-a capacity in phase modifiers.

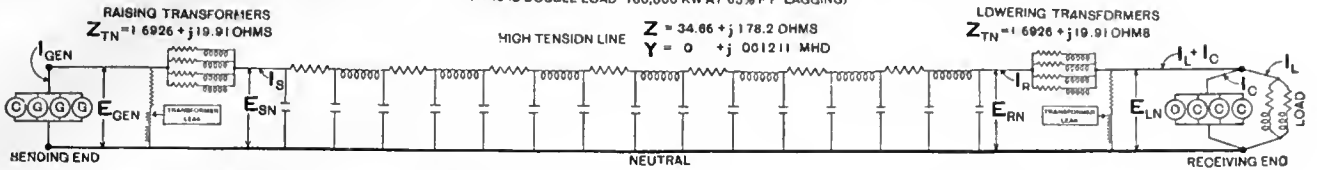
capacities of phase modifiers in parallel with different receiving end loads for drawing curves such as shown by Fig. 70 are most readily obtained by the following graphical procedure. After auxiliary constants *A* and *B* for the circuit under investigation have been determined (preferably through the medium of both the Wilkinson and Kennelly charts) a tabulation of the current to neutral corresponding to each load for which curves are desired is made. A further tabulation of current to neutral for condensers of various capacities is made. The current to neutral which represents the loss in the various condensers, is also tabulated. The resist-

*Such curves were suggested by Mr. F. W. Peek, Jr. in an article on "Practical Calculations of Long Distance Transmission Line Characteristics" in the *General Electrical Review* for June, 1913, p. 430.

CHART XXV—220 KV PROBLEM—EMERGENCY LOAD

(COMPLETE SOLUTION)

(THIS IS DOUBLE LOAD—160,000 KW AT 85% P.F. LAGGING)



EMERGENCY LOAD

PER 3 PHASE CIRCUIT	PER PHASE TO NEUTRAL
$KV-A_L = 178.470$	$KV-A_{LN} = 58.924$
$KW_L = 160,000$	$KW_{LN} = 60,000$
$PF_L = 85\% \text{ LAG.}$	$PF_{LN} = 95\% \text{ LAO.}$
$E_L = 220,000$	$E_{LN} = 127,020$
$I_L = 483.1$	$I_{LN} = 483.1$

60 CYCLES

TRANSFORMERS
(FOUR BANKS IN PARALLEL AT EACH END OF THE LINE)

$R_{TN} = 1.6925 \text{ OHMS RESISTANCE}$ $X_{TN} = 19.91 \text{ OHMS REACTANCE}$

MAGNETIZING CURRENT = 27.8 AMPS. AT 127,020 VOLTS
 IRDN LOSS = 3.70 AMPS. AT 127,020 VOLTS
 IRON LOSS = 470 KW TO NEUTRAL

MAGNETIZING CURRENT = 26.48 AMPS. AT 133,662 VOLTS
 IRON LOSS = 3.62 AMPS. AT 133,662 VOLTS

LINEAR CONSTANTS

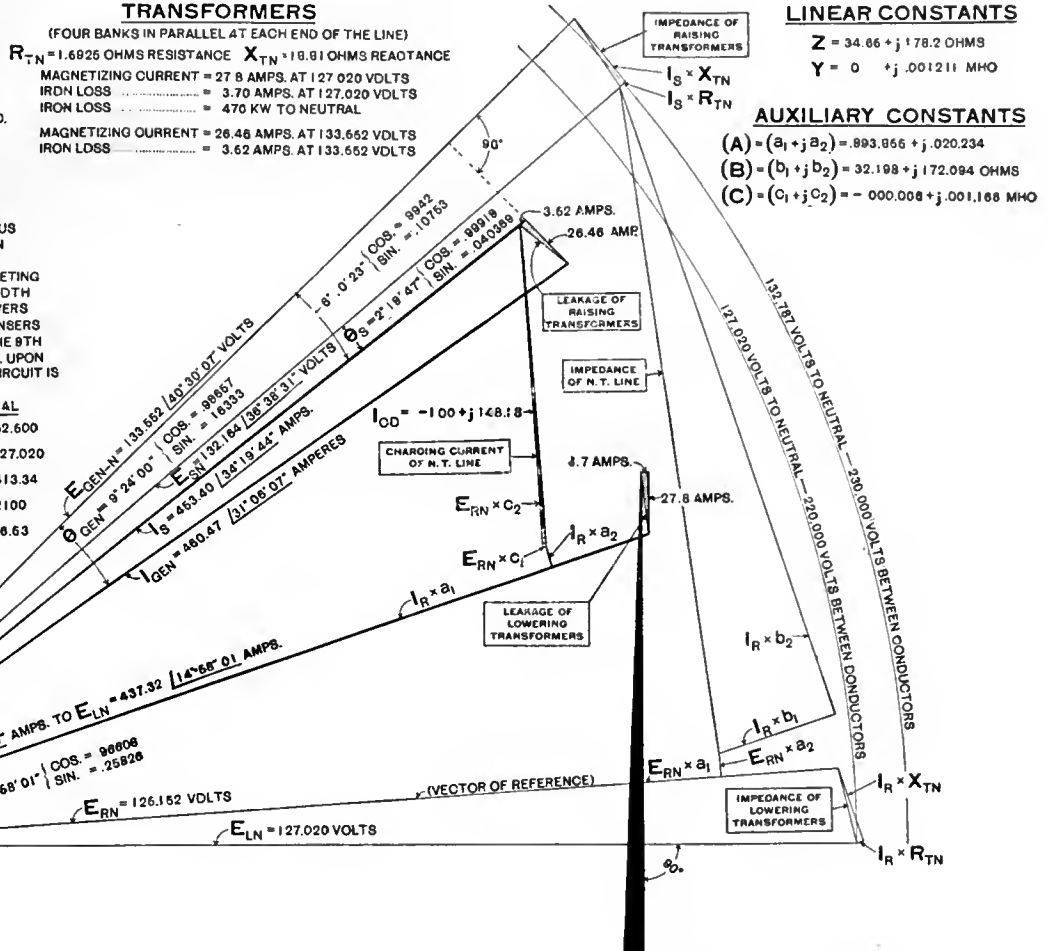
$Z = 34.66 + j178.2 \text{ OHMS}$
 $Y = 0 + j.001211 \text{ MHO}$

AUXILIARY CONSTANTS

(A) = $(a_1 + j a_2) = .893,856 + j .020,234$
 (B) = $(b_1 + j b_2) = 32.198 + j 172.094 \text{ OHMS}$
 (C) = $(c_1 + j c_2) = -.000,008 + j .001,168 \text{ MHO}$

CONDENSERS
FOUR 46,000 KV-A SYNCHRONOUS CONDENSERS PER 3 PHASE CIRCUIT WHEN OPERATING AT FULL LOAD PROVIDE MORE COMPENSATION THAN REQUIRED FOR MEETING THE EMERGENCY CONDITIONS—SINCE BOTH 3 PHASE CIRCUITS ON THE SAME TOWERS WILL BE OPERATED IN PARALLEL 7 CONDENSERS MAY BE USED FOR THE TWO CIRCUITS, THE 8TH CONDENSER BEING AVAILABLE AS A SPARE. UPON THIS BASIS THE CONDENSER DATA PER CIRCUIT IS AS FOLLOWS:

3 PHASE	TO NEUTRAL
$KV-A_C = 167,500$	$KV-A_{CN} = 62,600$
$E_C = 220,000$	$E_{CN} = 127,020$
$I_C = 413.34$	$I_{CN} = 413.34$
$KW_{C-LOSS} = 8300$	$KW_{C-LOSS-N} = 2100$
$I_{C-LOSS} = 18.63$	$I_{C-LOSS-N} = 18.63$

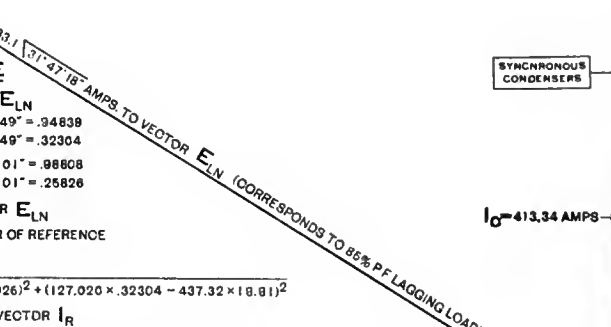


CALCULATION FOR RECEIVING-END CURRENT AND VOLTAGE

$I_L = 383.84 - j243.94 \text{ AMPS. TO VECTOR } E_{LN}$
 $I_C = 16.63 + j413.01$ $\text{COS } 18^\circ 60' 49'' = .94839$
 $I_T = 3.70 - j27.90$ $\text{SIN } 18^\circ 60' 49'' = .32304$
 $I_R = 413.97 + j141.27$ $\text{COS } 14^\circ 58' 01'' = .98608$
 $\text{SIN } 14^\circ 58' 01'' = .25826$

= 437.32 / $18^\circ 60' 49''$ AMPS. TO VECTOR E_{LN}
 = 437.32 / $14^\circ 58' 01''$ AMPS. TO VECTOR OF REFERENCE
 = 422.49 + j112.94

$E_{RN} = \sqrt{(127,020 \times .94839 + 437.32 \times 1.6926)^2 + (127,020 \times .32304 - 437.32 \times 19.91)^2}$
 = 125.152 / $14^\circ 58' 01''$ VOLTS TO VECTOR I_R



DETERMINATION OF LOSSES

$KW_{LN} = 60,000$
 $KW_{RN} = 126.162 \times 437.32 (\text{COS } 14^\circ 58' 01'') = 62,874$
 $KW_{BN} = 132.184 \times 463.40 (\text{COS } 2^\circ 18' 47'') = 69,874$
 $KW_{GEN-N} = 133.662 \times 480.47 (\text{COS } 9^\circ 24' 00'') = 60,871$

LOSSES TO NEUTRAL

LOWERING TRANSFORMERS AND CONDENSERS = 62,874 - 60,000 = 2874
 HIGH TENSION LINE = 69,874 - 62,874 = 7000
 RAISING TRANSFORMERS = 60,871 - 69,874 = -797
 TOTAL LOSS TO NEUTRAL = 10,871

CALCULATION FOR HIGH TENSION CIRCUIT

$E_{RN}(A) = 111.890 + j2.632$ $I_R(A) = 375.42 + j109.61$
 $I_R(B) = -6.833 + j78.344$ $E_{RN}(C) = -1.00 + j149.19$
 $E_{SN} = 106.047 + j79.878$ $I_S = 374.42 + j255.89$
 = 132.164 / $36^\circ 38' 31''$ VOLTS = 463.40 / $34^\circ 18' 44''$ AMPERES

AS A PARTIAL CHECK

LOWERING TRANSFORMERS
 (IRON LOSS 3.7×127.02) = 470
 (COPPER LOSS $437.32^2 \times 1.6926$) = 304
 SYNCHRONOUS CONDENSERS = 2100
 RAISING TRANSFORMERS
 (IRON LOSS 3.62×133.662) = 470
 (COPPER LOSS $463.4^2 \times 1.6926$) = 327
 HIGH TENSION LINE
 (VALUE ABOVE ASSUMED AS CORRECT) = 7000
 413.01 AMPS 10.871
 16.33 AMPS

CALCULATION FOR GENERATOR VOLTAGE AND CURRENT

$E_{GEN-N} = \sqrt{(132.184 \times .98918 + 463.4 \times 1.6926)^2 + (132.184 \times .040359 + 463.4 \times 19.91)^2}$
 = 133.562 / $8^\circ 10' 23''$ VOLTS TO VECTOR I_S
 = 133.562 / $40^\circ 30' 07''$ VOLTS

$KV-A_S = 132.164 \times 463.40 \times 3 = 179,769$ PER 3 PHASE CIRCUIT
 $KV-A_{OEN} = 133.562 \times 480.47 \times 3 = 184,490$ PER 3 PHASE CIRCUIT
 $KV-A_{CC} = 132.164 \times 146.18 \times 3 = 67,958$ PER 3 PHASE CIRCUIT

$I_{GEN-N} = 454.29 - j76.21$
 = 460.47 / $8^\circ 24' 00''$ AMPERES TO VECTOR E_{GEN}
 = 480.47 / $31^\circ 06' 07''$ AMPERES

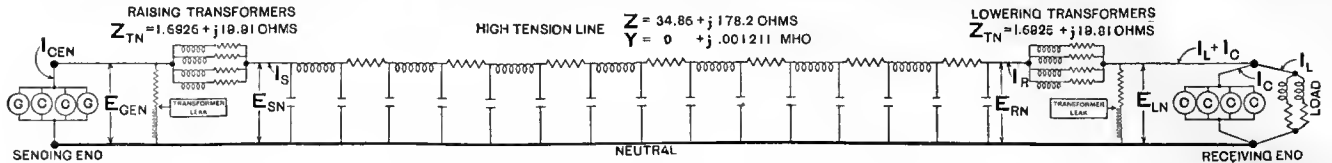
EFFICIENCY

EFFICIENCY (HIGH TENSION LINE) = $\frac{62,874}{69,874} = 68.31\%$
 EFFICIENCY (GENERATORS TO LOAD) = $\frac{60,000}{60,871} = 62.41\%$

CHART XXVI—220 KV PROBLEM—EMERGENCY LOAD

(APPROXIMATE SOLUTION)

THIS APPROXIMATE SOLUTION ASSUMES THAT THE IMPEDANCE OF THE LOWERING TRANSFORMERS MAY BE ADDED TO THE LINE IMPEDANCE AND TREATED AS THOUGH IT WERE DISTRIBUTED LINE IMPEDANCE—THIS ASSUMPTION SIMPLIFIES THE SOLUTION AT THE EXPENSE OF ACCURACY (SEE LOWER RIGHT HAND CORNER OF PAGE; ALSO TEXT)—THE SOLUTION BELOW IS BASED UPON THE VOLTAGE BEING HELD CONSTANT AT THE LOAD SIDE OF THE LOWERING TRANSFORMERS AND AT THE HIGH TENSION SIDE OF THE RAISING TRANSFORMERS—IF THE VOLTAGE IS TO BE HELD CONSTANT AT THE GENERATOR BUS, THE IMPEDANCE OF THE RAISING TRANSFORMERS MUST ALSO BE ADDED TO THAT OF THE LINE—ALL LOW TENSION VALUES ARE REFERRED TO THE HIGH TENSION CIRCUIT.



EMERGENCY LOAD
PER 3 PHASE CIRCUIT

$KV-A_L = 178,470$	$KV-A_{LN} = 68,824$
$KW_L = 160,000$	$KW_{LN} = 60,000$
$PF_L = 85\% \text{ LAG.}$	$PF_{LN} = 86\% \text{ LAG.}$
$E_L = 220,000$	$E_{LN} = 127,020$
$I_L = 483.1$	$I_{LN} = 483.1$

80 CYCLES

LINEAR CONSTANTS
 $Z = 37.835 + j218.02$ OHMS *
 $Y = 0 + j.001211$ MHO

TRANSFORMERS
(FOUR BANKS IN PARALLEL AT EACH END OF THE LINE)
 $R_{TN} = 1.6826$ OHMS RESISTANCE
 $X_{TN} = 19.81$ OHMS REACTANCE

* THIS INCLUDES IMPEDANCE OF LOWERING TRANSFORMERS

AUXILIARY CONSTANTS

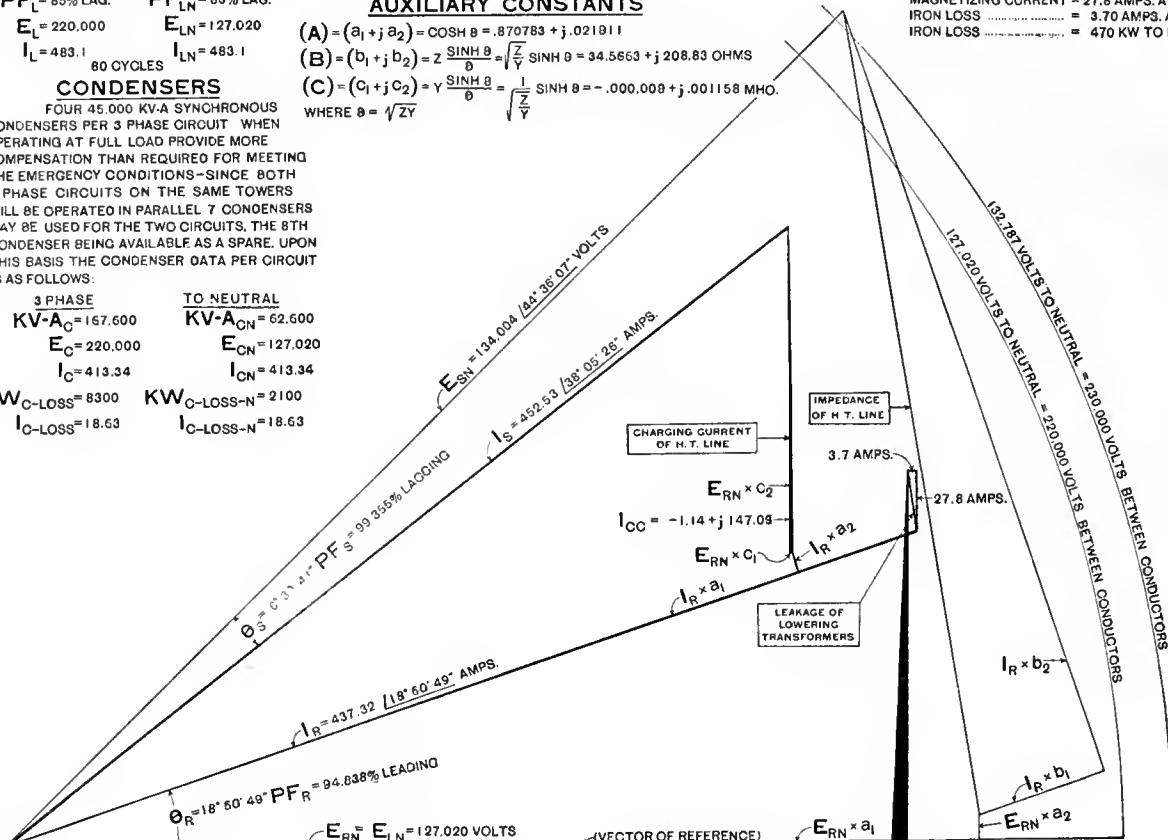
(A) $= (a_1 + j a_2) = \cosh \theta = .870783 + j .021911$
 (B) $= (b_1 + j b_2) = z \frac{\sinh \theta}{\theta} = \sqrt{\frac{Z}{Y}} \sinh \theta = 34.5663 + j 208.83$ OHMS
 (C) $= (c_1 + j c_2) = y \frac{\sinh \theta}{\theta} = \frac{1}{Z} \sinh \theta = -.000,008 + j .001158$ MHO.
 WHERE $\theta = \sqrt{ZY}$

MAGNETIZING CURRENT = 27.8 AMPS. AT 127,020 VOLTS
 IRON LOSS = 3.70 AMPS. AT 127,020 VOLTS
 IRON LOSS = 470 KW TO NEUTRAL

CONDENSERS

FOUR 45,000 KV-A SYNCHRONOUS CONDENSERS PER 3 PHASE CIRCUIT WHEN OPERATING AT FULL LOAD PROVIDE MORE COMPENSATION THAN REQUIRED FOR MEETING THE EMERGENCY CONDITIONS—SINCE BOTH 3 PHASE CIRCUITS ON THE SAME TOWERS WILL BE OPERATED IN PARALLEL 7 CONDENSERS MAY BE USED FOR THE TWO CIRCUITS, THE 8TH CONDENSER BEING AVAILABLE AS A SPARE. UPON THIS BASIS THE CONDENSER DATA PER CIRCUIT IS AS FOLLOWS:

3 PHASE	TO NEUTRAL
$KV-A_C = 167,600$	$KV-A_{CN} = 62,600$
$E_C = 220,000$	$E_{CN} = 127,020$
$I_C = 413.34$	$I_{CN} = 413.34$
$KW_{C-LOSS} = 8300$	$KW_{C-LOSS-N} = 2100$
$I_{C-LOSS} = 18.53$	$I_{C-LOSS-N} = 18.53$



DETERMINATION OF LOSSES

$KW_{LN} = 60,000$
 $KW_{RN} = 127,020 \times 437.32 (\cos 18^\circ 50' 49'') = 62,670$
 $KW_{SN} = 134,004 \times 462.63 (\cos 5^\circ 30' 41'') = 80,248$

LOSSES TO NEUTRAL

CONDENSERS = 2100
 LOWERING TRANSFORMERS (IRON LOSS) = 470
 (COPPER LOSS $347.32^2 \times 1.6826$) = 306
 RAISING TRANSFORMERS (IRON LOSS) = 470
 (COPPER LOSS) $462.63^2 \times 1.6826$ = 320
 HIGH TENSION LINE $80,248 - (62,670 + 306) = 7373$
 TOTAL LOSS TO NEUTRAL = 11,044

EFFICIENCY

EFFICIENCY (HIGH TENSION LINE) = $\frac{62,875}{80,248} = 67.76\%$
 EFFICIENCY (GENERATORS TO LOAD) = $\frac{60,000}{81,044} = 81.61\%$

% ERROR IN RESULTS

VOLTAGE AT SENDING END = + 1.38%
 CURRENT AT SENDING END = - 0.20%
 POWER FACTOR AT SENDING END = - 0.68%
 KV-A AT SENDING END = + 1.18%
 LOSS IN HIGH TENSION LINE = + 5.33%

NOTE—THESE PERCENTAGES OF ERROR ARE BASED UPON THE ASSUMPTION THAT THE COMPLETE METHOD PRODUCES 100% VALUES—THE MINUS SIGNS SIGNIFY RESULTS TOO LOW; THE PLUS SIGNS RESULTS TOO HIGH, 10.53 AMPS

CALCULATION FOR RECEIVING-END CURRENT

$I_L = 393.84 - j 243.84$ AMPS. TO VECTOR E_{RN}
 $I_C = 18.53 + j 413.01$
 $I_T = 3.70 - j 27.80$
 $I_R = 413.87 + j 141.27$
 = 437.32 / $18^\circ 50' 49''$ AMPERES

CALCULATION FOR HIGH TENSION CIRCUIT

$E_{RN}(A) = 110,607 + j 2,783$ $I_R(A) = 367.30 + j 132.08$
 $I_R(B) = -16.198 + j 91.311$ $E_{RN}(C) = -1.14 + j 147.08$
 $E_{SN} = 96,411 + j 94,094$ $I_S = 358.18 + j 279.17$
 = 134,004 / $44^\circ 36' 07''$ VOLTS = 452.53 / $38^\circ 05' 28''$ AMPERES

$KV-A_S = 134,004 \times 462.63 \times 3 = 181,022$ PER 3 PHASE CIRCUIT
 $KV-A_{CC} = 134,004 \times 147.08 \times 3 = 59,132$ PER 3 PHASE CIRCUIT

NOTE:—Linear constant Z, as used in this chart, incorrectly includes impedance of two banks, whereas it should have included four banks of transformers. This error will not, however, materially affect the result.

ance, reactance, iron loss and magnetizing currents of the transformer banks to neutral should also be determined for all capacity transformer banks required. With the above data tabulated any draughtsman can be instructed how to draw vector diagrams of the circuit to determine the sending end voltages corresponding to

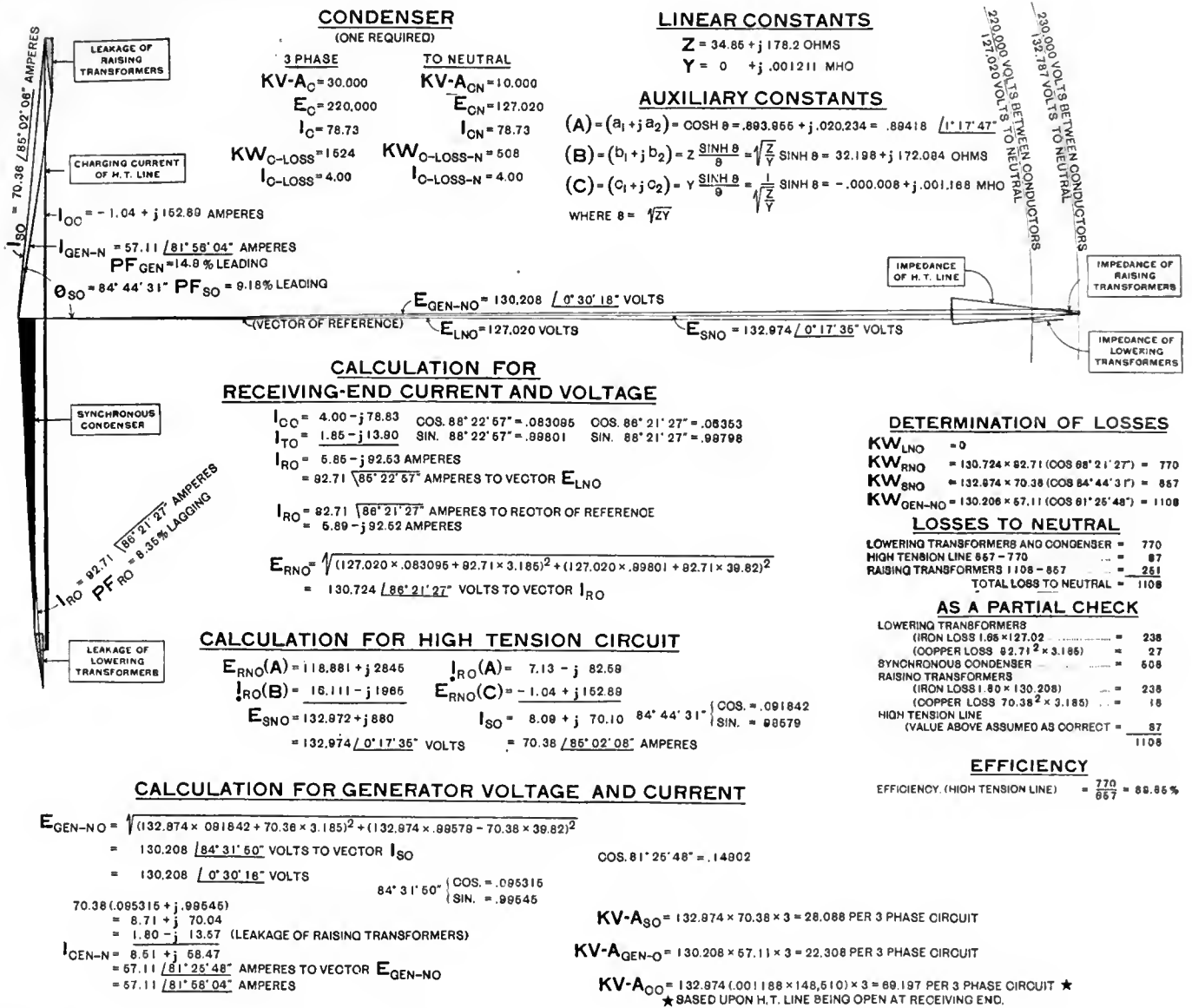
efficient to locate the curve, although more points were calculated for drawing the curves of Fig. 70. This method of obtaining condenser capacities corresponding to sending end voltages is a cut and try method. It has one important advantage in its favor. That is, the results check each other, so that an error in one

**CHART XXVII—220 KV PROBLEM—ZERO LOAD
(COMPLETE SOLUTION)**

(THIS CORRESPONDS TO NORMAL LOAD CONNECTIONS)

AT ZERO LOAD, WITH 230,000 VOLTS MAINTAINED BETWEEN CONDUCTORS (132,787 VOLTS TO NEUTRAL). AT THE HIGH TENSION SIDE OF THE RAISING TRANSFORMERS THE VOLTAGE AT THE HIGH TENSION SIDE OF THE LOWERING TRANSFORMERS (NEGLECTING THE EFFECT OF THE LAGGING MAGNETIZING CURRENT OF THE LOWERING TRANSFORMERS) WILL RISE TO 230,000 DIVIDED BY (A) = 230,000 DIVIDED BY .89418 = 257,219 VOLTS BETWEEN CONDUCTORS (132,187 VOLTS TO NEUTRAL). ACTUALLY THE GREATLY INCREASED LAGGING MAGNETIZING CURRENT OF THE LOWERING TRANSFORMERS WHEN EXCITED BY ABNORMALLY HIGH VOLTAGE WILL NOT PERMIT OF THE RECEIVING END VOLTAGE REACHING SUCH A HIGH VOLTAGE UNLESS THE GENERATOR VOLTAGE RAISES MOMENTARILY TO A VALUE GREATLY IN EXCESS OF 230,000 VOLTS. IF HOWEVER THE LOWERING TRANSFORMERS ARE DISCONNECTED FROM THE CIRCUIT, THE INCREASED LEADING CHARGING CURRENT OF THE LINE, REACTING UPON THE GENERATOR FIELDS, COMBINED WITH A MOMENTARY OVER SPEED OF THE GENERATORS MAY CAUSE THE RECEIVING END VOLTAGE TO GREATLY EXCEED THE ABOVE VALUE.

IN ORDER TO HOLD THE VOLTAGE AT THE RECEIVING END CONSTANT AT 220,000 VOLTS BETWEEN CONDUCTORS (127,020 VOLTS TO NEUTRAL) AT ZERO LOAD IT WILL BE NECESSARY TO PLACE AN ARTIFICIAL LAGGING LOAD AT THE LOAD END OF THE LINE--THIS IS ACCOMPLISHED BY OPERATING ONE OF THE SYNCHRONOUS CONDENSERS WITH ITS FIELDS UNDER EXCITATION--BY CONSTRUCTING SEVERAL VECTOR DIAGRAMS FOR THIS CIRCUIT EACH BASED UPON DIFFERENT VALUES OF REACTOR LOAD, A CURVE MAY BE DRAWN BY PLOTTING THE REACTOR LOADS AGAINST THE CORRESPONDING SENDING END VOLTAGES--FROM THIS CURVE THE REACTOR CAPACITY CORRESPONDING TO 230,000 VOLTS BETWEEN CONDUCTORS AT THE SENDING END WILL BE SEEN TO BE APPROXIMATELY 30,000 KV-A.



the various receiving end loads and different phase modifier capacities.

The graphical method used in determining the values to plot the curves of Fig. 70, is illustrated by Fig. 71. Three solutions are illustrated, two with condensers of different size and one without condensers. Three such solutions for each load will usually be suf-

ficient to locate the curve, although more points were calculated for drawing the curves of Fig. 72. This method of obtaining condenser capacities corresponding to sending end voltages is a cut and try method. It has one important advantage in its favor. That is, the results check each other, so that an error in one

CAPACITY OF PHASE MODIFIERS

The curves of Fig. 70 show that, for a constant delivered load, power-factor and voltage, the leading

capacity of phase modifiers required goes down as the line drop increases. For instance 75 000 kw at 85 percent power-factor and 220 kv can be delivered over this line with 230 kv sending end voltage, if 43 000 kv-a condenser capacity is placed in parallel with the load. If, however, a line drop of 20 kv is selected in place of 10 kv, the sending end voltage will be 240 kv and the corresponding condenser load will be reduced to approximately 30 000 kv-a. On the other hand this increased line drop will necessitate a greater capacity

The dotted line in Fig. 70 is simply the zero load line thrown over to the leading load side to facilitate study in phase modifier capacity. For instance, projection from the points where the dotted line intersects a load curve will give the minimum capacity of phase modifier on the bottom scale and the corresponding sending end voltage on the vertical scale to the left. Thus with a load of 75 000 kw, intersection of the dotted line with this load curve indicates that 33 000 kv-a phase modifier capacity will be required both at this load and at zero

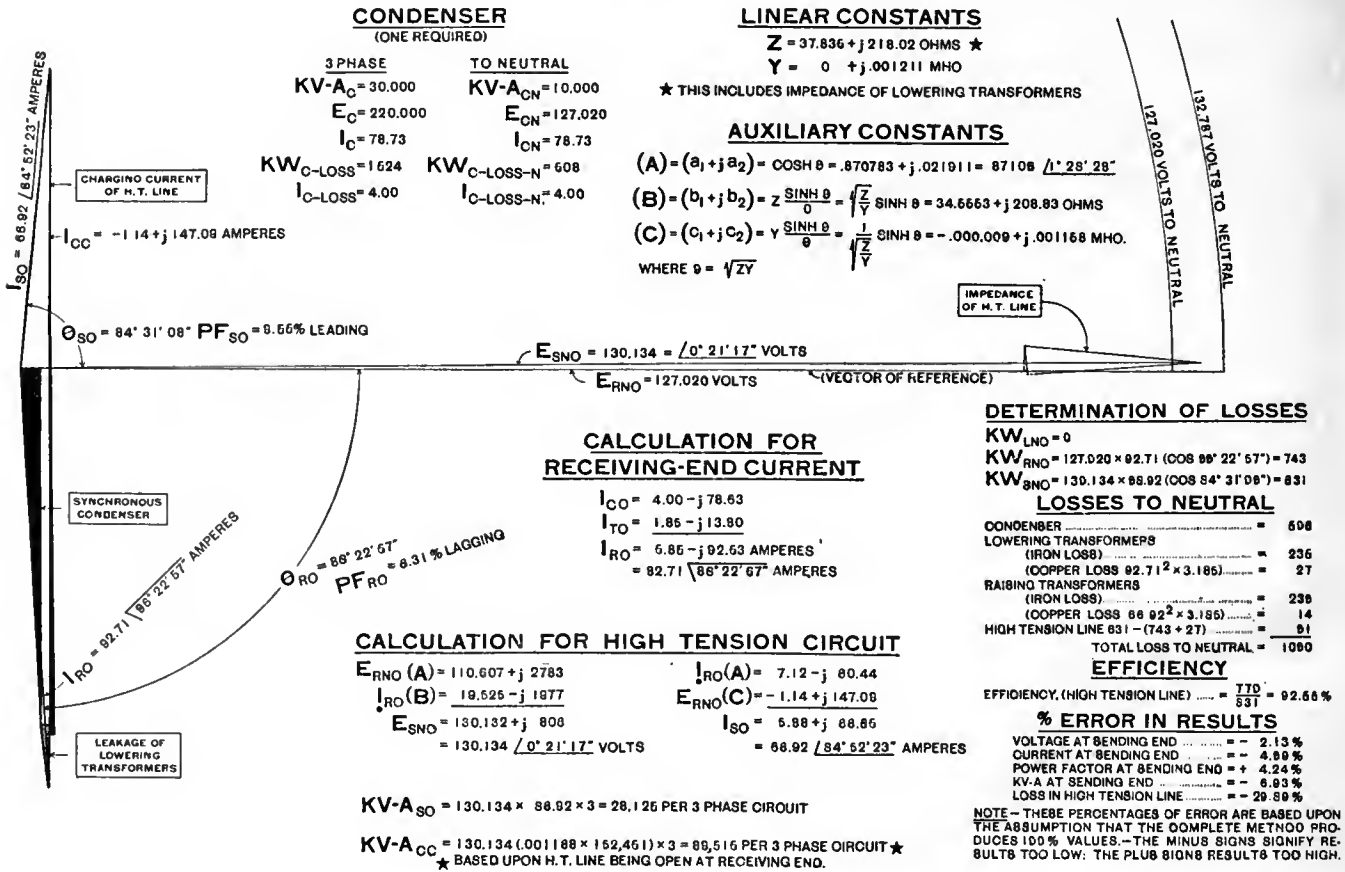
**CHART XXVIII—220 KV PROBLEM—ZERO LOAD
(APPROXIMATE SOLUTION)**

(THIS CORRESPONDS TO THE NORMAL LOAD CONNECTIONS)

THIS APPROXIMATE SOLUTION ASSUMES THAT THE IMPEDANCE OF THE LOWERING TRANSFORMERS MAY BE ADDED TO THE LINE IMPEDANCE AND TREATED AS THOUGH IT WERE DISTRIBUTED LINE IMPEDANCE—THIS ASSUMPTION SIMPLIFIES THE SOLUTION AT THE EXPENSE OF ACCURACY (SEE LOWER RIGHT HAND CORNER OF PAGE). ALSO TEXT—THE SOLUTION BELOW IS BASED UPON THE VOLTAGE BEING HELD CONSTANT AT THE LOAD SIDE OF THE LOWERING TRANSFORMERS AND AT THE HIGH TENSION SIDE OF THE RAISING TRANSFORMERS—IF THE VOLTAGE IS TO BE HELD CONSTANT AT THE GENERATOR BUS, THE IMPEDANCE OF THE RAISING TRANSFORMERS MUST ALSO BE ADDED TO THAT OF THE LINE—ALL LOW TENSION VALUES ARE REFERRED TO THE HIGH TENSION CIRCUIT.

AT ZERO LOAD, WITH 230,000 VOLTS MAINTAINED BETWEEN CONDUCTORS (132,787 VOLTS TO NEUTRAL) AT THE HIGH TENSION SIDE OF THE RAISING TRANSFORMERS THE VOLTAGE AT THE HIGH TENSION SIDE OF THE LOWERING TRANSFORMERS (NEGLECTING THE EFFECT OF THE LAGGING MAGNETIZING CURRENT OF THE LOWERING TRANSFORMERS) WILL RISE TO 230,000 DIVIDED BY (A)=230,000 DIVIDED BY 87108 = 264,046 VOLTS BETWEEN CONDUCTORS 132,451 VOLTS TO NEUTRAL. ACTUALLY THE GREATLY INCREASED LAGGING MAGNETIZING CURRENT OF THE LOWERING TRANSFORMERS WHEN EXCITED BY APPROXIMATELY HIGH VOLTAGE WILL NOT PERMIT OF THE RECEIVING END VOLTAGE REACHING SUCH A HIGH VOLTAGE UNLESS THE GENERATOR VOLTAGE RAISES MOMENTARILY TO A VALUE GREATLY IN EXCESS OF 230,000 VOLTS. IF HOWEVER THE LOWERING TRANSFORMERS ARE DISCONNECTED FROM THE CIRCUIT, THE INCREASED LEADING CHARGING CURRENT OF THE LINE, REACTING UPON THE GENERATOR FIELDS, COMBINED WITH A MOMENTARY OVER SPEED OF THE GENERATORS, MAY CAUSE THE RECEIVING END VOLTAGE TO GREATLY EXCEED THE ABOVE VALUE.

IN ORDER TO HOLD THE VOLTAGE AT THE RECEIVING END CONSTANT AT 220,000 VOLTS BETWEEN CONDUCTORS (127,020 VOLTS TO NEUTRAL) AT ZERO LOAD IT WILL BE NECESSARY TO PLACE AN ARTIFICIAL LAGGING LOAD AT THE LOAD END OF THE LINE—THIS IS ACCOMPLISHED BY OPERATING ONE OF THE SYNCHRONOUS CONDENSERS WITH ITS FIELDS UNDER EXCITATION BY CONSTRUCTING SEVERAL VECTOR DIAGRAMS FOR THIS CIRCUIT EACH BASED UPON DIFFERENT VALUES OF REACTOR LOAD. A CURVE MAY BE DRAWN BY PLOTTING THE REACTOR LOADS AGAINST THE CORRESPONDING SENDING END VOLTAGES.—FROM THIS CURVE THE REACTOR CAPACITY CORRESPONDING TO 230,000 VOLTS BETWEEN CONDUCTORS AT THE SENDING END WILL BE SEEN TO BE APPROXIMATELY 30,000 KV-A.



at zero load in order to maintain 240 kw constant at the sending end. Thus with 230 kv at the sending end, about 30 000 kv-a reactor load will be required at zero load, whereas with 240 kv at the sending end, about 40 000 kv-a reactor load will be required at zero load.

Obviously the smallest phase modifier capacity possible to maintain regulation is one in which full capacity leading will be required under maximum load and full capacity lagging under zero load. At half load such a phase modifier would operate at near zero kv-a.

load and that the corresponding sending end voltage will be approximately 236 kv. At 100 000 kw load, nearly 50 000 kv-a phase modifier capacity will be required, and the corresponding sending end voltage would be 250 kv.

As previously stated, phase modifiers which may be operated at rated load both lagging and leading are special, and cost more than standard phase modifiers. On account of unstable operation due to weakened field, standard condensers usually cannot be operated at lag-

ging loads above approximately 70 percent of their full load leading rating. To deliver 75 000 kv-a at 85 percent power-factor requires approximately 42 000 kv-a in phase modifier capacity with 230 kv at the sending end. To maintain the sending end voltage of 230 kv at zero

which determines the total capacity of phase modifiers, for the 220 kv problem. For instance at normal load, 43 000 kv-a in capacity is required, whereas for the double or emergency load 157 000 kv-a capacity (nearly four times) is required. This large increase is due to the fact that the line charging current (which tends to reduce phase modifier capacity under load) has not changed, and that the line impedance volts has become twice as much, making it necessary to turn the line impedance triangle through a large angle in the counter-clockwise direction in order that the sending end voltage be not increased.

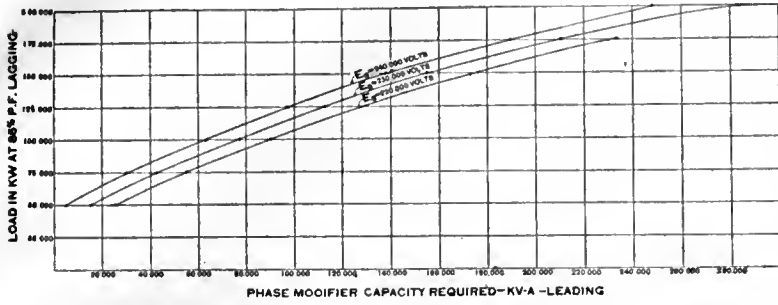


FIG. 72—PHASE MODIFIER CAPACITY REQUIRED FOR THE VARIOUS LOADS

These curves are plotted from values read from the curves of Fig. 70 and are on the basis of a constant load voltage of 220 kv.

load requires approximately 30 000 kv-a lagging. This is 70 percent of the capacity leading, thus permitting of employing a standard 43 000 kv-a condenser. To provide margin a 45 000 kv-a standard condenser might be selected for this normal load condition.

Under emergency conditions (that is, double or 150 000 kw load at 85 percent power-factor) 157 000 kv-a phase modifier capacity will be required if 230 kv is not to be exceeded at the sending end. If the generator can be operated during the emergency condition at increased voltage of, for instance, 240 kv, the phase modifier capacity could be reduced to approximately 140 000 kv-a. However, too much liberty in variation of generator operating voltage should not be taken. If the voltage is held constant at the high-voltage side of the raising transformers, the generator operating voltage will have to be varied to compensate for the regulation of the sending end transformers, and to provide a still greater range in generator operating voltage might impose a hardship on the generator designers. The voltage drop through the transformers is small under load conditions, since the power-factor will be near unity, but under zero load condition the drop will be considerable, due to the low power-factor, especially if a large phase modifier load is required at zero load. It will be seen that it is the emergency condition

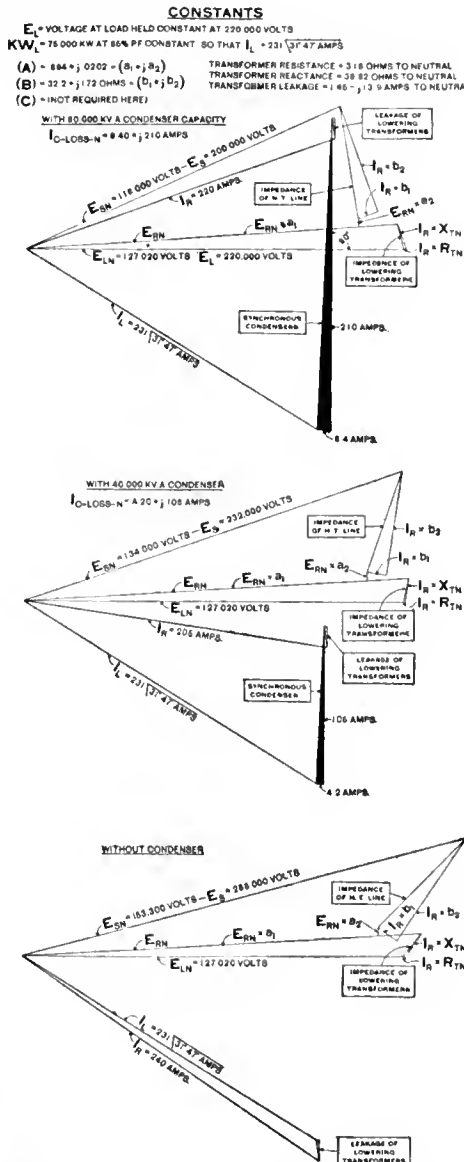


FIG. 71—GRAPHIC METHOD FOR DETERMINING THE VOLTAGE AT THE SENDING END.

Corresponding to different condenser loads in parallel with a constant power load of 75 000 kw at 85 percent power-factor and 220 kv. The results as plotted in Fig. 70 were obtained by similar constructions.

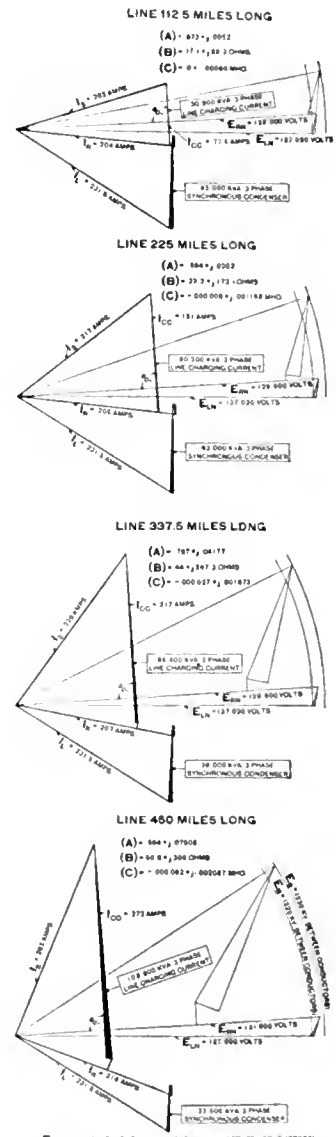


FIG. 73—VECTOR DIAGRAMS SHOWING THE EFFECT OF THE LENGTH OF THE LINE ON THE PHASE MODIFIER CAPACITY REQUIRED

The diagrams represent a three-phase, 60 cycle circuit, consisting of three 605 000 circ. mil aluminum steel reinforced conductors, when delivering 75 000 kw at 85 percent lagging power-factor at a load voltage of 220 kv with a sending end voltage of 230 kv.

The zero load curve on Fig. 70 is drawn for the normal load connection; that is, for two 50 000 kv-a transformer banks in parallel. For the emergency load four transformer banks in parallel will be required. The result of the increased magnetizing current consumed by four in place of two transformer banks will be to reduce the capacity of phase modifiers required under zero load. A second zero load line could be added, covering four transformer banks. Such a line would lie directly above the one for two transformer banks but would not materially affect the results. For load conditions of 100 000 kw at 85 percent power-

this feeds a net work on which condensers are required for voltage control.

It may be desired to investigate the effect of line charging current on phase modifier capacity for lines of different lengths. For this purpose the vector diagrams Fig. 73, and the phase modifier curves, Fig. 74, were prepared. These vector diagrams and curves are based upon a constant load of 75 000 kw at 85 percent power-factor delivered at 220 kv and a line drop of 10 kv. In other words the only variable for the four different lines is the length and this varies in equal increments.

The vector diagrams of Fig. 73 show the influence of line charging current upon condenser capacity. As the length of the line increases, the influence of the in-

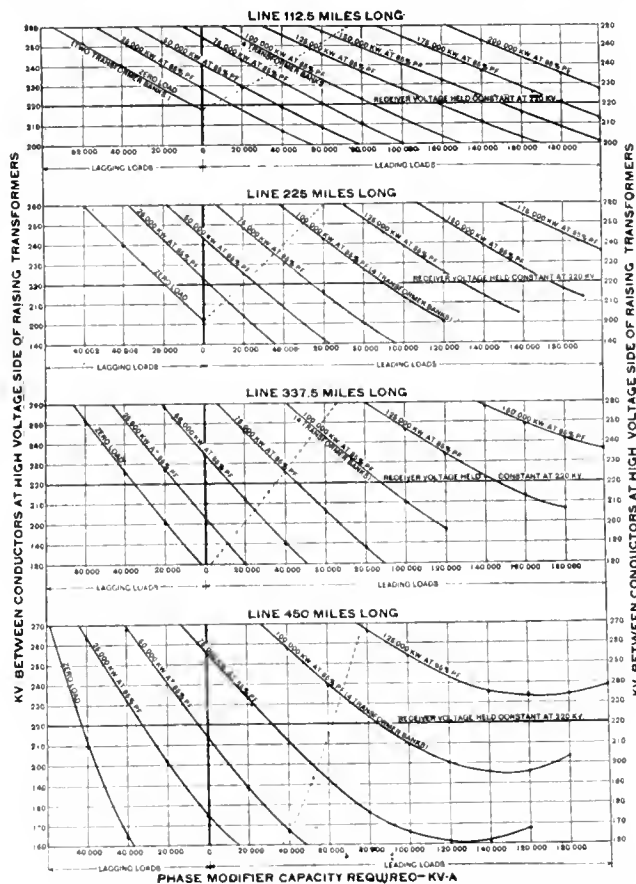


FIG. 74—CURVE SHOWING THE RELATION BETWEEN PHASE MODIFIER CAPACITY AND SENDING END VOLTAGE

For various receiving end loads of 85 percent lagging power-factor and a constant load voltage of 220 kv. These curves apply to a three phase, 60 cycle circuit consisting of three 605 000 circ. mil aluminum steel conductors. The vector construction of these four lines is shown in Fig. 73.

tor and above, the points for the curves were determined on the basis of four transformer banks.

In the above it was assumed that the power-factor of the load would be 85 percent lagging. A long line such as this would probably feed into an extended distribution net work, having numerous load centers. At these load centers synchronous condensers would probably be located for the purpose of holding the voltage constant. This would necessitate operating the condenser leading at heavy loads thus raising the power-factor of the entire system under load, and in effect reducing the capacity of phase modifiers required for voltage control at the receiving end of the line. This point should be investigated where a long line such as

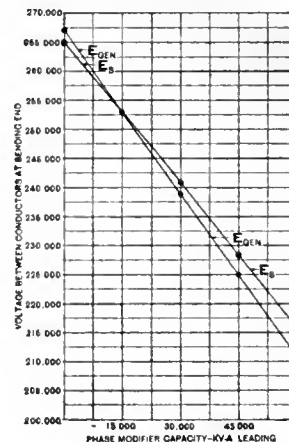


FIG. 75—CURVES SHOWING THE VOLTAGE ON EACH SIDE OF THE RAISING TRANSFORMERS

Corresponding to condenser loads of various capacities in parallel with a constant load of 75 000 kw at 85 percent power factor lagging and 220 kv. The vertical distance between the two voltage lines is the voltage drop or voltage rise through the raising transformers. For condenser loads up to 15 000 kv-a there is a drop in voltage through the raising transformers. For condenser loads above 15 000 kv-a there is a rise in voltage through the raising transformers.

creased line charging current is toward a reduction in condenser capacity; that is the line itself furnishes a large part of the leading current necessary to maintain the proper line voltage drop. If this line were longer than 450 miles, the line charging current at a certain length would be sufficient in itself to maintain the desired voltages at the two ends of the line without the aid of condensers. In such a case, however, a large reactor capacity would be required at zero and low loads to hold the receiving end voltage at a constant value.

The reason that a short line may necessitate more condenser capacities for voltage control than a long line is simple. For the 112.5 mile line the charging current will be about one half as much as for a 225 mile line. Since the line is only half as long this smaller charging current will flow through only half the inductance so that the net result of half the line charging current and half the inductance will be about one fourth the voltage

boosting effect due to line charging current. On the other hand the line impedance will be only half as great, but the net result will be more condenser capacity for the short line. A large part of the condenser capacity is required for neutralizing the lagging reactor component of the load.

Auxiliary constant A , as previously explained, accounts for the effect of the line charging current flowing through the impedance of the circuit; that is, the voltage boosting effect of the charging current. Thus for the 112.5 mile line (Fig. 73) a_1 which accounts for the line charging current flowing through the inductance of the circuit is near unity and a_2 near zero, but for the 450 mile line a_1 drops to 0.594 and a_2 increases to 0.07508. As the length of line increases, constant A moves the line impedance triangle to the left and raises its toe somewhat. The increased line impedance and

slightly increased current at the receiving end increases the size of the line impedance triangle.

The curves of Fig. 74 show the relation between phase modifier capacity and sending end voltage for different receiving end loads of 85 percent lagging power-factor and a constant load voltage of 220 kv. It is interesting to note the effect of distance for fixed size conductors upon the maximum amount of power which can be transmitted over a circuit, as evidenced by the load curves bending upward as the line length increase. It is also interesting to note the decrease in phase modifiers leading capacity and increase in phase modifier lagging capacity as the line becomes larger, as evidenced of the load curves shifting to the right. The curves, Fig. 75, show the voltage at each side of the raising transformer, corresponding to various condenser capacities in parallel with a constant load of 75 000 kw at 85 percent lagging power-factor and 220 kv.

H. B. DWIGHT'S METHOD.

In the various methods for determining the performance of transmission lines which are described above, current and voltage vectors or corresponding vector quantities have been employed throughout. It was believed that solutions embodying the use of current and voltage vectors would be the more easily followed by the young engineer, for the assistance of whom this book has been primarily written.

H. B. Dwight worked out and published in book form formulas for determining the complete performance of circuits by the employment of quantities not generally employed in the methods described above. These quantities require a new set of symbols applicable to his method. Partly to prevent confusion in symbols but principally because his method has been so completely and clearly set forth and illustrated with numerous examples worked out in the two books referred to his method has not been detailed in this book. To include it here would simply be a duplication of what is already available in very complete form.

THE CIRCLE DIAGRAM

Various forms of circle diagrams as an aid in determining the performance of *short* transmission lines have been frequently described by writers, notably by R. A. Philip thru the medium of the A. I. E. E. transactions of February 1911. Following this H. B. Dwight worked out a solution and construction for a circle diagram which accurately takes into account the effect of capacitance in transmission lines that is, a circle diagram for *long* high voltage lines. This circle diagram consists of curves which indicate the phase modi-

fier capacity (leading or lagging) required to maintain a certain receiving end voltage corresponding to all values of delivered load up to the maximum capacity of the line. In other words it gives data such as is given by the curves of Fig. No. 70.

The next step in the development of the circle diagram was to so alter the constants upon which it is constructed that it will take accurately into account the localized impedance and loss in raising or lowering transformers or in both. Of course the transformer impedance may be added to the line impedance as is frequently done and considered as distributed line impedance. Such procedure, will, however, in the case of the circle diagram for the line alone result in objectionable errors in the results. In order to correctly apply the circle diagram to *long* lines so as to accurately include the effect of transformers in the circuit it is necessary to develop new formulas for obtaining values for the constants by which the circle diagram is constructed. See articles on transmission line constants by R. D. Evans and H. K. Sels in the Electric Journal, page 306 July 1921, page 356 August 1921 and page 530 December 1921.

To the expert who spends much time investigating transmission problems the general use of the circle diagram should be of great assistance. It indicates performance at all loads, which with other methods would have to be obtained by a separate calculation or vector diagram construction for each load.

*Transmission Line Formulas, 1913, D. Van Nostrand Co., New York City and Constant-voltage Transmission, 1915, John Wiley & Sons Inc., New York City.

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